

Time Diversity Techniques for Frequency-Selective Multipath Channels

EE161: Digital Communication Systems

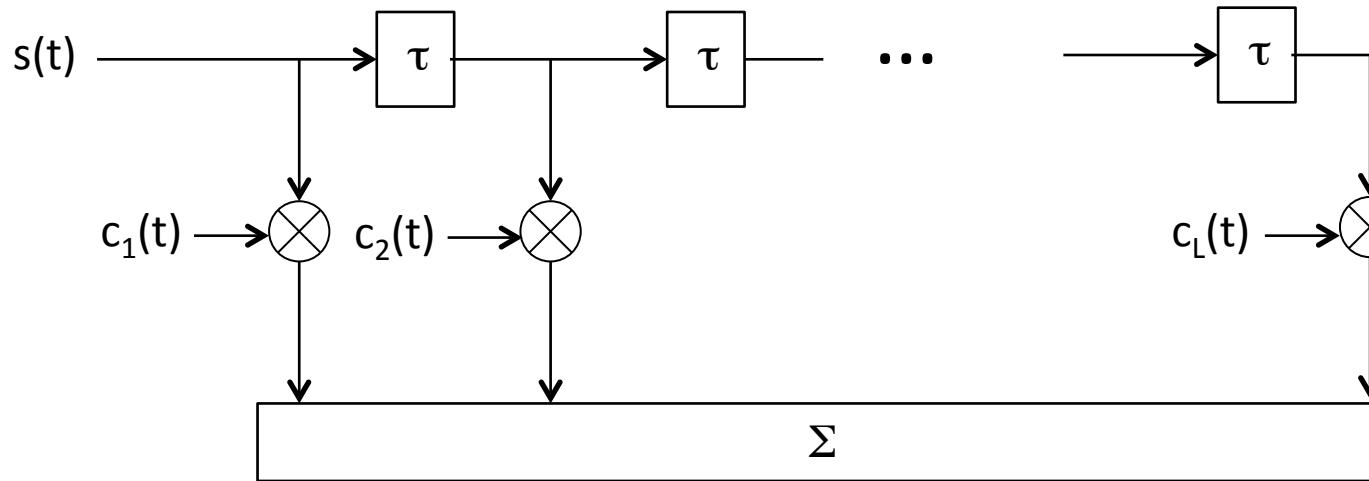
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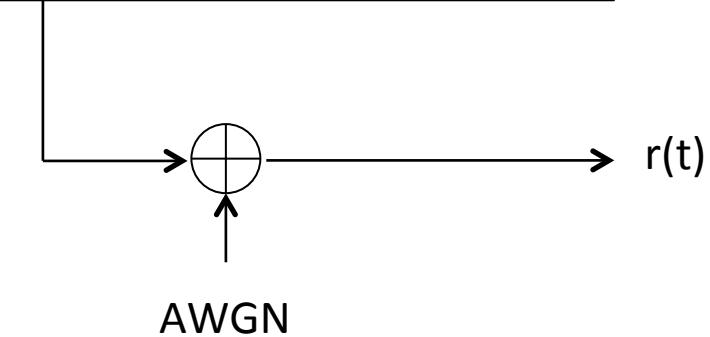
Time-diversity techniques

- Time-diversity techniques can be classified according to the frequency selectivity of the multipath channel
- Flat fading channels
 - Error correcting coding & interleaving (lecture notes)
 - Diversity order is equal to the minimum Hamming distance of the code
- Frequency-selective channels
 - RAKE demodulation
 - Linear adaptive equalization

