

Why Transform?

- ① Makes important information more accessible
- ② Operations on signals become easier to perform

- high frequency will correspond to abrupt changes - sharp transitions

- low frequency will correspond to smooth changes - smooth transition.

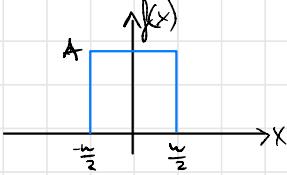
convolution in the Fourier Domain is just multiplication

- The Fourier transform's biggest shortcoming is that it does not give access to the time/spatial & frequency simultaneously

Continuous Fourier Transform of a signal, $f(x)$

$$\tilde{f}(w) = \int_{-\infty}^{\infty} f(x) e^{-j2\pi wx} dx \quad \begin{array}{l} \text{x: spatial coordinate (units of length)} \\ \text{w: spatial frequency (units of cycles/length)} \end{array}$$

↑
[spatial domain in signal] (Image)
Fourier frequency Domain Signal

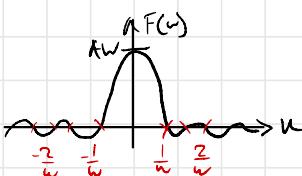


$$\tilde{f}(w) = \int_{-\frac{w}{2}}^{\frac{w}{2}} A e^{-j2\pi wx} dx = A w \frac{\sin(\pi w)}{\pi w}$$

! $\frac{\sin(x)}{x} \approx \sin(x)$

$$= Aw \operatorname{sinc}(\pi w)$$

- The Fourier transform of a rectangular pulse is a sinc function



- The Fourier transform makes frequency accessible.

- Z, Laplace, Fourier, wavelet all make frequency accessible.
- frequency is very important because all our senses depend on it touch, hearing, vision

Quiz 1 Notes: Will never ask something we have not discussed directly in class. Will always apply topic to new question in test

Question 1: a)

- Histogram equalization integrates and is summing so the curve should reflect that
 - monotonically increasing, $c_l < c_{l-1}$ doesn't matter
- The sum should increase rapidly where the histogram values are high
- The slope will slow down in the sum where the histogram values are low

Question 1: b) The mapping shows the average intensity value must increase because the low values of the map get set high.

Question 2:) The image will be binary because

$$S_0 = \frac{N_{r_0}}{N} (L-1) \quad S_1 = \frac{N_{r_0} + N_{r_1}}{N} (L-1) = L-1$$

Image will be
binary but not
the same

Question 3:) Use the gradient direction and mask the pixels with that direction.

-
- The Fourier transform has a duality property.
 - \mathcal{F} of squarewave is sinc func
 - \mathcal{F}^{-1} of squarewave is sinc func

Proving

$$F(\omega) = F^*(\omega)$$

1. $F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$

2. $F(\omega) = \int_{-\infty}^{\infty} f(t) \cdot e^{-j(-\omega)t} dt$

3. $F(\omega) = \int_{-\infty}^{\infty} f(t) \cdot e^{j\omega t} dt$

$\overline{F(\omega)} = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$

$\overline{F(\omega)} = \int_{-\infty}^{\infty} f(t) e^{j\omega t} dt$

$\therefore F(\omega) = \overline{F(\omega)}$

complex component
real component (a)

$$f(x) \xrightarrow{\text{FT}} F(u)$$

$\hookrightarrow F(u)$ is generally a complex quantity \rightarrow The sinc function example above is only real

$$|F(u)| = \sqrt{a^2 + b^2} \leftarrow \text{magnitude} \leftarrow (\text{spectrum})$$

$$\angle F(u) = \tan^{-1}\left(\frac{b}{a}\right) \leftarrow \text{phase (Angle)}$$

① If $f(x)$ is real then $F(-u) = F^*(u)$ then $|F(u)| = |F(u)|$

Prove this

Even symmetric

$$a+bi = a-bi$$

- Realness and even symmetry are fourier transform pairs

- We start with real images

② $f(ax) \xrightarrow{\text{FT}} \frac{1}{|a|} F\left(\frac{u}{a}\right)$ $a > 1$ contracts frequency
 $a < 1$ expands frequency

1-D Discrete Fourier Transform

Can be in either DFT or IDFT but not both

$$\text{DFT } F(u) = \frac{1}{M} \sum_{x=0}^{M-1} f(x) e^{-j2\pi ux/M} \quad u=0, 1, 2, \dots, L-1$$

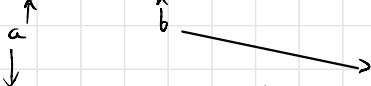
$$\text{IDFT } f(x) = \sum_{u=0}^{n-1} F(u) e^{j2\pi ux/M} \quad x=0, 1, 2, 3, \dots, L-1$$

- The input and output have the same length

- The Fourier transform of an image is almost always complex - why?

