

CONTROLLER DESIGN

Course Project - AE 308

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Problem Formulation:

The given transfer is

$$G(S) = \frac{S + 500}{S(S + 0.0325)(S^2 + 2.57S + 6667)} \quad (1)$$

Given conditions:

$$K_V = \text{Ramp Error Constant} = 100$$
$$\text{Phase Margin} > 35^\circ$$

Now, using these above conditions, we need to design a controller using Root Locus Method and we need to determine Resonant Frequency, Resonant Peak and Bandwidth of the Compensated System. Also determine the Transient Response of the System.

Theory:

Lets look how different types of controllers impact the closed loop behaviour of a given system.

a) P Controller:

- (i) Modifies overall loop gain and hence steady state behaviour.
- (ii) Change dominant pole location and thereby relative and transient response.

b) PD Controller:

- (i) It can reduce the overshoot of a proportional controller because PD controller takes into account the rate of change in error.
- (ii) It can improve system tolerances to external disturbances.

c) PI Controller:

- (i) It is mainly used for tracking of step and ramp inputs.
- (ii) Eliminates steady state error.

d) PID Controller:

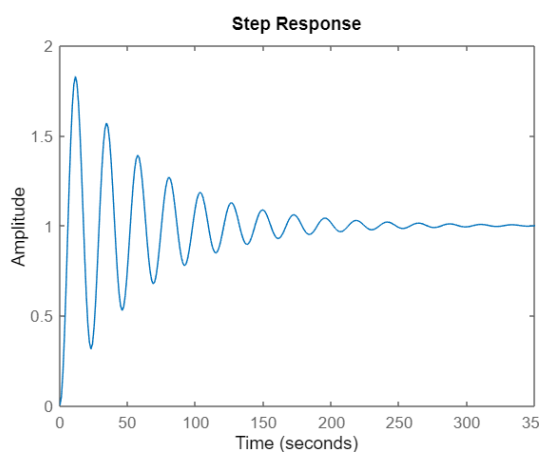
- (i) It can reduce overshoot of system.
- (ii) Improve system tolerances to external disturbances.
- (iii) Steady state error decreases.

e) Lead and Lag Compensator:

- (i) A Lead Compensator can increase the stability or speed of response of a system.
- (ii) A Lag Compensator can reduce the steady state error.

Design Strategy:

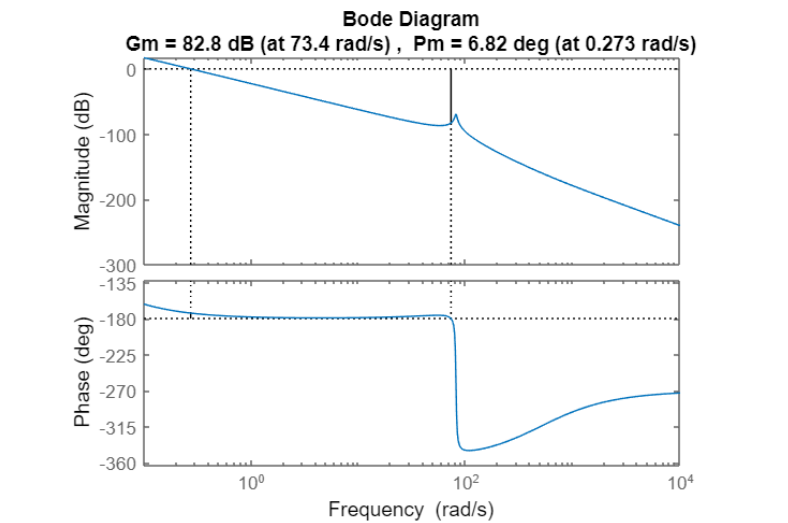
Let us first understand the system given to us observing the step response of the uncompensated system.