Complex baseband frequency-selective multipath channel model



Chapter 14: Fading Multipath Channels



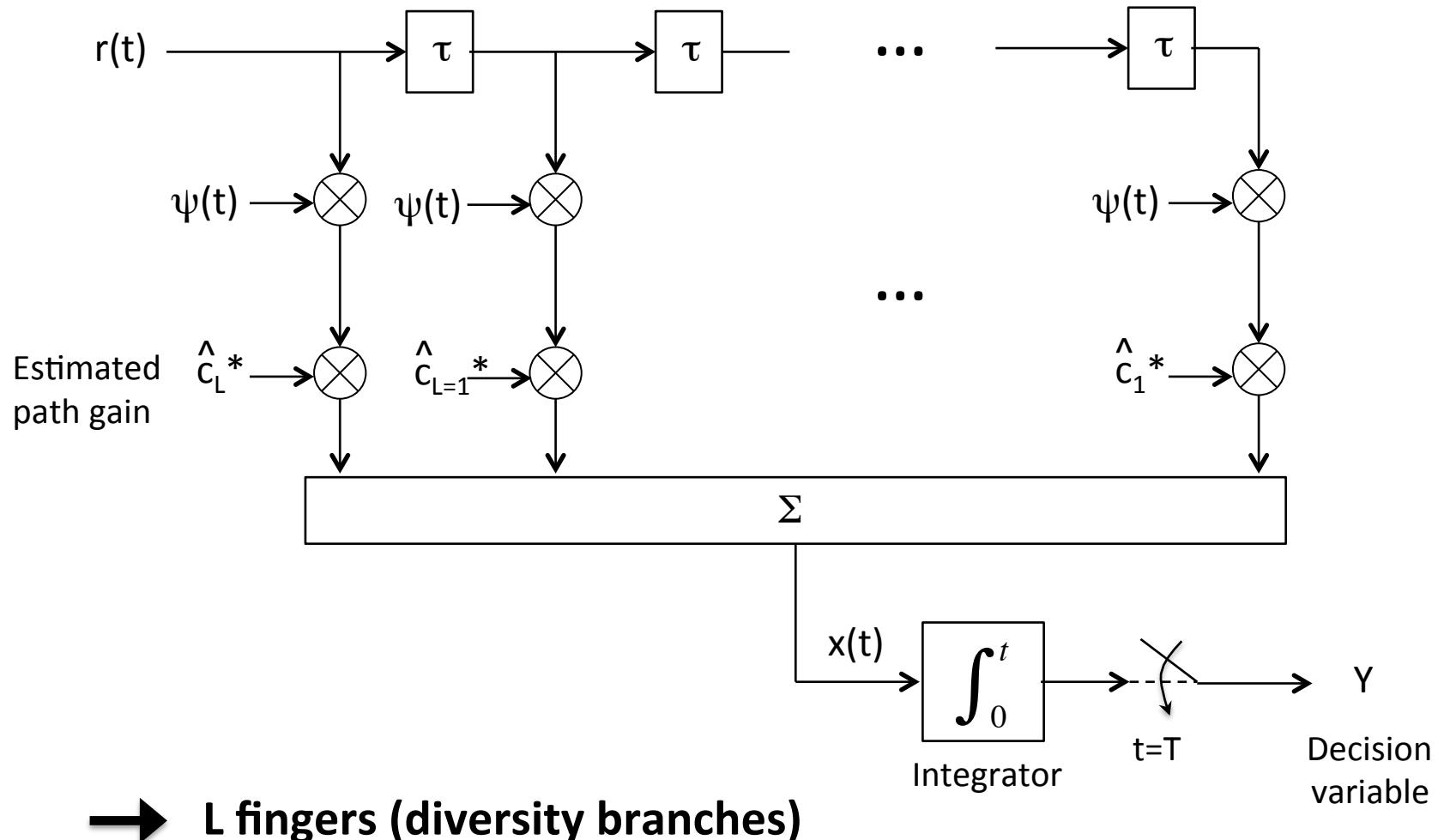
$$c_i(t) = a_i(t) e^{j\phi_i(t)} : i\text{-th path gain, } i=1,2,\dots,L \text{ (see lecture notes)}$$

