

TPS6281x-Q1 2.75V to 6V, Adjustable-Frequency, Automotive, Step-Down Converter

1 Features

- AEC-Q100 qualified for automotive applications
 - Device temperature grade 1:
–40°C to +125°C T_A
- Functional Safety-Capable
 - Documentation available to aid functional safety system design
- Input voltage range: 2.75V to 6V
- Family of 1A, 2A, 3A, and 4A
- Quiescent current 15 μ A typical
- Output voltage from 0.6V to 5.5V
- Output voltage accuracy $\pm 1\%$ (PWM operation)
- Adjustable soft start
- Forced PWM or PFM operation
- Adjustable switching frequency of 1.8MHz to 4MHz
- Precise ENABLE input allows
 - User-defined undervoltage lockout
 - Exact sequencing
- 100% duty cycle mode
- Active output discharge
- Spread spectrum clocking - optional
- Power-good output with window comparator
- Package with wettable flanks

2 Applications

- Infotainment head unit
- Hybrid and reconfigurable cluster
- Telematics control unit
- Surround view ECU, ADAS sensor fusion
- External amplifier

3 Description

The TPS6281x-Q1 is family of pin-to-pin 1A, 2A, 3A, and 4A synchronous automotive step-down DC/DC converters. All devices offer high efficiency and ease of use. The TPS6281x-Q1 family is based on a peak current mode control topology. The TPS6281x-Q1 is designed for automotive applications such as infotainment and advanced driver assistance systems. Low resistive switches allow up to 4A continuous output current at high ambient temperature. The switching frequency is externally adjustable from 1.8MHz to 4MHz and can also be synchronized to an external clock in the same frequency range. In PWM/PFM mode, the TPS6281x-Q1 automatically enter power save mode at light loads to maintain high efficiency across the whole load range. The TPS6281x-Q1 provide 1% output voltage accuracy in PWM mode which helps design a power supply with high output voltage accuracy. The SS/TR pin allows setting the start-up time or forming tracking of the output voltage to an external source. This feature allows external sequencing of different supply rails and limiting the inrush current during start-up.

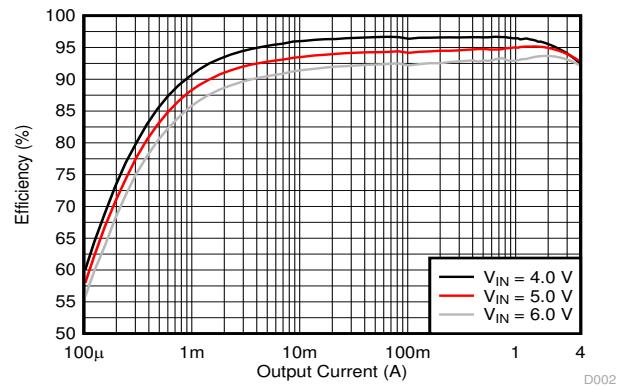
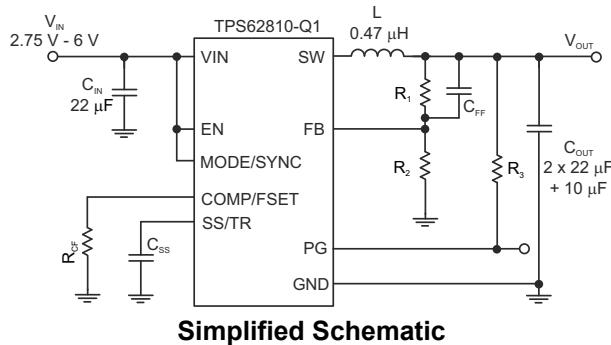
The TPS6281x-Q1 is available in a 3mm × 2mm VQFN package with wettable flanks.

Device Information

| PART NUMBER ⁽²⁾ | PACKAGE ⁽¹⁾ | BODY SIZE (NOM) |
|----------------------------|------------------------|-----------------|
| TPS62810-Q1 | RWY (VQFN, 9) | 3.00mm × 2.00mm |
| TPS62811-Q1 | | |
| TPS62812-Q1 | | |
| TPS62813-Q1 | | |

(1) For more information, see [Section 12](#).

(2) See the [Device Comparison Table](#).



Efficiency vs Output Current; $V_{OUT} = 3.3V$; PWM/PFM; $f_S = 2.25MHz$



An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.

Table of Contents

| | | | |
|---|-----------|--|-----------|
| 1 Features..... | 1 | 8.4 Device Functional Modes..... | 14 |
| 2 Applications..... | 1 | 9 Application and Implementation..... | 17 |
| 3 Description..... | 1 | 9.1 Application Information..... | 17 |
| 4 Device Comparison Table..... | 3 | 9.2 Typical Application..... | 19 |
| 5 Pin Configuration and Functions..... | 4 | 9.3 System Examples..... | 30 |
| 6 Specifications..... | 5 | 9.4 Power Supply Recommendations..... | 32 |
| 6.1 Absolute Maximum Ratings..... | 5 | 9.5 Layout..... | 32 |
| 6.2 ESD Ratings | 5 | 10 Device and Documentation Support..... | 34 |
| 6.3 Recommended Operating Conditions..... | 5 | 10.1 Device Support..... | 34 |
| 6.4 Thermal Information | 5 | 10.2 Documentation Support..... | 34 |
| 6.5 Electrical Characteristics..... | 6 | 10.3 Receiving Notification of Documentation Updates.. | 34 |
| 6.6 Typical Characteristics..... | 8 | 10.4 Support Resources..... | 34 |
| 7 Parameter Measurement Information..... | 9 | 10.5 Trademarks..... | 34 |
| 7.1 Schematic..... | 9 | 10.6 Electrostatic Discharge Caution..... | 34 |
| 8 Detailed Description..... | 11 | 10.7 Glossary..... | 34 |
| 8.1 Overview..... | 11 | 11 Revision History..... | 34 |
| 8.2 Functional Block Diagram..... | 11 | 12 Mechanical, Packaging, and Orderable | |
| 8.3 Feature Description..... | 12 | Information..... | 35 |

4 Device Comparison Table

| DEVICE NUMBER | OUTPUT CURRENT | V _{out} DISCHARGE | FOLDBACK CURRENT LIMIT | SPREAD SPECTRUM CLOCKING (SSC) | OUTPUT VOLTAGE |
|--------------------|----------------|----------------------------|------------------------|--------------------------------|----------------|
| TPS62811QWRWYRQ1 | 1 A | ON | OFF | OFF | adjustable |
| TPS6281120QWRWYRQ1 | 1 A | ON | OFF | ON | adjustable |
| TPS6281126QWRWYRQ1 | 1 A | ON | OFF | ON | fixed 1.0 V |
| TPS6281109QWRWYRQ1 | 1 A | ON | OFF | OFF | fixed 1.15 V |
| TPS628110AQWRWYRQ1 | 1 A | ON | OFF | OFF | fixed 1.2 V |
| TPS628112AQWRWYRQ1 | 1 A | ON | OFF | ON | fixed 1.2 V |
| TPS628112MQWRWYRQ1 | 1 A | ON | OFF | ON | fixed 1.8 V |
| TPS628113HQWRWYRQ1 | 1 A | ON | OFF | ON | fixed 3.3 V |
| TPS62812QWRWYRQ1 | 2 A | ON | OFF | OFF | adjustable |
| TPS6281220QWRWYRQ1 | 2 A | ON | OFF | ON | adjustable |
| TPS6281240QWRWYRQ1 | 2 A | OFF | OFF | ON | adjustable |
| TPS6281206QWRWYRQ1 | 2 A | ON | OFF | OFF | fixed 1.0 V |
| TPS6281208QWRWYRQ1 | 2 A | ON | OFF | OFF | fixed 1.1 V |
| TPS6281228QWRWYRQ1 | 2 A | ON | OFF | ON | fixed 1.1 V |
| TPS628122AQWRWYRQ1 | 2A | ON | OFF | ON | fixed 1.2 V |
| TPS628122GQWRWYRQ1 | 2 A | ON | OFF | ON | fixed 1.5 V |
| TPS628120MQWRWYRQ1 | 2 A | ON | OFF | OFF | fixed 1.8 V |
| TPS62812C0QWRWYRQ1 | 2A | disconnected | OFF | ON | adjustable |
| TPS62813QWRWYRQ1 | 3 A | ON | OFF | OFF | adjustable |
| TPS6281320QWRWYRQ1 | 3 A | ON | OFF | ON | adjustable |
| TPS6281326QWRWYRQ1 | 3 A | ON | OFF | ON | fixed 1.0 V |
| TPS628132DQWRWYRQ1 | 3 A | ON | OFF | ON | fixed 1.35 V |
| TPS628132MQWRWYRQ1 | 3 A | ON | OFF | ON | fixed 1.8 V |
| TPS628130AQWRWYRQ1 | 3 A | ON | OFF | OFF | fixed 1.2 V |
| TPS6281302QWRWYRQ1 | 3 A | ON | OFF | OFF | fixed 0.8 V |
| TPS6281322QWRWYRQ1 | 3A | ON | OFF | ON | fixed 0.8 V |
| TPS628133HQWRWYRQ1 | 3A | ON | OFF | ON | fixed 3.3 V |
| TPS6281389QWRWYRQ1 | 3A | ON | ON | OFF | fixed 1.15 V |
| TPS62810QWRWYRQ1 | 4 A | ON | OFF | OFF | adjustable |
| TPS6281020QWRWYRQ1 | 4 A | ON | OFF | ON | adjustable |
| TPS6281006QWRWYRQ1 | 4A | ON | OFF | OFF | fixed 1.0 V |
| TPS6281008QWRWYRQ1 | 4 A | ON | OFF | OFF | fixed 1.1 V |
| TPS628100MQWRWYRQ1 | 4 A | ON | OFF | OFF | fixed 1.8 V |
| TPS628101HQWRWYRQ1 | 4 A | ON | OFF | OFF | fixed 3.3 V |

5 Pin Configuration and Functions

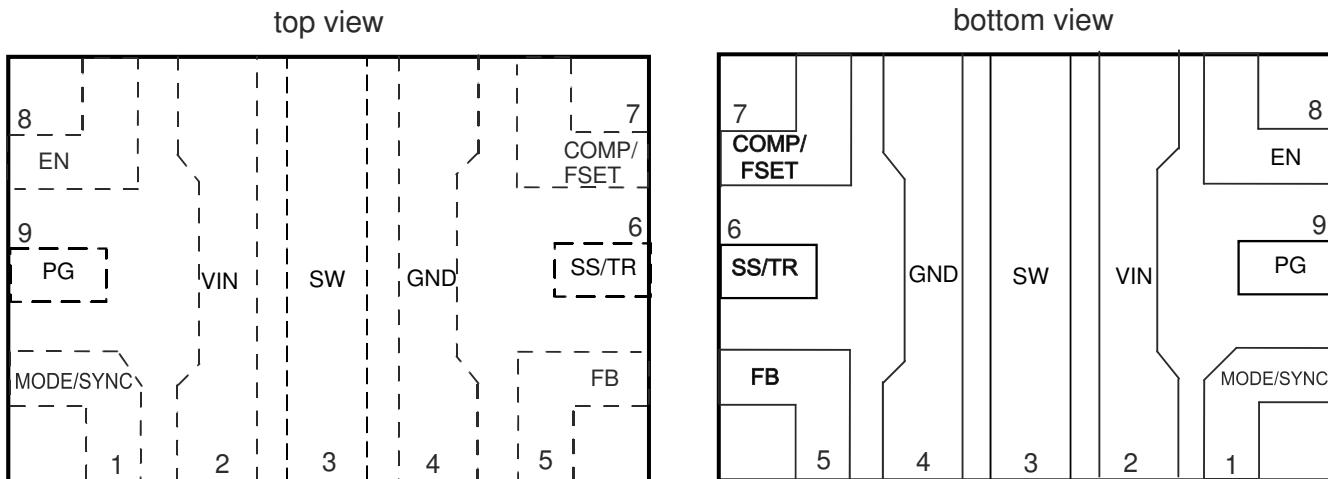


Figure 5-1. RWY Package 9 Pin (VQFN)

Table 5-1. Pin Functions

| PIN | | TYPE ⁽¹⁾ | DESCRIPTION |
|-----------|-----|---------------------|---|
| NAME | NO. | | |
| EN | 8 | I | This pin is the enable pin of the device. Connect to logic low to disable the device. Pull high to enable the device. Do not leave this pin unconnected. |
| FB | 5 | I | Voltage feedback input, connect the resistive output voltage divider to this pin. For the fixed voltage versions, connect the FB pin directly to the output voltage. |
| GND | 4 | | Ground pin |
| MODE/SYNC | 1 | I | The device runs in PFM/PWM mode when this pin is pulled low. If the pin is pulled high, the device runs in forced PWM mode. Do not leave this pin unconnected. The mode pin can also be used to synchronize the device to an external frequency. See the Section 6 for the detailed specification of the digital signal applied to this pin for external synchronization. |
| COMP/FSET | 7 | I | Device compensation and frequency set input. A resistor from this pin to GND defines the compensation of the control loop as well as the switching frequency if not externally synchronized. If the pin is tied to GND or VIN, the switching frequency is set to 2.25 MHz. Do not leave this pin unconnected. |
| PG | 9 | O | Open drain power good output. Low impedance when not "power good", high impedance when "power good". This pin can be left open or be tied to GND when not used. |
| SS/TR | 6 | I | Soft-Start / Tracking pin. A capacitor connected from this pin to GND defines the rise time for the internal reference voltage. The pin can also be used as an input for tracking and sequencing - see the Soft Start / Tracking (SS/TR) section. |
| SW | 3 | — | This pin is the switch pin of the converter and is connected to the internal Power MOSFETs. |
| VIN | 2 | — | Power supply input. Connect the input capacitor as close as possible between pin VIN and GND. |

(1) I = input, O = output

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

| | | MIN | MAX | UNIT |
|----------------------------------|---|------|--------------|------|
| Pin voltage range ⁽¹⁾ | VIN | -0.3 | 6.5 | V |
| | SW | -0.3 | $V_{IN}+0.3$ | V |
| | SW (transient for less than 10 ns) ⁽²⁾ | -3 | 10 | V |
| | FB | -0.3 | 4 | V |
| | PG, SS/TR, COMP/FSET | -0.3 | $V_{IN}+0.3$ | V |
| Pin voltage range ⁽¹⁾ | EN, MODE/SYNC | -0.3 | 6.5 | V |
| Storage temperature, T_{stg} | | -65 | 150 | °C |

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) While switching

6.2 ESD Ratings

| | | VALUE | UNIT |
|-------------|-------------------------|---|------------|
| $V_{(ESD)}$ | Electrostatic discharge | Human-body model (HBM), per AEC Q100-002 ⁽¹⁾ | ± 2000 |
| | | Charged-device model (CDM), per AEC Q100-011 | ± 750 |

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

| | | MIN | NOM | MAX | UNIT |
|-----------|--|------|------|------|------|
| V_{IN} | Supply voltage range | 2.75 | | 6 | V |
| V_{OUT} | Output voltage range | 0.6 | | 5.5 | V |
| L | Effective inductance for a switching frequency of 1.8 MHz to 3.5 MHz | 0.32 | 0.47 | 0.9 | μH |
| L | Effective inductance for a switching frequency of 3.5 MHz to 4 MHz | 0.25 | 0.33 | 0.9 | μH |
| C_{OUT} | Effective output capacitance for 1A and 2A version ⁽¹⁾ | 15 | 22 | 470 | μF |
| C_{OUT} | Effective output capacitance for 3A and 4A version ⁽¹⁾ | 27 | 47 | 470 | μF |
| C_{IN} | Effective input capacitance ⁽¹⁾ | 5 | 10 | | μF |
| R_{CF} | | 4.5 | | 100 | kΩ |
| T_J | Operating junction temperature | -40 | | +150 | °C |

(1) The values given for the capacitors in the table are effective capacitance, which includes the DC bias effect. Due to the DC bias effect of ceramic capacitors, the effective capacitance is lower than the nominal value when a voltage is applied. Please check the manufacturer's DC bias curves for the effective capacitance vs DC voltage applied. Further restrictions may apply. Please see the feature description for COMP/FSET about the output capacitance vs compensation setting and output voltage.

6.4 Thermal Information

| THERMAL METRIC ⁽¹⁾ | | TPS6281x-Q1 | UNIT |
|-------------------------------|--|-------------|------|
| | | RWY | |
| | | 9 PINS | |
| R_{JA} | Junction-to-ambient thermal resistance | 71.1 | °C/W |
| $R_{JC(\text{top})}$ | Junction-to-case (top) thermal resistance | 37.2 | °C/W |
| R_{JB} | Junction-to-board thermal resistance | 16.4 | °C/W |
| Ψ_{JT} | Junction-to-top characterization parameter | 0.9 | °C/W |
| Ψ_{JB} | Junction-to-board characterization parameter | 16.1 | °C/W |

6.4 Thermal Information (continued)

| THERMAL METRIC ⁽¹⁾ | | TPS6281x-Q1 | UNIT |
|-------------------------------|--|-------------|------|
| | | RWY | |
| | | 9 PINS | |
| $R_{\theta JC(bot)}$ | Junction-to-case (bottom) thermal resistance | n/a | °C/W |

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics application report](#).

