

Lecture #1 - A Review of SinusoidsSome Trigonometry

$$\sin(A+B) = \sin A \cos B + \cos A \sin B$$

$$\sin(A-B) = \sin A \cos B - \cos A \sin B$$

adding  $\sin A \cos B = \frac{1}{2} \sin(A+B) + \frac{1}{2} \sin(A-B)$

subtracting  $\cos A \sin B = \frac{1}{2} \sin(A+B) - \frac{1}{2} \sin(A-B)$

$$\cos(A+B) = \cos A \cos B - \sin A \sin B$$

$$\cos(A-B) = \cos A \cos B + \sin A \sin B$$

adding  $\cos A \cos B = \frac{1}{2} \cos(A+B) + \frac{1}{2} \cos(A-B)$

subtracting  $\sin A \sin B = \frac{1}{2} \cos(A-B) - \frac{1}{2} \cos(A+B)$

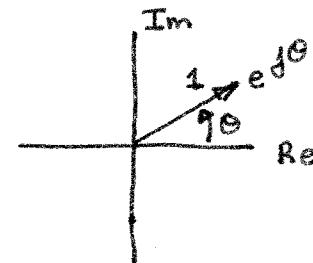
$$\cos^2 A = \frac{1 + \cos 2A}{2}$$

$$\sin^2 A = \frac{1 - \cos 2A}{2}$$

Complex Exponentials

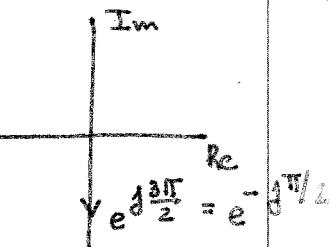
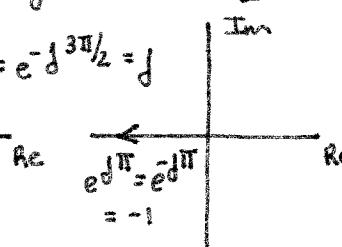
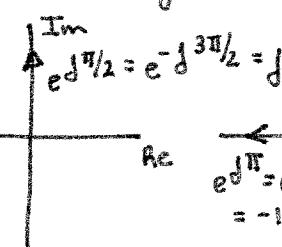
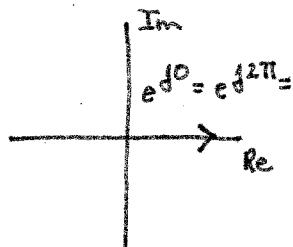
Euler Equation:  $e^{j\theta} = \cos \theta + j \sin \theta$

$$e^{-j\theta} = \cos \theta - j \sin \theta$$



adding  $\cos \theta = \frac{1}{2} [e^{j\theta} + e^{-j\theta}]$

subtracting  $\sin \theta = \frac{1}{2j} [e^{j\theta} - e^{-j\theta}]$

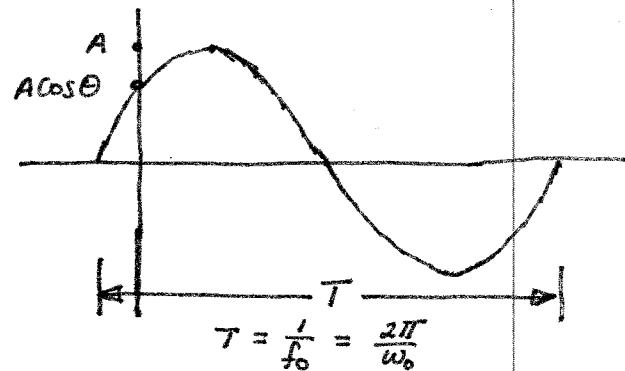


## Sinusoidal Voltages

A sinusoidal voltage is written as

$$v(t) = A \cos(\omega_0 t - \Theta)$$

$$\begin{aligned} v(t) &= \frac{A}{2} [e^{j(\omega_0 t - \Theta)} + e^{-j(\omega_0 t - \Theta)}] \\ &= \frac{A}{2} e^{-j\Theta} e^{j\omega_0 t} + \frac{A}{2} e^{j\Theta} e^{-j\omega_0 t} \end{aligned}$$

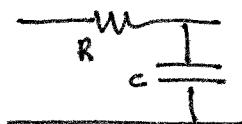


We like the exponential representation of a sinusoid.

