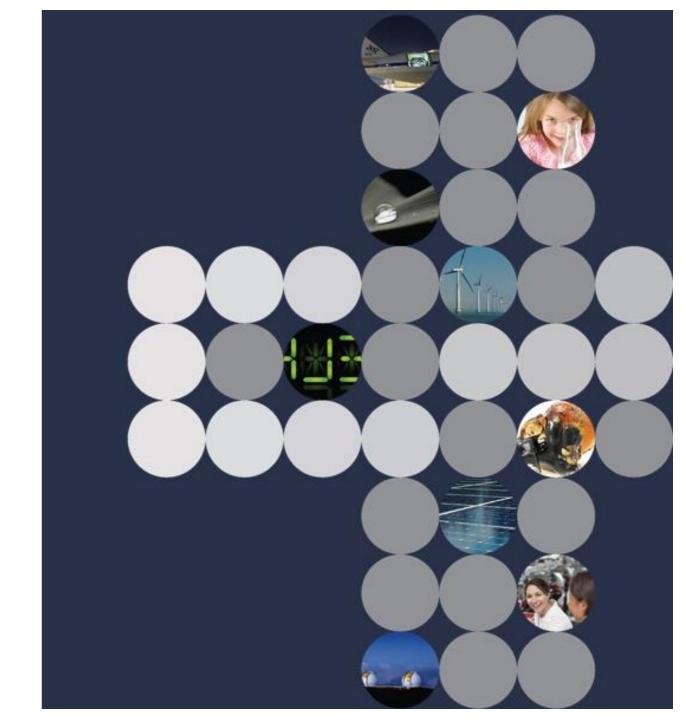
EE512 – Applied Biomedical Signal Processing

Linear Models I

João Jorge CSEM Signal Processing Group





Outline

Theory

- Intro/motivation
- Autoregressive (AR) models
 - Parameter estimation
 - Model design order selection
 - Signal description stability, structure

Practical session

- Based on Jupyter notebook
- See instructions in the notebook





Why use models?

- Make use of prior knowledge (physics, maths, biology) of the problem
- More efficient and/or robust estimation of signal properties (e.g. power spectral density)
- Get a **compact representation** of the signal structure (i.e. the model parameters)

... why not?

If the model is inappropriate, problems may follow

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Introduction

What are <u>linear</u> models?

 Models that rely on linear combinations of variables and the model parameters

Why study linear models?

- Quite common in nature (e.g. superposition of EM fields)
- Simple, attractive properties
- Well-studied, plenty of algorithms for diverse needs

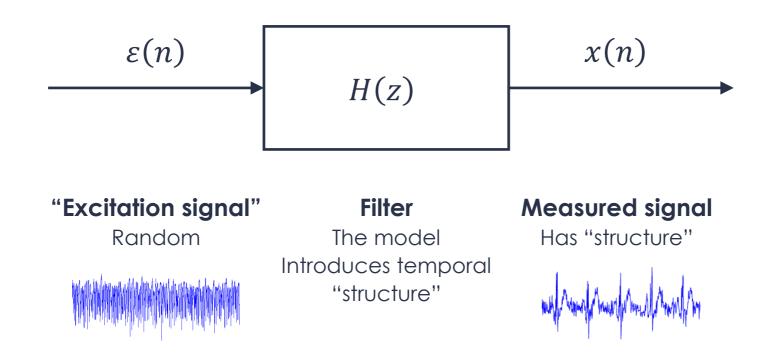






General framework

• We consider a discrete white noise process $\varepsilon(n)$ that enters a causal, linear shift-invariant filter H(z), resulting in a process x(n):



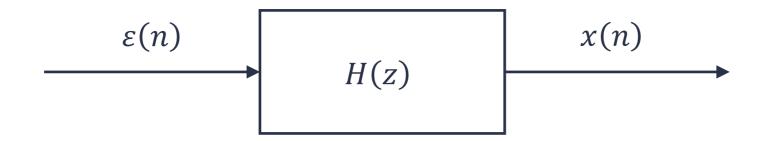




Autoregressive moving average (ARMA) models

General framework

• We consider a discrete white noise process $\varepsilon(n)$ that enters a causal, linear shift-invariant filter H(z), resulting in a process x(n):