- The image would have to be symmetric for the transform to be real

$$e^{j\omega t} = \cos(\omega t) + j \sin(\omega t)$$



\cos component is in phase \sin component is out of phase

Magnitude tells amplitude of freq components

Phase tells shift of freq components

- If a freq component has a phase shift it will be shown as a complex # because \sin & \cos are not aligned (forming a complex valued result)
- Images typically have spatial asymmetries (from lighting, edges, textures, etc) which don't exist perfect symmetry the irregularities result in a mix of positive or negative frequency components which manifest in the FT as complex #'s
- If an Image has perfect symmetry the FT may be real at some frequencies but in general, natural images do not have perfect symmetry.



In phase

Out of phase

- In phase components occur when the image has symmetry or smooth consistent patterns
- Alignment of frequency components is uniform across the image
- Out of phase components occur when sharp transitions like edges, corners, & abrupt changes in intensity occur
 - causes mis alignment of freq components leading to phase shift.

$$F(\omega) = \frac{1}{M} \sum_{x=1}^{M-1} f(x) \leftarrow \text{Mean values of signal}$$

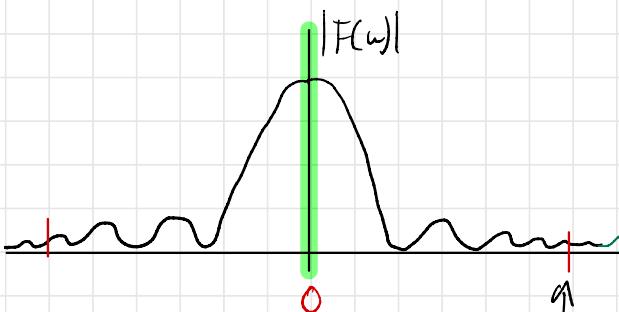
$$F(u+m) = F(u) \quad \leftarrow \text{Prove this, it is because } F(u) \text{ is periodic \& } M \text{ is the period}$$

$$f(x) \otimes g(x) \xrightarrow{\text{F}} F(u) \cdot G(u)$$

Instead of convolving
in the spatial Domain

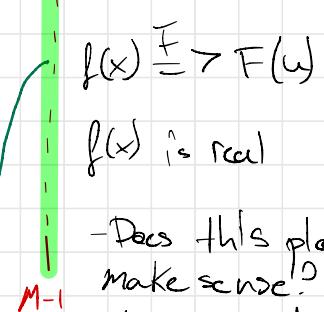
Multiply in the frequency Domain

$$\begin{aligned} f(x) &\stackrel{\text{F}}{\equiv} F(u) \\ g(x) &\stackrel{\text{F}}{\equiv} G(u) \end{aligned} \quad \xrightarrow{\text{Convolution}} f(x) \otimes g(x)$$



fft (fast fourier transform)
fft(F(u))

fft shift fixes this



- The formula computes
from $0 \leq u \leq M-1$

- Does this plot make sense? - yes
because of its symmetry. → real!

2D Discrete Fourier Transform

$$F(u,v) = \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x,y) e^{-j2\pi \left(\frac{ux}{M} + \frac{vy}{N} \right)} \quad \begin{array}{l} u = 0, 1, \dots, M-1 \\ v = 0, 1, \dots, N-1 \end{array}$$

Output Image
($M \times N$) $u: 0 \dots M-1$
 $v: 0 \dots N-1$

Input Img $M \times N$

$$F(u, v) = \sum_{x=0}^{N-1} \left(\sum_{y=0}^{M-1} f(x, y) e^{-j \frac{2\pi}{M} yv} \right) e^{-j \frac{2\pi}{N} ux}$$