## Linear Systems and Exponential Forcing Functions

$$\begin{aligned} x(t) &= c e^{j\omega_0 t} = |c| e^{j\phi} e^{j\omega_0 t} \xrightarrow{\text{H}(jw)} H(jw) = |H(jw)| e^{j\Theta(w)} \xrightarrow{\text{y}(t)} y(t) = |c| H(jw_0) e^{j\omega_0 t} \\ &= |c| |H(jw_0)| e^{j(\phi + \Theta(w_0))} e^{j\omega_0 t} \end{aligned}$$

### example #1 A lowpass RC circuit

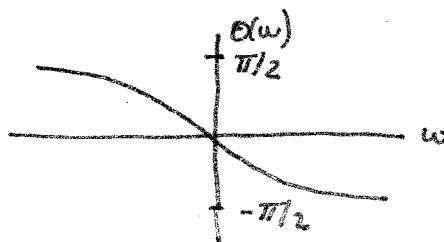
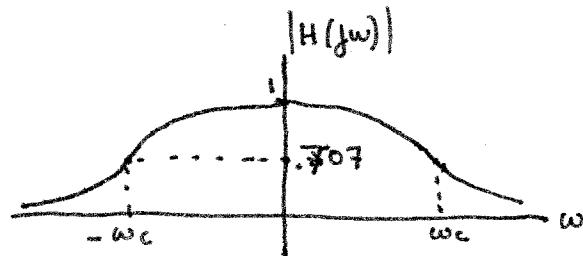


$$f_h(t) = \frac{1}{RC} e^{-t/RC} u(t)$$

$$\omega_c = 1/RC \quad H(s) = \frac{1}{s/\omega_c + 1}$$

$$H(jw) = \frac{1}{1 + j\omega/\omega_c} = \frac{1}{\sqrt{1 + (\omega/\omega_c)^2}} e^{-j \tan^{-1}(\omega/\omega_c)}$$

$$= |H(jw)| e^{j\Theta(w)}$$

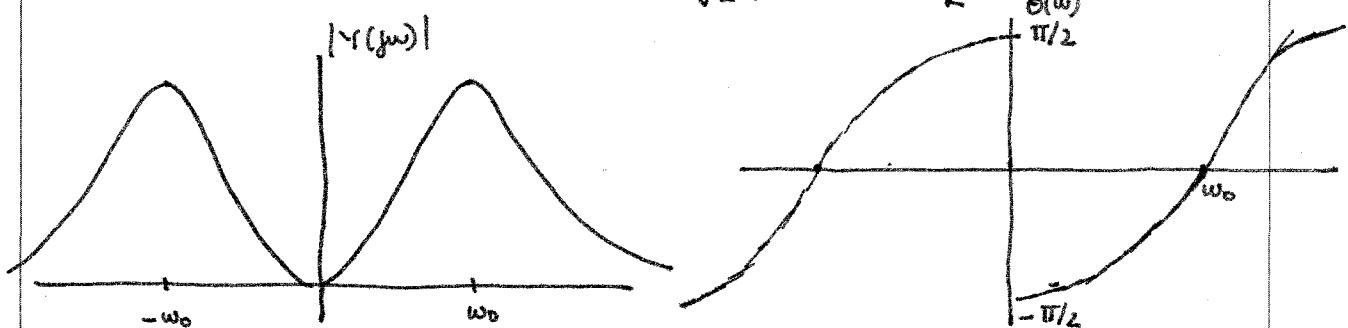


example #2 A simple bandpass circuit

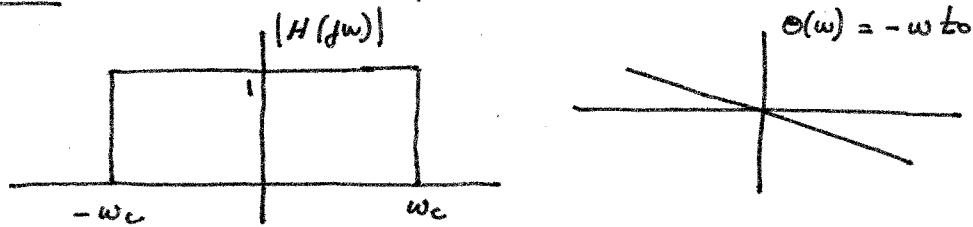


$$V(j\omega) = \frac{1}{R} + jQ_0 \left( \frac{\omega}{\omega_0} - \frac{\omega_0}{\omega} \right) = \frac{I(j\omega)}{Z(j\omega)}$$