6.5 Electrical Characteristics

over operating junction temperature ($T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$) and $V_{IN} = 2.75\text{ V}$ to 6 V . Typical values at $V_{IN} = 5\text{ V}$ and $T_J = 25^\circ\text{C}$. (unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---|---|--|------|------|----------|------|
| SUPPLY | | | | | | |
| I_Q | Operating Quiescent Current | EN = high, $I_{OUT} = 0\text{ mA}$, Device not switching, $T_J = 125^\circ\text{C}$ | | | 21 | µA |
| I_Q | Operating Quiescent Current | EN = high, $I_{OUT} = 0\text{ mA}$, Device not switching | | 15 | 30 | µA |
| I_{SD} | Shutdown Current | EN = 0 V, at $T_J = 125^\circ\text{C}$ | | | 18 | µA |
| I_{SD} | Shutdown Current | EN = 0 V, Nominal value at $T_J = 25^\circ\text{C}$, Max value at $T_J = 150^\circ\text{C}$ | | 1.5 | 26 | µA |
| V_{UVLO} | Undervoltage Lockout Threshold | Rising Input Voltage | 2.5 | 2.6 | 2.75 | V |
| | | Falling Input Voltage | 2.25 | 2.5 | 2.6 | V |
| T_{SD} | Thermal Shutdown Temperature | Rising Junction Temperature | 170 | | | °C |
| | Thermal Shutdown Hysteresis | | | 15 | | |
| CONTROL (EN, SS/TR, PG, MODE/SYNC) | | | | | | |
| V_{IH} | High Level Input Voltage for MODE/SYNC Pin | | 1.1 | | | V |
| V_{IL} | Low Level Input Voltage for MODE/SYNC Pin | | | 0.3 | | V |
| f_{SYNC} | Frequency Range on MODE/SYNC Pin for Synchronization | requires a resistor from COMP/FSET to GND, see application section | 1.8 | 4 | | MHz |
| | Duty Cycle of Synchronization Signal at MODE/SYNC Pin | | 20% | 50% | 80% | |
| | Time to Lock to External Frequency | | | 50 | | µs |
| V_{IH} | Input Threshold Voltage for EN pin; Rising Edge | | 1.06 | 1.1 | 1.15 | V |
| V_{IL} | Input Threshold Voltage for EN pin; Falling Edge | | 0.96 | 1.0 | 1.05 | V |
| I_{LKG} | Input Leakage Current for EN, MODE/SYNC | $V_{IH} = V_{IN}$ or $V_{IL} = \text{GND}$ | | | 150 | nA |
| | Resistance from COMP/FSET to GND for Logic Low | internal frequency setting with $f = 2.25\text{ MHz}$ | 0 | | 2.5 | kΩ |
| | voltage on COMP/FSET for logic high | internal frequency setting with $f = 2.25\text{ MHz}$ | | | V_{IN} | V |
| V_{TH_PG} | UVP Power Good Threshold Voltage; dc Level | Rising (% V_{FB}) | 92% | 95% | 98% | |
| | UVP Power Good Threshold Voltage; dc Level | Falling (% V_{FB}) | 87% | 90% | 93% | |
| | OVP Power Good Threshold; dc Level | Rising (% V_{FB}) | 107% | 110% | 113% | |
| | OVP Power Good Threshold; dc Level | Falling (% V_{FB}) | 104% | 107% | 111% | |
| | Power Good De-glitch Time | for a high level to low level transition on power good | | 40 | | µs |
| V_{OL_PG} | Power Good Output Low Voltage | $I_{PG} = 2\text{ mA}$ | | 0.07 | 0.3 | V |

6.5 Electrical Characteristics (continued)

over operating junction temperature ($T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$) and $V_{IN} = 2.75\text{ V}$ to 6 V . Typical values at $V_{IN} = 5\text{ V}$ and $T_J = 25^\circ\text{C}$. (unless otherwise noted)

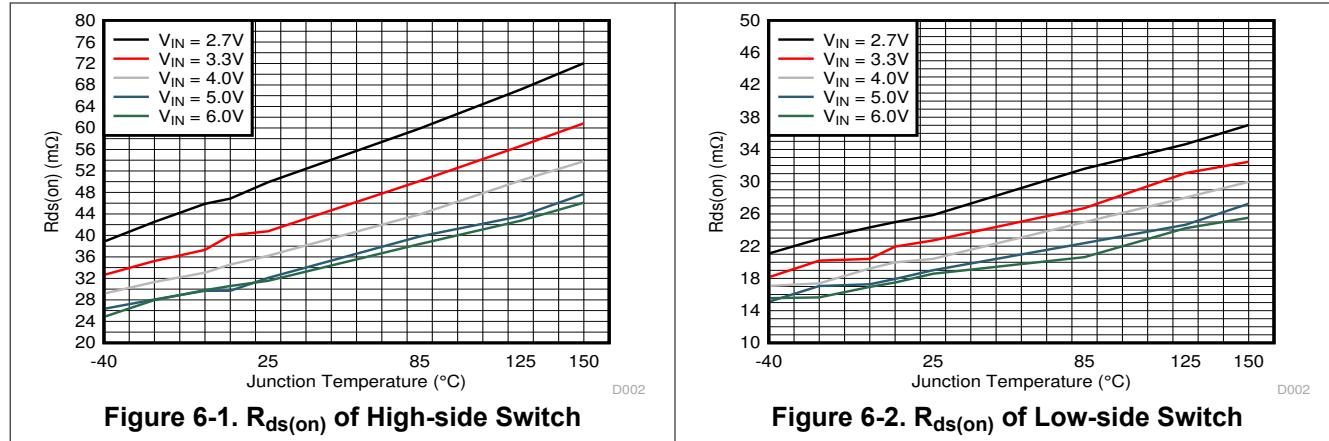
| PARAMETER | TEST CONDITIONS | | MIN | TYP | MAX | UNIT | |
|---------------------|-----------------------------------|--|-----|--------|------|------------------|-----|
| I_{LKG_PG} | $V_{PG} = 5\text{ V}$ | | | 100 | | nA | |
| $I_{SS/TR}$ | SS/TR Pin Source Current | | 2.1 | 2.5 | 2.8 | μA | |
| | Tracking Gain | | 1 | | | | |
| | Tracking Offset | | 17 | | | mV | |
| POWER SWITCH | | | | | | | |
| $R_{DS(ON)}$ | High-Side MOSFET ON-Resistance | $V_{IN} \geq 5\text{ V}$ | | 37 | 60 | $\text{m}\Omega$ | |
| $R_{DS(ON)}$ | Low-Side MOSFET ON-Resistance | $V_{IN} \geq 5\text{ V}$ | | 15 | 35 | $\text{m}\Omega$ | |
| | High-Side MOSFET leakage current | $V_{IN} = 6\text{ V}; V(\text{SW}) = 0\text{ V}$ | | 30 | | μA | |
| | Low-Side MOSFET leakage current | $V(\text{SW}) = 6\text{ V}$ | | 55 | | μA | |
| | SW leakage | $V(\text{SW}) = 0.6\text{ V};$ current into SW pin | | -0.025 | 30 | μA | |
| I_{LIMH} | High-Side MOSFET Current Limit | dc value, for TPS62810; $V_{IN} = 3\text{ V}$ to 6 V | | 4.8 | 5.6 | 6.55 | A |
| I_{LIMH} | High-Side MOSFET Current Limit | dc value, for TPS62813; $V_{IN} = 3\text{ V}$ to 6 V | | 3.9 | 4.5 | 5.25 | A |
| I_{LIMH} | High-Side MOSFET Current Limit | dc value, for TPS62812; $V_{IN} = 3\text{ V}$ to 6 V | | 2.8 | 3.4 | 4.2 | A |
| I_{LIMH} | High-Side MOSFET Current Limit | dc value, for TPS62811; $V_{IN} = 3\text{ V}$ to 6 V | | 2.0 | 2.6 | 3.25 | A |
| I_{LIMNEG} | Negative Valley Current Limit | dc value | | | -1.8 | | A |
| f_S | PWM Switching Frequency Range | | | 1.8 | 2.25 | 4 | MHz |
| f_S | PWM Switching Frequency | with COMP/FSET tied to VIN or GND | | 2.025 | 2.25 | 2.475 | MHz |
| | PWM Switching Frequency Tolerance | using a resistor from COMP/FSET to GND, $f_S = 1.8\text{ MHz}$ to 4 MHz | | -19% | | 18% | |
| $t_{on,min}$ | Minimum on-time of HS FET | $T_J = -40^\circ\text{C}$ to 125°C , $V_{IN} = 3.3\text{ V}$ | | 50 | 75 | ns | |
| $t_{on,min}$ | Minimum on-time of LS FET | $V_{IN} = 3.3\text{ V}$ | | 30 | | ns | |
| OUTPUT | | | | | | | |
| V_{FB} | Feedback Voltage | adjustable output voltage versions | | 0.6 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS6281302 | | 0.8 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS6281206, TPS6281126, TPS6281326, TPS6281006 | | 1.0 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS6281208, TPS6281008, TPS6281228 | | 1.1 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS6281109 | | 1.15 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS628110A, TPS628112A, TPS628122A, TPS628130A | | 1.2 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS628132D | | 1.35 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS628122G | | 1.5 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS628112M, TPS628120M, TPS628132D, TPS628100M, TPS628132M | | 1.8 | | V | |
| V_{FB} | Feedback Voltage | fixed output voltage TPS628113H | | 3.3 | | V | |

6.5 Electrical Characteristics (continued)

over operating junction temperature ($T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$) and $V_{IN} = 2.75\text{ V}$ to 6 V . Typical values at $V_{IN} = 5\text{ V}$ and $T_J = 25^\circ\text{C}$. (unless otherwise noted)

| PARAMETER | TEST CONDITIONS | | MIN | TYP | MAX | UNIT |
|---------------|---|--|--|--|------|---------------|
| I_{LKG_FB} | FB Input Leakage Current for Adjustable Voltage Versions | | $V_{FB} = 0.6\text{ V}$ | | 1 | 70 |
| I_{LKG_FB} | FB Input Current for Fixed Voltage Versions | | V_{FB} voltage at target output voltage | | 1 | μA |
| V_{FB} | Feedback Voltage Accuracy for Adjustable Voltage Versions | | $V_{IN} \geq V_{OUT} + 1\text{ V}$ | PWM mode | -1% | 1% |
| V_{FB} | Feedback Voltage Accuracy for Fixed Voltage Versions | | $V_{IN} \geq V_{OUT} + 1\text{ V}$ | PWM mode, $T_J = -40^\circ\text{C}$ to 125°C | -1% | 1% |
| V_{FB} | Feedback Voltage Accuracy for Fixed Voltage Versions | | $V_{IN} \geq V_{OUT} + 1\text{ V}$ | PWM mode | -1% | 1.3% |
| V_{FB} | Feedback Voltage Accuracy | | $V_{IN} \geq V_{OUT} + 1\text{ V}; V_{OUT} \geq 1.5\text{ V}$ | PFM mode; $C_{O,\text{eff}} \geq 22\text{ }\mu\text{F}, L = 0.47\text{ }\mu\text{H}$ | -1% | 2% |
| V_{FB} | Feedback Voltage Accuracy | | $1\text{ V} \leq V_{OUT} < 1.5\text{ V}$ | PFM mode; $C_{O,\text{eff}} \geq 47\text{ }\mu\text{F}, L = 0.47\text{ }\mu\text{H}$ | -1% | 2.5% |
| V_{FB} | Feedback Voltage Accuracy with Voltage Tracking | | $V_{IN} \geq V_{OUT} + 1\text{ V}; V_{SS/TR} = 0.3\text{ V}$ | PWM mode | -1% | 7% |
| | Load Regulation | | PWM mode operation | | 0.05 | %/A |
| | Line Regulation | | PWM mode operation, $I_{OUT} = 1\text{ A}, V_{IN} \geq V_{OUT} + 1\text{ V}$ | | 0.02 | %/V |
| | Output Discharge Resistance | | | | 50 | Ω |
| t_{delay} | Start-up Delay Time | | $I_{OUT} = 0\text{ mA}$, Time from EN=high to start switching; V_{IN} applied already | 135 | 250 | 470 |
| t_{ramp} | Ramp time; SS/TR Pin Open | | $I_{OUT} = 0\text{ mA}$, Time from first switching pulse until 95% of nominal output voltage; device not in current limit | 100 | 150 | 200 |

6.6 Typical Characteristics



7 Parameter Measurement Information

7.1 Schematic

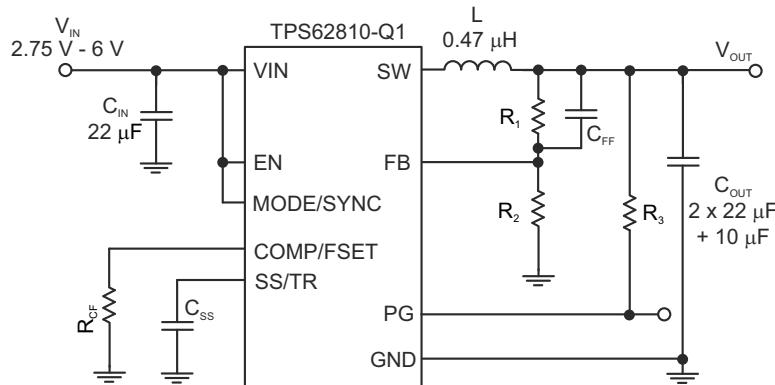


Figure 7-1. Measurement Setup for TPS62810-Q1 and TPS62813-Q1

Table 7-1. List of Components

| REFERENCE | DESCRIPTION | MANUFACTURER ⁽¹⁾ |
|------------------|---|-----------------------------|
| IC | TPS62810-Q1 or TPS62813-Q1 | Texas Instruments |
| L | 0.47-μH inductor; XEL4030-471MEB | Coilcraft |
| C _{IN} | 22 μF / 10 V; GCM31CR71A226KE02L | Murata |
| C _{OUT} | 2 × 22 μF / 10 V; GCM31CR71A226KE02L + 1 × 10 μF 6.3 V; GCM188D70J106ME36 | Murata |
| C _{SS} | 4.7 nF (equal to 1-ms start-up ramp) | Any |
| R _{CF} | 8.06 kΩ | Any |
| C _{FF} | 10 pF | Any |
| R ₁ | Depending on V _{OUT} | Any |
| R ₂ | Depending on V _{OUT} | Any |
| R ₃ | 100 kΩ | Any |

(1) See the [Third-party Products Disclaimer](#).

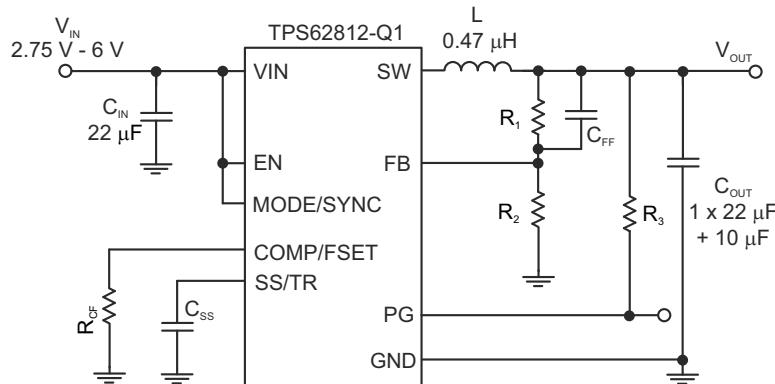


Figure 7-2. Measurement Setup for TPS62812-Q1 and TPS62811-Q1

Table 7-2. List of Components

| REFERENCE | DESCRIPTION | MANUFACTURER ⁽¹⁾ |
|------------------|--|-----------------------------|
| IC | TPS62812-Q1 or TPS62811-Q1 | Texas Instruments |
| L | 0.56- μ H inductor; XEL4020-561MEB | Coilcraft |
| C _{IN} | 22 μ F / 10 V; GCM31CR71A226KE02L | Murata |
| C _{OUT} | 1 \times 22 μ F / 10 V; GCM31CR71A226KE02L + 1 \times 10 μ F 6.3 V; GCM188D70J106ME36 | Murata |
| C _{SS} | 4.7 nF (equal to 1-ms start-up ramp) | Any |
| R _{CF} | 8.06 k Ω | Any |
| C _{FF} | 10 pF | Any |
| R ₁ | Depending on V _{OUT} | Any |
| R ₂ | Depending on V _{OUT} | Any |
| R ₃ | 100 k Ω | Any |

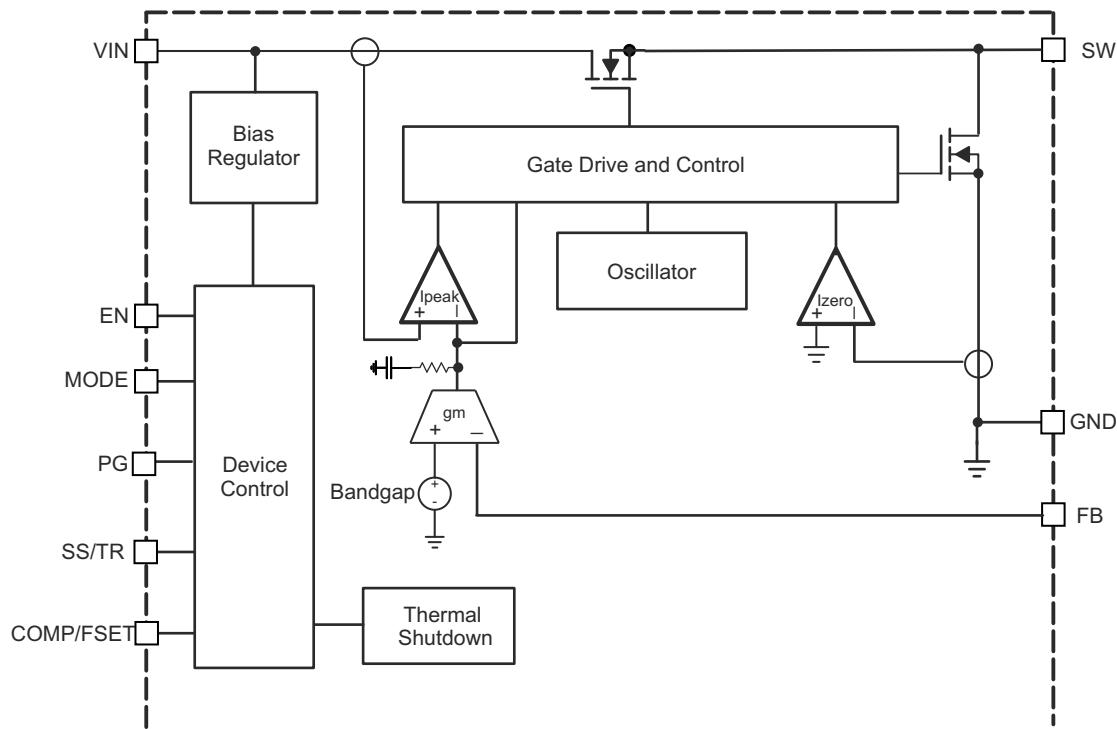
8 Detailed Description

8.1 Overview

The TPS6281x-Q1 synchronous switch mode DC/DC converters are based on a peak current mode control topology. The control loop is internally compensated. To optimize the bandwidth of the control loop to the wide range of output capacitance that can be used with TPS6281x-Q1, one of three internal compensation settings can be selected. See [Section 8.3.2](#). The compensation setting is selected either by a resistor from COMP/FSET to GND, or by the logic state of this pin. The regulation network achieves fast and stable operation with small external components and low ESR ceramic output capacitors. The device can be operated without a feedforward capacitor on the output voltage divider, however, using a typically 10-pF feedforward capacitor improves transient response.

The devices support forced fixed frequency PWM operation with the MODE pin tied to a logic high level. The frequency is defined as either 2.25 MHz internally fixed when COMP/FSET is tied to GND or VIN, or in a range of 1.8 MHz to 4 MHz defined by a resistor from COMP/FSET to GND. Alternatively, the devices can be synchronized to an external clock signal in a range from 1.8 MHz to 4 MHz, applied to the MODE pin with no need for additional passive components. External synchronization is only possible if a resistor from COMP/FSET to GND is used. If COMP/FSET is directly tied to GND or VIN, the TPS6281x-Q1 cannot be synchronized externally. An internal PLL allows to change from internal clock to external clock during operation. The synchronization to the external clock is done on a falling edge of the clock applied at MODE to the rising edge on the SW pin. This allows a roughly 180° phase shift when the SW pin is used to generate the synchronization signal for a second converter. When the MODE pin is set to a logic low level, the device operates in power save mode (PFM) at low output current and automatically transfers to fixed frequency PWM mode at higher output current. In PFM mode, the switching frequency decreases linearly based on the load to sustain high efficiency down to very low output current.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Precise Enable

The voltage applied at the Enable pin of the TPS6281x-Q1 is compared to a fixed threshold of 1.1 V for a rising voltage. This allows to drive the pin by a slowly changing voltage and enables the use of an external RC network to achieve a power-up delay.

The Precise Enable input provides a user-programmable undervoltage lockout by adding a resistor divider to the input of the Enable pin.

