







font





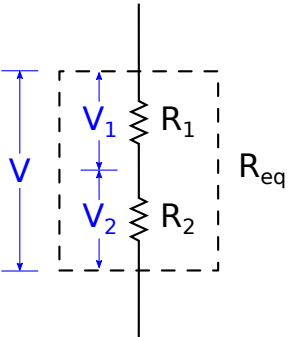




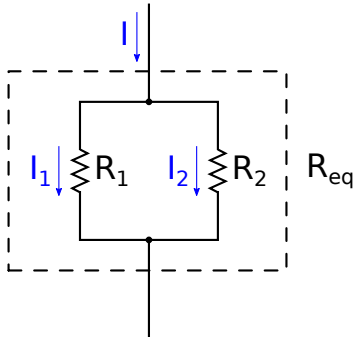




$$P_{\text{diss}} = I^2 R = \frac{V^2}{R}$$



Series  
**(a)**

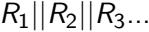


Parallel  
**(b)**

BR

Repeal RI + RI + RI + .

$$\frac{1}{R_{eq}} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} + \dots$$









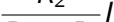
$R_1$  $V$  $R_1$  $+$  $R_2$



$R_2$  $V$  $R_1$  $+$  $R_2$



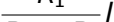
$R_2$



$R_1 + R_2$



$R_1$



$R_1 + R_2$

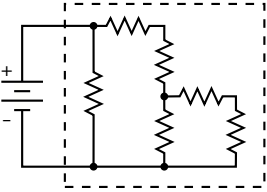




















## Resistor Network Analysis

1. Use the equivalent resistance expressions (Equations 3 and 4) to find the equivalent resistance of the entire network.
2. Use Ohm's law to calculate the current flowing through the voltage source.
3. Use the current division expressions (Equations 6) at each node to determine the currents in the network.
4. Use Ohm's law to determine the voltage drops across the resistors and from these, the voltages of all of the nodes.

$\Sigma$  $v_i$  $=$  $0$ 

loop



+

-



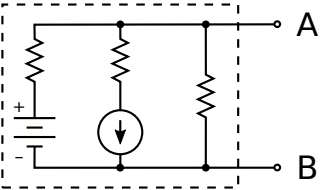
$\Sigma$ 

$$I_i = 0$$

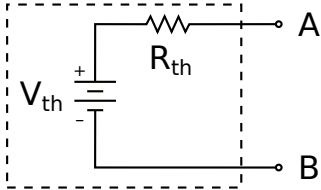
junction

## Circuit Analysis using Kirchhoff's Rules

1. Draw the circuit with arrows and labels representing all of the unique currents.
2. Construct a set of loop-rule equations including every device in the circuit at least once.
3. Construct a set junction-rule equations including every current in the circuit at least once.
4. Given enough known quantities (usually the properties of the devices in the circuit — resistances, capacitances, source voltages, e.g.), these equations can be combined algebraically to determine the unknown quantities (usually currents and potential differences in the circuit).



(a)



(b)

## Finding the Thévenin Equivalent of a Circuit

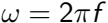
1. The Thévenin equivalent voltage  $V_{th}$  is the voltage across the output terminals  $A$  and  $B$  of the circuit with no load attached.
2. The Thévenin equivalent resistance  $R_{th}$  is the resistance of the circuit measured between the output terminals  $A$  and  $B$ , with all voltage and current sources replaced with their internal resistances. *Ideal voltage sources have zero resistance, and ideal current sources have infinite resistance.*



# CIVIL

The mnemonic CIVIL can be used to remember that the voltage across a capacitor lags  $\pi/2$  (half a period of oscillation) behind the current (CIV), and the voltage across an inductor leads the current by  $\pi/2$  (VIL). The voltage across a resistor is always in phase with the current.













$$X_C = - \frac{j}{\omega C}$$









$$\operatorname{Re}(v e^{i \omega t + \phi}) = v \cos(\omega t + \phi)$$



$$\operatorname{Re}(e^{i\omega t + \phi}) = \cos(\omega t + \phi)$$

















Revised 10/1/20

Real Estate Advisor





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$$\langle P \rangle = \frac{V}{2} \cos(\phi_v - \phi_i) = \frac{V^2}{2} \cos(\phi_v - \phi_i)$$

cos<sup>2</sup>wt

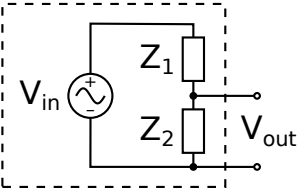






## Phasor Analysis

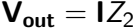
- Equivalent impedance calculations work the same way as equivalent resistance calculations (Equations 3 and 4).
- Kirchhoff's rules work for phasors  $\mathbf{V}$  and  $\mathbf{I}$ .
- Thévenin's theorem applies to AC circuits containing only linear components. The Thévenin voltage and impedance are found via the process outlined in Section 1.5 with impedance in place of resistance.