RiseTime: 3.9851
SettlingTime: 231.8658
SettlingMin: 0.3127
SettlingMax: 1.8291
Overshoot: 82.9073
Undershoot: 0
Peak: 1.8291
PeakTime: 11.4717

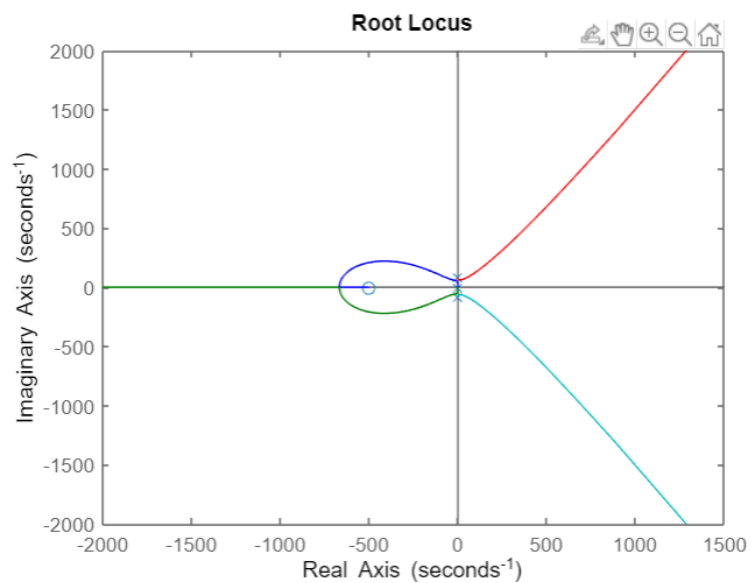
The uncompensated closed loop response has a settling time of 231.689 seconds and O.S of 82.7% which are very large and also has a k_v of 2.3 which is very little compared to our requirement and hence we need to add quite a bit of gain to the system.

Also let us look at the bode plot of the given plant system.



You observe that the phase margin of the uncompensated system is way too less than what is required.

Let us observe the root locus of the given plant.



From the root locus we can see that we need to shift the root locus to the left side and make our poles lie on the right. This is a classic example of Lead compensator.

Lets assume the controller is of the form,

$$K_c \cdot \frac{S + Z_c}{S + P_c} \quad (2)$$

The given uncompensated system,

$$G(S) = \frac{S + 500}{S(S + 0.0325)(S^2 + 2.57S + 6667)} \quad (3)$$

From the condition, $PM = 35^\circ$.

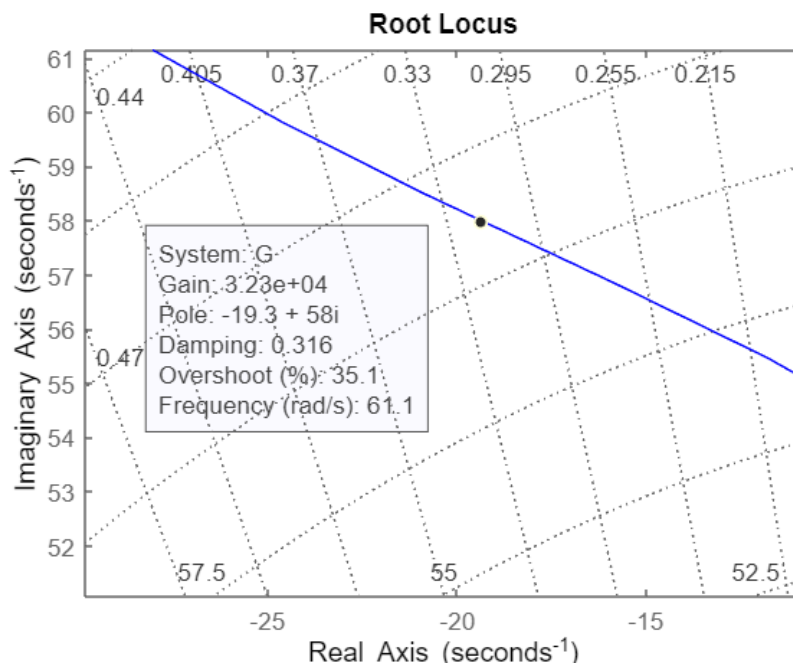
Lets find the damping ratio,

$$\begin{aligned} PM &= \tan^{-1}\left(\frac{2\tau}{\sqrt{-2\tau^2 + \sqrt{1 + 4\tau^4}}}\right) \\ 35^\circ &= \tan^{-1}\left(\frac{2\tau}{\sqrt{-2\tau^2 + \sqrt{1 + 4\tau^4}}}\right) \\ 0.7 &= \frac{2\tau}{\sqrt{-2\tau^2 + \sqrt{1 + 4\tau^4}}} \\ \tau &= 0.316 \end{aligned} \quad (4)$$