The enable input threshold for a falling edge is typically 100 mV lower than the rising edge threshold. The TPS6281x-Q1 starts operation when the rising threshold is exceeded. For proper operation, the EN pin must be terminated and must not be left floating. Pulling the EN pin low forces the device into shutdown, with a shutdown current of typically 1 µA. In this mode, the internal high-side and low-side MOSFETs are turned off and the entire internal control circuitry is switched off.

8.3.2 COMP/FSET

This pin allows to set two different parameters independently:

- Internal compensation settings for the control loop
- The switching frequency in PWM mode from 1.8 MHz to 4 MHz

A resistor from COMP/FSET to GND changes the compensation as well as the switching frequency. The change in compensation allows you to adapt the device to different values of output capacitance. The resistor must be placed close to the pin to keep the parasitic capacitance on the pin to a minimum. The compensation setting is sampled at start-up of the converter, so a change in the resistor during operation only has an effect on the switching frequency but not on the compensation.

To save external components, the pin can also be directly tied to VIN or GND to set a pre-defined switching frequency / compensation. Do not leave the pin floating.

The switching frequency has to be selected based on the input voltage and the output voltage to meet the specifications for the minimum on-time and minimum off-time.

For example: $V_{IN} = 5 \text{ V}$, $V_{OUT} = 1 \text{ V}$ --> duty cycle (DC) = $1 \text{ V} / 5 \text{ V} = 0.2$

- with $t_{on} = DC \times T \rightarrow t_{on,min} = 1 / f_{s,max} \times DC$
- --> $f_{s,max} = 1 / t_{on,min} \times DC = 1 / 0.075 \mu\text{s} \cdot 0.2 = 2.67 \text{ MHz}$

The compensation range has to be chosen based on the minimum capacitance used. The capacitance can be increased from the minimum value as given in [Table 8-1](#) and [Table 8-2](#), up to the maximum of 470 µF in all of the three compensation ranges. If the capacitance of an output changes during operation, for example, when load switches are used to connect or disconnect parts of the circuitry, the compensation has to be chosen for the minimum capacitance on the output. With large output capacitance, the compensation must be done based on that large capacitance to get the best load transient response. Compensating for large output capacitance but placing less capacitance on the output can lead to instability.

The switching frequency for the different compensation setting is determined by the following equations.

For compensation (comp) setting 1:

$$R_{CF}(k\Omega) = \frac{18\text{MHz} \cdot k\Omega}{f_s(\text{MHz})} \quad (1)$$

For compensation (comp) setting 2:

$$R_{CF}(k\Omega) = \frac{60MHz \cdot k\Omega}{f_s(MHz)} \quad (2)$$

For compensation (comp) setting 3:

$$R_{CF}(k\Omega) = \frac{180MHz \cdot k\Omega}{f_s(MHz)} \quad (3)$$

Table 8-1. Switching Frequency and Compensation for TPS62810-Q1 (4 A) and TPS62813-Q1 (3 A)

| COMPENSATION | R _{CF} | SWITCHING FREQUENCY | MINIMUM OUTPUT CAPACITANCE FOR V _{OUT} < 1 V | MINIMUM OUTPUT CAPACITANCE FOR 1 V ≤ V _{OUT} < 3.3 V | MINIMUM OUTPUT CAPACITANCE FOR V _{OUT} ≥ 3.3 V |
|--|-------------------------|--|---|---|---|
| for smallest output capacitance (comp setting 1) | 10 kΩ ... 4.5 kΩ | 1.8 MHz (10 kΩ) ... 4 MHz (4.5 kΩ) according to Equation 1 | 53 µF | 32 µF | 27 µF |
| for medium output capacitance (comp setting 2) | 33 kΩ ... 15 kΩ | 1.8 MHz (33 kΩ) ... 4 MHz (15 kΩ) according to Equation 2 | 100 µF | 60 µF | 50 µF |
| for large output capacitance (comp setting 3) | 100 kΩ ... 45 kΩ | 1.8 MHz (100 kΩ) ... 4 MHz (45 kΩ) according to Equation 3 | 200 µF | 120 µF | 100 µF |
| for smallest output capacitance (comp setting 1) | tied to GND | internally fixed 2.25 MHz | 53 µF | 32 µF | 27 µF |
| for large output capacitance (comp setting 3) | tied to V _{IN} | internally fixed 2.25 MHz | 200 µF | 120 µF | 100 µF |

Table 8-2. Switching Frequency and Compensation for TPS62812-Q1 (2 A) and TPS62811-Q1 (1 A)

| COMPENSATION | R _{CF} | SWITCHING FREQUENCY | MINIMUM OUTPUT CAPACITANCE FOR V _{OUT} < 1 V | MINIMUM OUTPUT CAPACITANCE FOR 1 V ≤ V _{OUT} < 3.3 V | MINIMUM OUTPUT CAPACITANCE FOR V _{OUT} ≥ 3.3 V |
|--|-------------------------|--|---|---|---|
| for smallest output capacitance (comp setting 1) | 10 kΩ ... 4.5 kΩ | 1.8 MHz (10 kΩ) ... 4 MHz (4.5 kΩ) according to Equation 1 | 30 µF | 18 µF | 15 µF |
| for medium output capacitance (comp setting 2) | 33 kΩ ... 15 kΩ | 1.8 MHz (33 kΩ) ... 4 MHz (15 kΩ) according to Equation 2 | 60 µF | 36 µF | 30 µF |
| for large output capacitance (comp setting 3) | 100 kΩ ... 45 kΩ | 1.8 MHz (100 kΩ) ... 4 MHz (45 kΩ) according to Equation 3 | 130 µF | 80 µF | 68 µF |
| for smallest output capacitance (comp setting 1) | tied to GND | internally fixed 2.25 MHz | 30 µF | 18 µF | 15 µF |
| for large output capacitance (comp setting 3) | tied to V _{IN} | internally fixed 2.25 MHz | 130 µF | 80 µF | 68 µF |

Refer to [Section 9.1.3.2](#) for further details on the output capacitance required depending on the output voltage.

A too high resistor value for R_{CF} is decoded as "tied to V_{IN}", a value below the lowest range is decoded as "tied to GND". The minimum output capacitance in [Table 8-1](#) and [Table 8-2](#) is for capacitors close to the output of the device. If the capacitance is distributed, a lower compensation setting can be required. All values are effective capacitance, including all tolerances, aging, dc bias effect, and so forth.

8.3.3 MODE / SYNC

When MODE/SYNC is set low, the device operates in PWM or PFM mode, depending on the output current. The MODE/SYNC pin allows to force PWM mode when set high. The pin also allows the user to apply an external clock in a frequency range from 1.8 MHz to 4 MHz for external synchronization. Similar to COMP/FSET, the specifications for the minimum on-time and minimum off-time must be taken into account when setting the external frequency. For use with external synchronization on the MODE/SYNC pin, the internal switching frequency must be set by R_{CF} to a similar value than the externally applied clock. This action makes sure of a fast settling to the external clock and, if the external clock fails, the switching frequency stays in the same range and the compensation settings are still valid. When there is no resistor from COMP/FSET to GND but the pin is pulled high or low, external synchronization is not possible.

8.3.4 Spread Spectrum Clocking (SSC)

For device versions with SSC enabled, the switching frequency is randomly changed in PWM mode when the internal clock is used. The frequency variation is typically between the nominal switching frequency and up to 288 kHz above the nominal switching frequency. When the device is externally synchronized by applying a clock signal to the MODE/SYNC pin, the TPS6281x-Q1 follows the external clock and the internal spread spectrum block is turned off. SSC is also disabled during soft start.

8.3.5 Undervoltage Lockout (UVLO)

If the input voltage drops, the undervoltage lockout prevents mis-operation of the device by switching off both the power FETs. The device is fully operational for voltages above the rising UVLO threshold and turns off if the input voltage trips below the threshold for a falling supply voltage.

8.3.6 Power Good Output (PG)

Power good is an open-drain output driven by a window comparator. PG is held low when the device is disabled, in undervoltage lockout, and in thermal shutdown. When the output voltage is in regulation hence, within the window defined in the electrical characteristics, the output is high impedance.

Table 8-3. PG Status

| EN | DEVICE STATUS | PG STATE |
|------|---|----------------|
| X | $V_{IN} < 2\text{ V}$ | undefined |
| low | $V_{IN} \geq 2\text{ V}$ | low |
| high | $2\text{ V} \leq V_{IN} \leq \text{UVLO}$ OR in thermal shutdown OR V_{OUT} not in regulation | low |
| high | V_{OUT} in regulation | high impedance |

8.3.7 Thermal Shutdown

The junction temperature (T_J) of the device is monitored by an internal temperature sensor. If T_J exceeds 170°C (typ), the device goes into thermal shutdown. Both the high-side and low-side power FETs are turned off and PG goes low. When T_J decreases by the hysteresis amount of typically 15°C , the converter resumes normal operation, beginning with soft start. During a PFM pause, the thermal shutdown is not active. After a PFM pause, the device needs up to $9\text{ }\mu\text{s}$ to detect a too high junction temperature. If the PFM burst is shorter than this delay, the device does not detect a too high junction temperature.

8.4 Device Functional Modes

8.4.1 Pulse Width Modulation (PWM) Operation

TPS6281x-Q1 has two operating modes: Forced PWM mode (discussed in this section) and PWM/PFM (discussed in [Section 8.4.2](#)).

With the MODE/SYNC pin set to high, the TPS6281x-Q1 operates with pulse width modulation in continuous conduction mode (CCM). The switching frequency is either defined by a resistor from the COMP pin to GND or by an external clock signal applied to the MODE/SYNC pin. With an external clock is applied to MODE/SYNC, the TPS6281x-Q1 follows the frequency applied to the pin. To maintain regulation, the frequency must be in a range the TPS6281x-Q1 can operate at, taking the minimum on-time into account.

8.4.2 Power Save Mode Operation (PWM/PFM)

When the MODE/SYNC pin is low, power save mode is allowed. The device operates in PWM mode as long as the peak inductor current is above the PFM threshold of about 1.2 A. When the peak inductor current drops below the PFM threshold, the device starts to skip switching pulses. In power save mode, the switching frequency decreases with the load current maintaining high efficiency.

8.4.3 100% Duty-Cycle Operation

The duty cycle of a buck converter operated in PWM mode is given as $D = V_{OUT} / V_{IN}$. The duty cycle increases as the input voltage comes close to the output voltage and the off-time gets smaller. When the minimum off-time of typically 30 ns is reached, the TPS6281x-Q1 skips switching cycles while approaching 100% mode. In 100% mode, the device keeps the high-side switch on continuously. The high-side switch stays turned on as long as the output voltage is below the target. In 100% mode, the low-side switch is turned off. The maximum dropout voltage in 100% mode is the product of the on-resistance of the high-side switch plus the series resistance of the inductor and the load current.

8.4.4 Current Limit and Short-Circuit Protection

The TPS6281x-Q1 is protected against overload and short circuit events. If the inductor current exceeds the current limit I_{LIMH} , the high-side switch is turned off and the low-side switch is turned on to ramp down the inductor current. The high-side switch turns on again only if the current in the low-side switch has decreased below the low-side current limit. Due to internal propagation delay, the actual current can exceed the static current limit. The dynamic current limit is given as:

$$I_{peak(typ)} = I_{LIMH} + \frac{V_L}{L} \cdot t_{PD} \quad (4)$$

where

- I_{LIMH} is the static current limit as specified in the electrical characteristics
- L is the effective inductance at the peak current
- V_L is the voltage across the inductor ($V_{IN} - V_{OUT}$)
- t_{PD} is the internal propagation delay of typically 50 ns

The current limit can exceed static values, especially if the input voltage is high and very small inductances are used. The dynamic high-side switch peak current can be calculated as follows:

$$I_{peak(typ)} = I_{LIMH} + \frac{V_{IN} - V_{OUT}}{L} \cdot 50\text{ns} \quad (5)$$

8.4.5 Foldback Current Limit and Short-Circuit Protection

Foldback current limit and short-circuit protection is valid for devices where foldback current limit is enabled.

When the device detects current limit for more than 1024 subsequent switching cycles, the device reduces the current limit from the nominal value to typically 1.8 A. Foldback current limit is left when the current limit indication goes away. For the case that device operation continues in current limit, the device can, after 3072 switching cycles, try again full current limit for again 1024 switching cycles.

8.4.6 Output Discharge

The purpose of the discharge function is to make sure of a defined down-ramp of the output voltage when the device is being disabled but also to keep the output voltage close to 0 V when the device is off. The output discharge feature is only active after TPS6281x-Q1 has been enabled at least once after the supply voltage was applied. The discharge function is enabled as soon as the device is disabled, in thermal shutdown, or in undervoltage lockout. The minimum supply voltage required for the discharge function to remain active typically is 2 V. Output discharge is not activated during a current limit or foldback current limit event. When supply voltage is applied but the device has not been enabled yet, the discharge function is undefined. For applications

where the output voltage discharge has to be off after supply voltage is applied and the enable pin is still at low level, use the versions that are stated as *Output discharge disconnected* in the [Device Comparison Table](#).

8.4.7 Soft Start / Tracking (SS/TR)

The internal soft-start circuitry controls the output voltage slope during start-up. This control avoids excessive inrush current and makes sure of a controlled output voltage rise time. This control also prevents unwanted voltage drops from high impedance power sources or batteries. When EN is set high to start operation, the device starts switching after a delay of about 200 µs then the internal reference and hence V_{OUT} rises with a slope controlled by an external capacitor connected to the SS/TR pin.

Leaving the SS/TR pin unconnected provides the fastest start-up ramp with 150 µs typically. A capacitor connected from SS/TR to GND is charged with 2.5 µA by an internal current source during soft start until reaching the reference voltage of 0.6 V. The capacitance required to set a certain ramp-time (t_{ramp}) therefore is:

$$C_{SS} [nF] = \frac{2.5 \mu A \cdot t_{ramp} [ms]}{0.6V} \quad (6)$$

If the device is set to shutdown (EN = GND), undervoltage lockout, or thermal shutdown, an internal resistor pulls the SS/TR pin to GND to make sure of a proper low level. Returning from those states causes a new start-up sequence.

A voltage applied at SS/TR can be used to track a controller voltage. The output voltage follows this voltage in both directions up and down in forced PWM mode. In PFM mode, the output voltage decreases based on the load current. The SS/TR pin must not be connected to the SS/TR pin of other devices. An external voltage applied on SS/TR is internally clamped to the feedback voltage (0.6 V). TI recommends to set the target for the external voltage on SS/TR slightly above the feedback voltage. Given the tolerances of the resistor divider R_5 and R_6 on SS/TR, this action makes sure the device *switches* to the internal reference voltage when the power-up sequencing is finished. See [Figure 9-58](#).

9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

9.1 Application Information

9.1.1 Programming the Output Voltage

The output voltage of the TPS6281x-Q1 is adjustable. The output voltage can be programmed for output voltages from 0.6 V to 5.5 V using a resistor divider from VOUT to GND. The voltage at the FB pin is regulated to 600 mV. The value of the output voltage is set by the selection of the resistor divider from [Equation 7](#). TI recommends to choose resistor values which allow a current of at least 2 μ A, meaning the value of R_2 must not exceed 400 k Ω . TI recommends lower resistor values for highest accuracy and most robust design.

$$R_1 = R_2 \cdot \left(\frac{V_{OUT}}{V_{FB}} - 1 \right) \quad (7)$$

9.1.2 External Component Selection

9.1.2.1 Inductor Selection

The TPS6281x-Q1 is designed for a nominal 0.47- μ H inductor with a switching frequency of typically 2.25 MHz. Larger values can be used to achieve a lower inductor current ripple but can have a negative impact on efficiency and transient response. Smaller values than 0.47 μ H cause a larger inductor current ripple which causes larger negative inductor current in forced PWM mode at low or no output current. For a higher or lower nominal switching frequency, the inductance must be changed accordingly. See [Recommended Operating Conditions](#) for details.

The inductor selection is affected by several effects like inductor ripple current, output ripple voltage, PWM-to-PFM transition point, and efficiency. In addition, the inductor selected has to be rated for appropriate saturation current and DC resistance (DCR). [Equation 8](#) calculates the maximum inductor current.

$$I_{L(\max)} = I_{OUT(\max)} + \frac{\Delta I_{L(\max)}}{2} \quad (8)$$

$$\Delta I_{L(\max)} = \frac{V_{OUT} \cdot \left(1 - \frac{V_{OUT}}{V_{IN}} \right)}{L_{\min}} \cdot \frac{1}{f_{sw}} \quad (9)$$

where

- $I_{L(\max)}$ is the maximum inductor current
- $\Delta I_{L(\max)}$ is the peak-to-peak inductor ripple current
- L_{\min} is the minimum inductance at the operating point

Table 9-1. Typical Inductors

| TYPE | INDUCTANCE [μ H] | CURRENT [A] ⁽¹⁾ | FOR DEVICE | NOMINAL SWITCHING FREQUENCY | DIMENSIONS [LxBxH] mm | MANUFACTURER ⁽²⁾ |
|------------------|--------------------------|-------------------------------|---|-----------------------------------|-------------------------------|-----------------------------|
| XFL4015-471ME | 0.47 μ H, $\pm 20\%$ | 3.5 | TPS62813-Q1, TPS62812-Q1 | 2.25 MHz | 4 \times 4 \times 1.6 | Coilcraft |
| XEL4020-561ME | 0.56 μ H, $\pm 20\%$ | 9.9 | TPS62810-Q1, TPS62813-Q1, TPS62812-Q1 | 2.25 MHz | 4 \times 4 \times 2.1 | Coilcraft |
| XEL4030-471ME | 0.47 μ H, $\pm 20\%$ | 12.3 | TPS62810-Q1, TPS62813-Q1, TPS62812-Q1 | 2.25 MHz | 4 \times 4 \times 3.1 | Coilcraft |
| XEL3515-561ME | 0.56 μ H, $\pm 20\%$ | 4.5 | TPS62813-Q1, TPS62812-Q1 | 2.25 MHz | 3.5 \times 3.2 \times 1.5 | Coilcraft |
| XFL3012-331MEB | 0.33 μ H, $\pm 20\%$ | 2.6 | TPS62811-Q1 TPS62812-Q1 | ≥ 3.5 MHz | 3 \times 3 \times 1.3 | Coilcraft |
| XPL2010-681ML | 0.68 μ H, $\pm 20\%$ | 1.5 | TPS62811-Q1 | 2.25 MHz | 2 \times 1.9 \times 1 | Coilcraft |
| DFE252012PD-R47M | 0.47 μ H, $\pm 20\%$ | see data sheet | TPS62812-Q1, TPS62813-Q1 | 2.25 MHz | 2.5 \times 2 \times 1.2 | Murata |

(1) Lower of I_{RMS} at 20°C rise or I_{SAT} at 20% drop.(2) See the [Third-party Products Disclaimer](#).

Calculating the maximum inductor current using the actual operating conditions gives the minimum saturation current of the inductor needed. TI recommends a margin of about 20% to add. A larger inductor value is also useful to get lower ripple current, but increases the transient response time and size as well.