↓ 1D DFT along columns
 ↓ 1D DFT along rows

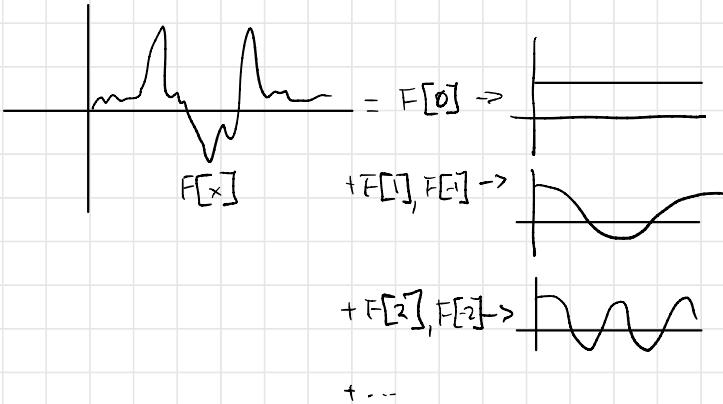
The 2D DFT is separable

$$f(x, y) = \frac{1}{MN} \sum_{u=0}^{N-1} \sum_{v=0}^{M-1} F[u, v] e^{\frac{j 2\pi u x}{N} + j \frac{2\pi v y}{M}}$$

! if $n=N$ you may see N^2

As with the 1D DFT, 2D DFT is like a decomposition of an image into complex exponentials (sines & cosines)

$$e^{-j \frac{2\pi u x}{N}} = \cos\left(\frac{2\pi u x}{N}\right) - j \sin\left(\frac{2\pi u x}{N}\right)$$



FT Properties

shift: $g(x, y) = F(x-a, y-b)$

$$G(u, v) = F(u, v) e^{-2\pi j \left[\frac{au}{N} + \frac{bv}{M} \right]}$$

Complex phase shift but the Magnitude remains the same

$$|G(u, v)| = |F(u, v)|$$

Scaling/Flip: $g(x, y) = a F(x, y)$ $g(x, y) = F(ax, by)$

$$G(u, v) = a F(u, v) \quad G(u, v) = \frac{1}{|ab|} F\left(\frac{u}{a}, \frac{v}{b}\right)$$

If $a \text{ or } b = -1 \rightarrow$ Flip in spatial = Flip in freq. Domain
 If $a \text{ or } b < 1 \& ab > 0$, compress spatial, expand freq
 If $a \text{ or } b > 1$, expand spatial, compress freq

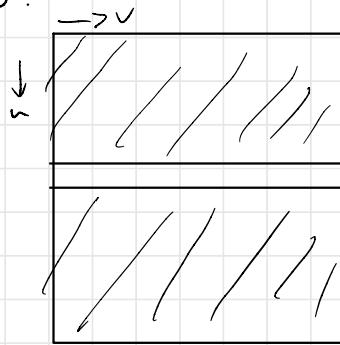
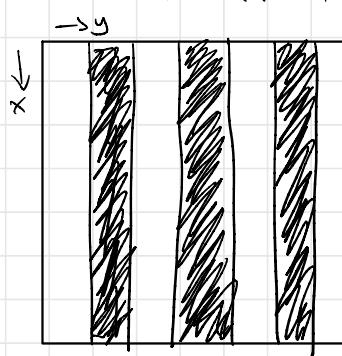
Rotation: If $g(x, y) = F(x, y)$ rotated by θ° (cw)

Then $G(u, v) = F(u, v)$ rotated by θ° (cw)

Convolution: $h(x, y) = F(x, y) \otimes g(x, y)$

$$H(u, v) = F(u, v) \cdot G(u, v)$$

Why Do Strong Edges in images at θ° show a strong contribution to the DFT at $\theta^\circ + 90^\circ$?



$$\begin{bmatrix} 1 & 255 & 0 & 255 & 0 & 255 & 0 \\ 1 & & & & & & \\ 1 & & & & & & \\ 1 & & & & & & \\ 1 & & & & & & \\ 1 & & & & & & \end{bmatrix}$$

$\xrightarrow{\text{DFT}}$

! DFT is separable

$$\begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & & & & & & \\ 0 & & & & & & \end{bmatrix}$$