Uncompensated:

Search along the line $\tau = 0.316$ and

find the operating point $= -19.3 + 58i$ and Gain $= 3.23 \times 10^4$



$$G(S) = \frac{3.23 \times 10^4(S + 500)}{S(S + 0.0325)(S^2 + 2.57S + 6667)} \quad (5)$$

$$\begin{aligned} K_v &= \lim_{S \rightarrow 0} G(S) \\ &= \frac{32300 \times 500}{0.0325 \times 6667} = 74500 \end{aligned} \quad (6)$$

But our required value = 100.

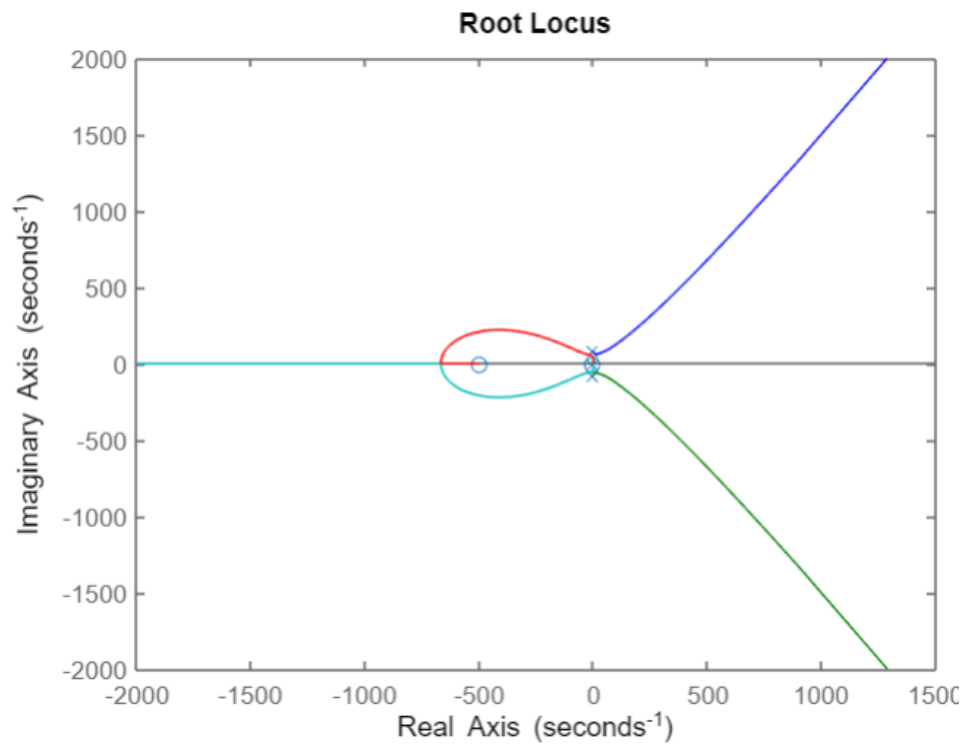
\therefore The improvement of K_v from uncompensated system to compensated system,

$$\frac{100}{74500} = 0.00134 \quad (7)$$

Our Controller $T.f. = \frac{K(S+Z_c)}{S+P_c}$.