9.1.3 Capacitor Selection

9.1.3.1 Input Capacitor

For most applications, 22 μ F nominal is sufficient and recommended. The input capacitor buffers the input voltage for transient events and also decouples the converter from the supply. TI recommends a low-ESR multilayer ceramic capacitor (MLCC) for best filtering and must be placed between VIN and GND as close as possible to those pins.

9.1.3.2 Output Capacitor

The architecture of the TPS6281x-Q1 allows the use of tiny ceramic output capacitors with low equivalent series resistance (ESR). These capacitors provide low output voltage ripple and are recommended. To keep the low resistance up to high frequencies and to get narrow capacitance variation with temperature, TI recommends to use dielectric X7R, X7T, or an equivalent. Using a higher value has advantages like smaller voltage ripple and a tighter DC output accuracy in power save mode. By changing the device compensation with a resistor from COMP/FSET to GND, the device can be compensated in three steps based on the minimum capacitance used on the output. The maximum capacitance is 470 μ F in any of the compensation settings.

The minimum capacitance required on the output depends on the compensation setting as well as on the current rating of the device. TPS62810-Q1 and TPS62813-Q1 require a minimum output capacitance of 27 μ F while the lower current versions TPS62812-Q1 and TPS62811-Q1 requires 15 μ F at minimum. The required output capacitance also changes with the output voltage.

For output voltages below 1 V, the minimum increases linearly from 32 μ F at 1 V to 53 μ F at 0.6 V for the TPS62810-Q1, the TPS62813-Q1 with the compensation setting for smallest output capacitance. Other compensation ranges for TPS62811-Q1 and TPS62812-Q1, or both are equivalent. See [Table 8-1](#) and [Table 8-2](#) for details.

9.2 Typical Application

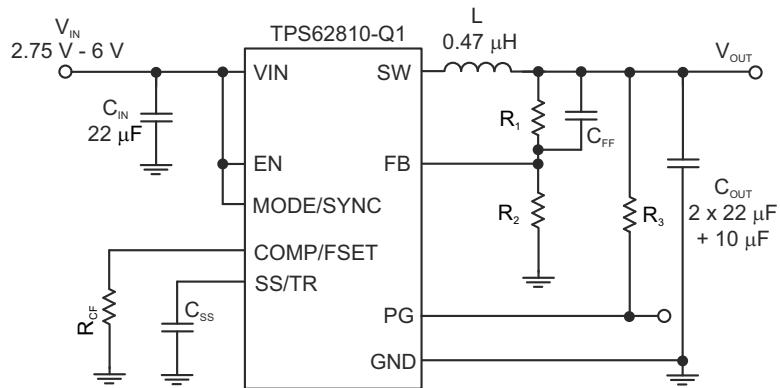


Figure 9-1. Typical Application

9.2.1 Design Requirements

The design guidelines provide a component selection to operate the device within the recommended operating conditions.

9.2.2 Detailed Design Procedure

$$R_1 = R_2 \cdot \left(\frac{V_{OUT}}{V_{FB}} - 1 \right) \quad (10)$$

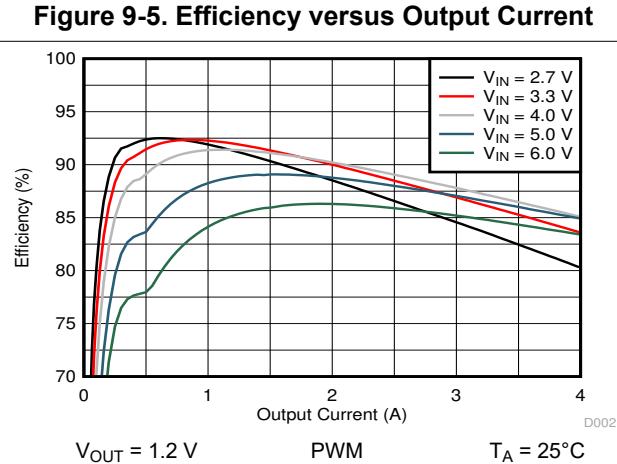
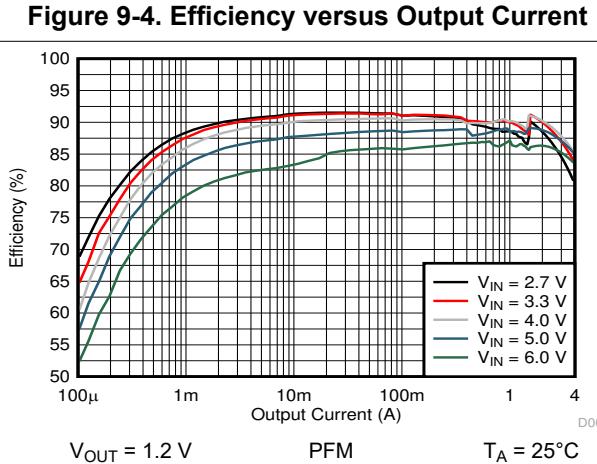
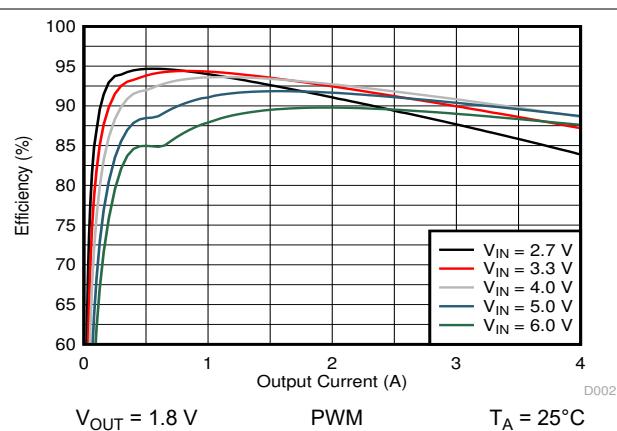
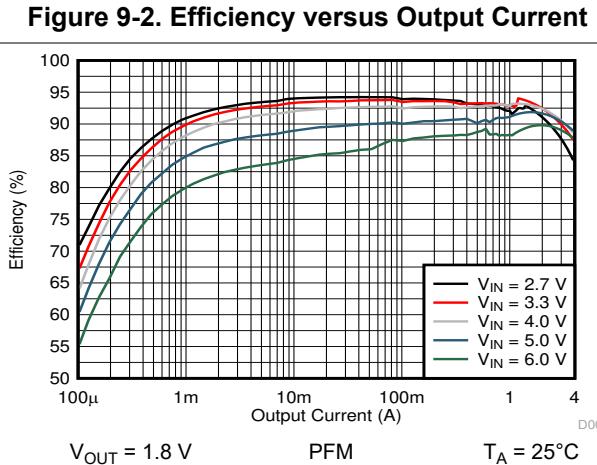
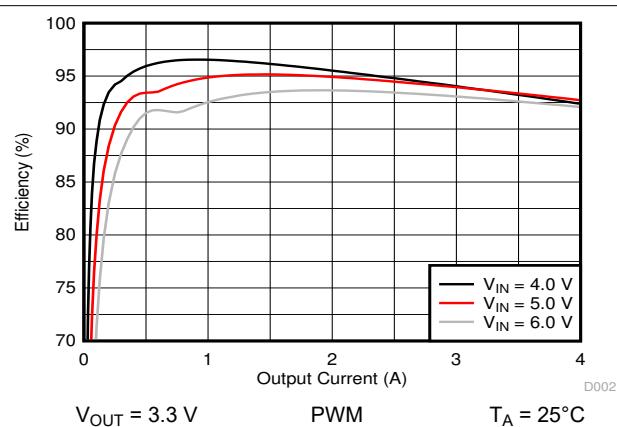
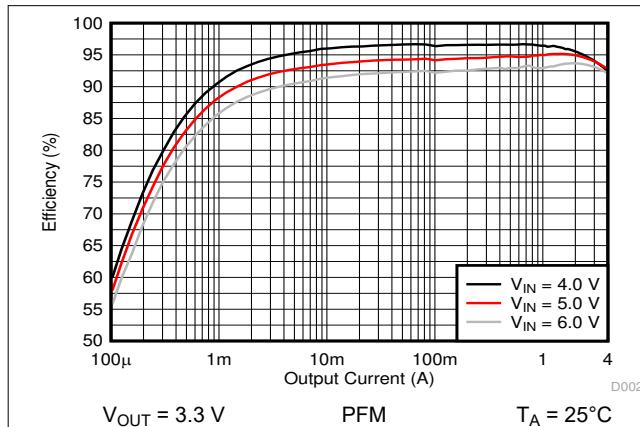
With $V_{FB} = 0.6$ V:

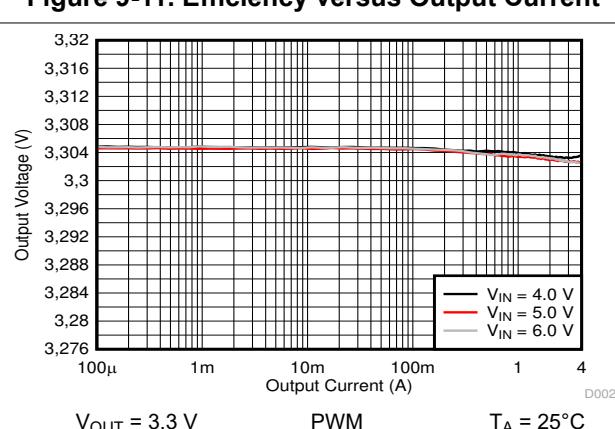
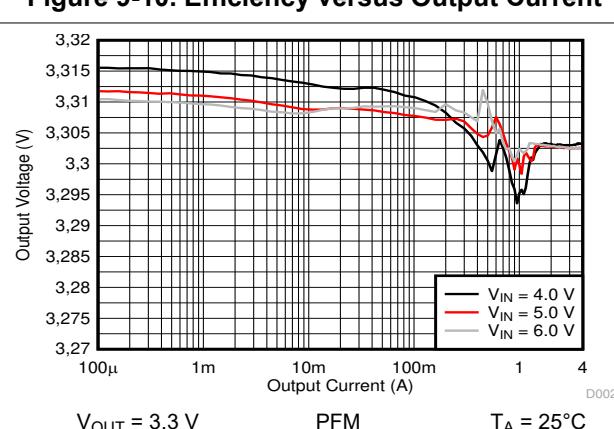
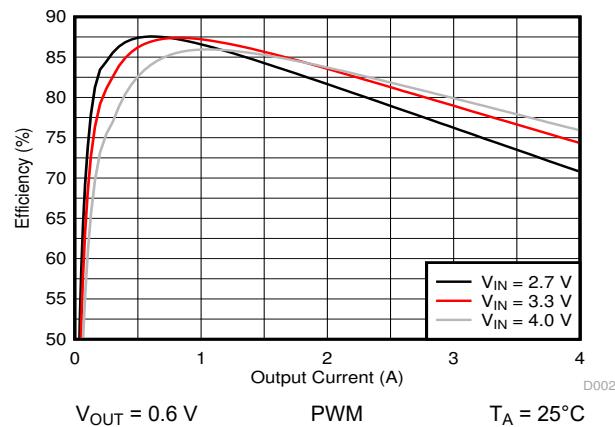
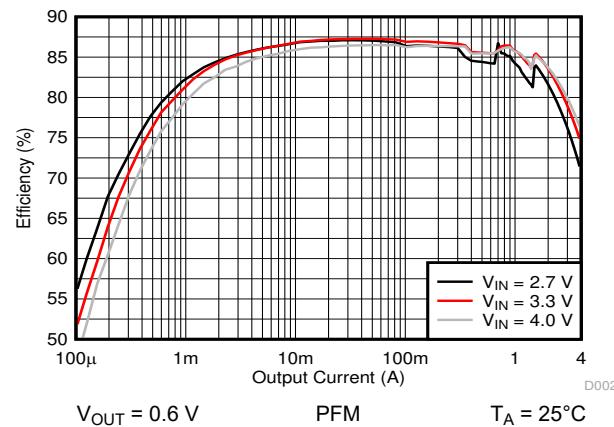
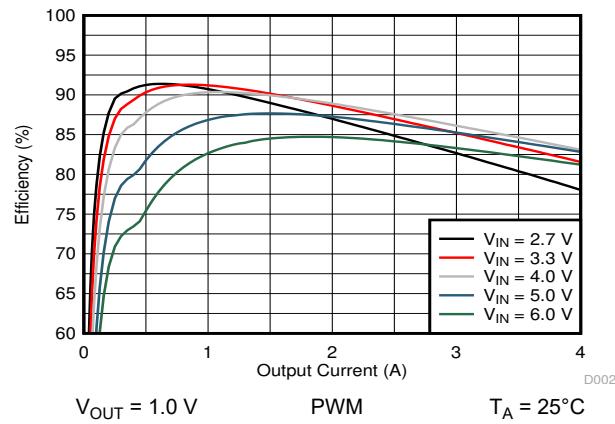
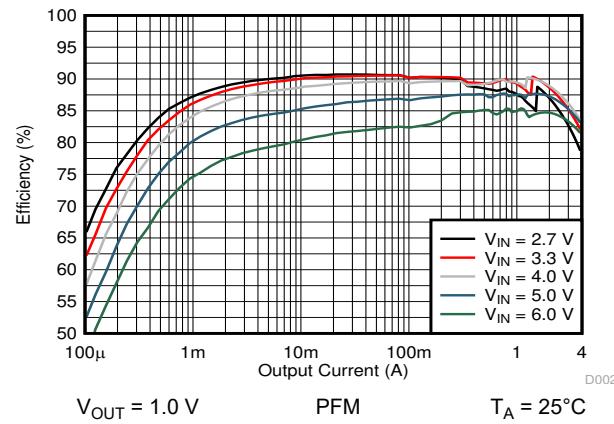
Table 9-2. Setting the Output Voltage

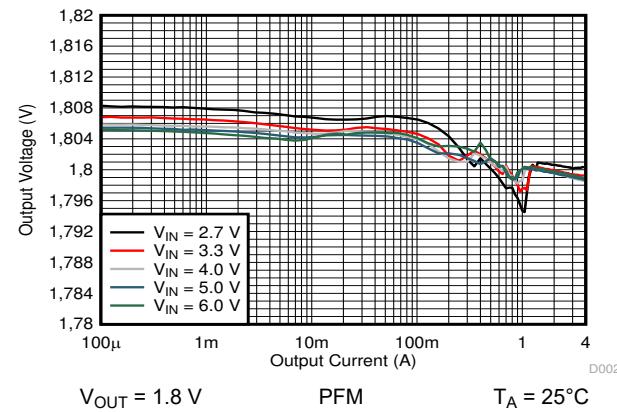
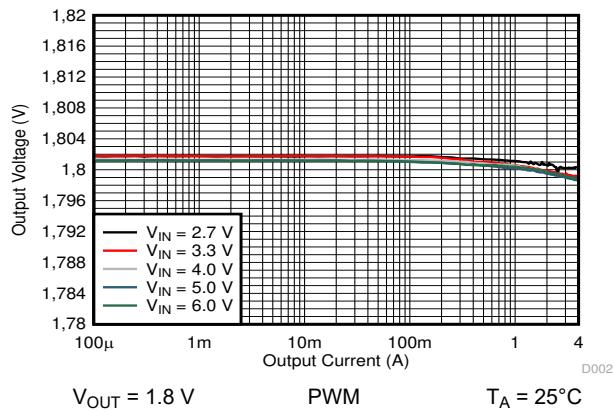
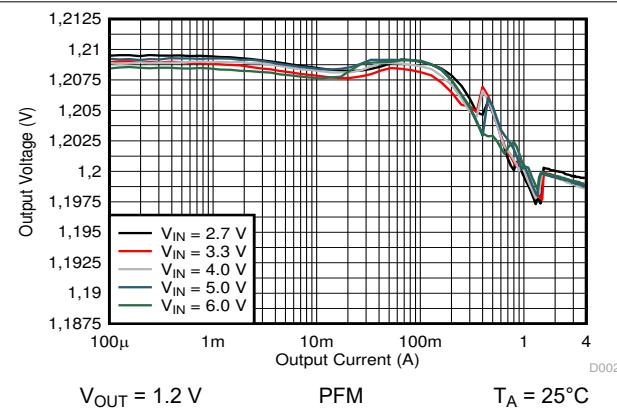
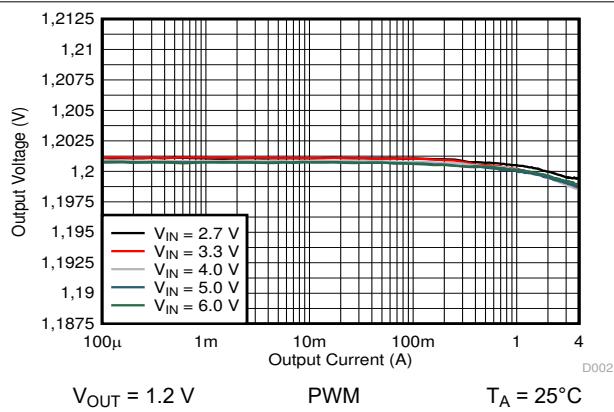
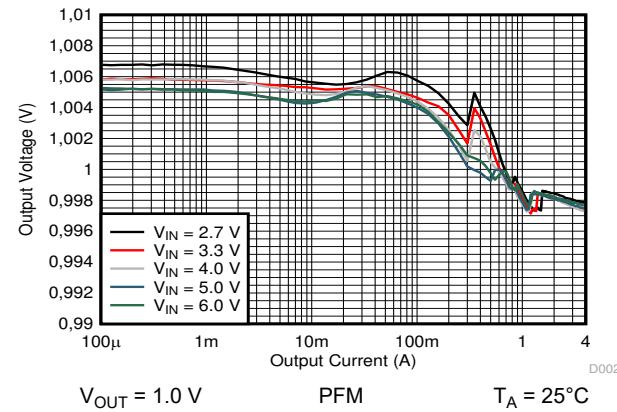
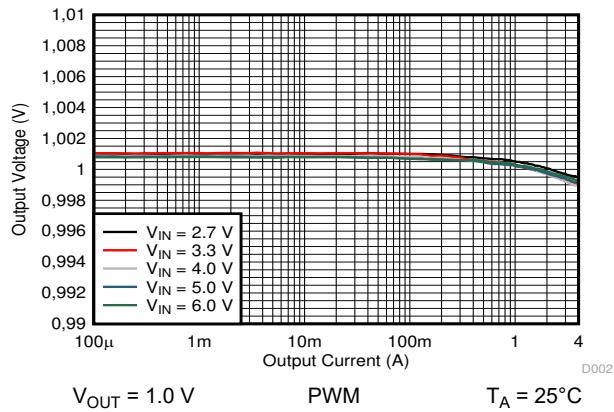
| NOMINAL OUTPUT VOLTAGE V_{OUT} | R_1 | R_2 | C_{FF} | EXACT OUTPUT VOLTAGE |
|-------------------------------------|---------|---------|----------|----------------------|
| 0.8 V | 16.9 kΩ | 51 kΩ | 10 pF | 0.7988 V |
| 1.0 V | 20 kΩ | 30 kΩ | 10 pF | 1.0 V |
| 1.1 V | 39.2 kΩ | 47 kΩ | 10 pF | 1.101 V |
| 1.2 V | 68 kΩ | 68 kΩ | 10 pF | 1.2 V |
| 1.5 V | 76.8 kΩ | 51 kΩ | 10 pF | 1.5 V |
| 1.8 V | 80.6 kΩ | 40.2 kΩ | 10 pF | 1.803 V |
| 2.5 V | 47.5 kΩ | 15 kΩ | 10 pF | 2.5 V |
| 3.3 V | 88.7 kΩ | 19.6 kΩ | 10 pF | 3.315 V |

9.2.3 Application Curves

All plots have been taken with a nominal switching frequency of 2.25 MHz when set to PWM mode, unless otherwise noted. The BOM is according to [Table 7-1](#).





