

## Received signal model

Receive  $L$  noisy  $M$ -ary phase-shifted-keyed ( $M$ -PSK) symbols of the form

$$y_i = a_0 s_i + w_i, \quad i = 1, \dots, L.$$

- $a_0 = \rho_0 e^{j\theta_0}$  is the unknown carrier phase  $\theta_0$  and amplitude  $\rho_0$ ,
  - $s_1, \dots, s_L$  are transmitted  $M$ -PSK symbols,
  - $w_1, \dots, w_L \in \mathbb{C}$  are i.i.d. circularly symmetric complex random variables representing noise.
- Interested in estimating  $\rho_0$  and  $\theta_0$ .

## Least squares estimator

If all symbols  $s_1, \dots, s_L$  are known

$$\hat{a}_{uc} = \arg \min_{a \in \mathbb{C}} \sum_{i=1}^L |y_i - as_i|^2 = \frac{1}{L} \sum_{i=1}^L y_i s_i^*.$$

More interested in the practical situation where symbols are not known,

$$\hat{a} = \arg \min_{a \in \mathbb{C}} \min_{s_1, \dots, s_L} \sum_{i=1}^L |y_i - as_i|^2.$$

This estimator can be computed in  $O(L \log L)$  operations.

## Theorem (Almost sure convergence)

Let  $R_i \geq 0$  and  $\Phi_i \in [-\pi, \pi)$  be real random variables satisfying

$$R_i e^{j\Phi_i} = 1 + \frac{w_i}{a_0 s_i},$$

and define the continuous function

$$G(x) = \mathbb{E} R_1 \cos(x + \Phi_1).$$

If  $G(x)$  is uniquely maximised at  $x = 0$  over the interval  $[-\frac{\pi}{M}, \frac{\pi}{M})$ , then:

- $\langle \hat{\theta} - \theta_0 \rangle \rightarrow 0$  almost surely as  $L \rightarrow \infty$ ,
- $\hat{\rho} \rightarrow \rho_0 G(0)$  almost surely as  $L \rightarrow \infty$ ,

where  $\langle \cdot \rangle$  takes its argument 'modulo  $\frac{2\pi}{M}$ ' into  $[-\frac{\pi}{M}, \frac{\pi}{M})$ .

## Theorem (Asymptotic normality)

Let  $f(r, \phi)$  be the joint pdf of  $R_1$  and  $\Phi_1$ , and let

$$g(\phi) = \int_0^\infty r f(r, \phi) dr.$$

Put  $\hat{\lambda}_L = \langle \hat{\theta} - \theta_0 \rangle$  and  $\hat{m}_L = \hat{\rho} - \rho_0 G(0)$ . The distribution of  $(\sqrt{L} \hat{\lambda}_L, \sqrt{L} \hat{m}_L)$  converges to the bivariate normal with zero mean and covariance matrix

$$\begin{pmatrix} H^{-2}A & 0 \\ 0 & \rho_0^2 B \end{pmatrix}$$

as  $L \rightarrow \infty$ , where

$$H = G(0) - 2 \sin\left(\frac{\pi}{M}\right) \sum_{k=0}^{M-1} g\left(\frac{2\pi}{M}k + \frac{\pi}{M}\right),$$

$$A = \mathbb{E} R_1^2 \sin^2 \langle \Phi_1 \rangle, \quad B = \mathbb{E} R_1^2 \cos^2 \langle \Phi_1 \rangle - G^2(0).$$

## Simulations

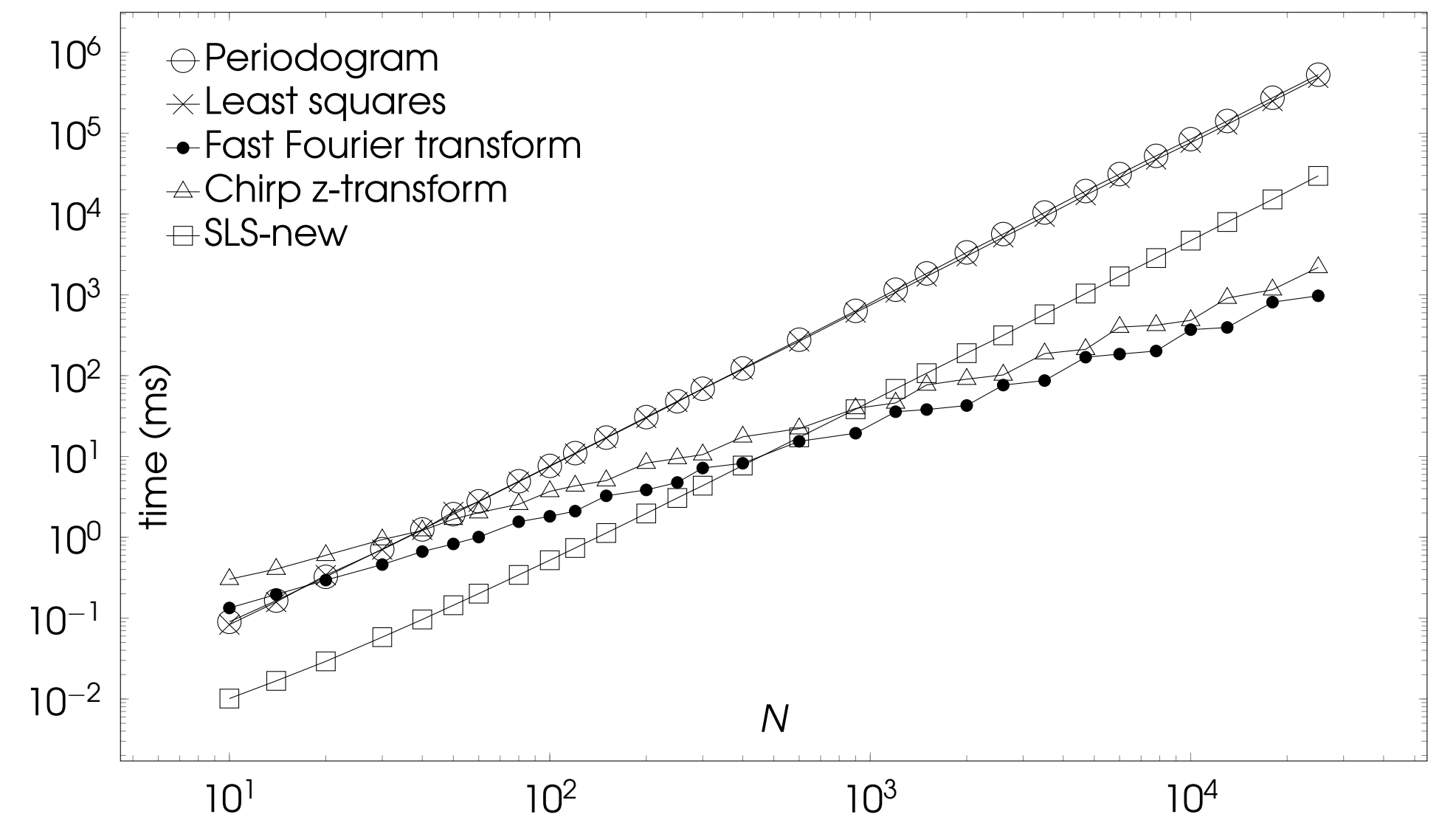


Figure : Computation time in milliseconds versus  $N$  for the periodogram, least squares, and quantized periodogram estimators computed using the chirp z-transform and a single fast Fourier transform.