

The concept of diversity and diversity combining

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EE161: Digital Communication Systems

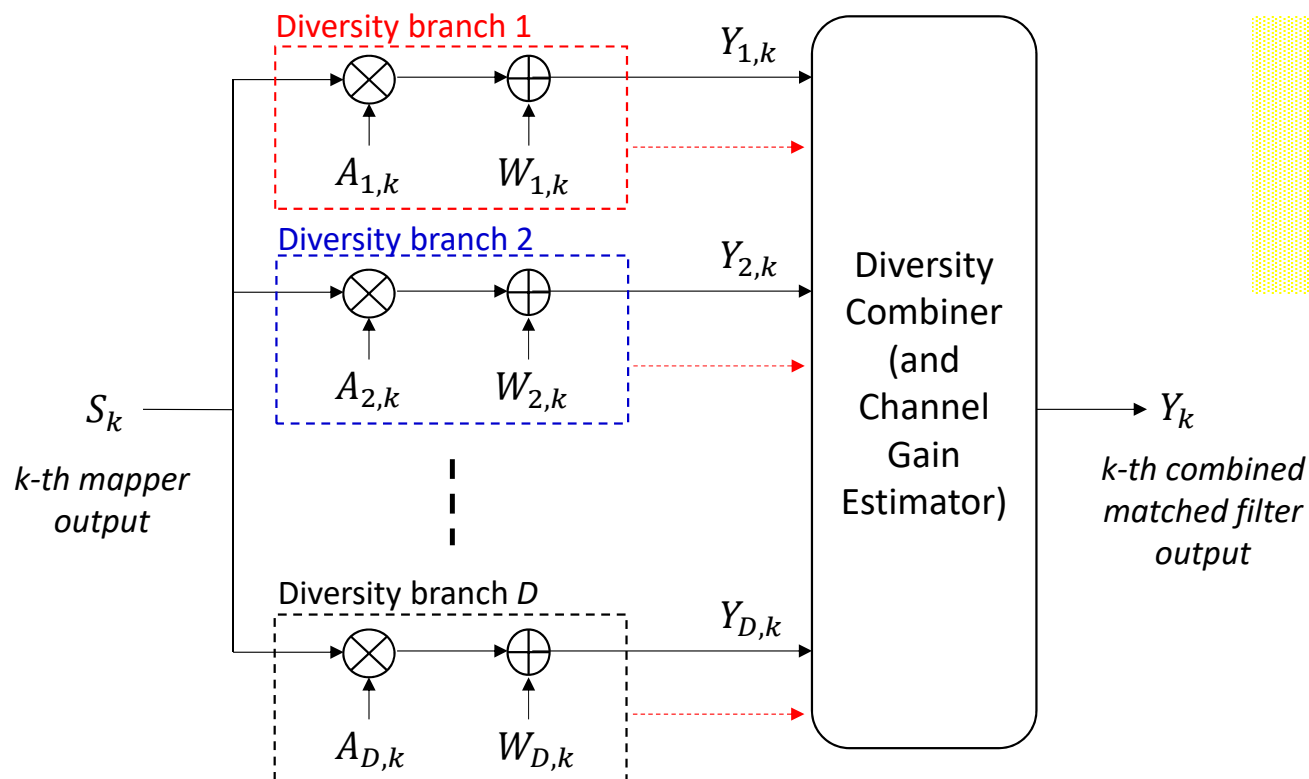
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Diversity

- Diversity in wireless communications is the process of creating D statistically independent (flat Rayleigh) fading channels
- Each channel is referred to as a *diversity branch*
- D is known as the *diversity order* of the system
- In general, each independent channel can have different types of fading
 - In this class focus is on *slow flat Rayleigh fading*

Complex baseband model with diversity



Each diversity branch i has
 $A_{i,k}$: Channel gain (complex)
 $W_{i,k}$: AWGN sample (complex)
 $Y_{i,k}$: MF output (complex)

$$A_{i,k} = \alpha_{i,k} e^{-j\phi_{i,k}}$$

(as in textbook)

Signal energy under Rayleigh fading

- The *instantaneous* signal energy (or power)-to-noise ratio,

$$\Gamma = \frac{E_s}{N_0} |A|^2,$$

is an **exponential** r.v. of parameter equal to the average signal energy-to-noise ratio γ_0 , i.e., its pdf is given by

$$p_{\Gamma}(\gamma) = (1/\gamma_0) \exp(-\gamma/\gamma_0), \quad \gamma > 0,$$

where $\gamma_0 = (E_s / N_0) E\{A^2\}$, and A is the Rayleigh fading amplitude.