• Moreover, we define the filter H(z) as a rational system function with p poles and q zeros:

$$H(z) = \frac{A(z)}{B(z)} = \frac{\sum_{k=0}^{q} b(k)z^{-k}}{1 + \sum_{k=1}^{p} a(k)z^{-k}}$$





Autoregressive moving average (ARMA) models

• Considering the relationship $X(z) = H(z)E(z) \Leftrightarrow X(z)B(z) = A(z)E(z)$, and returning to the time domain, we obtain:

$$x(n) + \sum_{l=1}^{p} a(l)x(n-l) = \sum_{l=0}^{q} b(l)\varepsilon(n-l)$$

• Or as a function of x(n):

$$x(n) = \sum_{l=0}^{q} b(l)\varepsilon(n-l) - \sum_{l=1}^{p} a(l)x(n-l)$$
 Linear combination of present & past inputs ε of past outputs x



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Autoregressive moving average (ARMA) models

• The ARMA model can be simplified when appropriate:

Autoregressive moving average (ARMA)

$$x(n) = \sum_{l=0}^{q} b(l)\varepsilon(n-l) - \sum_{l=1}^{p} a(l)x(n-l)$$

$$a(l) = 0 b(l) = \delta(l)$$

Moving average (MA)

$$x(n) = \sum_{l=0}^{q} b(l)\varepsilon(n-l)$$

$$x(n) = \varepsilon(n) - \sum_{l=1}^{p} a(l)x(n-l)$$



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Autoregressive (AR) models

Why use AR models?

- Practical experience shows good performance in diverse situations;
- Well suited to highlight oscillations in a signal;
- Easy estimation of the AR parameters;
- Direct link with linear signal prediction.







Autoregressive (AR) models

In general:

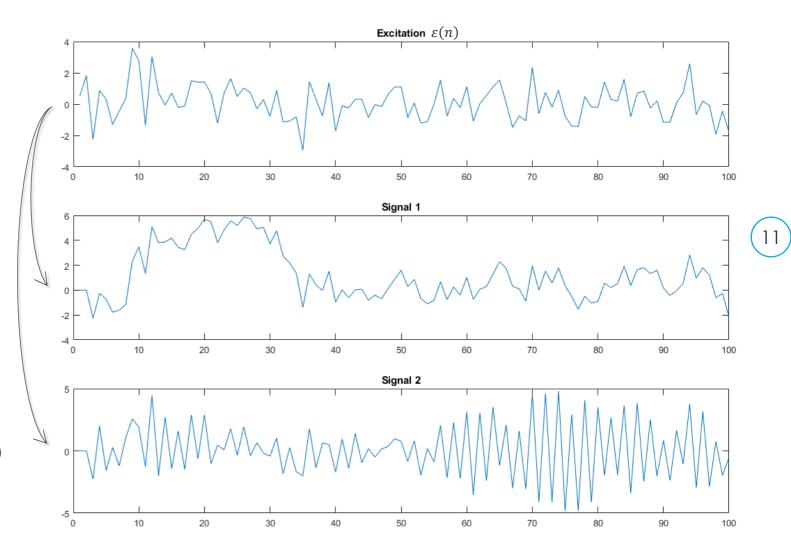
$$x(n) = -\sum_{l=1}^{p} a_l x(n-l) + \varepsilon(n)$$