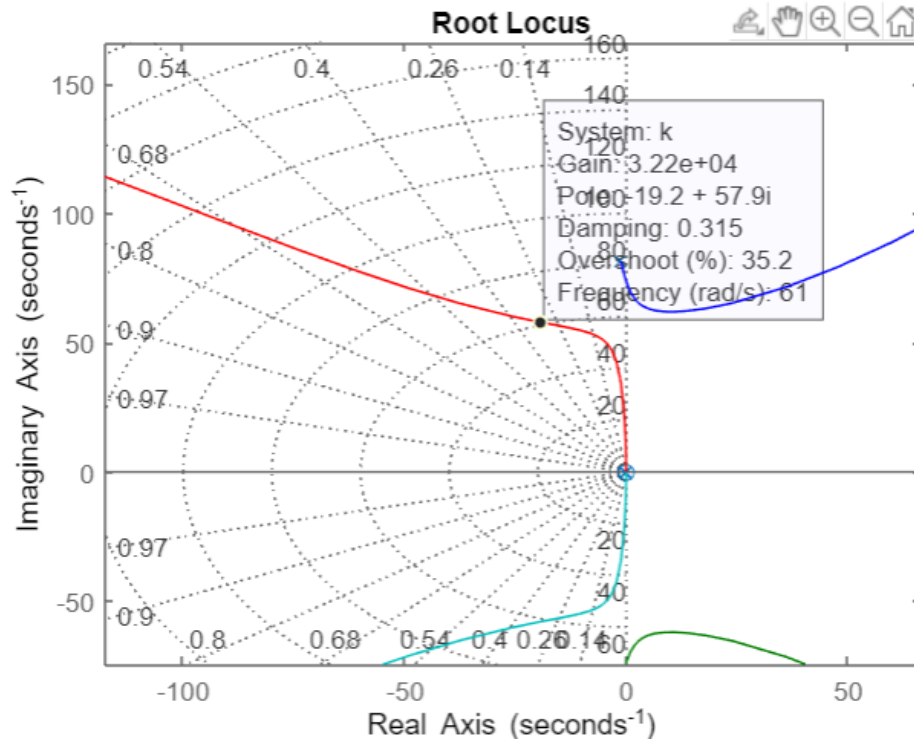
Let us arbitrarily select $Z_c = 0.000134$, $P_c = 0.1$

Now, our $G_c(S) = \frac{K(S+0.000134)}{(S+0.1)}$.



Compensated:

We added a pole at -0.1 and zero at -0.000134. Search along $\tau = 0.316$ line.



We find operating point as $-19.2 + 57.9i$ and Gain $= 3.22 \times 10^4$.

\therefore Our Final Transfer Function becomes,

$$G(S) = 3.23 \times 10^4 \cdot \frac{(S + 0.000134)}{(S + 0.1)} \cdot \frac{(S + 500)}{S(S + 0.0325)(S^2 + 2.57S + 6667)} \quad (8)$$

$$\begin{aligned} K_v &= \lim_{S \rightarrow 0} G(S) \\ &= \frac{3.22 \times 10^4 \times 0.000134 \times 500}{0.1 \times 0.0325 \times 6667} \\ &= 99.56 \end{aligned} \quad (9)$$