RAKE demodulator: Assumptions

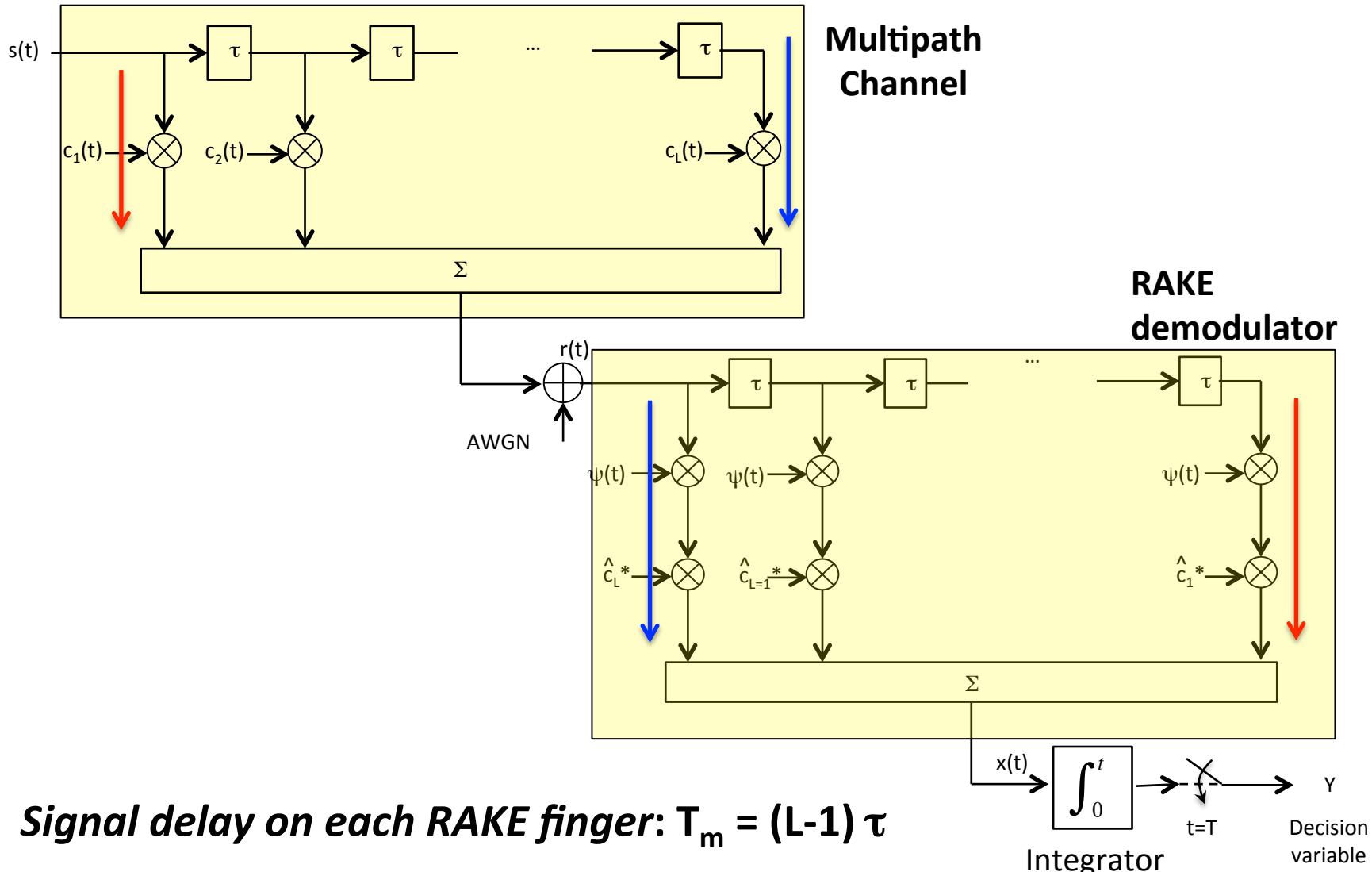
- Slow fading: $T \ll T_c \Rightarrow c_i(t) = c_i, i = 1, \dots, L$
- Frequency-selective fading: $W \gg B_c$ (1)
- No ISI: $T \gg T_m$ (2)
 - Assumptions (1) and (2) are satisfied by *wideband pulses*, such as PPM or spread-spectrum
 - *Examples (in classroom board)*
- Path gains known
 - Need channel estimation techniques (*finger search*)
- Binary modulation ($N=1$ pulse)

Assuming same delay between replicas in CIR

RAKE demodulator: Structure



Maximal-ratio combining property



Maximal-ratio combining property (cont.)

