

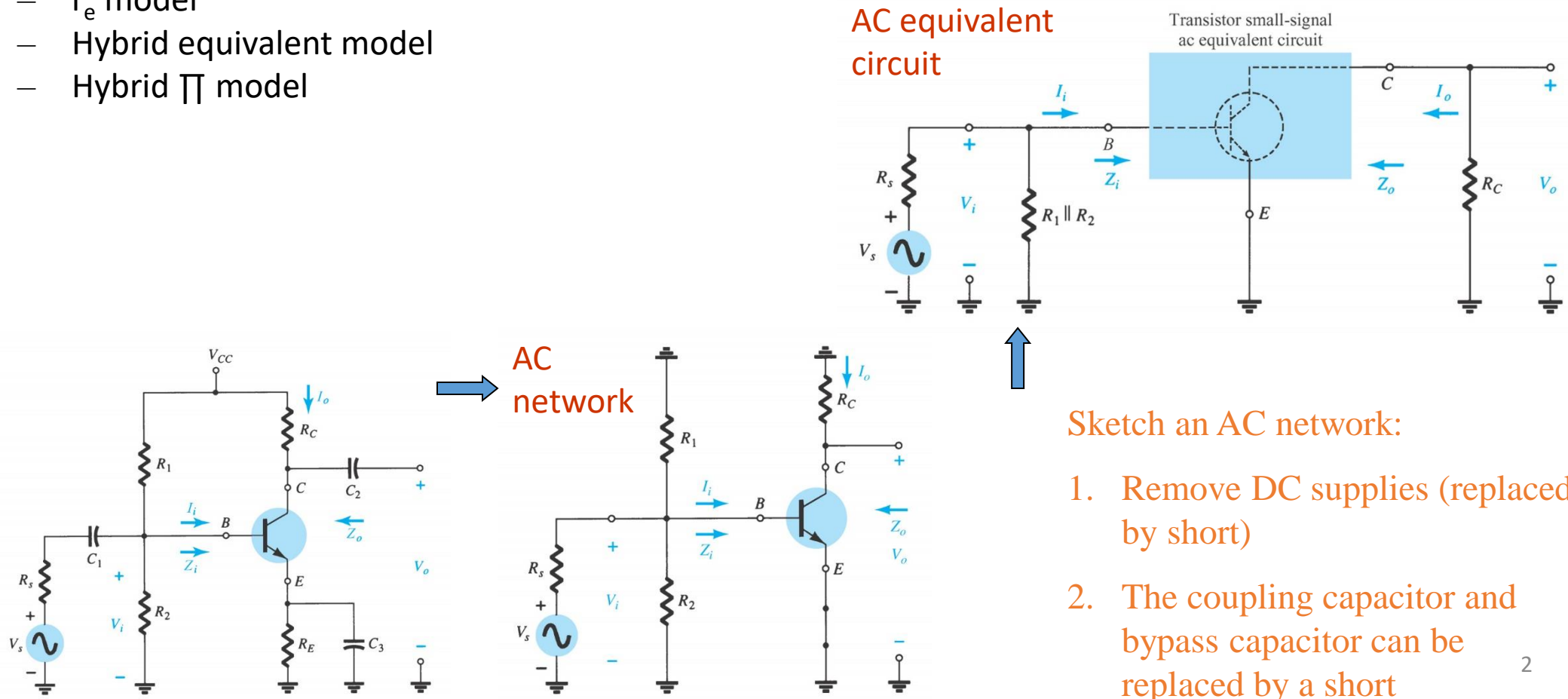
BJT Hybrid Equivalent Model

Semiconductor Devices and Circuits
(ECE 181302)