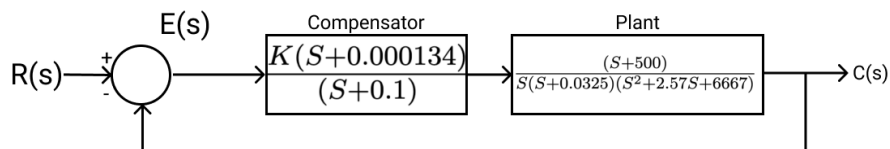
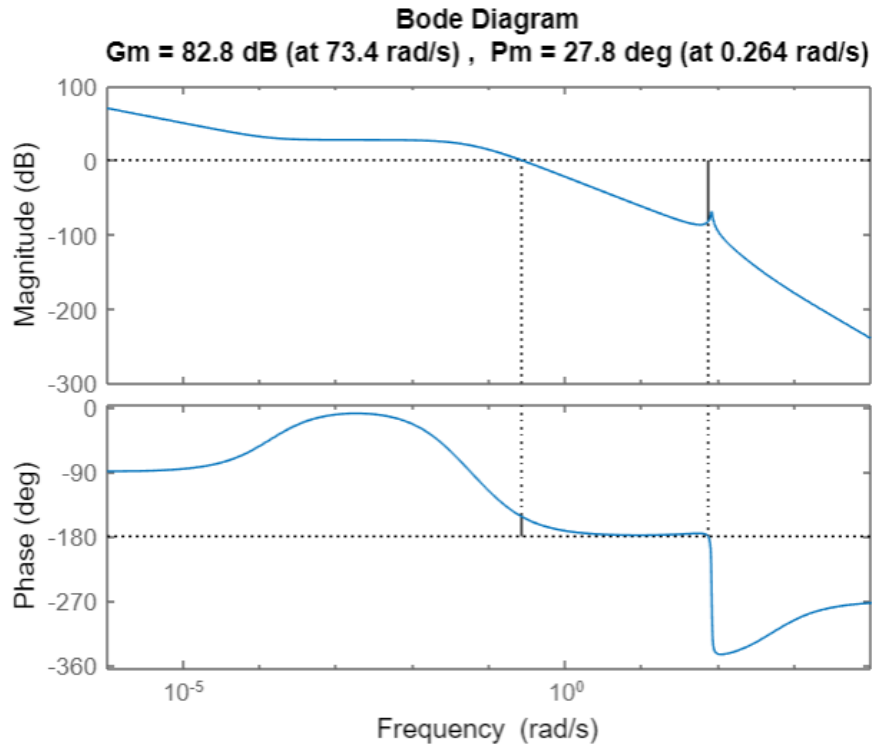


Figure 1: Closed-loop Block Diagram

Let us look at the Bode diagram of the Compensated system.



Parameter	Uncompensated	Compensated
Plant and Compensator	$\frac{K(S+500)}{S(S+0.0325)(S^2+2.57S+6667)}$	$\frac{K(S+0.000134)}{(S+0.1)} \frac{(S+500)}{S(S+0.0325)(S^2+2.57S+6667)}$
K	32300	32200
K_v	74500	99.56
Dominant Second Order Pole	$-19.3 \pm 58i$	$-19.2 \pm 57.9i$

$$\therefore \text{Our Controller} = G(S) = 3.23 \times 10^4 \cdot \frac{(S+0.000134)}{(S+0.1)}$$

with Poles = $-19.2 + 57.9i$

From above Dominant Poles,

$$19.2 = \sigma = \tau\omega_n \quad (10)$$

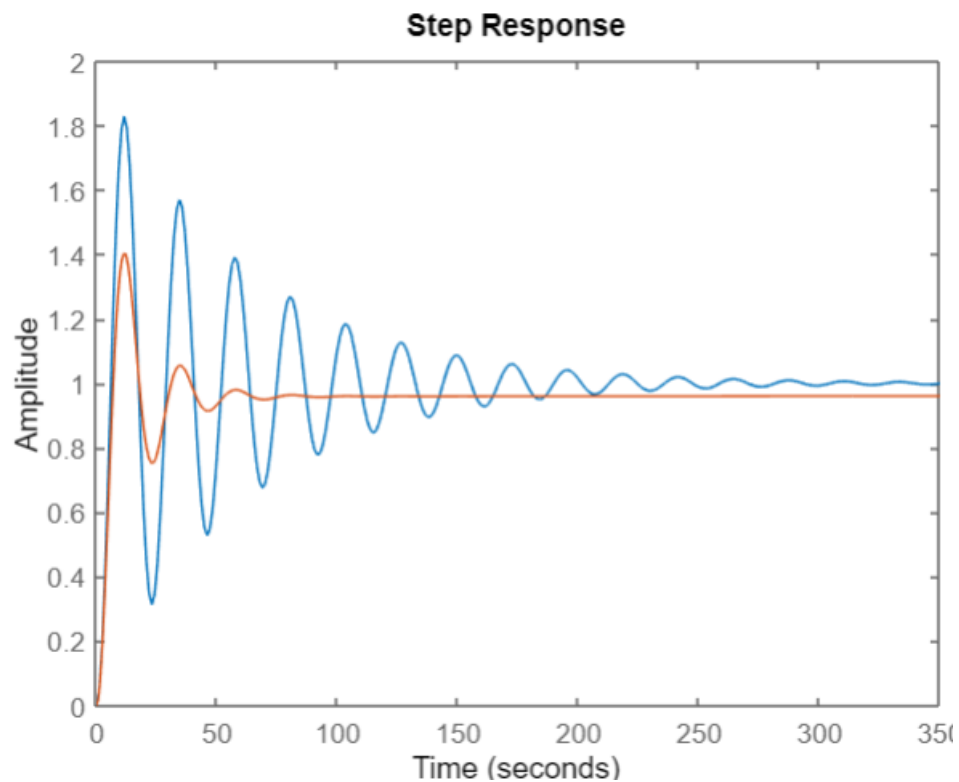
$$\omega_n = 60.79 \text{ rad/s} \quad (11)$$