Note: We use $\gamma_0 = \bar{\Gamma}$ in these notes

Outage probability under slow flat Rayleigh fading

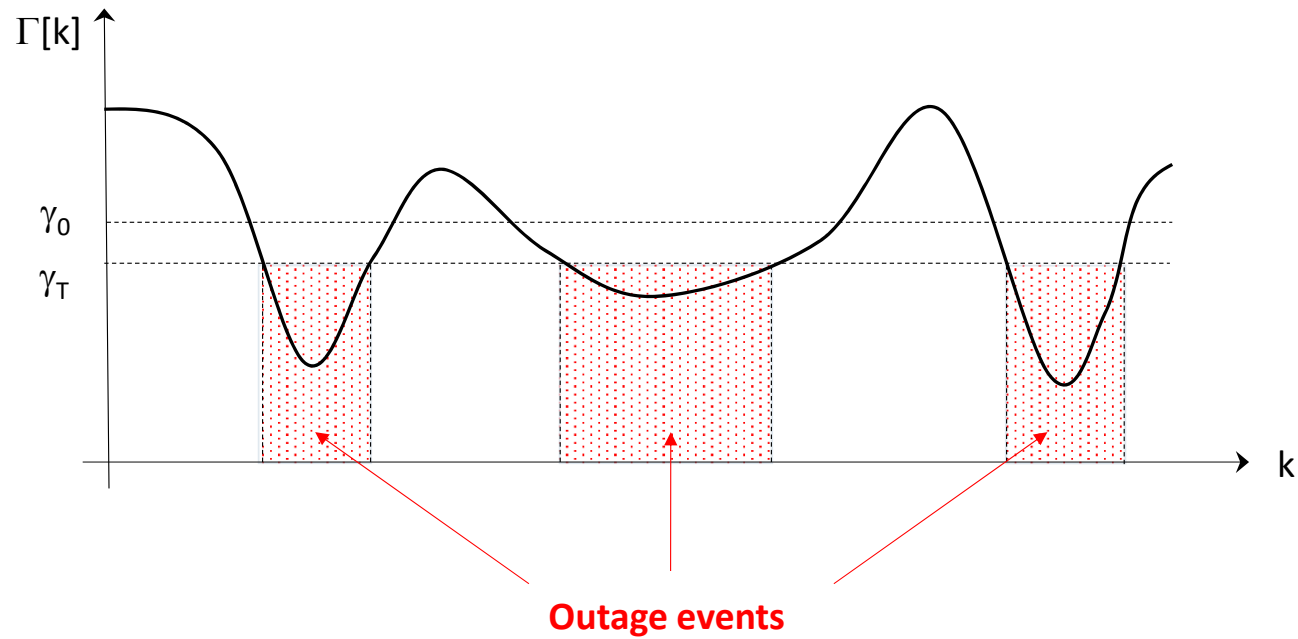
- The outage probability is defined as the probability that Γ falls below a threshold γ_T on the signal energy required to achieve a maximum average bit error probability:

$$P[\Gamma \leq \gamma_T] = F_T(\gamma_T) = 1 - \exp\left(-\frac{\gamma_T}{\gamma_0}\right)$$