31st December 2021

BJT Transistor Modeling

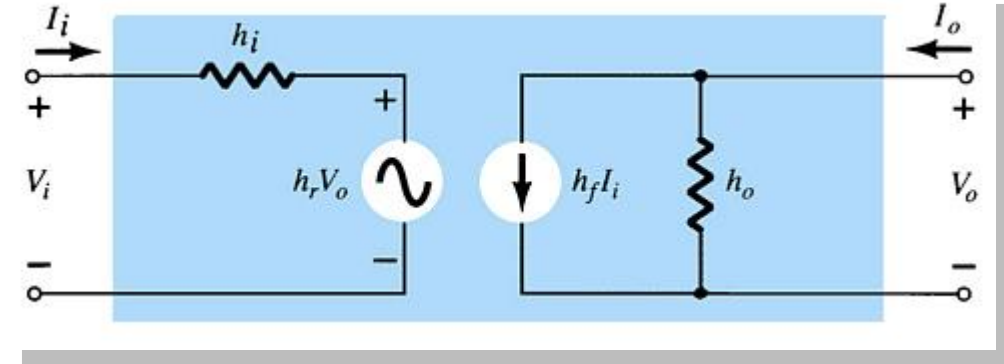
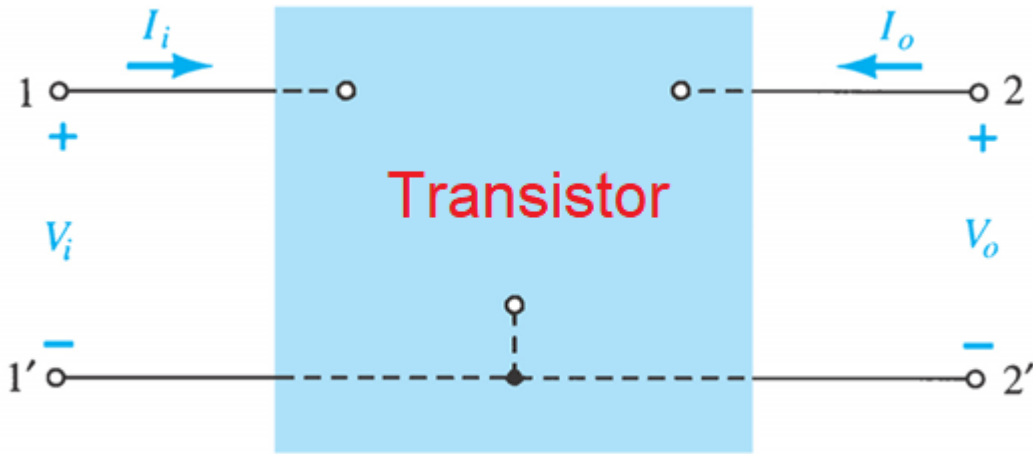
- An equivalent circuit that represents the AC characteristics of the transistor.
- Uses circuit elements that approximate the behavior of the transistor.
- There are three models commonly used in small signal AC analysis of a transistor:
 - r_e model
 - Hybrid equivalent model
 - Hybrid π model



Sketch an AC network:

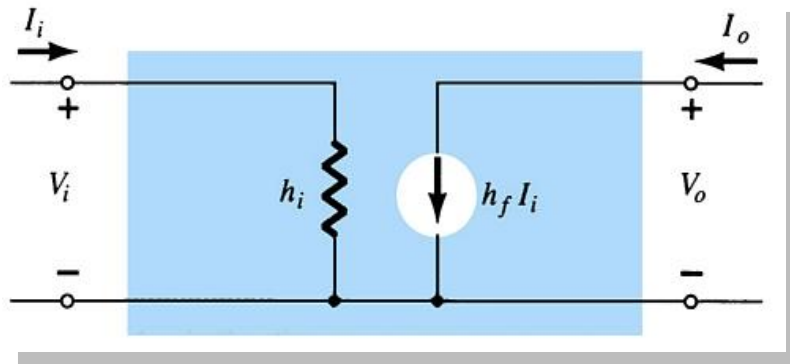
1. Remove DC supplies (replaced by short)
2. The coupling capacitor and bypass capacitor can be replaced by a short

The Hybrid Equivalent Model



- h_i = input resistance
- h_r = reverse transfer voltage ratio ($V_i/V_o \cong 0$)
- h_f = forward transfer current ratio (I_o/I_i)
- h_o = output conductance $\cong \infty$