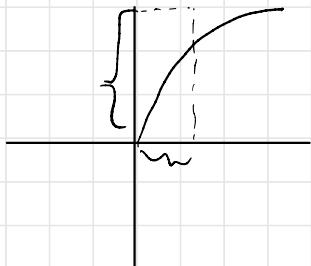
DC

$$\rightarrow \begin{bmatrix} & & & & & & 0 \\ & & & & & \hline \text{Content} & & & & & \\ & & & & & \hline & & & & & 0 \end{bmatrix}$$

$$f: M \times N \rightarrow \text{FFT} \rightarrow F: M \times N$$

↑ you can not directly display this because it is a complex number

you can view the spectrum (magnitude) or the phase



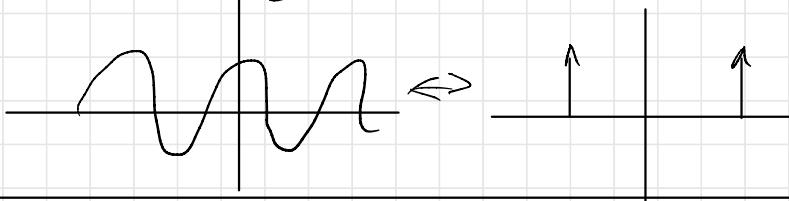
- To view magnitude, you must scale it (use linear scale)
- The low frequencies typically overwhelm the high frequencies
so use logarithmic scaling

$$\log(1 + \text{abs}(\text{fftshift}(\text{fft2}(t))))$$

- Typically phase isn't used visually
but it does have important info

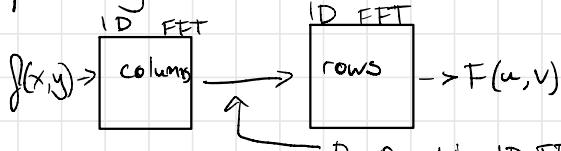
Do MATLAB Exercises w/ Synthetic Images

$$\cos(\omega x) \xleftrightarrow{\text{FT}} [\delta(\omega + \omega_0) + \delta(\omega - \omega_0)]$$



DFT vs. FFT

- Separability means no-one does a 2D DFT, you do two 1D DFTs



Do row wise 1D FFT to the column wise 1D FFT

- Don't Do 1D FFT, it's fast, and precise (not approximating)

↳ This reduces computation from $M \times M$ to $M \cdot 2$

frequency Domain filtering (blurring & sharpening)

Convolution \xrightarrow{F} Multiplying

$$g(x) = f(x) \otimes h(x)$$

out in kernel

\rightarrow FT both sides

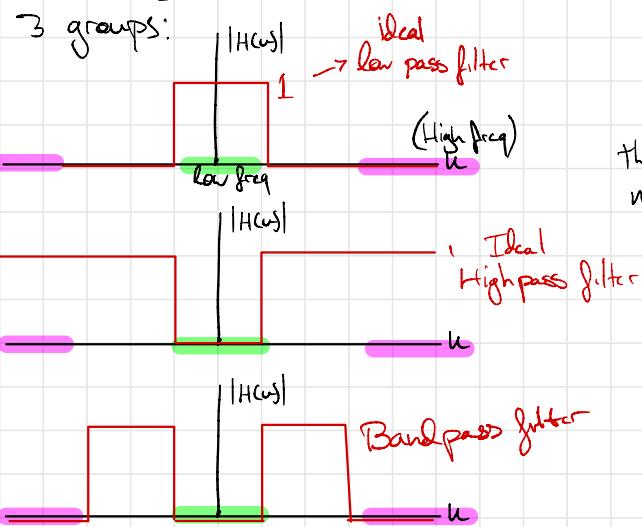
$$G(u) = F(u) \odot H(u)$$

Element by element multiplication.

! Multiply to complex quantity not the
Magnitude of phase!

What does $H(u)$ look like?

3 groups:



These values do not
need to be zero & 1

Otsu's thresholding technique \rightarrow assumes bimodal hist &
 \hookrightarrow Binary thresh technique.

Places thresh in valley of the
two modes

- Binarization first was important because it
isolates the object right away which reduces
noise.

$$f(x,y) \xrightarrow{F} F(x,y) \triangleq |F(u,v)|, |F(u,v)| \Rightarrow 1. \quad \boxed{|F(u,v)| \rightarrow f(x,y)}$$

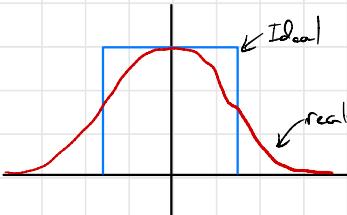
Phase gives structure info

\hookrightarrow shows phase structure

Magnitude gives intensity information

$\hookrightarrow |F(u,v)| \cdot 10 \rightarrow f(x,y)$

\hookrightarrow shows structure less magnitude.