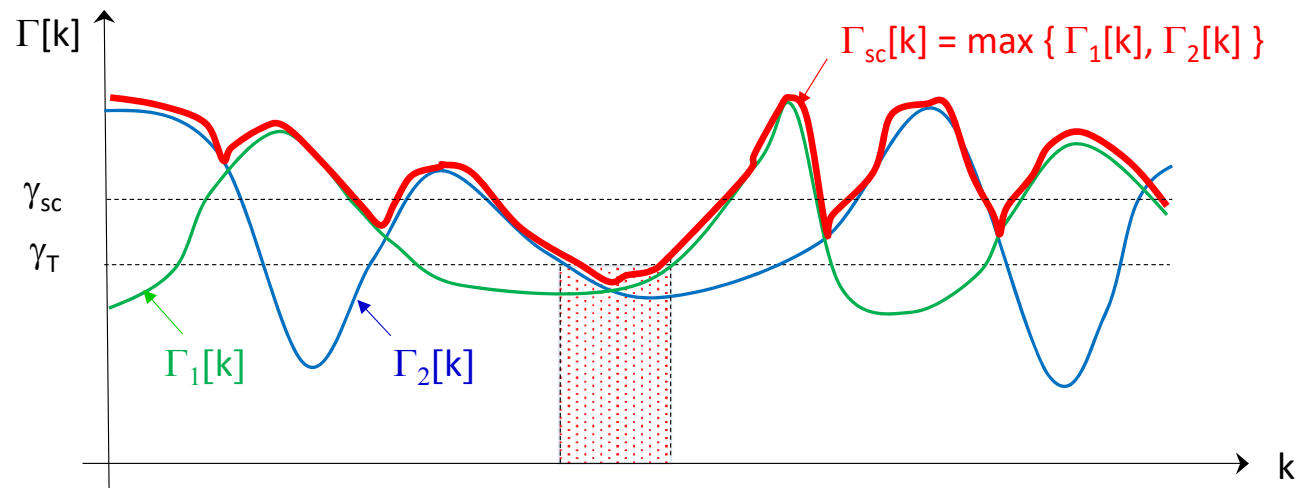
- How to set the threshold?
 - Under slow fading conditions, the threshold γ_T is determined based on the mapping (e.g., PAM, PSK, QAM) and the average bit error probability as a function of γ_T under AWGN (Q-function)

Whiteboard

Illustration of outage events



Selection combining: Two-branch ($D = 2$) example



- The selection combiner chooses the diversity branch with the *largest value* of Γ :

$$\Gamma_{sc} = \max \{ \Gamma_1, \Gamma_2 \}$$

- The average γ_{sc} increases
- The variance decreases

→

$$P[\Gamma_{sc} \leq \gamma_T] = \left(1 - \exp\left(-\frac{\gamma_T}{\gamma_0}\right) \right)^2 < P[\Gamma \leq \gamma_T]$$

Diversity combining techniques

- Complex baseband model: Operate on *matched-filter (or correlator) outputs*. Let the gain of each branch be $A_i = \alpha_i e^{-j\varphi_i}$
- Three popular techniques to combine the outputs:

1. Selection combining (SC):

$$Y_{SC} = \arg \max \{|Y_i|\}$$

(Worst performance)

2. Equal-gain combining (EGC):

$$Y_{EGC} = \sum_{i=1}^D Y_i e^{j\varphi_i}$$

3. Maximal-ratio combining (MRC):

$$Y_{MRC} = \sum_{i=1}^D \alpha_i Y_i e^{j\varphi_i}$$

(Best performance)

