

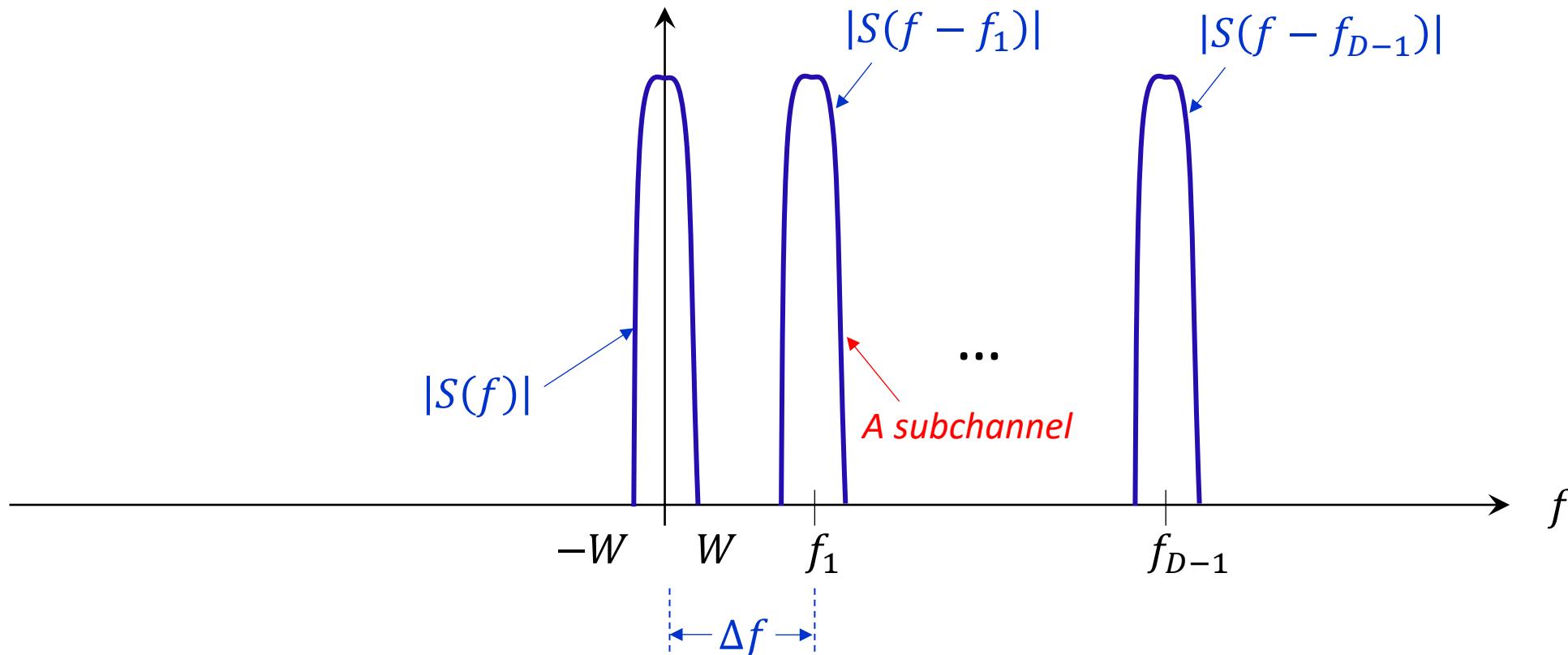
# Multicarrier Signals

EE161: Digital Communication Systems

San José State University

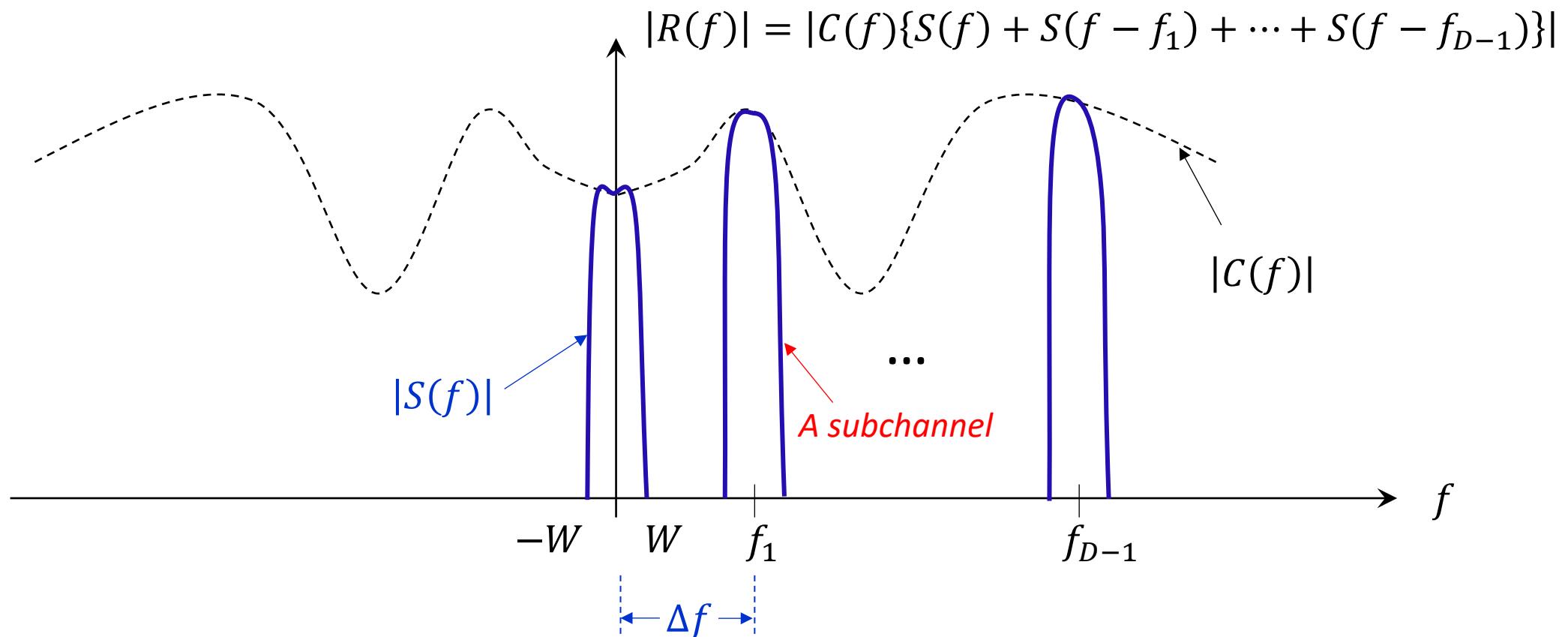
# Frequency domain diversity

- Pure diversity is achieved by sending the same signal over multiple frequencies.  
Example of a diversity order  $D$  scheme:



# Frequency domain diversity (cont.)

- Received signal:



# Some observations

- The time signal of this “pure-diversity” frequency domain scheme is *complex valued*:

$$S(f) + S(f - f_1) + \cdots + S(f - f_{D-1}) \xrightarrow{F} s(t) + s(t)e^{j2\pi f_1 t} + \cdots + s(t)e^{j2\pi f_{D-1} t}$$

$$= s(t)[1 + e^{j2\pi f_1 t} + \cdots + e^{j2\pi f_{D-1} t}]$$

- This is similar to a  $(D,1,D)$  repetition code
- Spectral efficiency* decreases by a factor of at least  $D$  (Coding rate:  $R = 1/D$ )
- To achieve full diversity, it is needed to have a spectral separation equal to  $\Delta f = f_{i+1} - f_i$ , for  $i = 0, 1, D - 2$ , (assumed constant) greater than the coherence bandwidth:

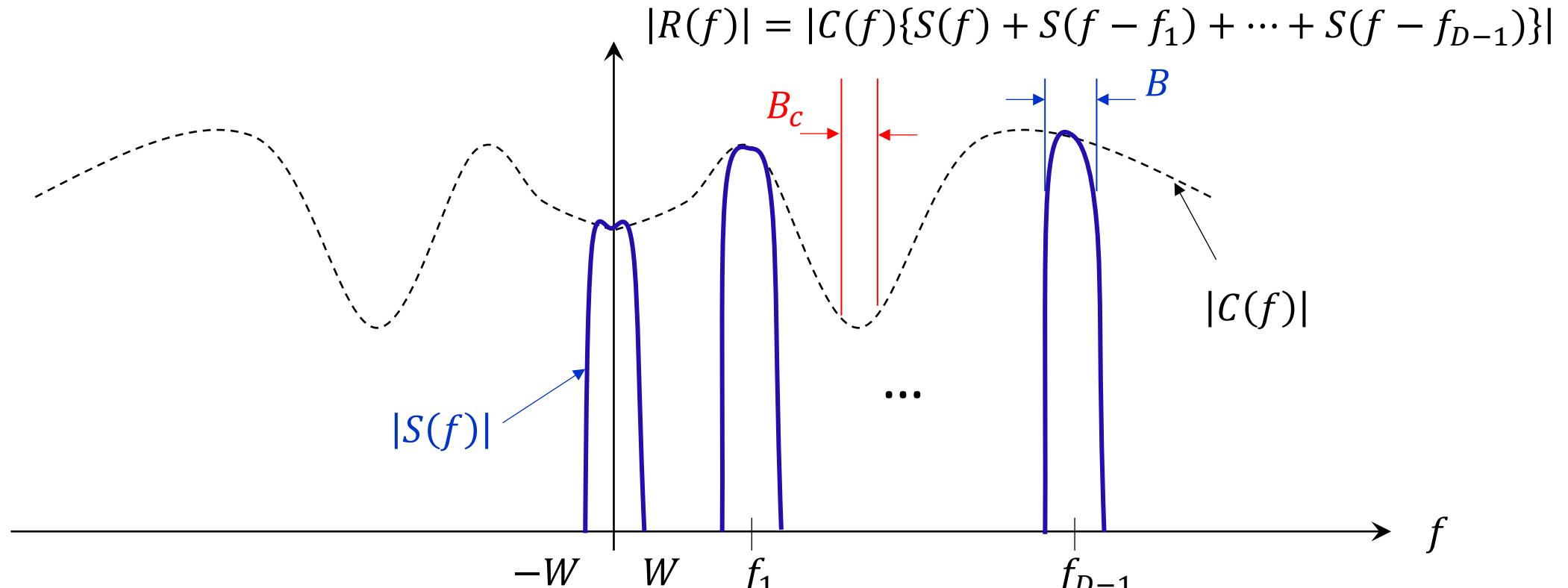
$$\Delta f > B_c$$

# Some observations (cont.)

- *Diversity* is achieved only if the signal bandwidth  $B = 2W$  satisfies

$$B \ll B_c$$

where  $B_c$  is the coherence bandwidth: *Narrowband (low rate) signal under flat fading*