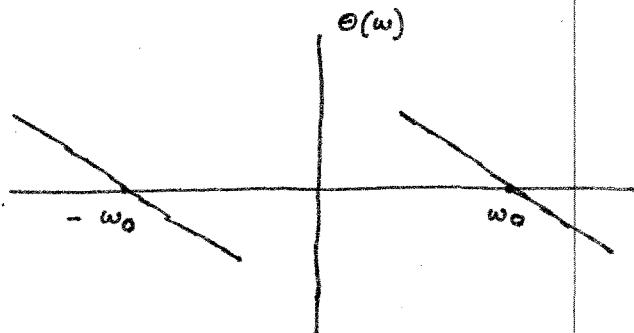
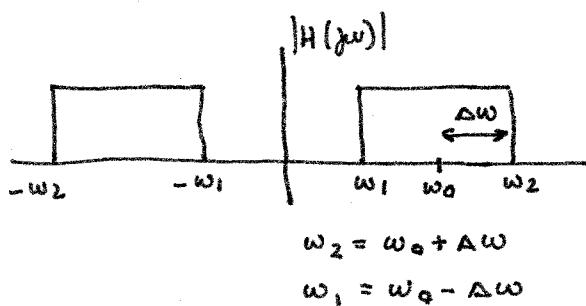
$$\omega_0 = \frac{1}{\sqrt{LC}} \quad Q_0 = \frac{\omega_0 L}{R}$$



example #3 An ideal lowpass filter



example #4 An ideal bandpass filter



Note  $|H(j\omega)|$  is even (symmetric) around  $\omega=0$

$\theta(\omega)$  is odd (antisymmetric) around  $\omega=0$

## The Average Power of a Sinusoid

### One Sinusoid

$$v(t) = A \cos \omega_0 t \quad \omega_0 = \frac{2\pi}{T}$$

$$p(t) = v^2(t) = A^2 \cos^2 \omega_0 t = \frac{A^2}{2} [1 + \cos 2\omega_0 t]$$

$$P_{AV} = \frac{1}{T} \int_{-T/2}^{T/2} p(t) dt = \frac{1}{T} \frac{A^2}{2} \int_{-T/2}^{T/2} [1 + \cos 2\omega_0 t] dt = \frac{A^2}{2}$$

### Two Sinusoids

$$v(t) = A \cos \omega_0 t + B \cos n \omega_0 t$$

$$p(t) = v^2(t) = A^2 \cos^2 \omega_0 t + B^2 \cos^2 n \omega_0 t$$

$$+ 2AB \cos \omega_0 t \cos n \omega_0 t$$

$$= \frac{A^2}{2} [1 + \cos 2\omega_0 t] + \cancel{\frac{B^2}{2} \cos 2n\omega_0 t} + \frac{B^2}{2} [1 + \cos 2n\omega_0 t]$$

$$+ AB \cos(n+1)\omega_0 t + AB \cos(n-1)\omega_0 t$$

$$P_{AV} = \frac{A^2}{2} + \frac{B^2}{2}$$

But if we have  $v(t) = A \cos \omega_0 t + B \cos \beta \omega_0 t$

where  $\beta \neq n$

then  $P_{AV} = \frac{1}{T} \int_{-T/2}^{T/2} p^2(t) dt \neq \frac{A^2}{2} + \frac{B^2}{2}$

This is the issue of orthogonality.

$\omega_0$  is a fundamental

$n\omega_0$  is the  $n^{\text{th}}$  harmonic

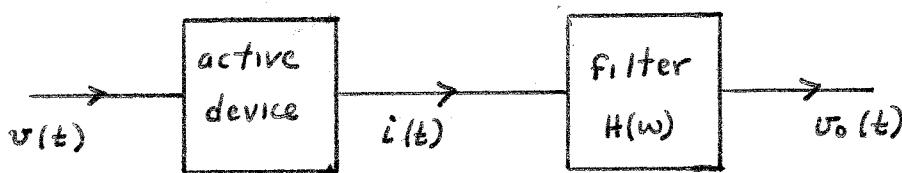
harmonics and fundamentals are orthogonal over the period of the fundamental.

## Sinusoids and Nonlinearities

An active device (like a transistor) has a transfer characteristic

$$i(t) = a_0 + a_1 v(t) + a_2 v^2(t) + a_3 v^3(t) + \dots$$

where  $a_0 > a_1 > a_2 > a_3 > \dots$ . Typically, the active device is followed by a filter in order to be able to extract the desired set of frequencies.