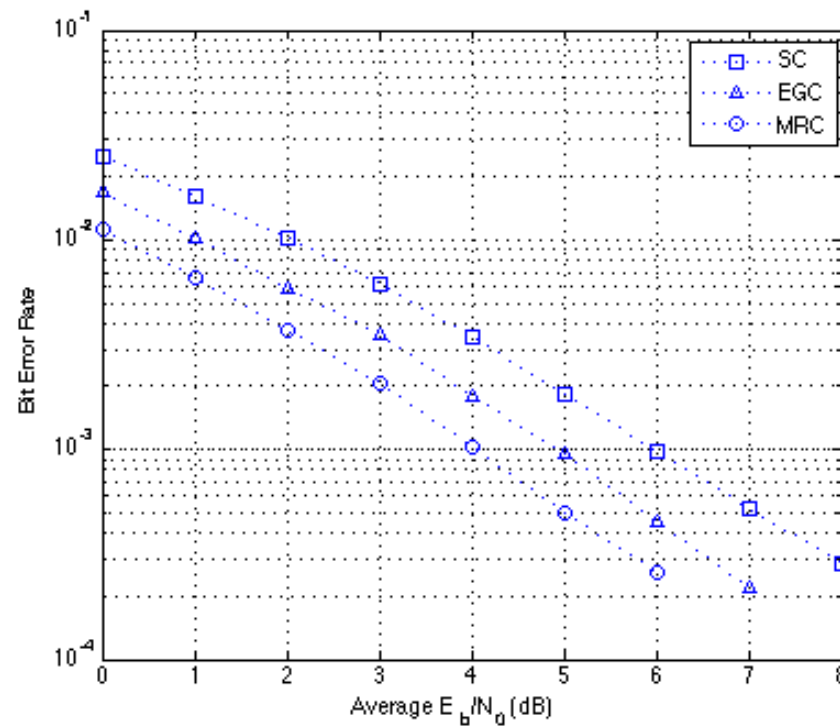
D-branch diversity performance

- Remarkably, once D diversity branches are created, **all** diversity combining techniques achieve the same diversity order D !!!
- That is, the average bit error probability is inversely proportional to the average signal energy-to-noise ratio, E_s/N_0 , raised to the D th power:

$$P_b \approx \frac{K_D}{(4\gamma_0)^D},$$

where K_D depends on the specific combining technique.

MATLAB simulation: BPSK with 4-branch diversity



Improvement in *post-combiner* average E_s / N_0 [1]

- SC:

$$\frac{\gamma_{\text{SC}}}{\gamma_0} = \sum_{i=1}^D \frac{1}{i}$$

- EGC:

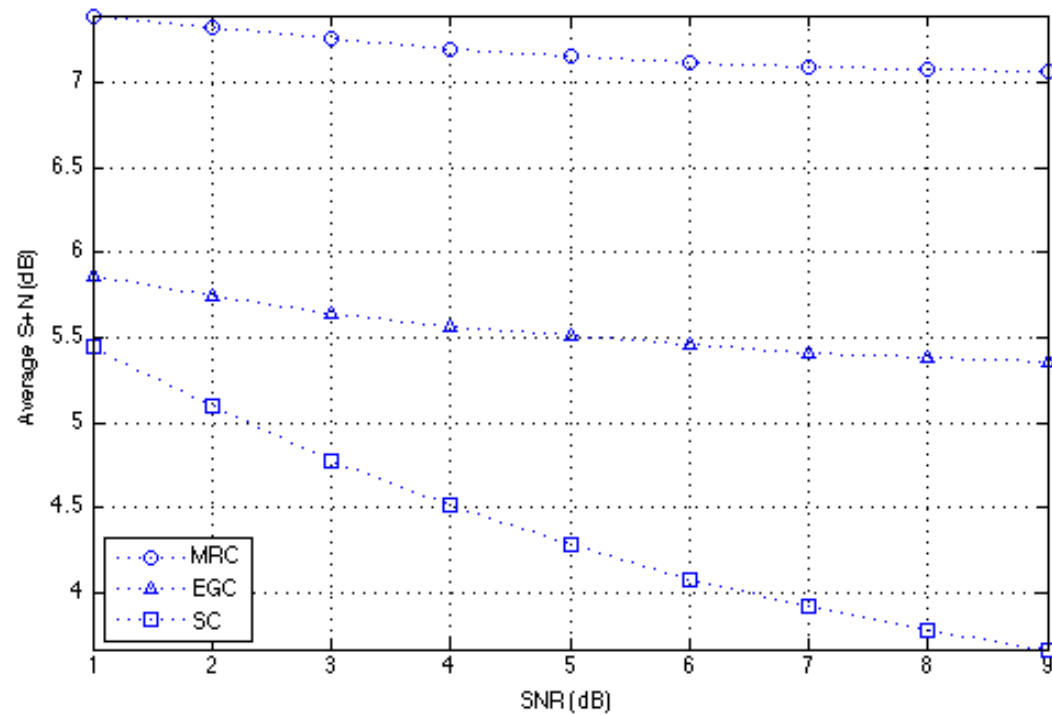
$$\frac{\gamma_{\text{EGC}}}{\gamma_0} = 1 + \frac{\pi}{4} (D - 1)$$

- MRC:

$$\frac{\gamma_{\text{MRC}}}{\gamma_0} = D$$

Whiteboard

MATLAB simulation: BPSK, 4-branch diversity (cont.)



Reference

1. P.M. Shankar, *Introduction to Wireless Systems*, Wiley, 2002.