$$\begin{aligned}
 \text{Resonant Frequency} &= \omega_r = \omega_n \sqrt{1 - 2\tau^2} \\
 &= 60.79 \sqrt{1 - 2(0.316)^2} \\
 &= 54.3 \text{ rad/s}
 \end{aligned} \quad (12)$$

$$\begin{aligned}
 \text{Resonant Peak} &= \frac{1}{2\tau\sqrt{1-\tau^2}} \\
 &= \frac{1}{2 \times 0.316\sqrt{1-(0.316)^2}} \\
 &= 1.667
 \end{aligned} \tag{13}$$

$$\begin{aligned}
 \text{Bandwidth} &= \omega_n(1 - 2\tau^2 + \sqrt{4\tau^4 - 4\tau^2 + 2})^{\frac{1}{2}} \\
 &= 87.69 \text{ rad/s}
 \end{aligned} \tag{14}$$

Let's see both Compensated and Uncompensated transient responses.



RiseTime: 3.9851	RiseTime: 4.6472
SettlingTime: 231.8658	SettlingTime: NaN
SettlingMin: 0.3127	SettlingMin: 0.7518
SettlingMax: 1.8291	SettlingMax: 1.4033
Overshoot: 82.9073	Overshoot: 40.3287
Undershoot: 0	Undershoot: 0
Peak: 1.8291	Peak: 1.4033
PeakTime: 11.4717	PeakTime: 11.8176

```

1 %code for plot1
2 clear all
3 clc
4 s = zpk(0,[],1);
5 Gu = (s+500)/(s*(s+0.0325)*(s^2+ 2.57*s+ 6667));
6 stepinfo(Gu/(1+Gu))
7
8
9 %code for plot2
10 clear all
11 clc
12 s = zpk(0,[],1);
13 k = (s+500)/(s*(s+0.0325)*(s^2+ 2.57*s+ 6667));
14 bode(k)
15
16
17 %code for plot3
18 clear all
19 clc
20 s = zpk(0,[],1);
21 k = (s+500)/(s*(s+0.0325)*(s^2+ 2.57*s+ 6667));
22 rlocus(k)
23
24
25 %Code for plot4
26 clear all
27 clc
28 s = zpk(0,[],1);
29 Gu = (s+0.000134)(s+500)/((s+0.1)*s(s+0.0325)*(s^2+ 2.57*s+ 6667));

```

```
30 rlocus(Gu)
31
32
33 %code for plot5
34 clear all
35 clc
36 s = zpk(0,[],1);
37 k = (s+0.000134)(s+500)/((s+0.1)*s(s+0.0325)*(s^2+ 2.57*s+ 6667));
38 margin(k)
39
40
41 %code for plot6
42 clear all
43 clc
44 s = zpk(0,[],1);
45 Gu = (s+500)/(s*(s+0.0325)*(s^2+ 2.57*s+ 6667));
46
47 Gc = (s+0.000134)/(s+0.1);
48
49 Gce=Gu*Gc;
50 Tu=feedback(Gu,1);
51 Tc=feedback(Gce,1);
52 step(Tu)
53 hold
54 step(Tc)
```