**Figure 9-14. Output Voltage versus Output Current****Figure 9-15. Output Voltage versus Output Current****Figure 9-16. Output Voltage versus Output Current****Figure 9-17. Output Voltage versus Output Current****Figure 9-18. Output Voltage versus Output Current****Figure 9-19. Output Voltage versus Output Current**

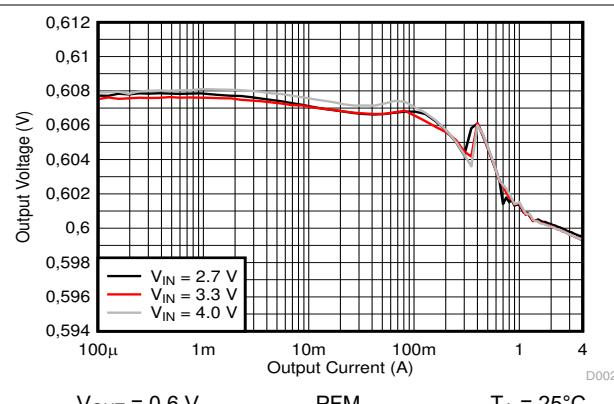


Figure 9-20. Output Voltage versus Output Current

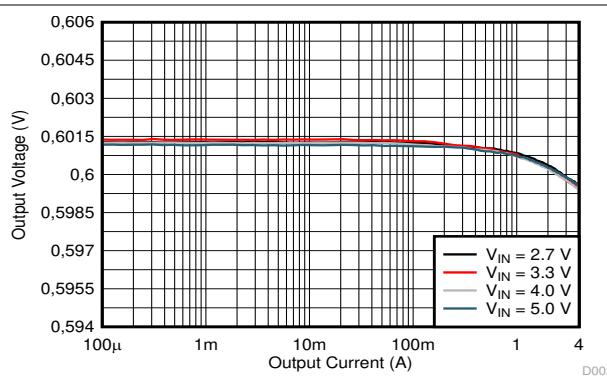


Figure 9-21. Output Voltage versus Output Current

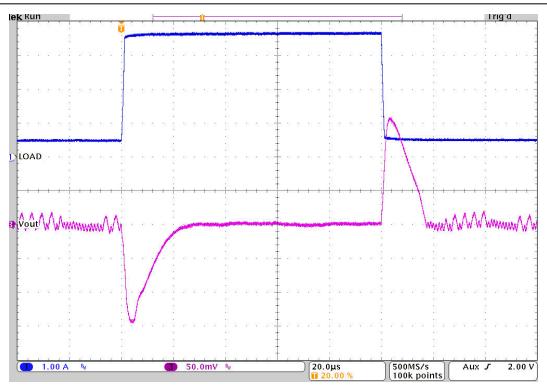


Figure 9-22. Load Transient Response

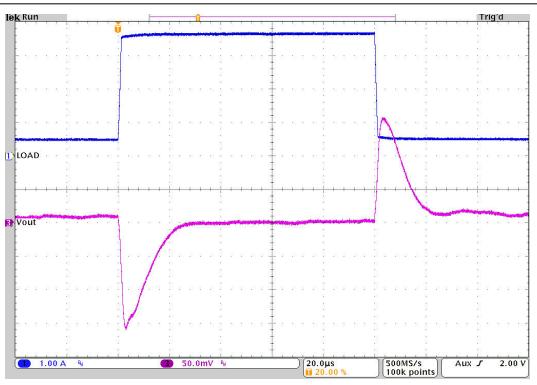


Figure 9-23. Load Transient Response

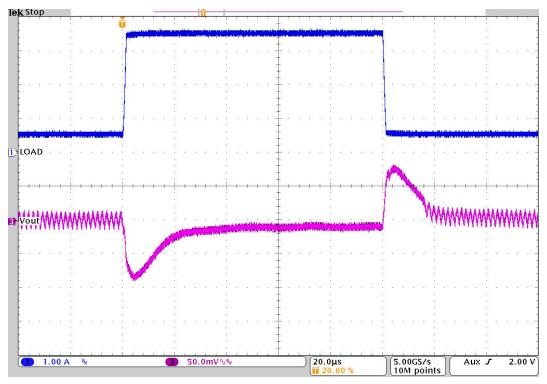


Figure 9-24. Load Transient Response

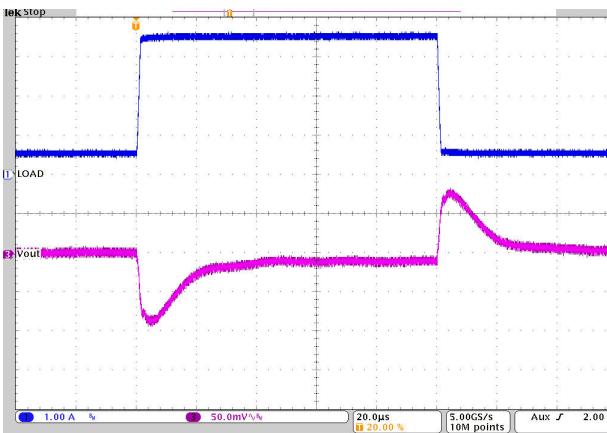
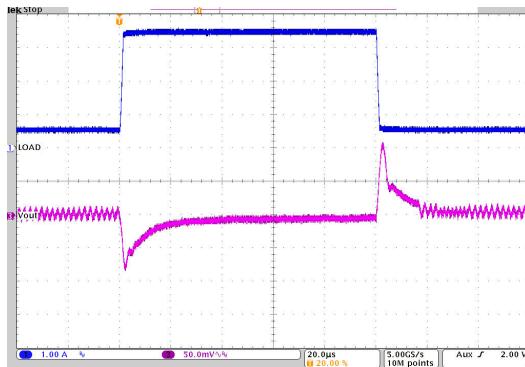
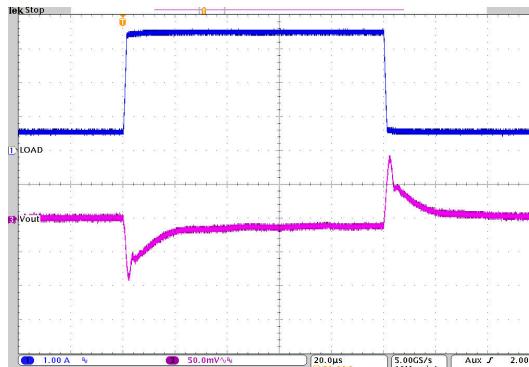


Figure 9-25. Load Transient Response



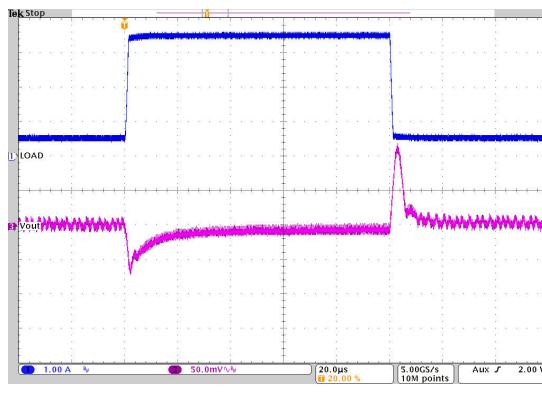
$V_{OUT} = 1.2 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $V_{IN} = 5.0 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-26. Load Transient Response



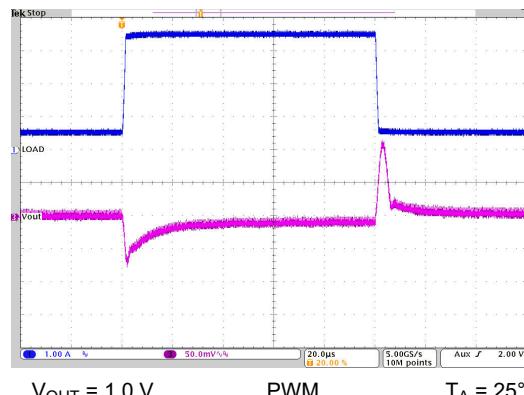
$V_{OUT} = 1.2 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $V_{IN} = 5.0 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-27. Load Transient Response



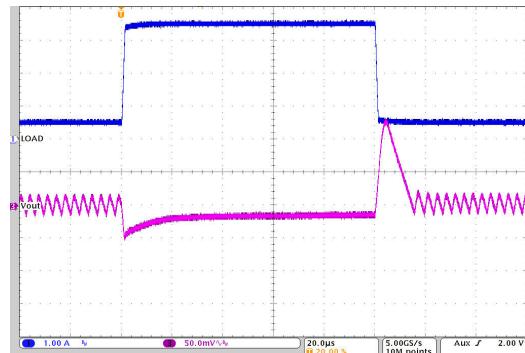
$V_{OUT} = 1.0 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $V_{IN} = 5.0 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-28. Load Transient Response



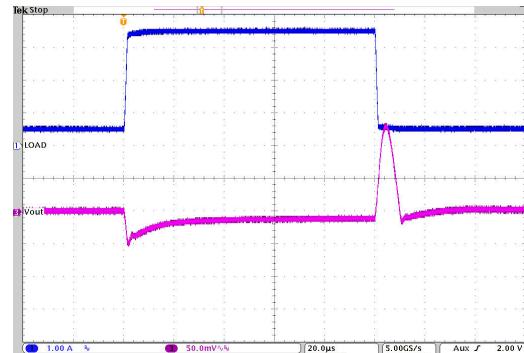
$V_{OUT} = 1.0 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $V_{IN} = 5.0 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-29. Load Transient Response



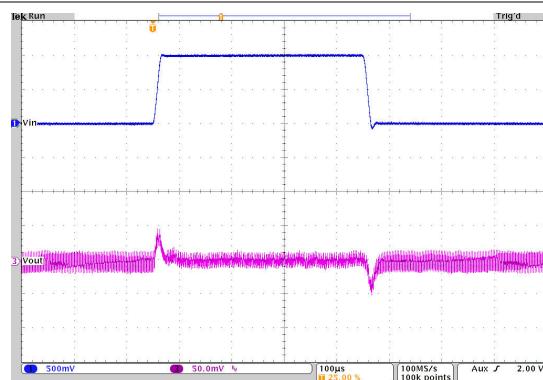
$V_{OUT} = 0.6 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $V_{IN} = 3.3 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-30. Load Transient Response



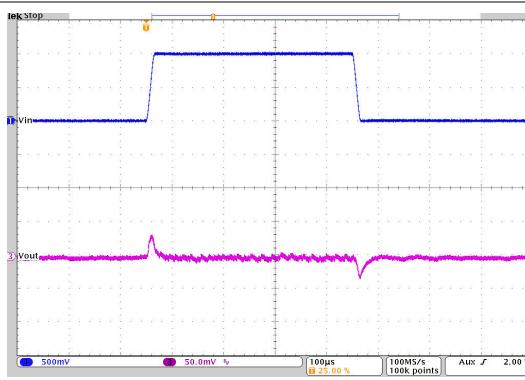
$V_{OUT} = 0.6 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $V_{IN} = 3.3 \text{ V}$ $I_{OUT} = 0.4 \text{ A to } 3.6 \text{ A to } 0.4 \text{ A}$

Figure 9-31. Load Transient Response



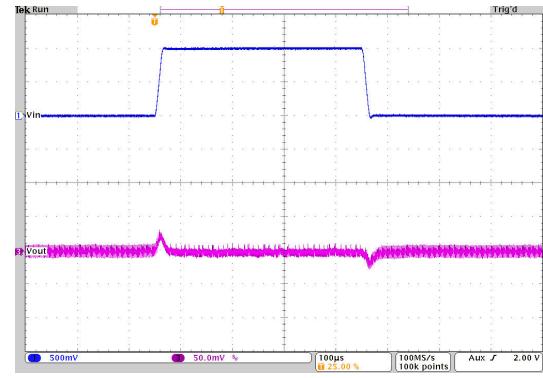
$V_{OUT} = 3.3 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-32. Line Transient Response



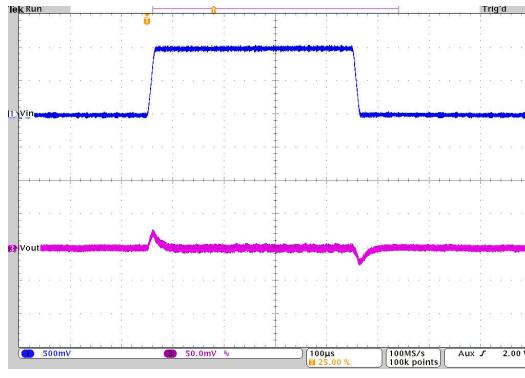
$V_{OUT} = 3.3 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-33. Line Transient Response



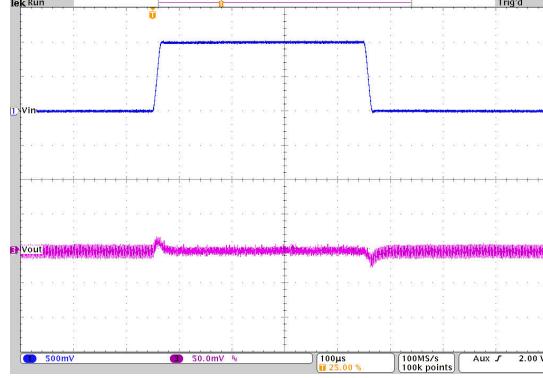
$V_{OUT} = 1.8 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-34. Line Transient Response



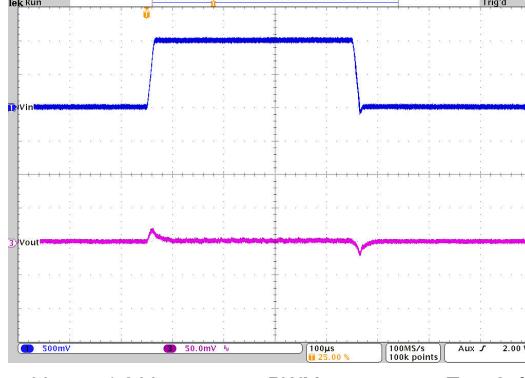
$V_{OUT} = 1.8 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-35. Line Transient Response



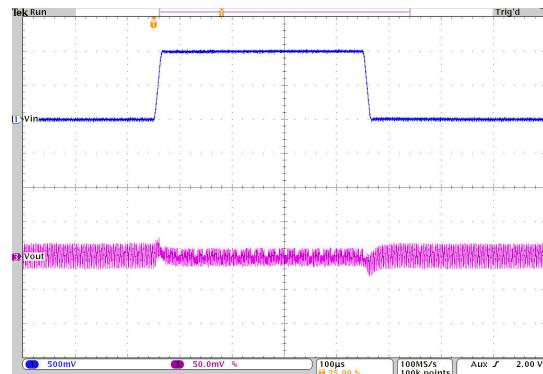
$V_{OUT} = 1.2 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-36. Line Transient Response



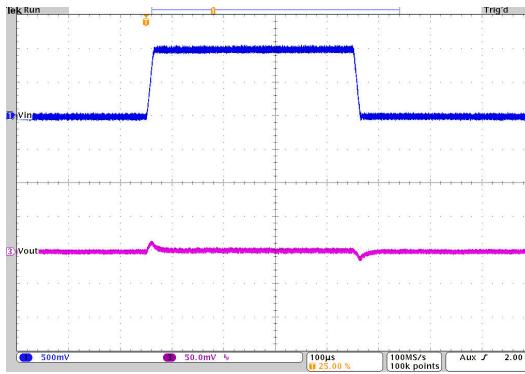
$V_{OUT} = 1.2 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-37. Line Transient Response



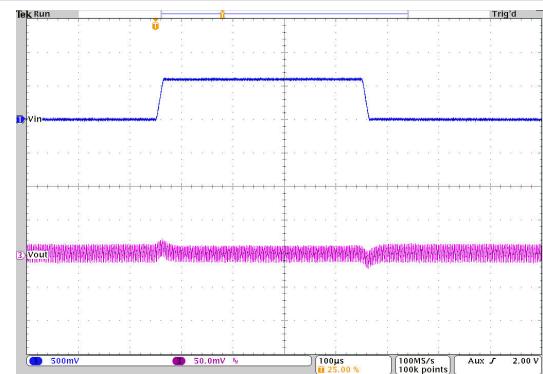
$V_{OUT} = 1.0 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-38. Line Transient Response



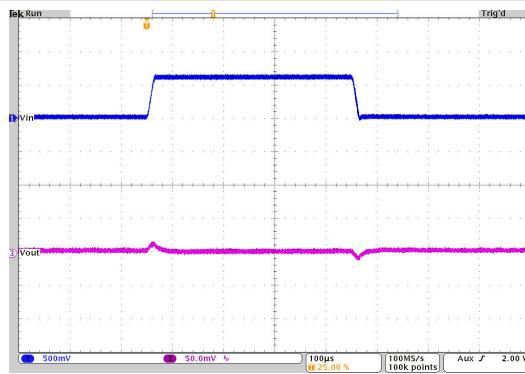
$V_{OUT} = 1.0 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 4.5 \text{ V to } 5.5 \text{ V to } 4.5 \text{ V}$

Figure 9-39. Line Transient Response



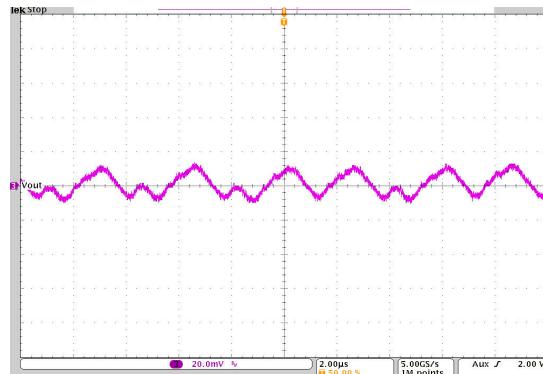
$V_{OUT} = 0.6 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 3.0 \text{ V to } 3.6 \text{ V to } 3.0 \text{ V}$

Figure 9-40. Line Transient Response



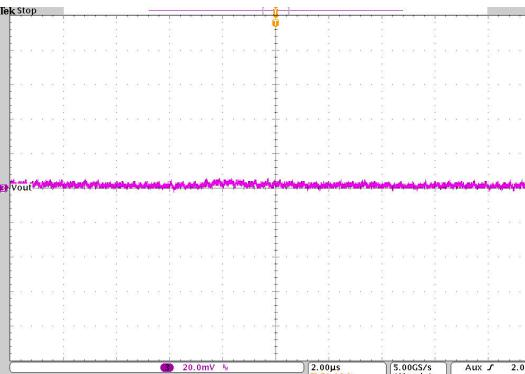
$V_{OUT} = 0.6 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 3.0 \text{ V to } 3.6 \text{ V to } 3.0 \text{ V}$