- computers can do the discretized Ideal filter
but we don't actually want to do that.

Integration w/ F.T. and filters

$$\text{integration of } f(x) \xrightarrow{\int} F(\omega)$$

$$\text{differentiation of } f(x) \xrightarrow{?} j\omega F(\omega)$$

$$G(\omega) = F(\omega) \cdot \frac{1}{j\omega}$$

↑ would be integral

$$G(\omega) = F(\omega) \cdot j\omega$$



- integration is a lowpass filter
- Reduces noise, abrupt change

- Differentiation is a highpass filter
- Details, texture has high frequency

to make band pass shift by ω_0

Niquist theorem decides frequency range to analyze based on sample rate

Magnitude spectrum is not histogram

Real signal in spatial makes even symmetry in Fourier

You want to design filters symmetrical around the origin

Apply Gaussian for filter design

Compass filter Design

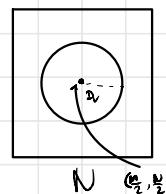
$H(u, v)$ is filter, D_0 is cutoff freq., n is order

$$D(u, v) = \left[\left(u - \frac{M}{2} \right)^2 + \left(v - \frac{N}{2} \right)^2 \right]^{1/2}$$

for $M \times N$ image

↑

Equation of a circle in u, v plane



Ideal

$$H(u, v) = \begin{cases} 1 & \text{if } D(u, v) \leq D_0 \\ 0 & \text{if } D(u, v) > D_0 \end{cases}$$

your filter's image, with disk in center, outside disk is 0
inside disk is 1

if things have been shifted
so center is DC component
this is a low pass filter

- Ideal is transition between 1 and 0
is abrupt

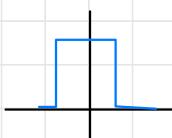
FT → shift → multiply by filter → shift → IFT

- Smaller circle gets more blurring (smaller freq. range passed)

- Lowpass blurs and removes details

- Artifacts are occurring w/ stronger butterworth because of the ideal filter

Ideal →



FT →



as this expands

Side bands cause the artifacts
"ringing artifacts"



This shrinks
so the ring artifact shrinks when the ideal filter expands

Butterworth filter:

$$H(u,v) = \frac{1}{1 + [D(u,v)/D_0]^{2n}}$$

As n approaches ∞ , the filter becomes ideal

D_0 is radius

* How would you know if this is high pass or low pass

Gaussian:

$$H(u,v) = e^{-D^2(u,v)/(2D_0^2)}$$

↑ why does D_0 appear here? *

Highpass filter:

$$H(u,v) = \begin{cases} 0 & \text{if } D(u,v) \leq D_0 \\ 1 & \text{if } D(u,v) > D_0 \end{cases}$$

Ideal

$$H(u,v) = 1 - e^{-D^2(u,v)/(2D_0^2)}$$

Gaussian

$$H(u,v) = \frac{1}{1 + [D_0/D(u,v)]^{2n}}$$

Butterworth

Video Segment:

for High pass filter design, subtract LP from Identity filter

$$\dots \begin{bmatrix} 0 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix} \dots - LP$$

↑ gaussian or Ideal

Anti-Aliasing:

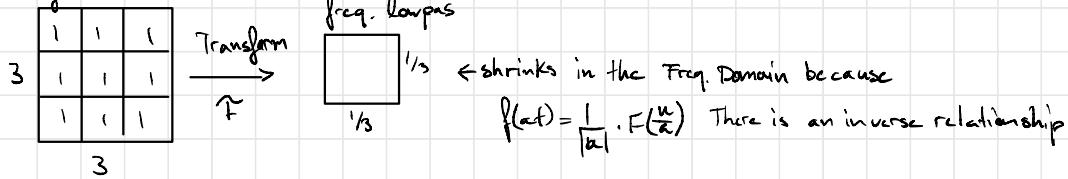
- It's not removing aliasing - that's not possible