Example 1

$$x(n) = 0.5 x(n-1) + 0.4 x(n-2) + \varepsilon(n)$$

Example 2

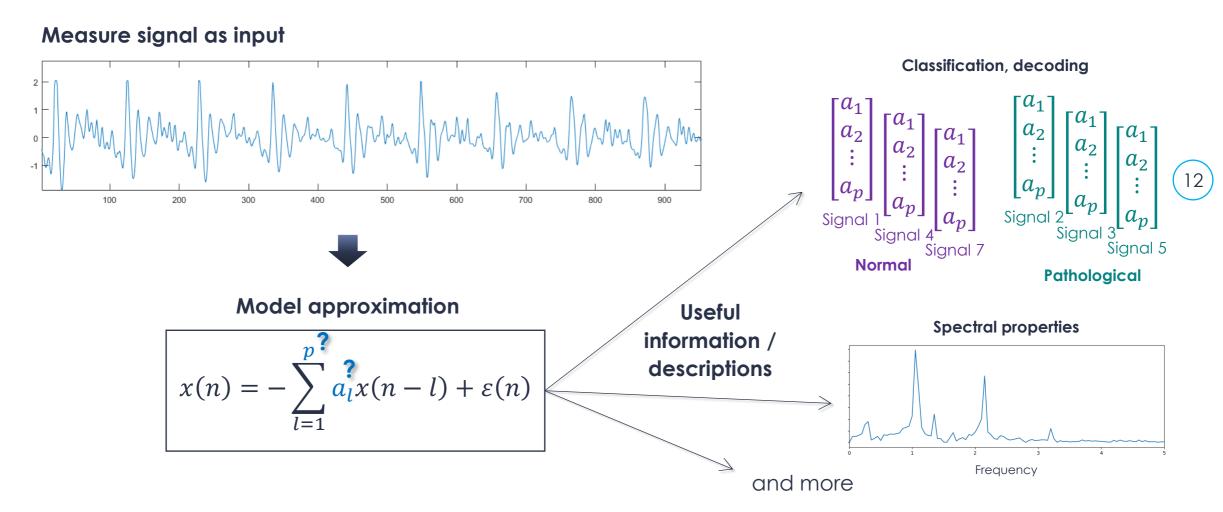
$$x(n) = -0.5 x(n-1) + 0.4 x(n-2) + \varepsilon(n)$$





Autoregressive (AR) models

• Typically, in practice, we go in the "opposite direction":





How can we estimate the parameters?

- An important relationship governing AR models is expressed in the socalled Yule-Walker equations.
- We start by noting that:
 - The autocorrelation of a signal x is defined as: $R_{xx}(n,k) = E\{x(n)x(n-k)\}$
 - In a stationary signal, $R_{\chi\chi}(n,k)$ does not depend on $n: R_{\chi\chi}(n,k) \to R_{\chi\chi}(k)$
 - In an AR process, since ε is stationary and H(z) is time-invariant, then x is also stationary
 - For now we assume we know $R_{xx}(k)$ and the desired model order p



- Starting from the base relationship: $x(n) = -\sum_{l=1}^{\infty} a_l x(n-l) + \varepsilon(n)$
- Then, with $k \geq 0$:

$$R_{xx}(k) = E\{x(n)x(n-k)\}\$$

$$= -\sum_{l=1}^{p} a_{l}E\{x(n-l)x(n-k)\} + E\{\varepsilon(n)x(n-k)\}\$$

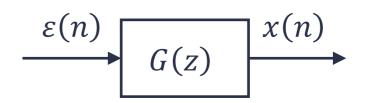
$$= -\sum_{l=1}^{p} a_{l}R_{xx}(k-l) + E\{\varepsilon(n)x(n-k)\}\$$

What can we say about this term?





Considering the transfer function:



• Then, for the previous term:

$$x(n) = \sum_{j=0}^{\infty} g(j) \, \varepsilon(n-j)$$

$$E\{\varepsilon(n)x(n-k)\} = \sum_{j=0}^{\infty} g(j)E\{\varepsilon(n)\varepsilon(n-k-j)\}$$

$$= \sum_{j=0}^{\infty} g(j)R_{\varepsilon\varepsilon}(k+j)$$

$$= \begin{cases} g(0)\sigma_{\varepsilon}^{2} & \text{if } k=0\\ 0 & \text{otherwise} \end{cases}$$

Note that with:
$$G(z) = \sum_{j=0}^{\infty} g(j)z^{-k}$$

Then:

$$g(0) = \lim_{z \to \infty} G(z)$$

= $\lim_{z \to \infty} \frac{1}{1 + \sum_{l=1}^{p} a(l)z^{-l}} = 1$

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We then obtain the

Yule-Walker equations:
$$R_{xx}(k) = \begin{cases} -\sum_{l=1}^{p} a_l R_{xx}(k-l) & \text{for } k > 0 \\ -\sum_{l=1}^{p} a_l R_{xx}(k-l) + \sigma_{\varepsilon}^2 & \text{for } k = 0 \end{cases}$$

- These linear equations relate the signal autocorrelation R_{xx} with the AR model parameters a_1 ;
- If R_{xx} is known, the parameters a_l can be estimated with only p equations;
- Once the a_l are estimated, the equation for k=0 will give σ_{ε}^2 .