win 1200



$$v_{out} = \frac{z_2}{z_1 + z_2} v_{in}$$

W

o

ut

$v_{out}$

$=$

$\sqrt{v_{out} * v_{out}}$



$$\tan \phi = \frac{\operatorname{Im} \left( \frac{Z_2}{Z_1 + Z_2} \right)}{\operatorname{Re} \left( \frac{Z_2}{Z_1 + Z_2} \right)}$$





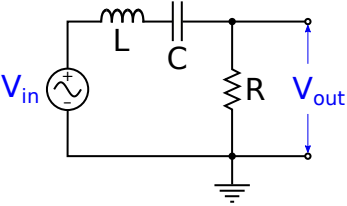
$$A_v = \frac{V_{out}}{V_{in}} = \frac{|V_{out}|}{|V_{in}|} = \left| \frac{Z_2}{Z_1 + Z_2} \right| = \sqrt{\left( \frac{Z_2}{Z_1 + Z_2} \right)^* \left( \frac{Z_2}{Z_1 + Z_2} \right)}$$



$A_v$

$=$

$\frac{1}{\sqrt{2}}$





$$Z_1 = j \left( \omega L - \frac{1}{\omega C} \right)$$

$$\frac{V_{out}}{V_{in}} = \sqrt{\frac{R^2}{R^2 + (\omega L - 1/\omega C)^2}}$$





$$\frac{Z_2}{Z_1 + Z_2} = \frac{R}{R + j\left(\omega L - \frac{1}{\omega C}\right)} = \frac{R^2 - jR\left(\omega L - \frac{1}{\omega C}\right)}{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}$$

$$\tan \phi = \frac{\frac{1}{\omega C} - \omega L}{R}$$

voilà

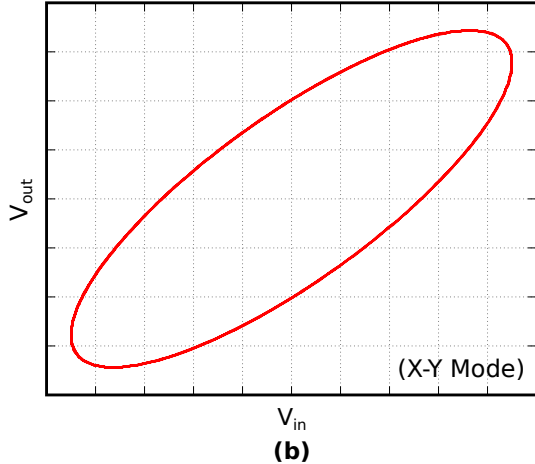
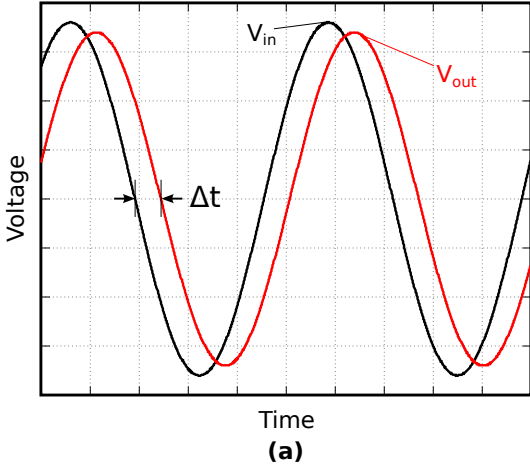
Wavelength

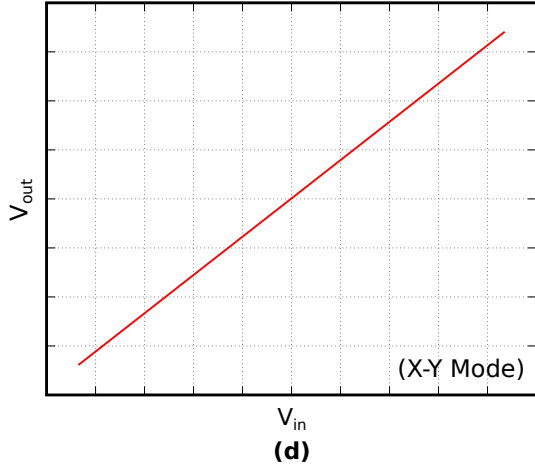
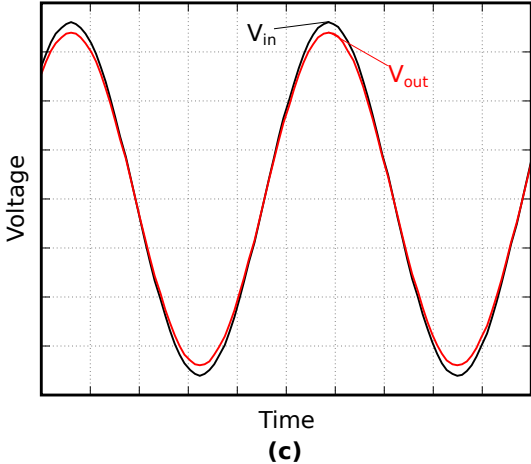