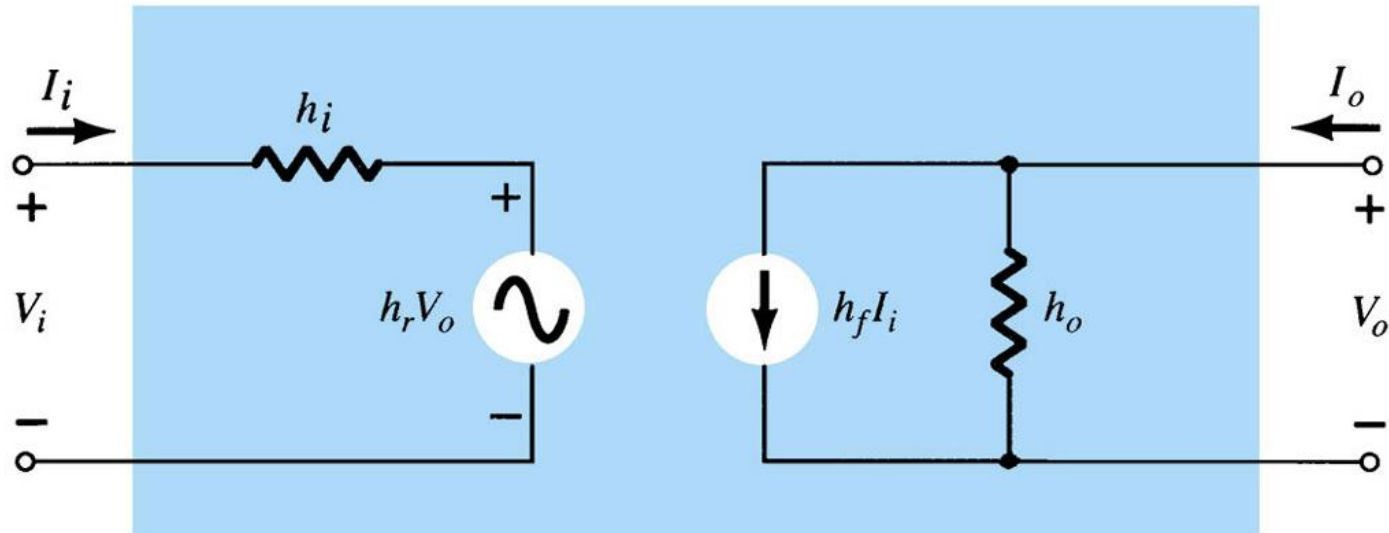
Simplified General H-Parameter Model: Approximate hybrid equivalent model



- h_i = input resistance
- h_r = reverse transfer voltage ratio ($V_i/V_o \cong 0$)
- h_f = forward transfer current ratio (I_o/I_i)
- h_o = output conductance $\cong \infty$

Hybrid Equivalent Model

- General h-Parameters: h_{ie} , h_{re} , h_{fe} , h_{oe} are developed and used to model the transistor.
- The h-Parameters can be found in a specification sheet for a transistor.



h_i = input resistance

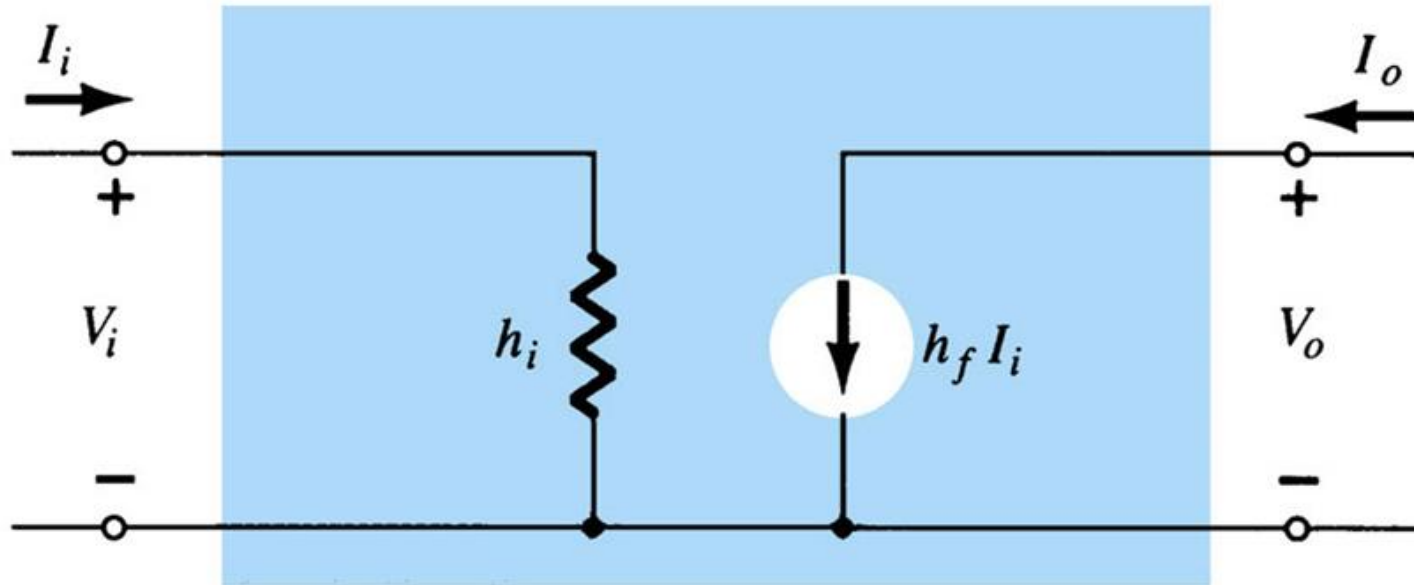
h_r = reverse transfer voltage ratio (V_i/V_o)