• The p first Yule-Walker equations (k = 1, ..., p) can be expressed in matrix form (benefitting from the symmetry of R_{xx}) as:

$$\begin{bmatrix} R_{xx}(0) & R_{xx}(1) & \cdots & R_{xx}(p-1) \\ R_{xx}(1) & R_{xx}(0) & \cdots & R_{xx}(p-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_{xx}(p-1) & R_{xx}(p-2) & \cdots & R_{xx}(0) \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_p \end{bmatrix} = - \begin{bmatrix} R_{xx}(1) \\ R_{xx}(2) \\ \vdots \\ R_{xx}(p) \end{bmatrix}$$

$$R\mathbf{a} = -\mathbf{b}$$

- The matrix R is the autocorrelation matrix of x:
 - In practice, usually invertible \Rightarrow allows us to solve for a;
 - Hermitian/symmetric and Toeplitz (same value along each diagonal) matrix;



• A simple example: p = 1 model

Y.W. for
$$k = 1$$
: $[R_{\chi\chi}(0)][a_1] = -[R_{\chi\chi}(1)] \Leftrightarrow a_1 = -\frac{R_{\chi\chi}(1)}{R_{\chi\chi}(0)}$

Y.W. for
$$k=0$$
: $R_{\chi\chi}(0)=-a_1R_{\chi\chi}(1)+\sigma_{\varepsilon}^2 \Leftrightarrow \sigma_{\varepsilon}^2=(1-a_1^2)R_{\chi\chi}(0)$

Thus giving a_1 and σ_{ε}^2 .



In general, how to solve the Yule-Walker equations for a?

- The Toeplitz property of R allows applying the Levinson-Durbin algorithm;
- It computes the AR parameters recursively for increasing order p.
- We denote by a_{pi} the i^{th} coefficient, and by σ_p^2 the variance of ε for the p-order model.

$$\begin{bmatrix} R_{\chi\chi}(0) & R_{\chi\chi}(1) & \cdots & R_{\chi\chi}(p-1) \\ R_{\chi\chi}(1) & R_{\chi\chi}(0) & \cdots & R_{\chi\chi}(p-2) \\ \vdots & & \vdots & \ddots & \vdots \\ R_{\chi\chi}(p-1) & R_{\chi\chi}(p-2) & \cdots & R_{\chi\chi}(0) \end{bmatrix} \begin{bmatrix} a_{p1} \\ a_{p2} \\ \vdots \\ a_{pp} \end{bmatrix} = - \begin{bmatrix} R_{\chi\chi}(1) \\ R_{\chi\chi}(2) \\ \vdots \\ R_{\chi\chi}(p) \end{bmatrix}$$





AR models: Levinson-Durbin algorithm

• The algorithm:

- 1. Initialize with the 1st-order model:
 - a) $a_{11} = -R_{xx}(1)/R_{xx}(0)$
 - b) $\sigma_1^2 = (1 a_{11}^2) R_{\chi\chi}(1)$
- 2. For increasing order j = 2, 3, ..., p:
 - $\gamma_j = -\left(R_{xx}(j) + \sum_{l=1}^{j-1} a_{j-1,l} R_{xx}(j-l)\right) / \sigma_{j-1}^2$
 - b) For l = 1, 2, ..., j 1: $a_{j,l} = a_{j-1,l} + \gamma_j a_{j-1,j-l}$
 - $a_{j,j} = \gamma_j$
 - $\sigma_i^2 = \sigma_{i-1}^2 \left(1 \gamma_i^2 \right)$

By the end we obtain the a_p parameters and the σ_p^2 value.

A derivation can be found in:

Hayes, Statistical Digital Signal Processing and Modeling, Wiley, 1996, chapter 5.2







Notes:

- At each iteration (order j), the estimated parameters are exactly those that would be obtained directly from Y.W. for that order;
- It can be shown that $|\gamma_j| \leq 1$ for a stable signal; as a consequence, $\sigma_j^2 \leq \sigma_{j-1}^2$ the excitation variance decreases when j increases;
- If x is generated by an AR(m) process, then for j > m, we will obtain: $\gamma_j = 0$, $\sigma_j^2 = \sigma_m^2$, $a_{j,m+i} = 0$, i.e. it is useless to go beyond m.

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What about the autocorrelation $R_{\chi\chi}(k)$?