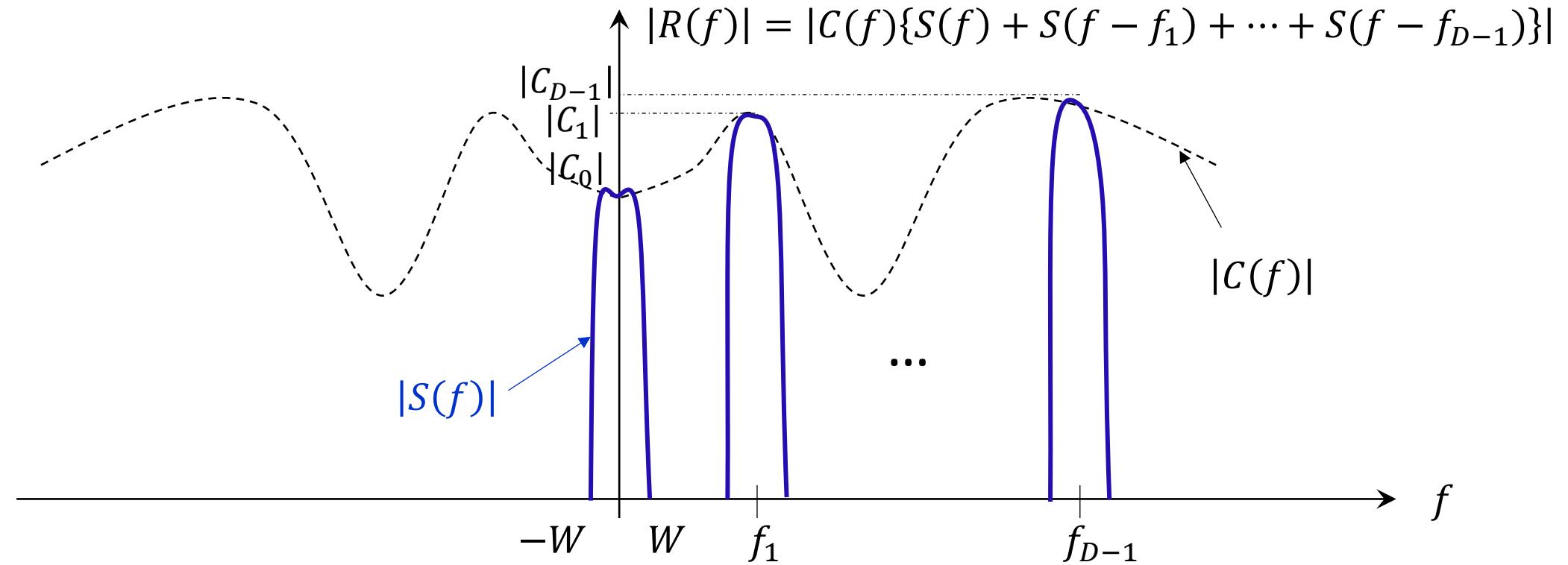


# Some observations (cont.)

- If the signal bandwidth  $B$  is not sufficiently small, then subchannels become frequency-selective
- ~~The larger the number  $K$  of subchannel frequencies  $f_i, i = 1, 2, \dots, K$ , the less frequency selective they become~~
- In the general case with a relatively large number of subchannels, the diversity order may be smaller:  $\mathbf{D} \leq \mathbf{K}$

# Channel estimation

- For diversity combining (e.g., MRC), the subchannel (complex-valued) gains estimates are needed:  $\hat{C}_i$ , where  $C_i = C(f_i)$  for  $i = 0, 1, \dots, D - 1$

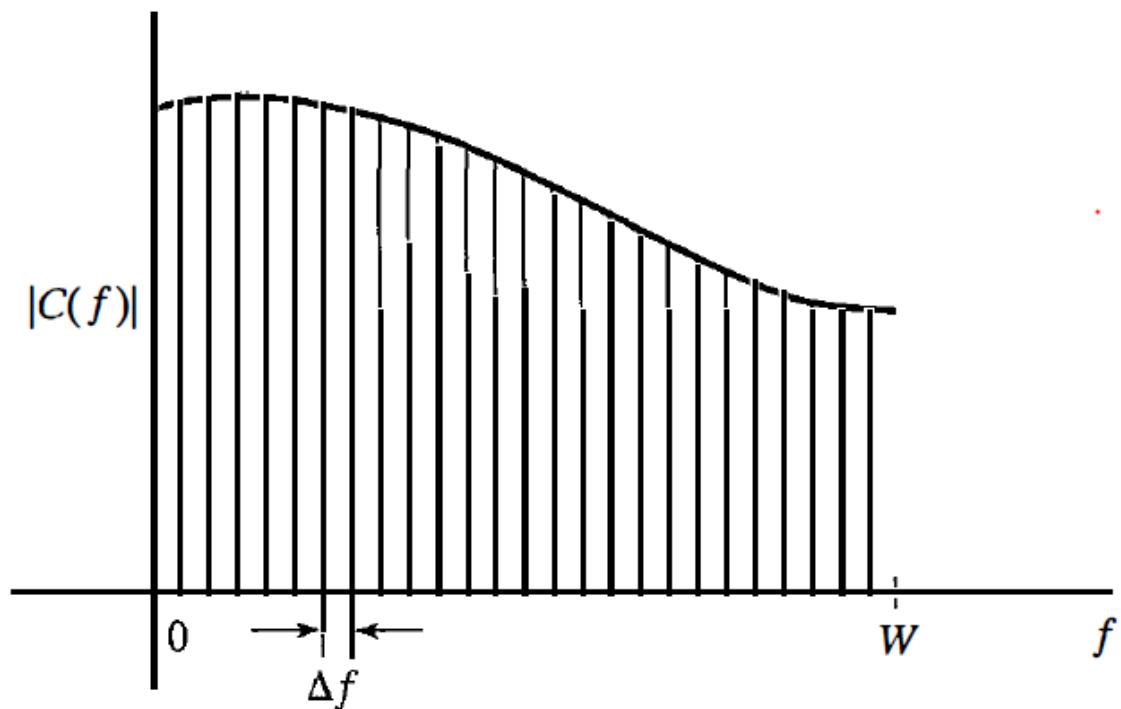


- This means that **pilot symbols** known at the receiver need to be transmitted

# Signal spectrum in the pure diversity scheme

- The spectrum  $S(f)$  of the signal can be either
  1. *Bandwidth-limited* (like with square-root raised-cosine transmit filters), or
  2. *Raised-cosine pulses in the time domain*, provided that the complex exponentials  $e^{j2\pi f_i t}, i = 1, 2, \dots, K$ , are orthogonal
- The second option is common in a scheme known as **orthogonal frequency-division multiplexing** (OFDM)
- Most of modern wireless networks today use OFDM