- RAKE output (assuming no AWGN):

$$x(t) = r(t)\psi(t)c_L^* + r(t-\tau)\psi(t)c_{L-1}^* + \cdots + r(t-(L-1)\tau)\psi(t)c_1^*$$

$$= \left[\sum_{i=1}^L c_i s(t - (i-1)\tau) \right] \psi(t)c_L^* + \left[\sum_{i=1}^L c_i s(t - i\tau) \right] \psi(t)c_{L-1}^*$$

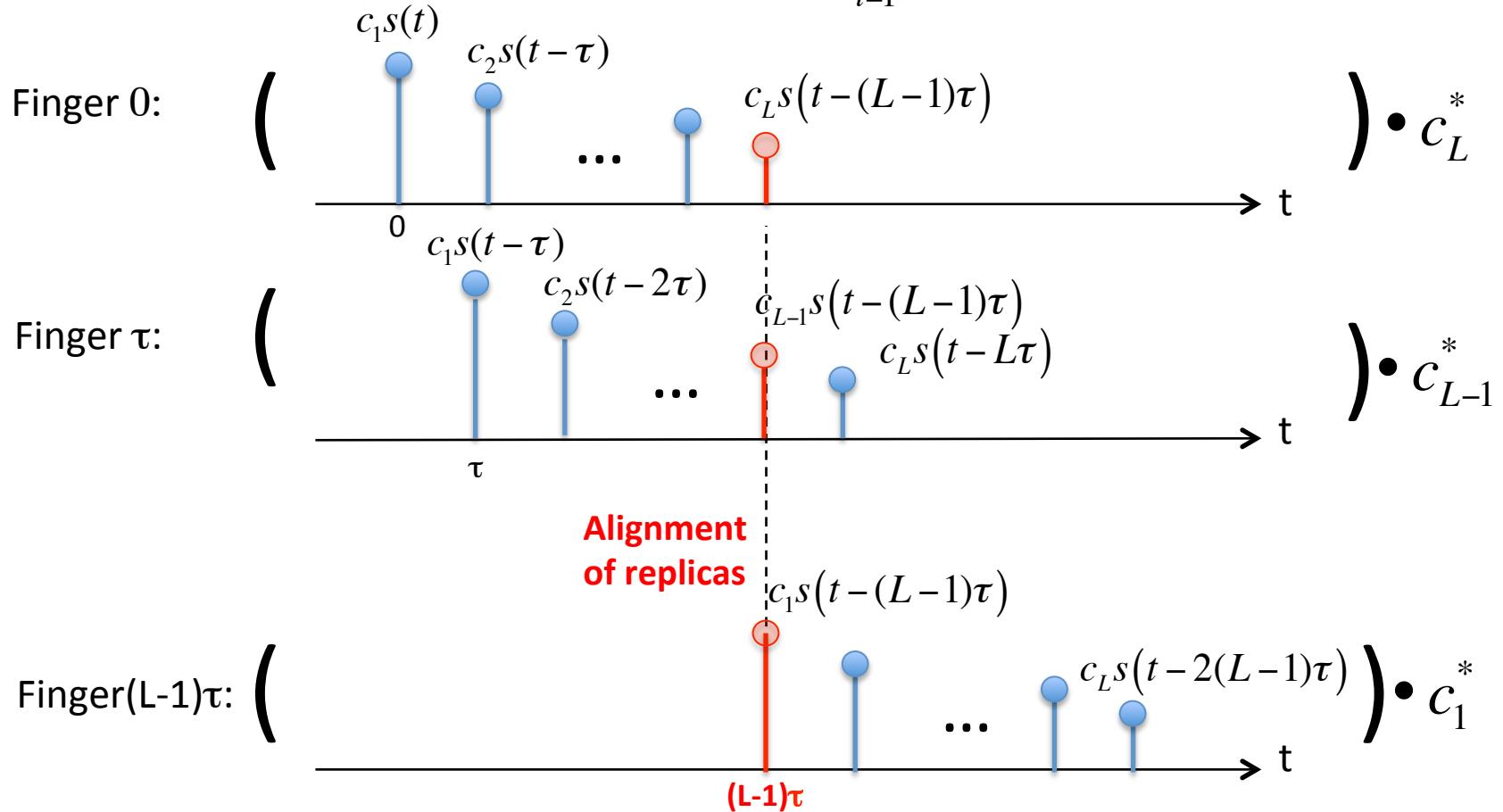
$$+ \cdots + \left[\sum_{i=1}^L c_i s(t - (i+L-2)\tau) \right] \psi(t)c_1^*$$

$$\Rightarrow x(t - (L-1)\tau) = \underbrace{\sum_{i=1}^L |c_i|^2}_{\text{MRC performance}} s(t - (L-1)\tau) \psi(t)$$

MRC performance

Maximal-ratio combining property (cont.)