- In most applications, the **statistical autocorrelation** $R_{xx}(k)$ is unknown, and must be estimated from a sample realization of the process;
- Given a sample x(n) with length N ($1 \le n < N$), we may estimate the sample autocorrelation $\hat{R}_{xx}(k)$:

$$\hat{R}_{xx}(k) = \frac{1}{N} \sum_{n=1}^{N-k} x(n)x(n+k)$$

Using this biased version has some convenient properties and guarantees the estimated AR model is stable.





$$x_1 = 2, x_2 = 1, x_3 = -2$$
 (bpm)

- **1.** Calculate the sample autocorrelation (biased estimate) of x, $\hat{R}_{xx}(k)$, for k = 0,1,2, based on these samples.
- **2.** Suppose we seek to model this signal with an AR(2) model. Write its Yule-Walker equations in terms of the AR model parameters, and with the numerical values obtained in **1**.
- 3. Solve the Yule-Walker equations to estimate <u>all</u> the AR model parameters.

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Exam-like question

$$x_1 = 2, x_2 = 1, x_3 = -2$$
 (bpm)

1. Calculate the sample autocorrelation (biased estimate) of x, $\hat{R}_{xx}(k)$, for k = 0,1,2, based on these samples.

The sample autocorrelation (biased form) is given by:

Therefore, for this sample, we have:

$$\hat{R}_{xx}(k) = \frac{1}{N} \sum_{n=1}^{N-k} x(n)x(n+k)$$

$$\widehat{R}_{xx}(0) = \frac{1}{3}(4+1+4) = 3$$

$$\hat{R}_{xx}(1) = \frac{1}{3}(2-2) = 0$$

$$\widehat{R}_{xx}(2) = -\frac{1}{3}4 = -\frac{4}{3}$$

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Exam-like question

2. Suppose we seek to model this signal with an AR(2) model. Write its Yule-Walker equations in terms of the AR model parameters, and with the numerical values obtained in 1.

The Yule-Walker equations are given by:

$$R_{xx}(k) = \begin{cases} -\sum_{l=1}^{p} a_l R_{xx}(k-l) & \text{for } k > 0 \\ -\sum_{l=1}^{p} a_l R_{xx}(k-l) + \sigma_{\varepsilon}^2 & \text{for } k = 0 \end{cases} \qquad k = 1: \qquad -R_{xx}(1) = a_1 R_{xx}(0) + a_2 R_{xx}(-1) \\ \leftrightarrow -R_{xx}(1) = a_1 R_{xx}(0) + a_2 R_{xx}(1) \end{cases}$$

For an AR(2) model, this becomes (note the minus sign was moved to the left side):

$$k = 2$$
: $-R_{xx}(2) = a_1 R_{xx}(1) + a_2 R_{xx}(0)$
 $\frac{4}{3} = 3a_2$

$$k = 1$$
: $-R_{xx}(1) = a_1 R_{xx}(0) + a_2 R_{xx}(-1)$
 $\leftrightarrow -R_{xx}(1) = a_1 R_{xx}(0) + a_2 R_{xx}(1)$
 $\mathbf{0} = \mathbf{3}a_1$

$$k = 0: -R_{xx}(0) = a_1 R_{xx}(-1) + a_2 R_{xx}(-2) - \sigma_{\varepsilon}^2$$

$$\leftrightarrow -R_{xx}(0) = a_1 R_{xx}(1) + a_2 R_{xx}(2) - \sigma_{\varepsilon}^2$$

$$-3 = -\frac{4}{3} a_2 - \sigma_{\varepsilon}^2$$



Exam-like question

3. Solve the Yule-Walker equations to estimate <u>all</u> the AR model parameters.

From the previous answer, we have:

$$k = 2$$
: $\frac{4}{3} = 3a_2$

$$k = 1$$
: $0 = 3a_1$

$$k = 0$$
: $-3 = -\frac{4}{3}a_2 - \sigma_{\varepsilon}^2$

- From the equation of k = 1 we see that $a_1 = 0$.
- From the eq. of k=2 we can derive $a_2=\frac{4}{9}$.
- From the eq. of k=0 , and having obtained a_2 above, we can derive:

$$\sigma_{\varepsilon}^2 = 3 - \frac{4}{3}a_2 = 3 - \frac{4}{3}\frac{4}{9} = 3 - \frac{16}{27} = \frac{81}{27} - \frac{16}{27} = \frac{65}{27}.$$