Figure 9-41. Line Transient Response



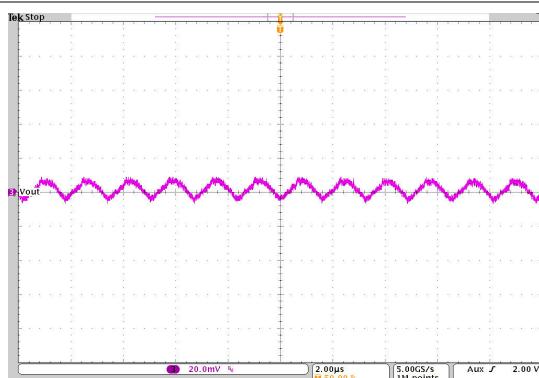
$V_{OUT} = 3.3 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-42. Output Voltage Ripple



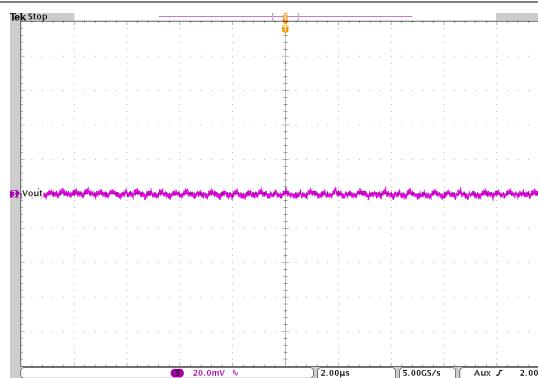
$V_{OUT} = 3.3 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-43. Output Voltage Ripple



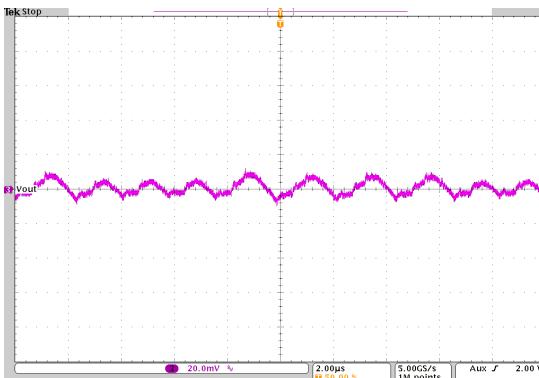
$V_{OUT} = 1.8 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-44. Output Voltage Ripple



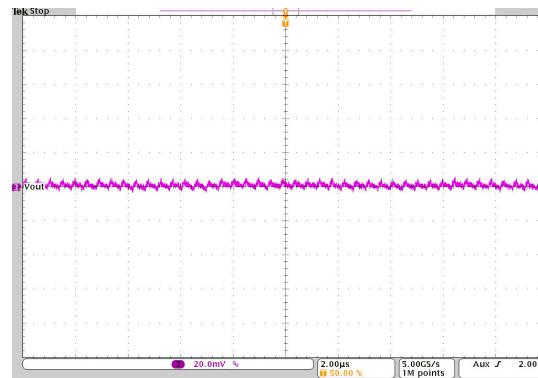
$V_{OUT} = 1.8 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-45. Output Voltage Ripple



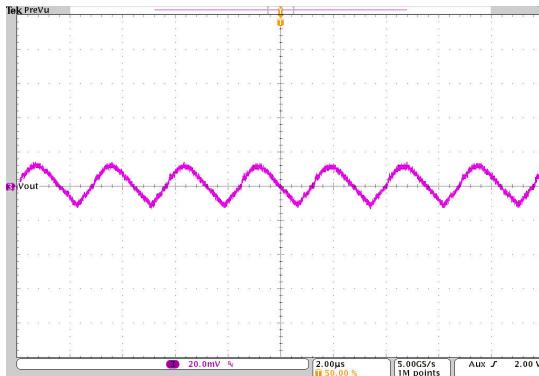
$V_{OUT} = 1.2 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-46. Output Voltage Ripple



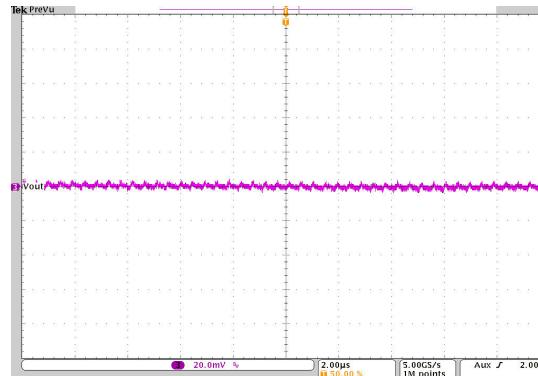
$V_{OUT} = 1.2 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-47. Output Voltage Ripple



$V_{OUT} = 1.0 \text{ V}$ PFM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 0.5 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-48. Output Voltage Ripple



$V_{OUT} = 1.0 \text{ V}$ PWM $T_A = 25^\circ\text{C}$
 $I_{OUT} = 4 \text{ A}$ $V_{IN} = 5.0 \text{ V}$ BW = 20 MHz

Figure 9-49. Output Voltage Ripple

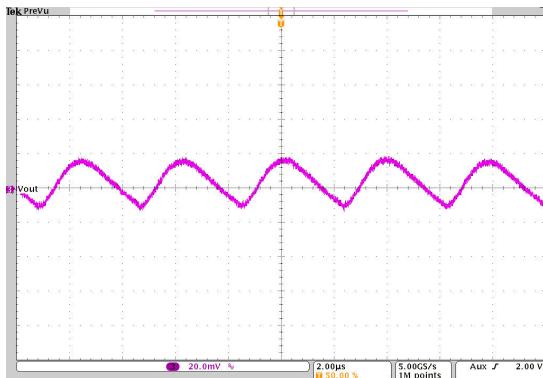


Figure 9-50. Output Voltage Ripple

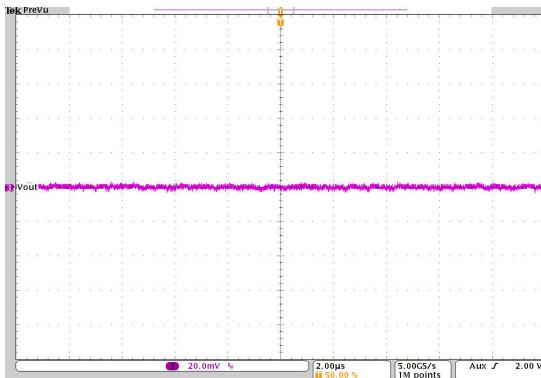


Figure 9-51. Output Voltage Ripple

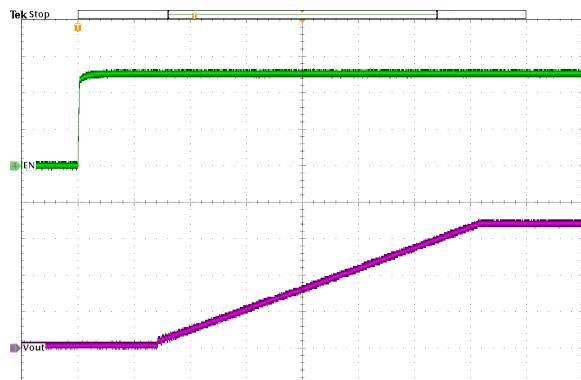


Figure 9-52. Start-Up Timing

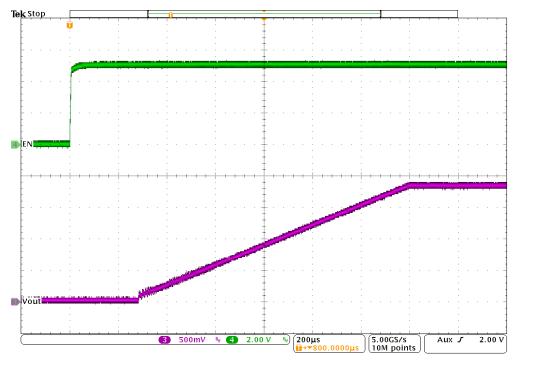
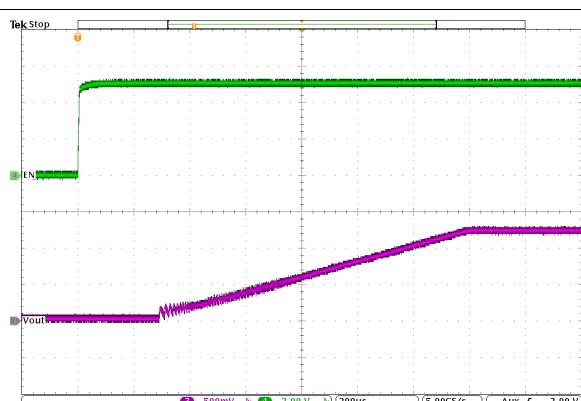
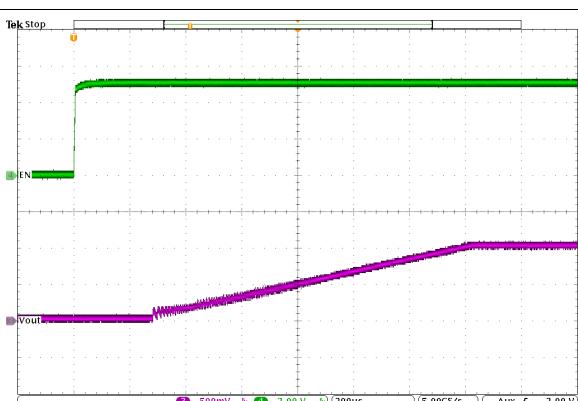


Figure 9-53. Start-Up Timing



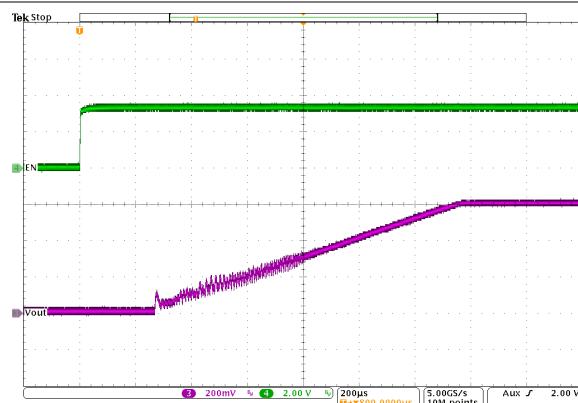
V_{OUT} = 1.2 V PWM T_A = 25°C
I_{OUT} = 4 A V_{IN} = 5 V C_{SS} = 4.7 nF

Figure 9-54. Start-Up Timing



V_{OUT} = 1.0 V PWM T_A = 25°C
I_{OUT} = 4 A V_{IN} = 5 V C_{SS} = 4.7 nF

Figure 9-55. Start-Up Timing



$V_{OUT} = 0.6 \text{ V}$
 $I_{OUT} = 4 \text{ A}$

PWM
 $V_{IN} = 3.3 \text{ V}$

$T_A = 25^\circ\text{C}$
 $C_{SS} = 4.7 \text{ nF}$

Figure 9-56. Start-up Timing

9.3 System Examples

9.3.1 Fixed Output Voltage Versions

Versions with an internally fixed output voltage allow the user remove the external feedback voltage divider. This ability not only allows the user reduce the total design size but also provides higher accuracy as there is no additional error caused by the external resistor divider. The FB pin must be tied to the output voltage directly as shown in [Figure 9-57](#). Independent of that, the application shown runs with an internally defined switching frequency of 2.25 MHz by connecting COMP/FSET to GND.

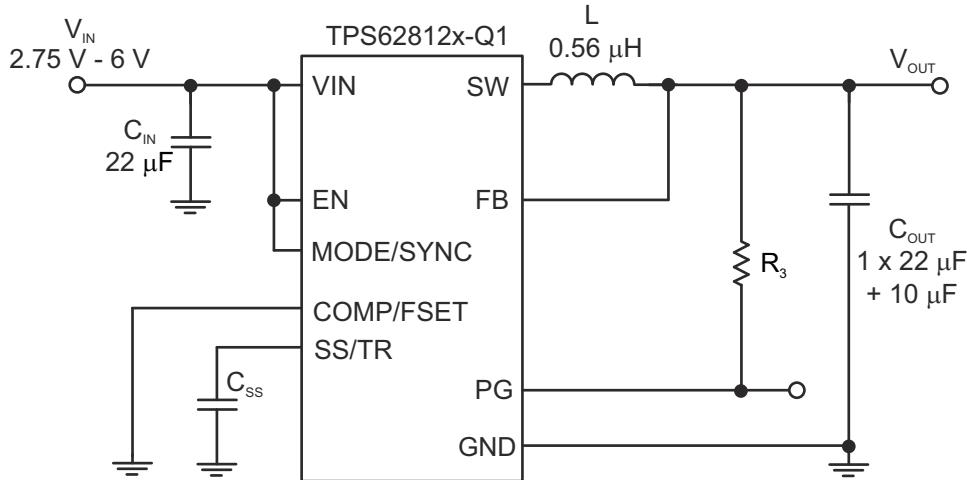


Figure 9-57. Schematic for Fixed Output Voltage Versions

9.3.2 Voltage Tracking

The TPS6281x-Q1 follows the voltage applied to the SS/TR pin. A voltage ramp on SS/TR to 0.6 V ramps the output voltage according to the 0.6 V feedback voltage.

Tracking the 3.3 V of device 1, such that both rails reach the target voltage at the same time, requires a resistor divider on SS/TR of device 2 equal to the output voltage divider of device 1. The output current of 2.5 μ A on the SS/TR pin causes an offset voltage on the resistor divider formed by R_5 and R_6 . The equivalent resistance of $R_5 // R_6$, so keep below 15 k Ω . The current from SS/TR causes a slightly higher voltage across R_6 than 0.6 V, which is desired because device 2 switches to the internal reference as soon as the voltage at SS/TR is higher than 0.6 V.

In case both devices must run in forced PWM mode, TI recommends to tie the MODE pin of device 2 to the output voltage or the power good signal of device 1, the controller device. The TPS6281x-Q1 has a duty cycle limitation defined by the minimum on-time. For tracking down to low output voltages, device 2 cannot follow after the minimum duty cycle is reached. Enabling PFM mode while tracking is in progress allows the user to ramp down the output voltage close to 0 V.

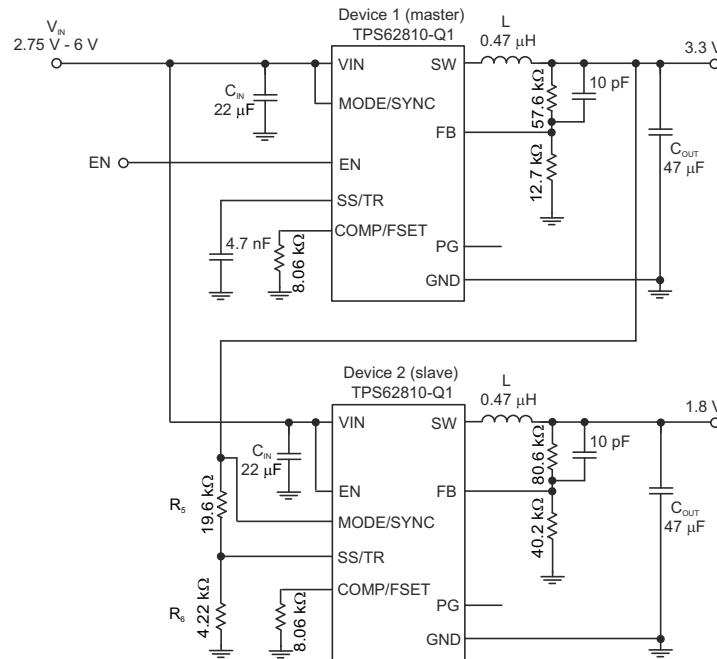


Figure 9-58. Schematic for Output Voltage Tracking

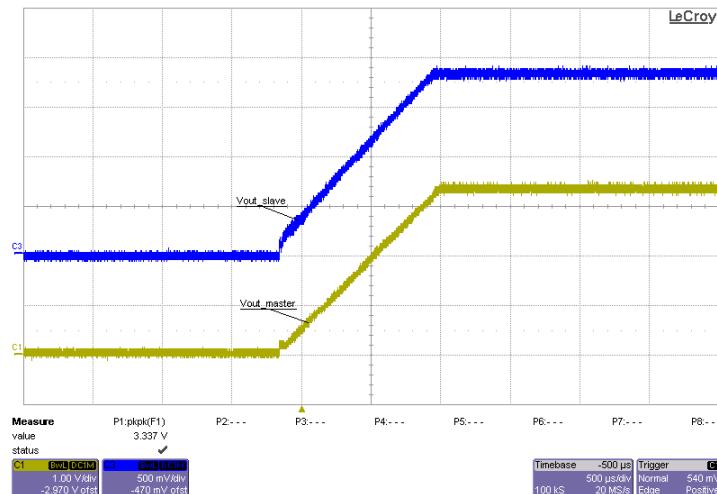


Figure 9-59. Scope Plot for Output Voltage Tracking

9.3.3 Synchronizing to an External Clock

The TPS6281x-Q1 can be externally synchronized by applying an external clock on the MODE/SYNC pin. There is no need for any additional circuitry as long as the input signal meets the requirements given in the electrical specifications. The clock can be applied, removed during operation, allowing the user to switch from an externally-defined fixed frequency to power-save mode or to internal fixed frequency operation. The value of the R_{CF} resistor must be chosen so that the internally defined frequency and the externally applied frequency are close to each other. This action makes sure of a smooth transition from internal to external frequency and vice versa.