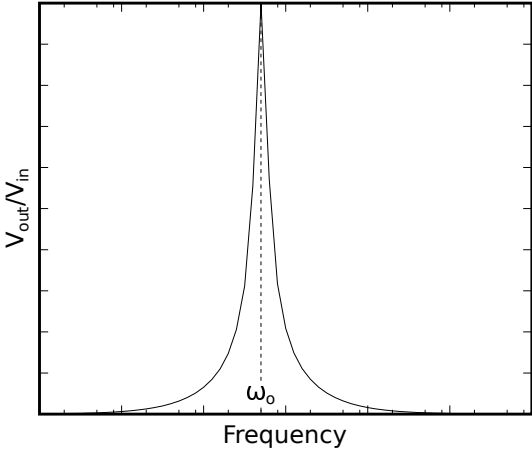
$$\Delta t = \frac{\phi}{\omega} = \frac{\phi}{2\pi f} = \frac{\phi}{2\pi} T$$



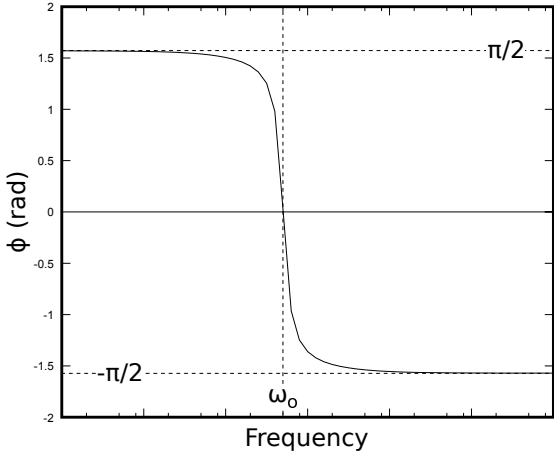








**(a)**



**(b)**

over

—

or





$$\omega_0 = \frac{1}{\sqrt{LC}}$$

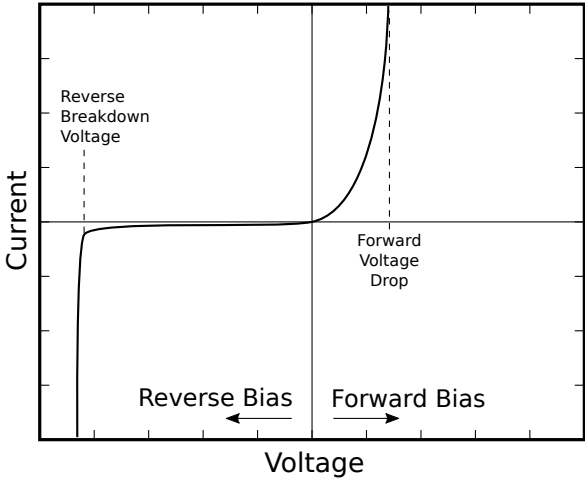


WAVE







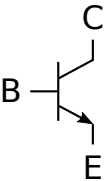






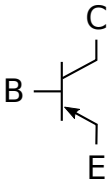






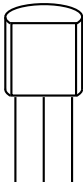
NPN

**(a)**



PNP

**(b)**



E B C

**(c)**



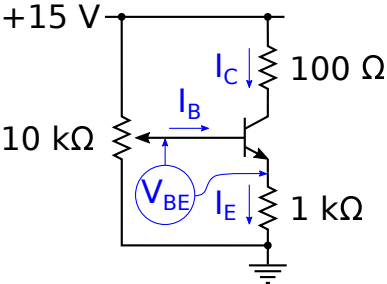


$$I_C = I_0 \left( e^{\frac{V_{BE}}{kT/e}} - 1 \right)$$



100% 100%

1930-2020









$$r_e = \frac{dV_{BE}}{dI_C} = \frac{kT/e}{I_C}$$

Learn from the best



W E

5

0

.