### example #1 Amplitude Modulation

Let  $v(t) = A \cos \omega_c t + B \cos \omega_i t$  where  $\omega_i \ll \omega_c$ .

$\nearrow$  carrier frequency (1 MHz)  
 $\nwarrow$  tone (1 KHz)

Ignore  $a_3, a_4, a_5, \text{etc} \dots$

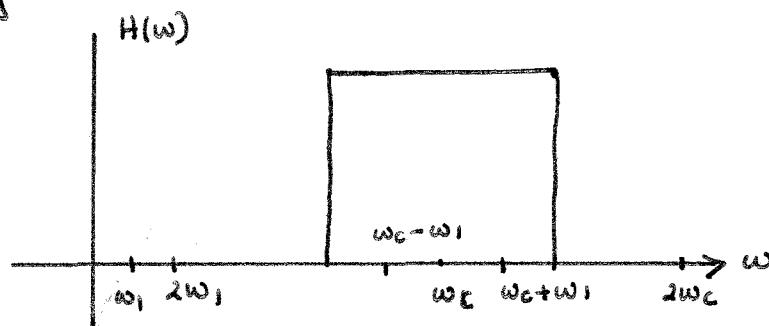
Then

$$i(t) = a_0 + a_1 (A \cos \omega_c t + B \cos \omega_i t)$$

$$\begin{aligned}
 &+ a_2 (A^2 \cos^2 \omega_c t + 2AB \cos \omega_c t \cos \omega_i t \\
 &+ B^2 \cos^2 \omega_i t)
 \end{aligned}$$

$$v(t) = \left[ a_0 + \frac{a_2 A^2}{2} + \frac{a_2 B^2}{2} \right] + a_1 B \cos \omega_1 t + \underline{\frac{a_2 B^2}{2} \cos 2\omega_1 t} \\ + a_1 A \cos \omega_1 t + 2a_2 AB \cos \omega_1 t \cos \omega_1 t + \underline{\frac{a_2 B^2}{2} \cos 2\omega_1 t}$$

A bandpass filter centered at  $\omega_c$ , ~~with~~ and extending a bit beyond  $\omega_c \pm \omega_1$  will include the underlined portion and exclude everything else.



The filter output can then be written as

$$v_o(t) = a_1 A \left[ 1 + 2B \frac{a_2}{a_1} \cos \omega_1 t \right] \cos \omega_1 t$$

This is an amplitude-modulated signal.

### example #2 Heterodyning or Mixing

This is similar to amplitude modulation except that

$$v(t) = A \cos \omega_1 t + B \cos \omega_2 t$$

where  $\omega_1$  and  $\omega_2$  are relatively close. For example,  $\omega_1 = 2 \text{ MHz}$  and  $\omega_2 = 5 \text{ MHz}$ .

In this instance, we want to use the term

$$2ABa_2 \cos \omega_1 t \cos \omega_2 t = a_2 AB \left\{ \cos(\omega_1 + \omega_2)t + \cos(\omega_1 - \omega_2)t \right\}$$

to generate the sum and the difference frequencies.

The filter is used to extract the desired one of these frequencies from the complicated set of non-linear device outputs.

### example #3      Intermodulation Distortion

In this example, we will focus on the 3rd order term in the non-linearity  $a_3 v^3(t)$

and we will have the input be composed of two frequencies that are very close to each other

$$v(t) = A \cos \omega_1 t + B \cos \omega_2 t \quad \text{where } \omega_2 = \omega_1 + \Delta$$

$$\Rightarrow v^3(t) = A^3 \cos^3 \omega_1 t + 3A^2 B \cos^2 \omega_1 t \cos \omega_2 t + 3AB^2 \cos \omega_1 t \cos^2 \omega_2 t + B^3 \cos^3 \omega_2 t$$

$$\text{Now } A^3 \cos^3 \omega_1 t = \frac{A^3}{2} \cos \omega_1 t (1 + \cos 2\omega_1 t)$$

$$= \frac{3A^3}{4} \cos \omega_1 t + \frac{A^3}{4} \cos 3\omega_1 t$$

$$3A^2 B \cos^2 \omega_1 t \cos \omega_2 t = \frac{3A^2 B}{2} (1 + \cos 2\omega_1 t) \cos 3\omega_2 t$$

$$= \frac{3A^2 B}{2} \cos \omega_2 t + \frac{3A^2 B}{4} \cos (2\omega_1 + \omega_2) t$$

$$+ \frac{3A^2 B}{4} \cos (2\omega_1 - \omega_2) t$$