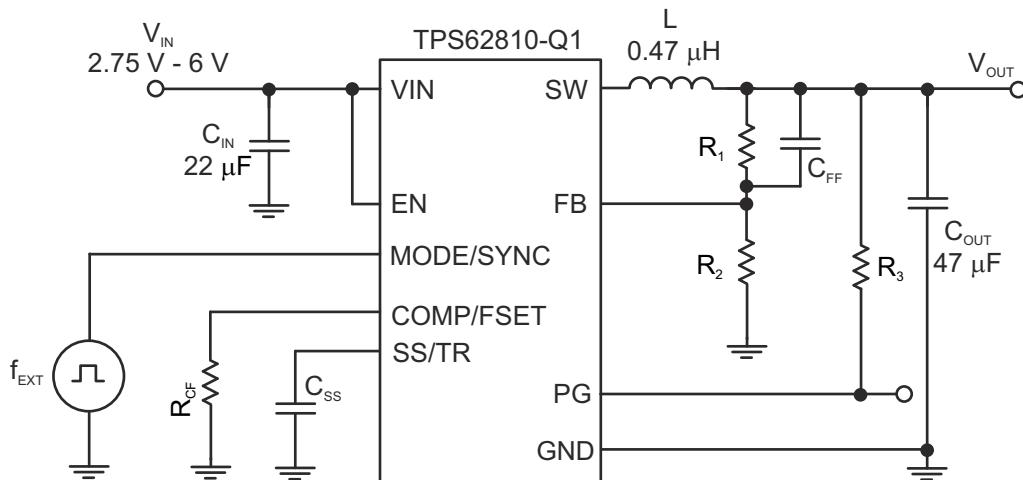
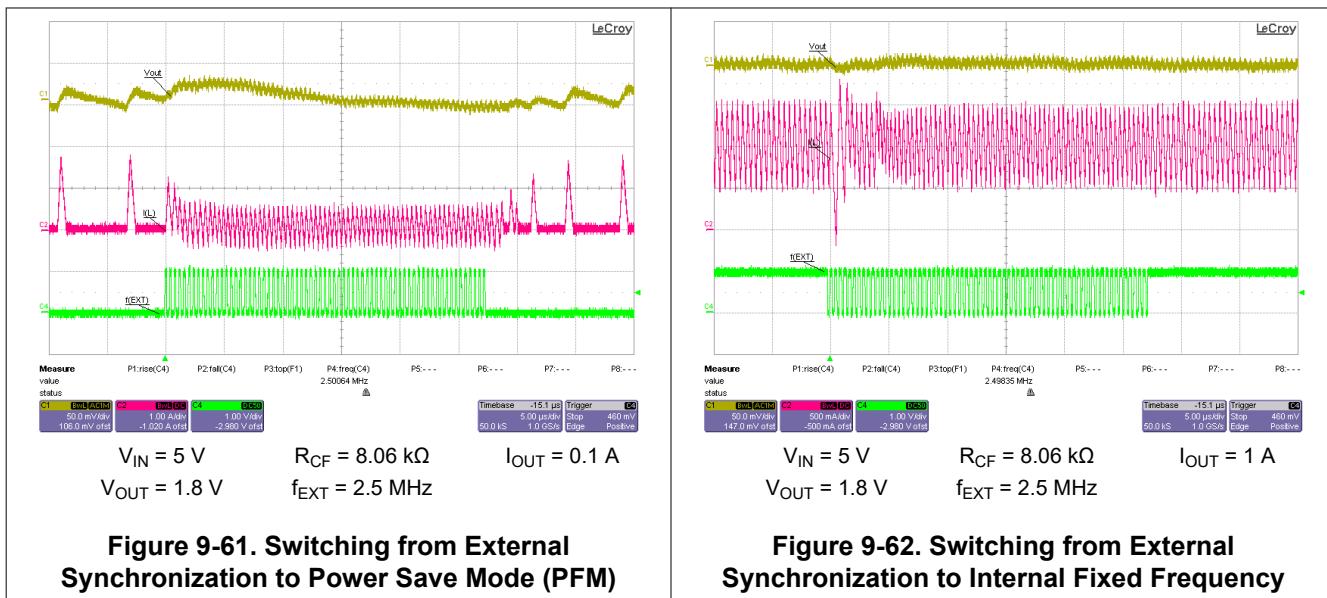


Figure 9-60. Schematic Using External Synchronization



9.4 Power Supply Recommendations

The TPS6281x-Q1 device family has no special requirements for the input power supply. The output current of the input power supply must be rated according to the supply voltage, output voltage, and output current of the TPS6281x-Q1.

9.5 Layout

9.5.1 Layout Guidelines

A proper layout is critical for the operation of a switched mode power supply, even more at high switching frequencies. Therefore, the PCB layout of the TPS6281x-Q1 demands careful attention to make sure of operation and to get the performance specified. A poor layout can lead to issues like poor regulation (both line and load), stability and accuracy weaknesses increased EMI radiation and noise sensitivity.

See [Layout Example](#) for the recommended layout of the TPS6281x-Q1, which is designed for common external ground connections. The input capacitor must be placed as close as possible between the VIN and GND pin.

Provide low inductive and resistive paths for loops with high di/dt. Therefore, paths conducting the switched load current must be as short and wide as possible. Provide low capacitive paths (with respect to all other nodes) for

wires with high dv/dt. Therefore, the input and output capacitance must be placed as close as possible to the IC pins and parallel wiring over long distances as well as narrow traces must be avoided. Loops that conduct an alternating current must outline an area as small as possible, as this area is proportional to the energy radiated.

Sensitive nodes like FB must be connected with short wires and not nearby high dv/dt signals (for example, SW). Because they carry information about the output voltage, they must be connected as close as possible to the actual output voltage (at the output capacitor). The capacitor on the SS/TR pin as well as the FB resistors, R₁ and R₂, must be kept close to the IC and connect directly to those pins and the system ground plane.

The package uses the pins for power dissipation. Thermal vias on the VIN and GND pins help spread the heat into the pcb.

The recommended layout is implemented on the EVM and shown in the [TPS62810EVM-015 Evaluation Module user's guide](#).

9.5.2 Layout Example

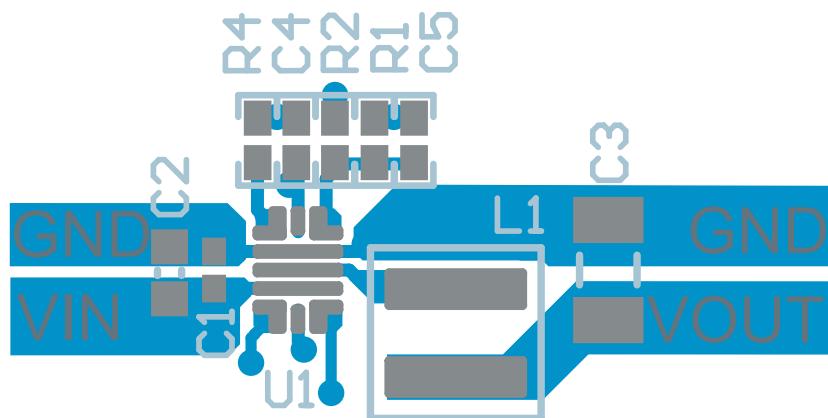


Figure 9-63. Example Layout

10 Device and Documentation Support

10.1 Device Support

10.1.1 Third-Party Products Disclaimer

TI'S PUBLICATION OF INFORMATION REGARDING THIRD-PARTY PRODUCTS OR SERVICES DOES NOT CONSTITUTE AN ENDORSEMENT REGARDING THE SUITABILITY OF SUCH PRODUCTS OR SERVICES OR A WARRANTY, REPRESENTATION OR ENDORSEMENT OF SUCH PRODUCTS OR SERVICES, EITHER ALONE OR IN COMBINATION WITH ANY TI PRODUCT OR SERVICE.

10.2 Documentation Support

10.2.1 Related Documentation

For related documentation see the following:

Texas Instruments, [TPS62810EVM-015 Evaluation Module user's guide](#)

10.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

10.4 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

10.5 Trademarks

TI E2E™ is a trademark of Texas Instruments.

All trademarks are the property of their respective owners.

10.6 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.



ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

10.7 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

11 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

| Changes from Revision J (March 2023) to Revision K (July 2025) | Page |
|---|------|
| • Updated the document title..... | 1 |
| • Added device spins to <i>Device Comparison Table</i> | 3 |
| • Added "output discharge disconnected" options to the <i>Output Discharge</i> section..... | 15 |

| Changes from Revision I (December 2021) to Revision J (March 2023) | Page |
|--|------|
| • Added planned device spins to <i>Device Comparison Table</i> | 3 |

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

| Orderable part number | Status (1) | Material type (2) | Package Pins | Package qty Carrier | RoHS (3) | Lead finish/ Ball material (4) | MSL rating/ Peak reflow (5) | Op temp (°C) | Part marking (6) |
|-----------------------|---------------|----------------------|-------------------|-----------------------|-------------|--------------------------------------|-----------------------------------|--------------|---------------------|
| TPS6281006QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81006Q |
| TPS6281006QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81006Q |
| TPS6281008QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81008Q |
| TPS6281008QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81008Q |
| TPS628100MQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8100MQ |
| TPS628100MQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8100MQ |
| TPS628101HQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8101HQ |
| TPS6281020QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81020Q |
| TPS6281020QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81020Q |
| TPS62810QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 810Q |
| TPS62810QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 810Q |
| TPS6281109QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81109Q |
| TPS6281109QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81109Q |
| TPS628110AQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8110AQ |
| TPS628110AQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8110AQ |
| TPS6281120QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81120Q |
| TPS6281120QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81120Q |
| TPS6281120QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81120Q |
| TPS6281126QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81126Q |
| TPS6281126QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81126Q |
| TPS628112AQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8112AQ |
| TPS628112AQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8112AQ |
| TPS628112MQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8112MQ |
| TPS628112MQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8112MQ |
| TPS628113HQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8113HQ |
| TPS628113HQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8113HQ |
| TPS628113HQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8113HQ |
| TPS628111QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 811Q |
| TPS628111QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 811Q |
| TPS6281206QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81206Q |
| TPS6281206QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81206Q |

| Orderable part number | Status (1) | Material type (2) | Package Pins | Package qty Carrier | RoHS (3) | Lead finish/ Ball material (4) | MSL rating/ Peak reflow (5) | Op temp (°C) | Part marking (6) |
|------------------------------------|---------------|----------------------|-------------------|-----------------------|-------------|--------------------------------------|-----------------------------------|--------------|---------------------|
| TPS6281208QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81208Q |
| TPS6281208QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81208Q |
| TPS628120MQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8120MQ |
| TPS628120MQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8120MQ |
| TPS6281220QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81220Q |
| TPS6281220QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81220Q |
| TPS6281228QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81228Q |
| TPS6281228QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81228Q |
| TPS628122AQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8122AQ |
| TPS628122AQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8122AQ |
| TPS628122GQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8122GQ |
| TPS628122GQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8122GQ |
| TPS6281240QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81240Q |
| TPS6281240QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81240Q |
| TPS62812C0QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 812C0Q |
| TPS62812C0QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 812Q |
| TPS62812QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 812Q |
| TPS6281302QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81302Q |
| TPS6281302QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81302Q |
| TPS628130AQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8130AQ |
| TPS628130AQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8130AQ |
| TPS6281320QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81320Q |
| TPS6281320QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 81320Q |
| TPS6281322QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81322Q |
| TPS6281326QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 120 | 81326Q |
| TPS6281326QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 120 | 81326Q |
| TPS628132DQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8132DQ |
| TPS628132DQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8132DQ |
| TPS628132MQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8132MQ |
| TPS628132MQWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 8132MQ |
| TPS628133HQWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 8133HQ |

| Orderable part number | Status (1) | Material type (2) | Package Pins | Package qty Carrier | RoHS (3) | Lead finish/ Ball material (4) | MSL rating/ Peak reflow (5) | Op temp (°C) | Part marking (6) |
|-----------------------|---------------|----------------------|-------------------|-----------------------|-------------|--------------------------------------|-----------------------------------|--------------|---------------------|
| TPS6281389QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 81389Q |
| TPS62813QWRWYRQ1 | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU SN | Level-2-260C-1 YEAR | -40 to 125 | 813Q |
| TPS62813QWRWYRQ1.A | Active | Production | VQFN-HR (RWY) 9 | 3000 LARGE T&R | Yes | NIPDAU | Level-2-260C-1 YEAR | -40 to 125 | 813Q |
| XPS62810QWRWYRQ1 | Obsolete | Preproduction | VQFN-HR (RWY) 9 | - | - | Call TI | Call TI | -40 to 125 | |

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

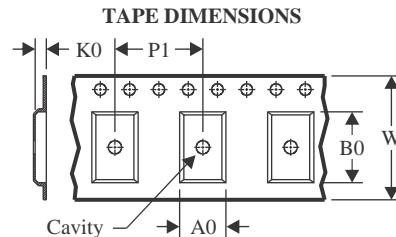
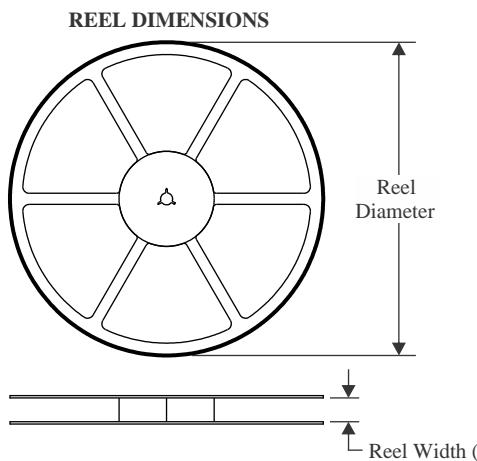
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF TPS62810-Q1, TPS62811-Q1, TPS62812-Q1, TPS62813-Q1 :

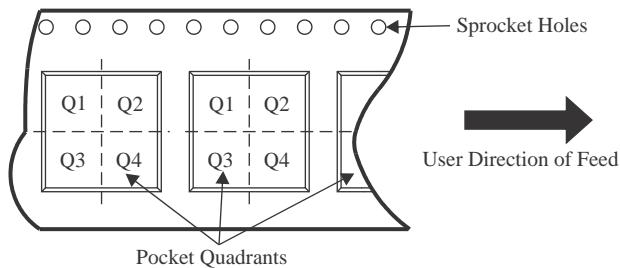
- Enhanced Product : [TPS62810-EP](#), [TPS62811-EP](#), [TPS62812-EP](#), [TPS62813-EP](#)

NOTE: Qualified Version Definitions:

- Enhanced Product - Supports Defense, Aerospace and Medical Applications

TAPE AND REEL INFORMATION

| | |
|----|---|
| A0 | Dimension designed to accommodate the component width |
| B0 | Dimension designed to accommodate the component length |
| K0 | Dimension designed to accommodate the component thickness |
| W | Overall width of the carrier tape |
| P1 | Pitch between successive cavity centers |

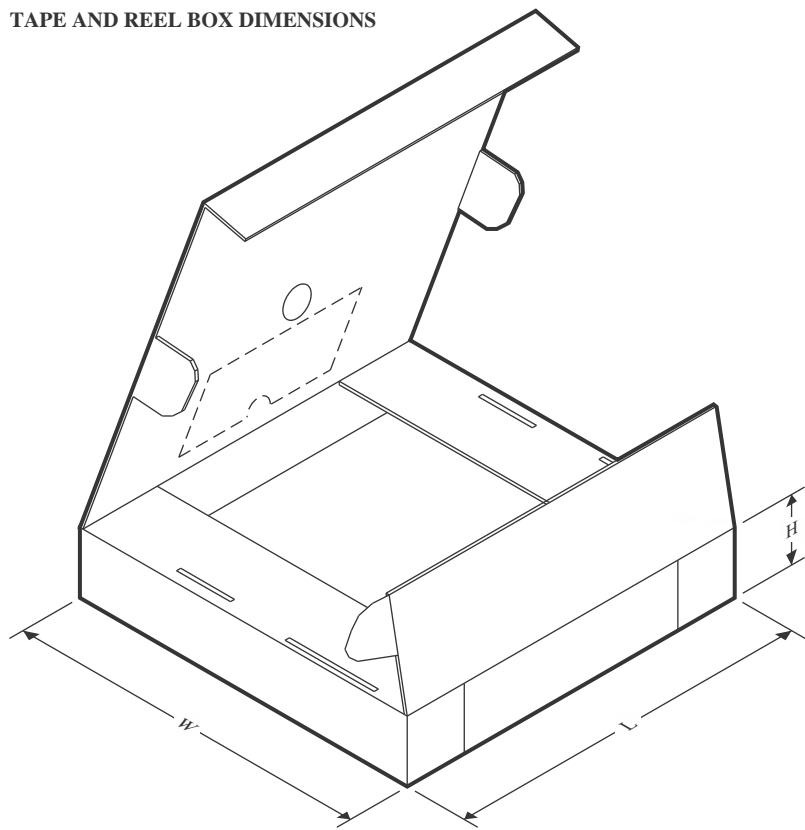
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|--------------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TPS6281006QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281008QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628100MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628101HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS628101HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281020QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS62810QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281109QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628110AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|--------------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TPS6281120QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281126QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628112AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628112MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628113HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628111QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281206QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281208QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628120MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281220QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281228QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628122AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628122GQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281240QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS62812C0QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS62812C0QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS62812QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281302QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS6281302QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628130AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281320QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281322QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281322QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS6281326QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|--------------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TPS628132DQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628132MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS628133HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS628133HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281389QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |
| TPS6281389QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.3 | 3.2 | 1.0 | 4.0 | 8.0 | Q1 |
| TPS62813QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 180.0 | 8.4 | 2.25 | 3.25 | 1.05 | 4.0 | 8.0 | Q1 |

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|--------------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TPS6281006QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281008QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628100MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628101HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS628101HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281020QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS62810QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281109QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628110AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281120QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281126QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628112AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628112MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628113HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628111QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281206QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281208QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628120MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|--------------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TPS6281220QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281228QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628122AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628122GQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281240QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS62812C0QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS62812C0QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS62812QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281302QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS6281302QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628130AQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281320QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281322QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281322QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS6281326QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628132DQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628132MQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS628133HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS628133HQWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281389QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |
| TPS6281389QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 213.0 | 191.0 | 35.0 |
| TPS62813QWRWYRQ1 | VQFN-HR | RWY | 9 | 3000 | 210.0 | 185.0 | 35.0 |