# OFDM: Divide and conquer approach



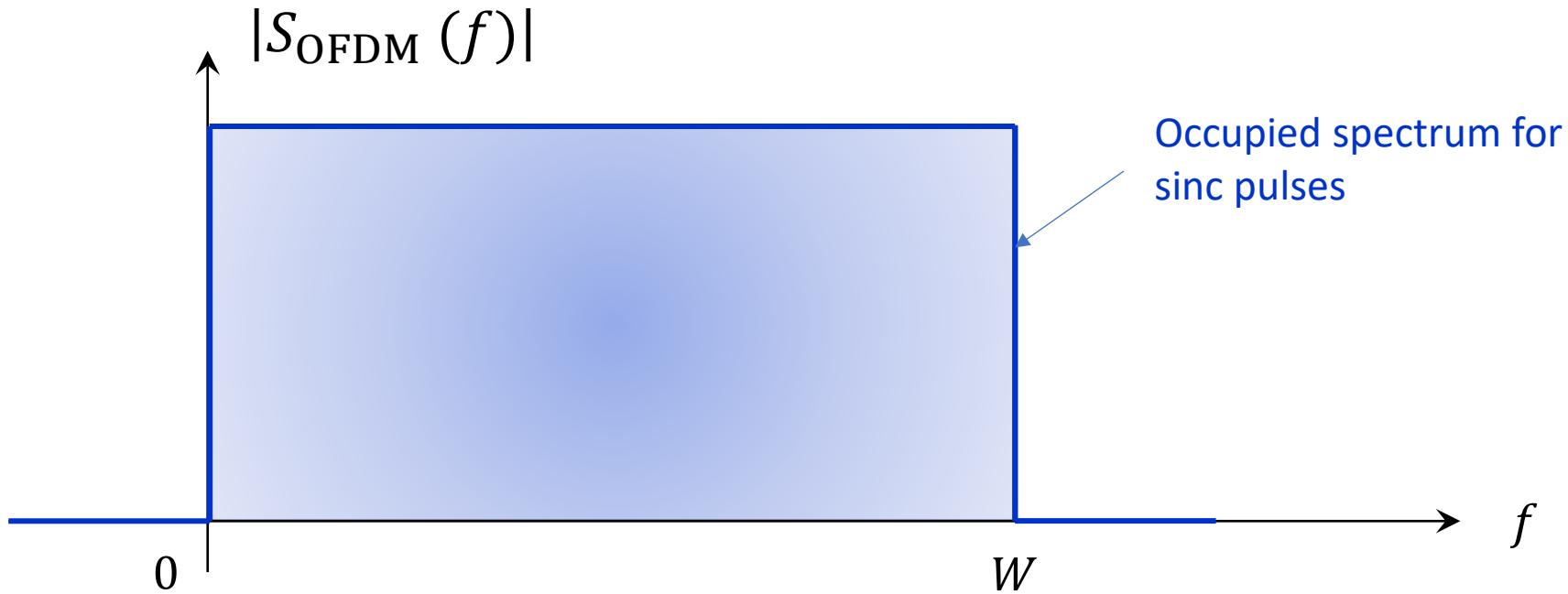
**Figure 11.1** Subdivision of the channel bandwidth  $W$  into narrowband subchannels of equal width  $\Delta f$ .

- **No spacing** between subchannels

Figure from textbook

# OFDM: Complex-signal spectrum

- The OFDM signal occupied spectrum is that of a complex signal, i.e., the amplitude spectrum is not even-symmetric !



- In other words, the RF bandwidth is the same as the lowpass bandwidth:

$$B = W$$

# OFDM: Signal multiplexing

- Unlike the pure-diversity scheme, in OFDM each subchannel carries an independent message signal:

$$s_{\text{OFDM}}(t) = s_0(t) + s_1(t)e^{j2\pi f_1 t} + \dots + s_{K-1}(t)e^{j2\pi f_{K-1} t}$$

- The number of subcarriers  $K$  is such that *each subchannel is relatively flat*
  - ✓ In this sense, OFDM is equivalent to quantizing the frequency domain