h_f = forward transfer current ratio (I_o/I_i)

h_o = output conductance

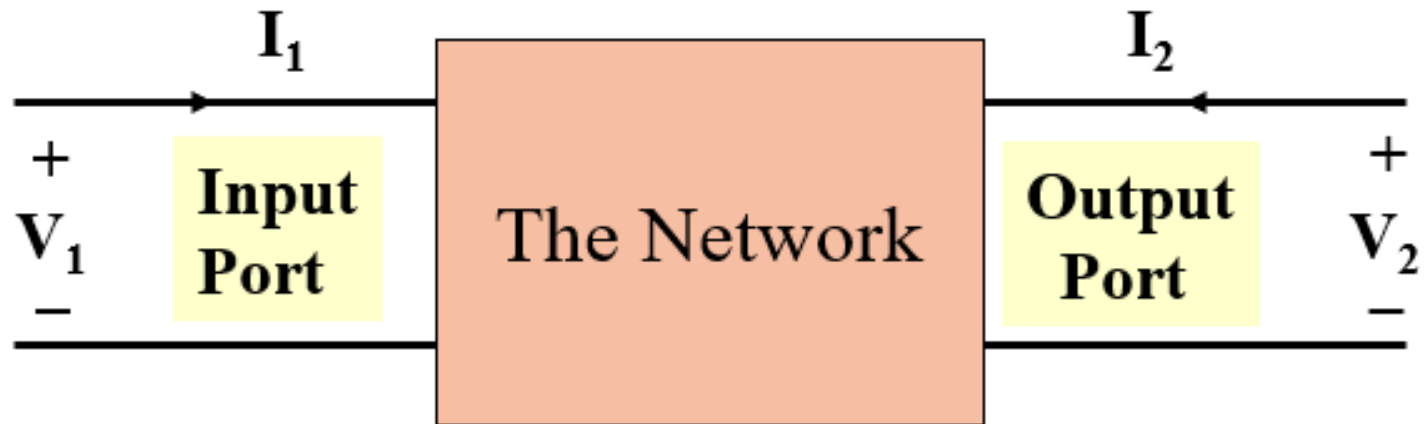
Simplified General h-parameter Model

- The general h-parameter model can be simplified based on approximations:
- $h_r \cong 0$ therefore $h_r V_o = 0$ and $h_o \cong \infty$



Two Port Networks

- The standard configuration of a two port:



Network Equations:

**Impedance
Z parameters**

$$\begin{aligned}V_1 &= z_{11}I_1 + z_{12}I_2 \\V_2 &= z_{21}I_1 + z_{22}I_2\end{aligned}$$

$$\begin{aligned}V_2 &= b_{11}V_1 - b_{12}I_1 \\I_2 &= b_{21}V_1 - b_{22}I_1\end{aligned}$$

**Admittance
Y parameters**

$$\begin{aligned}I_1 &= y_{11}V_1 + y_{12}V_2 \\I_2 &= y_{21}V_1 + y_{22}V_2\end{aligned}$$

**Hybrid
H parameters**

$$\begin{aligned}V_1 &= h_{11}I_1 + h_{12}V_2 \\I_2 &= h_{21}I_1 + h_{22}V_2\end{aligned}$$

**Transmission
A, B, C, D
parameters**

$$\begin{aligned}V_1 &= AV_2 - BI_2 \\I_1 &= CV_2 - DI_2\end{aligned}$$

$$\begin{aligned}I_1 &= g_{11}V_1 + g_{12}I_2 \\V_2 &= g_{21}V_1 + g_{22}I_2\end{aligned}$$

- From the four variables V_1 , V_2 , i_1 and i_2 , two can be selected as independent variables and the remaining two can be expressed in terms of these independent variables.
- This leads to various two part parameters out of which the following three are more important.

Hybrid Parameters or h-parameters

- If the input current i_1 and output Voltage V_2 are taken as independent variables,
- Then input voltage V_1 and output current i_2 can be written as:

$$\begin{aligned} V_1 &= h_{11} i_1 + h_{12} V_2 \\ i_2 &= h_{21} i_1 + h_{22} V_2 \end{aligned}$$

- The four hybrid parameters : h_{11} , h_{12} , h_{21} and h_{22} are defined as follows.
 - $h_{11} = [V_1 / i_1]$ with $V_2 = 0$, : I/P Resistance with output port short circuited.
 - $h_{12} = [V_1 / V_2]$ with $i_1 = 0$, : reverse voltage transfer ratio with i/p port open circuited.
 - $h_{21} = [i_2 / i_1]$ with $V_2 = 0$, : Forward current gain with output part short circuited.
 - $h_{22} = [i_2 / V_2]$ with $i_1 = 0$, : Output admittance with input part open circuited.