GENERIC PACKAGE VIEW

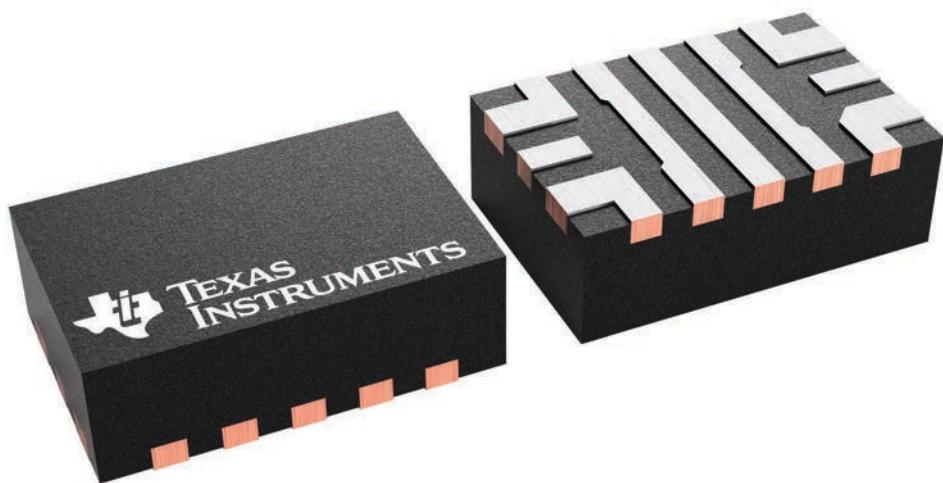
RWY 9

VQFN-HR - 1 mm max height

2 x 3, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4226729/A

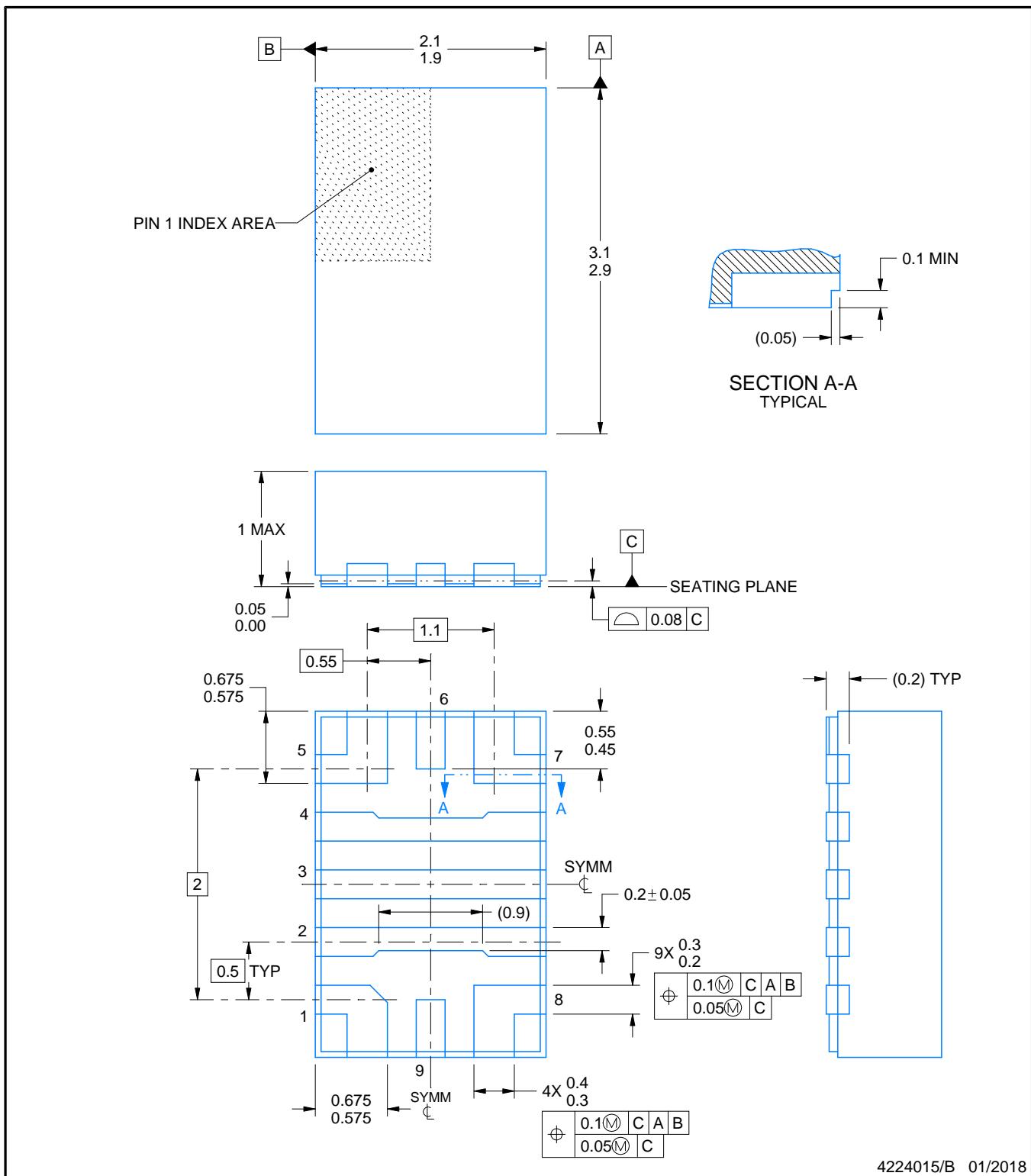
PACKAGE OUTLINE

RWY0009A



VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



4224015/B 01/2018

NOTES:

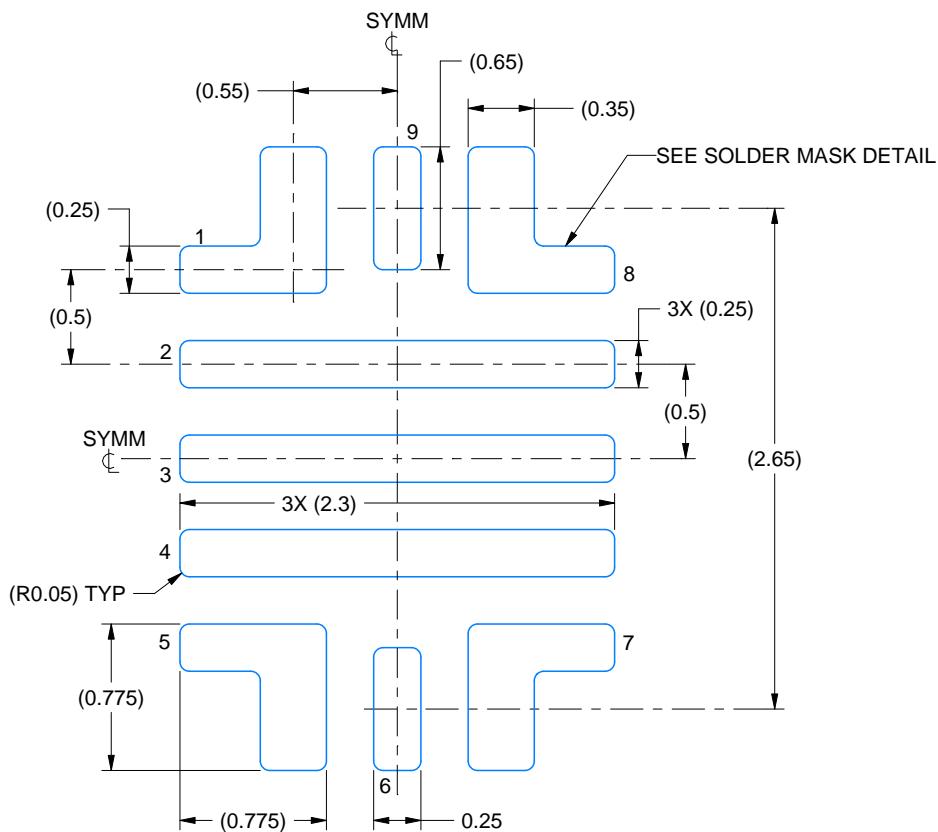
- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.

EXAMPLE BOARD LAYOUT

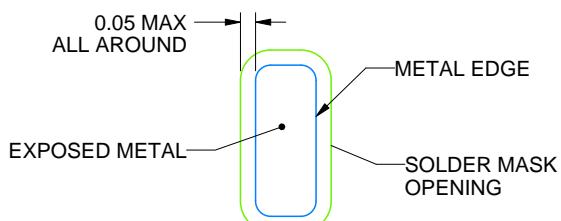
RWY0009A

VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 25X



SOLDER MASK DETAIL

4224015/B 01/2018

NOTES: (continued)

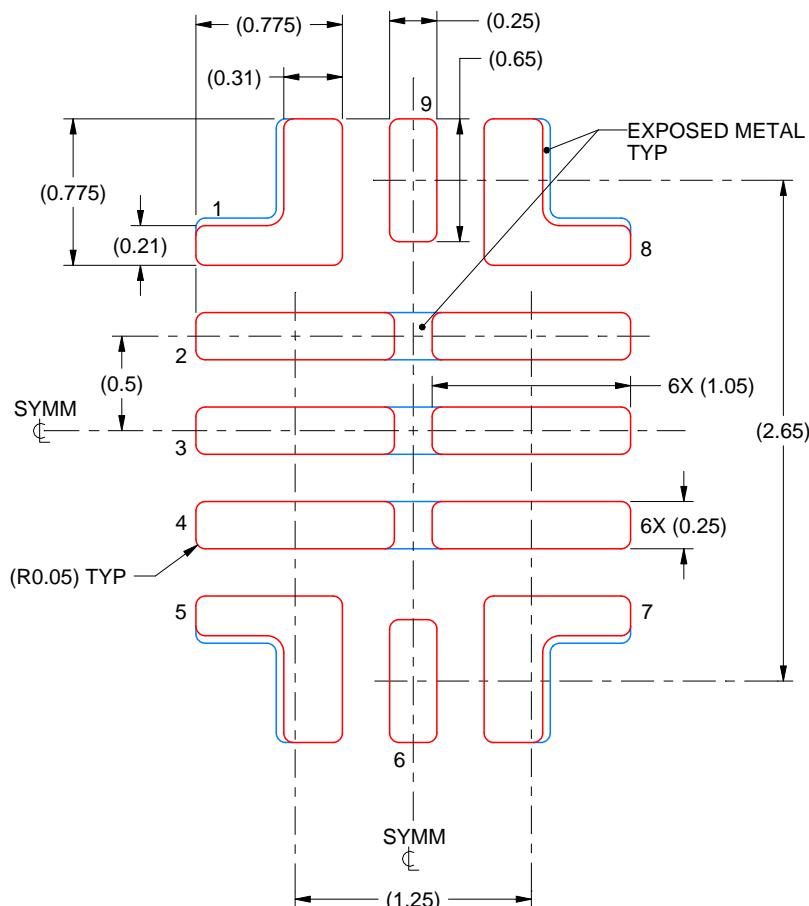
3. This package is designed to be soldered to thermal pads on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
4. Vias are optional depending on application, refer to device data sheet. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RWY0009A

VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE

BASED ON 0.1 mm THICK STENCIL

PADS 1, 5, 7 & 8:

90% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

SCALE: 25X

4224015/B 01/2018

NOTES: (continued)

5. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



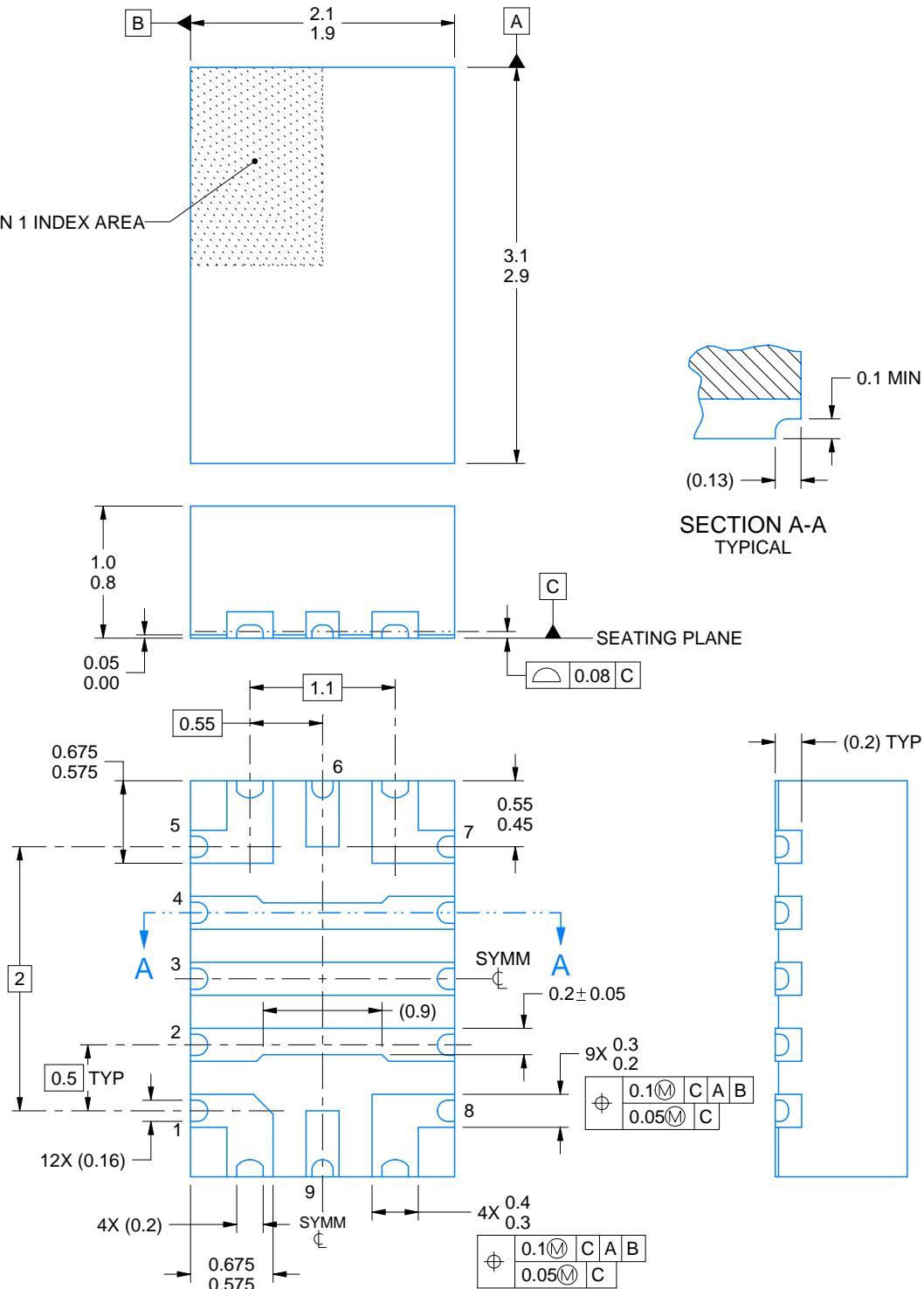
PACKAGE OUTLINE

RWY0009C



VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



4226803/A 05/2021

NOTES:

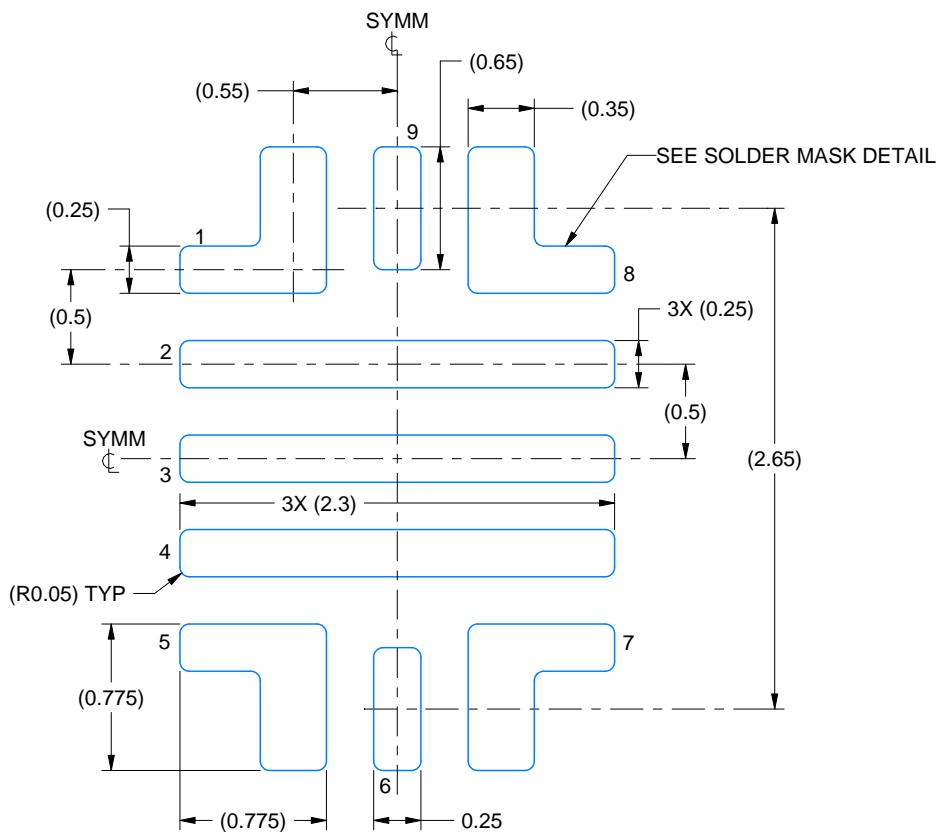
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.

EXAMPLE BOARD LAYOUT

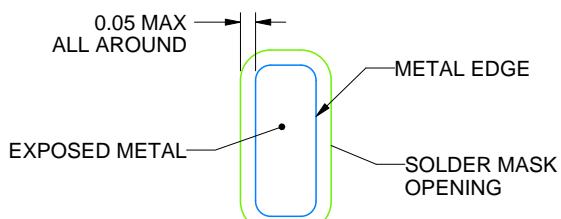
RWY0009C

VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 25X



NON SOLDER MASK
DEFINED

SOLDER MASK DETAIL

4226803/A 05/2021

NOTES: (continued)

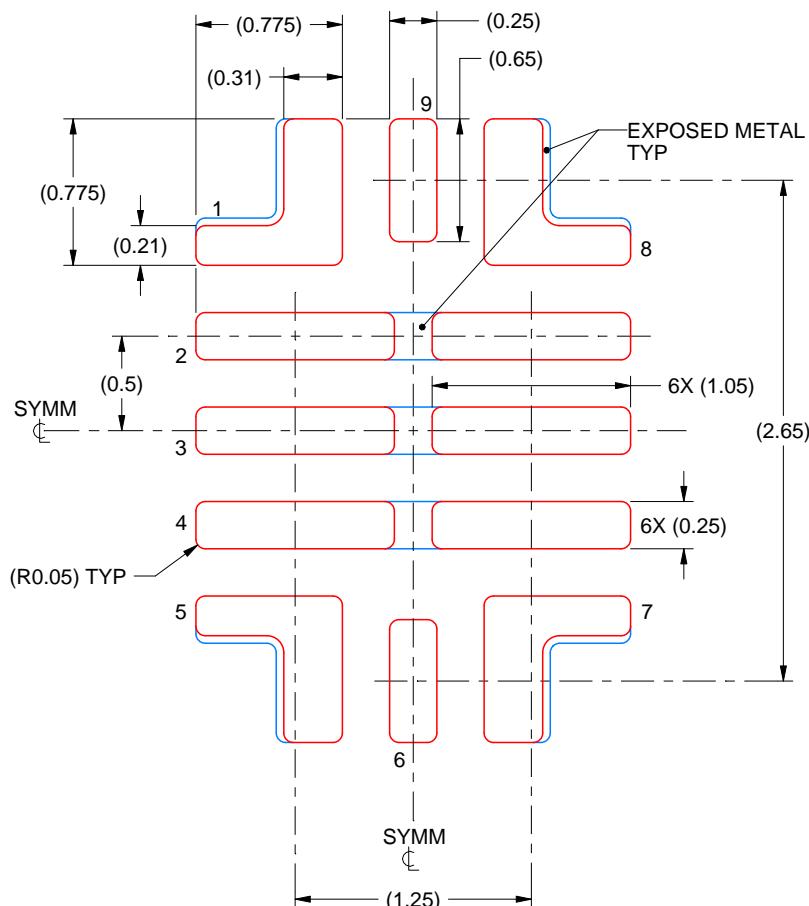
3. This package is designed to be soldered to thermal pads on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
4. Vias are optional depending on application, refer to device data sheet. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RWY0009C

VQFN-HR - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE

BASED ON 0.1 mm THICK STENCIL

PADS 1, 5, 7 & 8:

90% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

SCALE: 25X

4226803/A 05/2021

NOTES: (continued)

5. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATA SHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#) or other applicable terms available either on [ti.com](#) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

TI objects to and rejects any additional or different terms you may have proposed.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265
Copyright © 2025, Texas Instruments Incorporated