- It blurs a digital image before downsampling to prevent aliasing/Artifacts

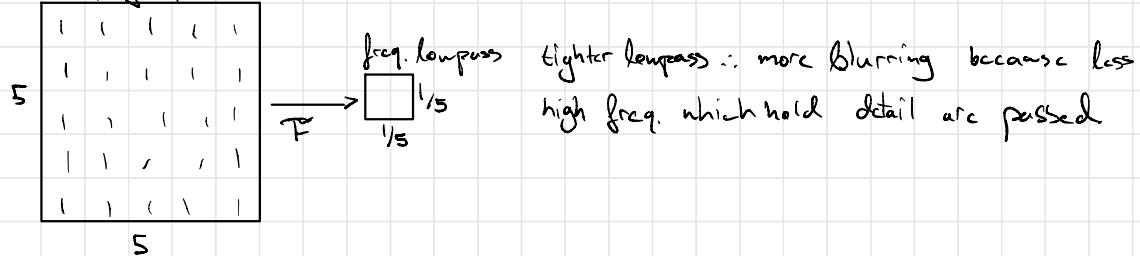
Original Img \rightarrow Blur LPF \rightarrow Downsample

Spatial \rightarrow Spatial Frequency Domain filter

A spatial low pass filter



Stronger spatial lowpass



Binary Image Morphology

Morphology → Structure, form, or shape of objects

- Let A be a set in \mathbb{Z}^2 . If $a = (a_1, a_2)$ is an element in A , then $a \in A$

0,0	.	.	0	0
.	.	0	0	
.
.

ordered pair of positive integers

\rightarrow 0's

\rightarrow 1's

$$A = \{(0,2), (0,3), (1,2), (1,3)\}$$

$$-\text{Expression } C = \{w \mid w = d \times d \in D\}$$

\hookrightarrow Means that C is a set w/ elements w such that w is formed by multiplying each of the two coordinates

of all elements of set D by -1.

Notation: $A \subseteq B$ $\rightarrow A$ is contained in B

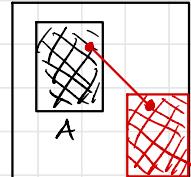
$A \cup B$ $\rightarrow A$ union B

$A \cap B$ $\rightarrow A$ intersect w/ B

Basic Operations: Let A & B be sets in \mathbb{Z}^2 with components $a = (a_1, a_2)$ and $b = (b_1, b_2)$, respectively

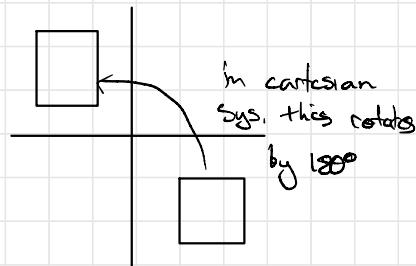
- Translation of A by $x = (x_1, x_2)$ is denoted by:

$$(A)_x = \{c | c = \underline{a} + \underline{x} \quad \forall a \in A\}$$



- Reflection of B is defined as

$$B = \{x | x = -b \quad \forall b \in B\}$$

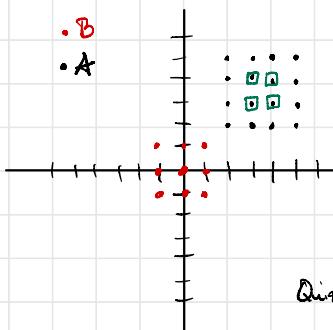
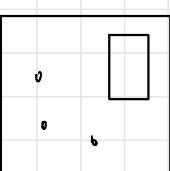


- Two fundamental operations in BIM

Think of ANDing to set origin

\rightarrow Erosion: $A \ominus B = \{x | (B)_x \subseteq A\}$ ← other equivalent definitions of erosion \ominus

\uparrow Erosion



“consider the translations where B is contained in A ”

only need to check x values in A
structuring element
usually not interchangeable

is $A \ominus B \subseteq A$?

Quiz: Is there always an erosion of A by B ? *

Is not if A is smaller than B

Quiz \Rightarrow If B was shifted to the left would you still get the same erosion? \rightarrow important def. & see

think of ORing street element to set origin

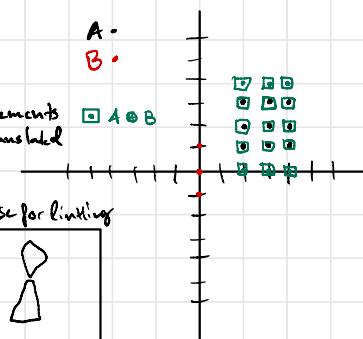
$$A \oplus B = \bigcup_{b \in B} (A)_b$$

↑
Dilated

Union
↓
Union of all elements when A is translated by B

A:
B:

$\square A \oplus B$



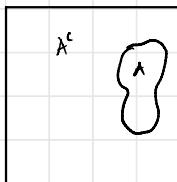
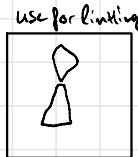
- Note: $A \oplus B = B \oplus A$

$$A \oplus B \neq B \oplus A$$

Duality of Erosion:

$$(A \ominus B)^c = A^c \oplus \hat{B}$$

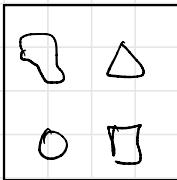
$$(A \oplus B)^c = A^c \ominus \hat{B}$$



you can erode A by
dilating A^c or erode
 A^c by dilating A

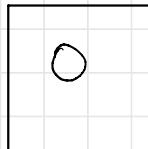
?

→ Connected Component Analysis

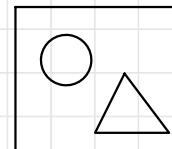


How many objects are there and how many pixels are in each object

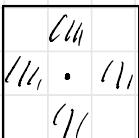
- Quiz Question *



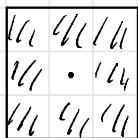
Easy



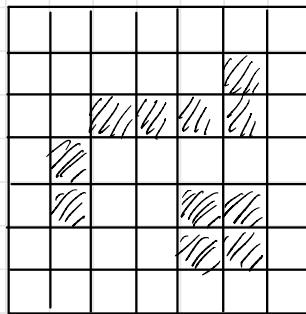
Hard



4 connected neighbors of center pixel



8 connected neighbors of center pixel



of objects?

a - conn = 3

b - conn = 2

- if you don't know the number of objects it is very tough

Algorithm for CCA

- Let A be a set containing one or more connected components. form a Matrix X_0 of the same size as A whose elements are zero except at each point known to correspond to a point in each connected component.

Following the iterative procedure:

$$X_k = (X_{k-1} \oplus B) \cap A \quad k=1, 2, 3$$

will find all connected components of A when

$$X_k = X_{k-1}$$

B is an appropriate structuring element

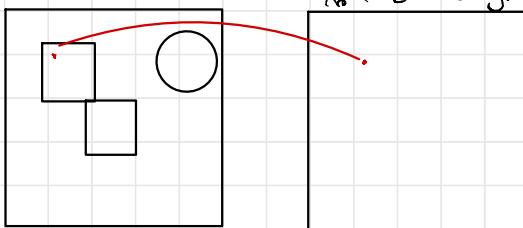
$$B_4 = \begin{bmatrix} 0 & 1 & 0 \\ 1 & 1 & 1 \\ 0 & 1 & 0 \end{bmatrix}$$

4 con

$$B_8 = \begin{bmatrix} 1 & 1 & 1 \\ 1 & 1 & 1 \\ 1 & 1 & 1 \end{bmatrix}$$

8 con

A



$X_0 < \text{Zero everywhere except for 1 point inside of one object}$

the process will keep expanding that point with dilation and intersects with the original object to limit the growth

$$\text{Opening } A \ominus B = (A \oplus B) \cap A$$

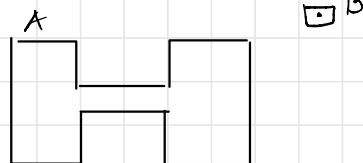
Eroding by B then dilating by B

$$= \bigcup \left\{ (B)_z \mid (B)_z \subseteq A \right\}$$

- the union of all translations of Z of B which are contained in A

Erosion: $\left\{ X \mid (B)_X \subseteq A \right\}$

- Doesn't keep just the origin point it keeps all the points where the struct. Pts



$A \ominus B$

$$(A \ominus B) \oplus B \rightarrow A \cdot B$$

- opening is idempotent

- if you erode \rightarrow dilate \rightarrow erode \rightarrow dilate, The idempotency holds,

but if you erode, erode, erode \rightarrow dilate, dilate, dilate it is not idempotent

Opening breaks narrow bridges and eliminates thin structures

Closing: $A \cdot B = (A \oplus B) \ominus B$

- Look into Example why structuring element ... origin

- The rounding in photo ex. occurs b.c. circular struct. elemnt

- Opening can help smoothing

- What kinds of shapes will open/dilate/erode differently

- Re-opening will give new results - look into why

- can create a skeleton of an object using erosion struct sizes

- you can have even or odd shaped structuring element
but pay attention to what the origin should be.