$$h_{11} = [V_1 / i_1] \text{ with } V_2 = 0$$

= Input Impedance with output part short circuited.

$$h_{22} = [i_2 / V_2] \text{ with } i_1 = 0$$

= Output admittance with input part open circuited.

$$h_{12} = [V_1 / V_2] \text{ with } i_1 = 0$$

= Reverse voltage transfer ratio with input part open circuited.

$$h_{21} = [i_2 / i_1] \text{ with } V_2 = 0$$

= Forward current gain with output part short circuited.

Dimensions of the h-parameters

h_{11} - Ω

h_{12} – dimension less

h_{21} – dimension less

h_{22} – mhos

- As the dimensions are not alike, (i.e) they are hybrid in nature, these parameters are called as hybrid parameters.

Equations for the Hybrid Parameters

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

$$h_{11} = \left. \frac{V_1}{I_1} \right|_{V_2 = 0}$$

$$h_{12} = \left. \frac{V_1}{V_2} \right|_{I_1 = 0}$$

$$h_{21} = \left. \frac{I_2}{I_1} \right|_{V_2 = 0}$$

$$h_{22} = \left. \frac{I_2}{V_2} \right|_{I_1 = 0}$$

$$V_1 = h_{11} i_1 + h_{12} V_2$$

↓

$$V_1 = h_i i_1 + h_r V_2$$

$$I_2 = h_{21} i_1 + h_{22} V_2$$

↓

$$I_2 = h_f i_1 + h_o V_2$$

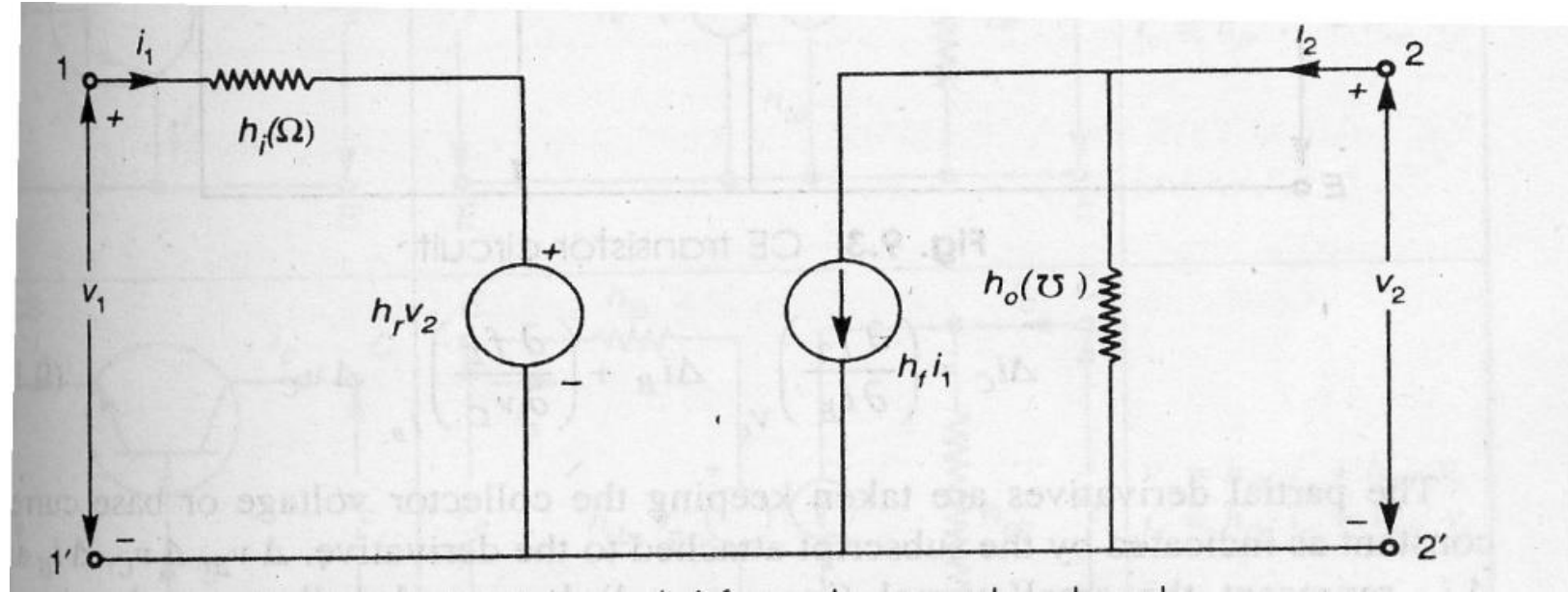
$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} h_i & h_r \\ h_f & h_o \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