- Matched-filter output: $Y = S \sum_{i=1}^L |c_i|^2 + N$



RAKE performance under Rayleigh fading

- The signal goes through L diversity branches
- Probability of error for BPSK (polar) and BFSK (binary orthogonal or 2-PPM):

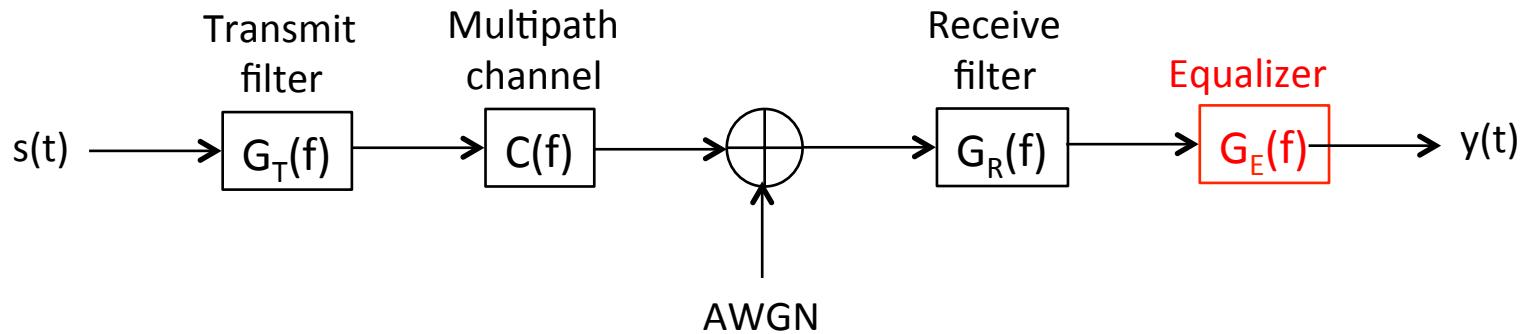
$$P_b = \frac{(2L-1)!}{L!(L-1)!} \prod_{i=1}^L \left[\frac{1}{2(1-c)\bar{E}_{s,i}/N_0} \right]^i$$

where $\bar{E}_{s,i}/N_0$ is the average signal energy-to-noise ratio of the i -th path, $i=1,2, \dots, L$, $c = -1$ for BPSK and $c = 0$ for BFSK.

Linear adaptive equalizer: Assumptions

- Slow fading: $T \ll T_c \Rightarrow c_i(t) \approx c_i, i = 1, \dots, L$
- Frequency-selective fading: $W \gg B_c$
- Presence of ISI: $T < T_m$
 - Signaling using *narrowband pulses*
 - *Example: Square-root raised-cosine (SRRC) pulses*
- Path gains unknown
 - Need pilot symbols (known at the receiver) and an adaptive algorithm to estimate gains
- Binary modulation ($N=1$ pulse)

Linear adaptive equalizer: System model



- Matched filter, assuming SRRRC pulses:

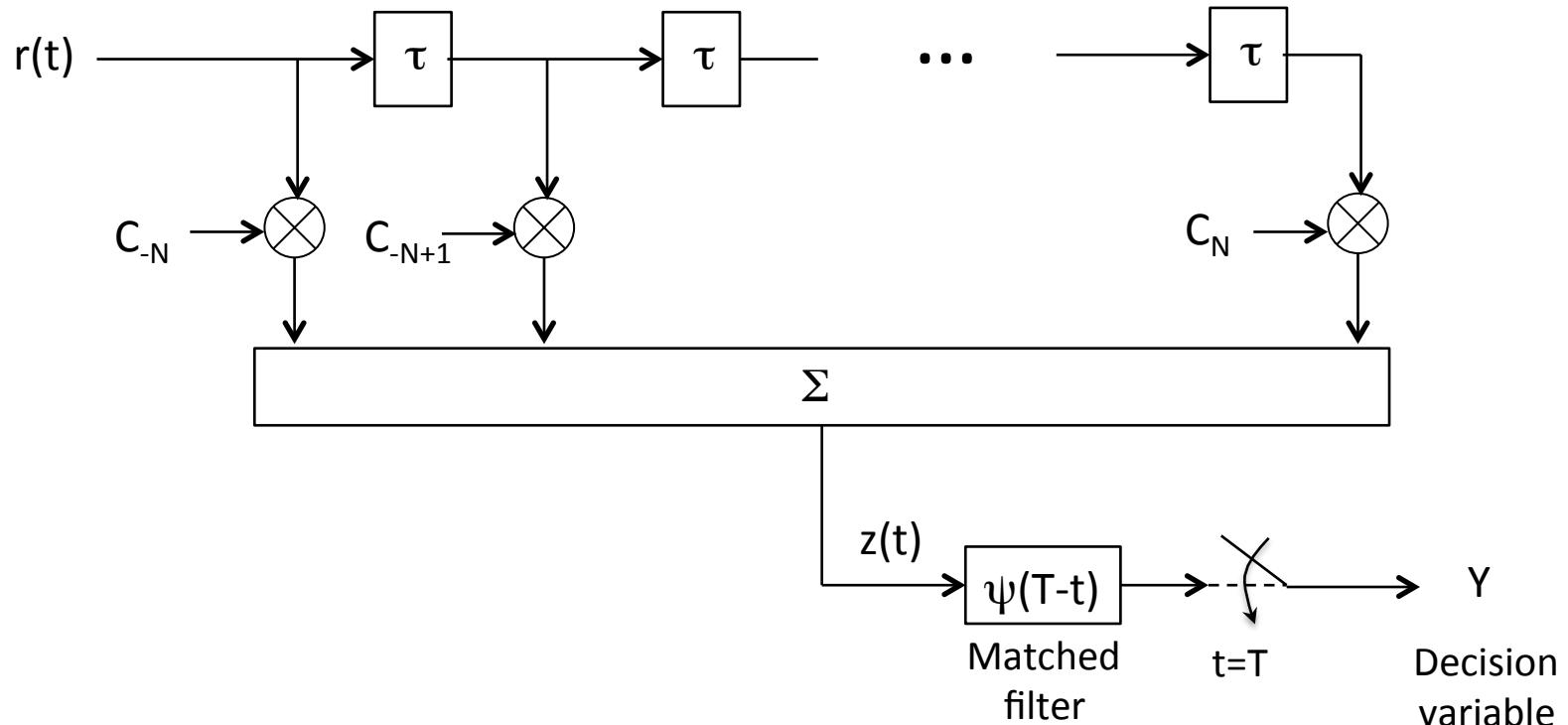
$$G_R(f) = G_T^*(f), \quad G_R(f)G_T(f) = X_{RC}(f) \quad (\text{Raised cosine})$$

- Equalizer can be interpreted as a filter “matched” to the channel:

$$G_E(f) = \frac{1}{C(f)}$$

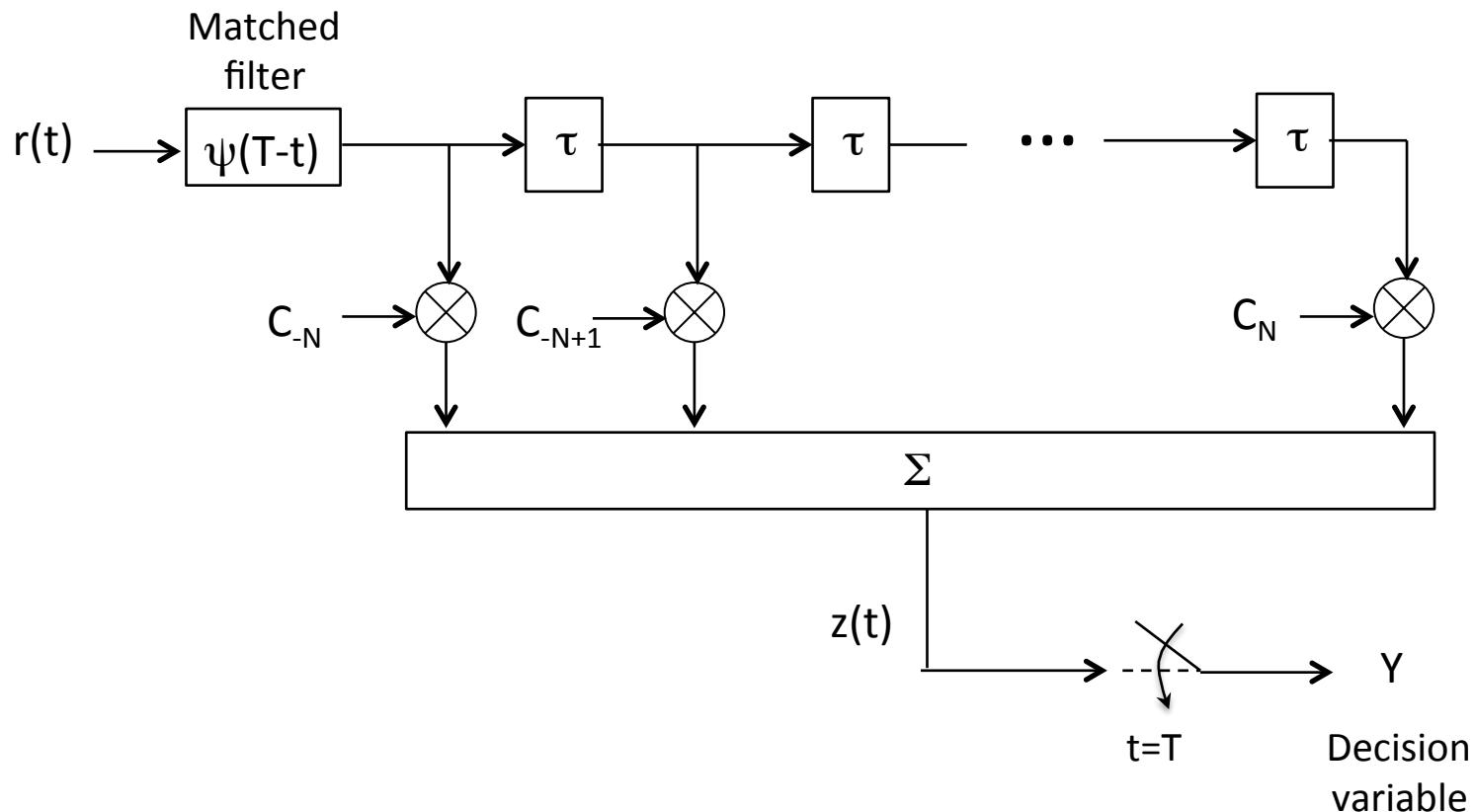
Linear adaptive equalizer: Pre-detector structure

- Similar to the RAKE demodulator:



- Equalizer (complex) coefficients may be updated based on an error signal between Y and pilot symbols known to the receiver: **MMSE criterion**. See Figure 9.22 of textbook

Linear adaptive equalizer: Post-detector structure



Typically, $\tau = T$ (symbol spaced) or $T/2$ or $T/4$ (fractionally spaced)

The LMS algorithm

- Coefficients update at time $t=(k+1)\tau$, for $n=-2N, \dots, 2N$:

$$C_n[k+1] = C_n[k] + \Delta e[k] \cdot r_n[k]$$

↑ ↑ ↑ ↑
Previous coefficient Step size Error Received (stored)
values

- Error signal (decision directed):

$$e[k] = s_P[k] - Y[k]$$

↑ ↑
Pilot symbol MF output
($a[k]$ in textbook)

- The adaptive equalizer can track ***slow variations*** of multipath fading. Moreover, if nulls in the frequency response are too pronounced then the BER is high

MATLAB demo

- Linear adaptive (LMS) equalizers work in two regimes
 - *Training*: Use *pilot symbols* to set coefficients values
 - *Equalization*: Combine replicas and estimate *data symbols*
- Matlab model ***Equalizer_Multipath.mdl*** from webpage of class
 - Demonstration in class
 - Post-detector structure
 - Frames of 3000 BPSK symbols (could be any number)
 - First 700 symbols (23% overhead) are pilot symbols (derived from a linear-feedback shift register or LFSR)
 - Remaining 2300 symbols are randomly generated data symbols