25







0

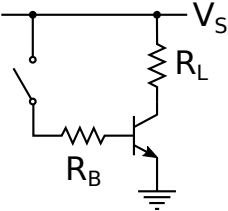
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12345



## Basic Transistor Behavior

- In order for a transistor to function, make  $V_C > V_E$ , and keep  $I_B$ ,  $I_C$ , and  $V_{CE}$  below the rated maximum values of the transistor.
- **On:** In saturation,  $V_{BE} \approx 0.7 \text{ V}$ ,  $V_{CE} \approx 0.25 \text{ V}$ , and  $I_C = h_{FE} I_B$ , where  $h_{FE} \approx 100$ .
- **Off:** If  $V_{BE} < 0.7 \text{ V}$  (significantly),  $I_C \approx 0$ .
- **Active Region:**  $V_{BE}$  is between “off” and saturation. The collector current is governed by the Ebers-Moll equation (Equation 29). The emitter resistance at room temperature is  $r_e = (25 \text{ } \Omega) / I_C [\text{mA}]$ .







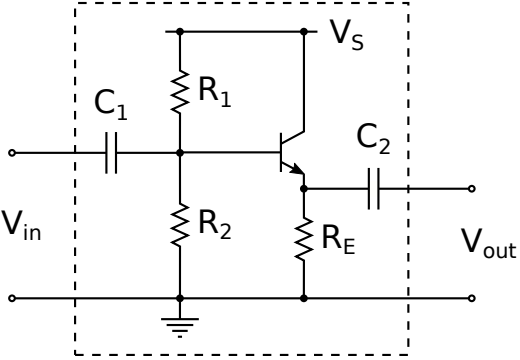




















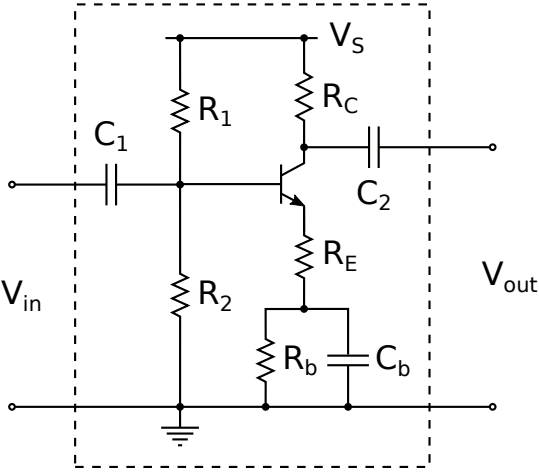
W E \* O O W



APPRESENTATION







$$A_v = -\frac{R_C}{R_E}$$











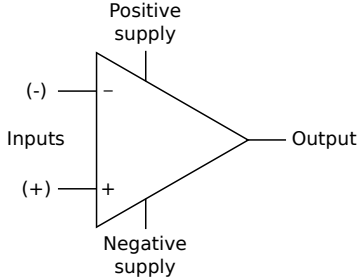




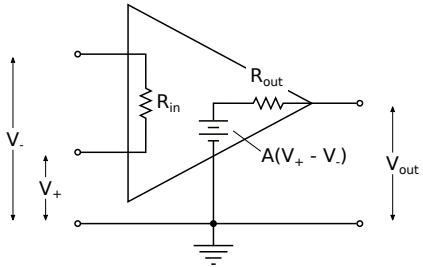
ARIZONA REFERENCE + ABILITY



REAR + REAR



**(a)**



**(b)**







WAVE LOVE

100% **over** 100%













1

0

9

1

0

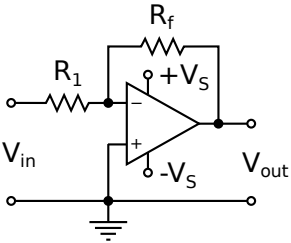
5



## Basic Op Amp Behavior

- $R_{in} = \infty$ ,  $A = \infty$ ,  $R_{out} = 0$
- **Linear region:**  $V_+ = V_-$ ,  $-V_S < V_{out} < V_S$
- **Saturation:**  $V_+ > V_- \implies V_{out} = V_S$  or  $V_+ < V_- \implies V_{out} = -V_S$