$$h_i = \frac{V_1}{I_1} \quad \left| \quad V_2 = 0 \right.$$

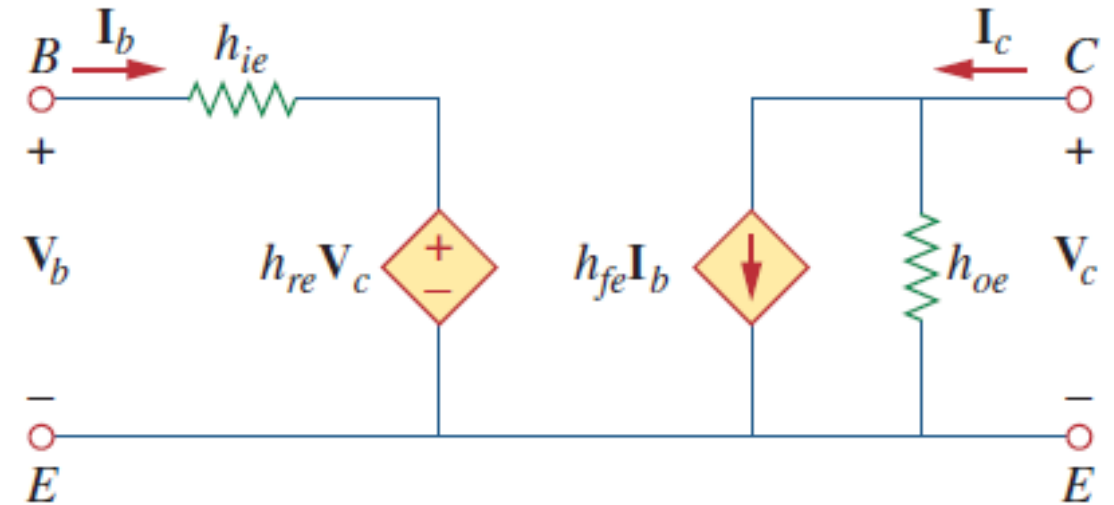
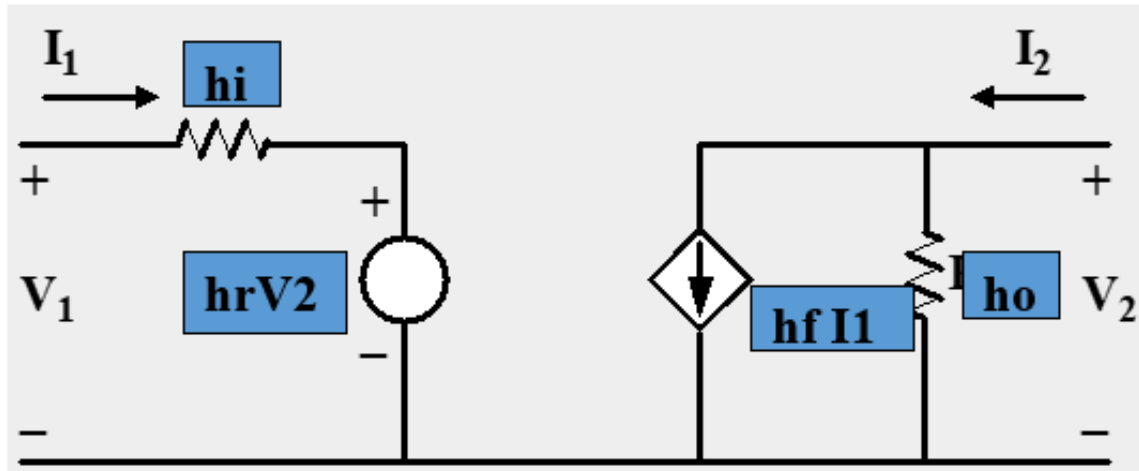
$$h_r = \frac{V_1}{V_2} \quad \left| \quad I_1 = 0 \right.$$

$$h_f = \frac{I_2}{I_1} \quad \left| \quad V_2 = 0 \right.$$

$$h_o = \frac{I_2}{V_2} \quad \left| \quad I_1 = 0 \right.$$



Transistor Model



- We can write the following equations:

$$\begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} h_i & h_r \\ h_f & h_o \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix}$$