$$I = \frac{V_{in}}{R_1} = \frac{-V_{out}}{R_f}$$

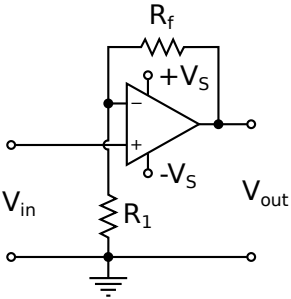


$$A_v = -\frac{R_f}{R_1}$$



Q2



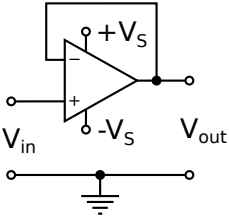


$$V_{in} = \frac{R_1}{R_1 + R_f} V_{out}$$

$$A_v = 1 + \frac{R_f}{R_1}$$

Real + Real

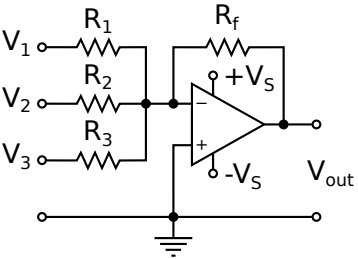












V1



R1

V2



R2





V3



R3



$V_{out}$

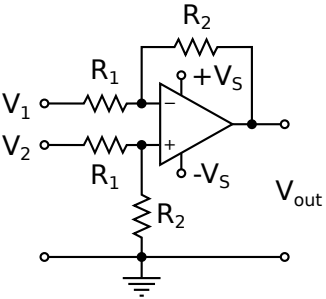


$R_f$



$$V_{out} = - \left( \frac{R_1}{R_f} V_1 + \frac{R_2}{R_f} V_2 + \frac{R_3}{R_f} V_3 \right)$$









$V_1$

—

$V_L$

---

$R_1$

$V_2$

$-$

$V_+$

---

$R_1$

$$V_L - V_{out}$$



$$R_2$$





$$V_+ = \frac{R_2}{R_1 + R_2} V_2$$

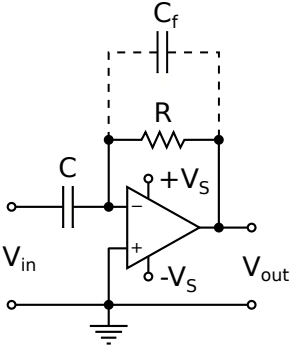


$$V_{out} = -\frac{R_2}{R_1}(V_1 - V_2)$$















$V_{in}$

$=$

$\frac{Q}{C}$

1

=

dQ

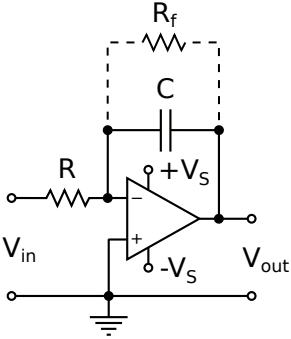
—

dt



$$V_{out} = -RC \frac{dV_{in}}{dt}$$





$I$

$=$

$$\frac{V_{in}}{R}$$

$$V_{out} = - \frac{Q}{C}$$



$$Q = \int dq = \frac{1}{R} \int v_{in} dt$$

$$v_{out} = -\frac{1}{RC} \int v_{in} dt$$



THE END

T H A N K S

$$dB = 10 \log \left( \frac{\text{Thing}_2}{\text{Thing}_1} \right)$$

Thinner



Therapy



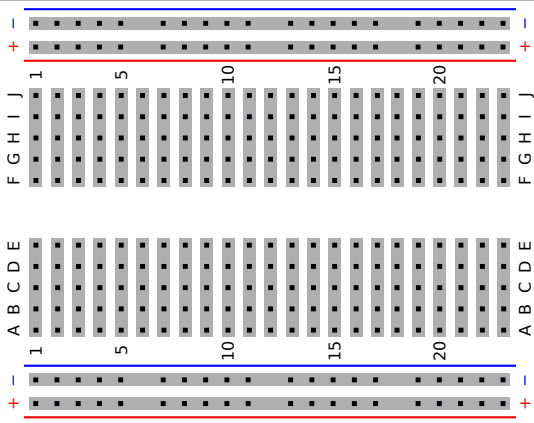




$$10 \log \left( \frac{P_{out}}{P_{in}} \right) = 10 \log \left( \frac{1}{2} \right) = 10(-0.3010) = -3.01$$







Distribution Strips

Terminal Strips

Channel

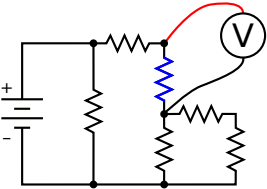
Terminal Strips

Distribution Strips

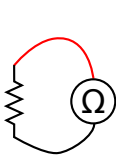


Resubmitted 10/2/2020

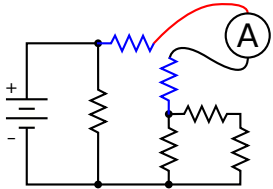
1200



**(a)**



**(b)**



**(c)**