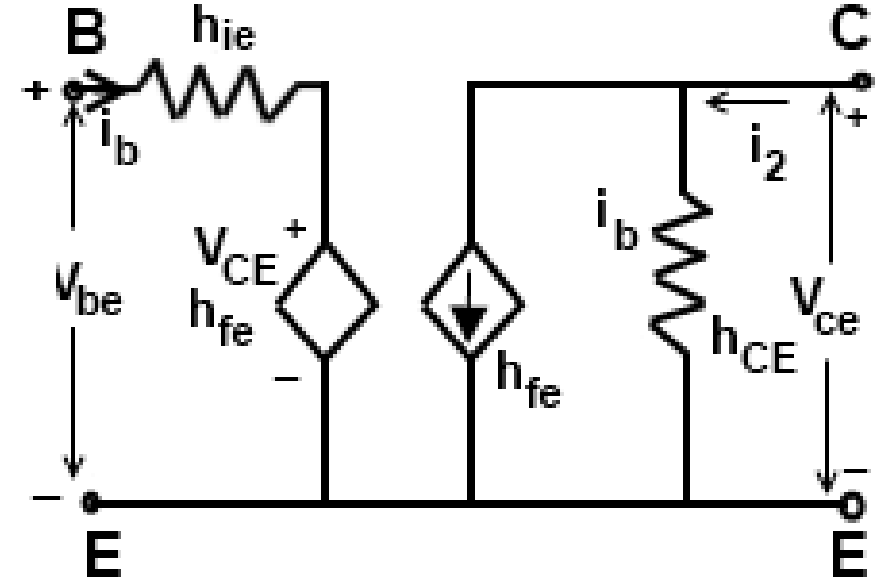
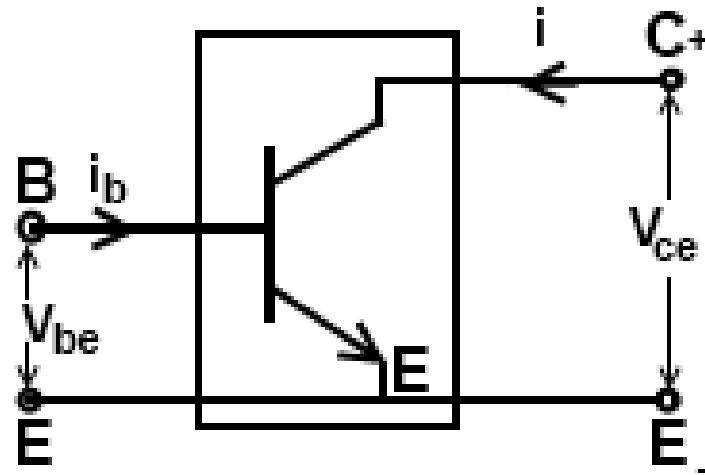
h-parameters Advantages

- h-parameters are real numbers up to radio frequencies.
- They are easy to measure.
- They can be determined from the transistor static characteristics curves.
- They are convenient to use in circuit analysis and design.
- Easily convert able from one configuration to other.
- Readily supplied by manufactories.

Hybrid Model CE Configuration

- In the common emitter transistor configuration, the input signal is applied between the base and emitter terminals of the transistor and output appears between the collector and emitter terminals.
- The input voltage (V_{be}) and the output current (i_c) are given by the following equations:

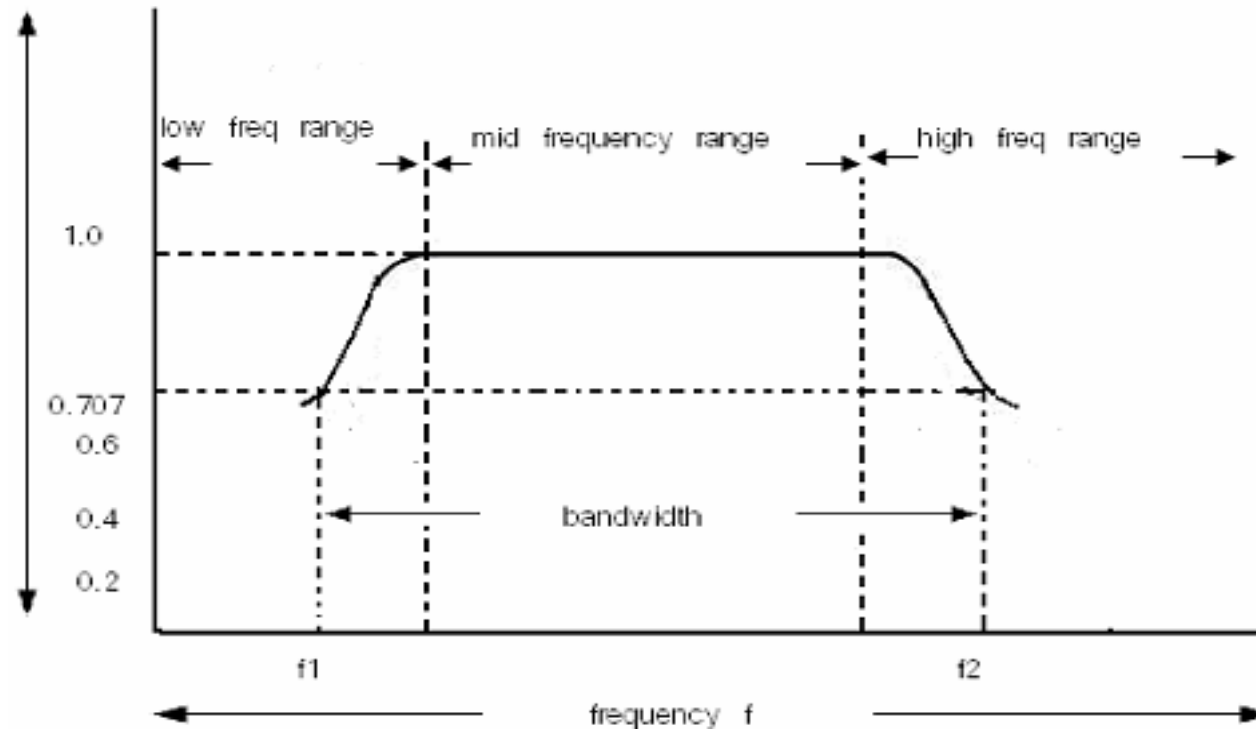
$$V_{be} = h_{ie} \cdot i_b + h_{re} \cdot V_c$$
$$i_e = h_{fe} \cdot i_b + h_{oe} \cdot V_c$$



Hybrid h-Parameter Model for an Amplifier

- The equivalent circuit of a transistor can be drawn using simple approximation by retaining its essential features.
- The equivalent circuit aids in analyzing transistor circuits easily and rapidly.
- A transistor can be treated as a two part network.
- The terminal behavior of any two part network can be specified by the terminal voltages V_1 & V_2 at parts 1 & 2 respectively and current i_1 and i_2 , entering parts 1 & 2, respectively, as shown in figure.

RC Coupled Amplifier



- Response of an amplifier depends on frequency considerations.
- There are 3 regions of frequency : low , mid and high
- The difference between high and low frequency is the bandwidth

Transistor Amplifier Analysis using Complete h-Parameter Model

- Small signal response is analyzed using the h-parameter model
- In the h-parameter model consider the load Resistance R_L and input signal V_s .
- The expressions for Current gain, Voltage gain ,input and output impedance are:

1. Current Gain:

$$A_i = -h_f / (1 + h_o R_L)$$

- Where A_i is the current amplification or current gain.

- The overall current gain taking source resistance is given by:

$$A_{is}=A_i * (R_s/Z_i + R_s)$$

- Where,
 - Z_i input impedance
 - R_s source resistance

2) Input Impedance(Z_i)

$$Z_i= h_i+h_rA_iR_L$$

3) Voltage Gain(A_v):

$$A_v=(A_i * R_L)/ Z_i$$

- Voltage gain taking source resistance is given by

$$A_{vs}=(A_v * Z_i)/(Z_i+R_s)$$

4) Output Admittance(Y_o)

$$Y_o=h_o-h_f * h_r/(h_i+R_s)$$

Analysis of Transistor Amplifier using Simplified h-Parameter Model

- Common Emitter Configuration

- Fixed Bias configuration:

$$\text{Input Impedance } Z_i = R_B \parallel h_{ie}$$

$$\text{Output Impedance } Z_o = R_C \parallel (1/h_{oe})$$

$$\text{Voltage gain } A_v = -h_{fe} * (R_C \parallel (1/h_{oe})) / h_{ie}$$

$$\text{Current Gain } A_i = h_{fe} * R_B / (R_B + h_{ie})$$

- Voltage Divider Configuration:

$$\text{Input impedance } Z_i = (R_{B1} \parallel R_{B2}) \parallel h_{ie}$$

$$\text{Output Impedance } Z_o = R_C \parallel (1/h_{oe})$$

$$\text{Voltage gain } A_v = -h_{fe} * [R_C \parallel (1/h_{oe})] / h_{ie}$$

$$\text{Current gain } A_i = h_{fe} * (R_{B1} \parallel R_{B2}) / (R_{B1} \parallel R_{B2} + h_{ie})$$



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