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AR models: linear prediction

• AR model estimation can also be seen as a **linear prediction** problem, where we are trying to predict x(n) based on p past samples:

$$\widehat{x}(n) = -\sum_{l=1}^{p} a_l x(n-l) = -\boldsymbol{a}^T \boldsymbol{x}$$

with:
$$a = [a_1 \ ... \ a_p]^T \ x = [x(n-1) \ ... \ x(n-p)]^T$$

 What do we consider a good prediction? A common choice is based on the mean squared error (MSE):

$$MSE = E\left\{\left(x(n) - \hat{x}(n)\right)^2\right\}$$

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AR models: linear prediction

Expanding this expression:

$$MSE = E\left\{\left(x(n) - \hat{x}(n)\right)^{2}\right\}$$

$$= E\left\{(x(n) + \boldsymbol{a}^{T}\boldsymbol{x})^{2}\right\}$$

$$= E\left\{x(n)^{2} + 2x(n)\boldsymbol{a}^{T}\boldsymbol{x} + \boldsymbol{a}^{T}\boldsymbol{x}\boldsymbol{x}^{T}\boldsymbol{a}\right\}$$

$$= E\left\{x(n)^{2}\right\} + 2\boldsymbol{a}^{T}E\left\{x(n)\boldsymbol{x}\right\} + \boldsymbol{a}^{T}E\left\{\boldsymbol{x}\boldsymbol{x}^{T}\right\}\boldsymbol{a}$$

$$= E\left\{x(n)^{2}\right\} + 2\boldsymbol{a}^{T}\boldsymbol{b} + \boldsymbol{a}^{T}\boldsymbol{a}$$
(see definitions next)



$$\boldsymbol{b} = E\{x(n)\boldsymbol{x}\}$$

$$= E \left\{ x(n) \begin{bmatrix} x(n-1) \\ \vdots \\ x(n-p) \end{bmatrix} \right\}$$

$$= \begin{bmatrix} E\{x(n)x(n-1)\} \\ \vdots \\ E\{x(n)x(n-p)\} \end{bmatrix}$$

$$= \begin{bmatrix} R_{xx}(1) \\ \vdots \\ R_{xx}(p) \end{bmatrix}$$

$$R = E\{xx^T\}$$

$$= E\left\{x(n)\begin{bmatrix}x(n-1)\\ \vdots\\ x(n-p)\end{bmatrix}\right\} \qquad = E\left\{\begin{bmatrix}x(n-1)^2 & \cdots & x(n-1)x(n-p)\\ \vdots & \ddots & \vdots\\ x(n-p)x(n-1) & \cdots & x(n-p)^2\end{bmatrix}\right\}$$

$$= \begin{bmatrix} R_{\chi\chi}(0) & R_{\chi\chi}(1) & \cdots & R_{\chi\chi}(p-1) \\ R_{\chi\chi}(1) & R_{\chi\chi}(0) & \cdots & R_{\chi\chi}(p-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_{\chi\chi}(p-1) & R_{\chi\chi}(p-2) & \cdots & R_{\chi\chi}(0) \end{bmatrix}^{29}$$

(which is the autocorrelation matrix)

$$MSE = E\{x(n)^2\} + 2\boldsymbol{a}^T\boldsymbol{b} + \boldsymbol{a}^TR\boldsymbol{a}$$

- It can be shown that R is a semi-positive definite matrix (i.e. $a^T R a \ge 0$), and typically it is positive definite;
- It can also be shown that this implies that the MSE has a global minimum;
- The global minimum can then be obtained by zero-ing the **gradient** of MSE with respect to a.





• The gradient is:
$$\frac{\partial MSE}{\partial a} = 2Ra + 2b$$

- Equaling to zero: Ra = -b
- This result matches the Yule-Walker equations. Thus, seeking the MSEoptimal predictor is the same as estimating the AR model via Y.W.
- Also, it can be shown that the prediction error $e(n) = x(n) \hat{x}(n)$ obeys $E\{e(n)x\} = 0$. Intuitively, we can interpret that the prediction error at sample n does not share any information with past samples of x.



Linear prediction with a real, finite sample

- Say we have a sample x(n) with length N ($0 \le n < N-1$);
- From MSE, we derive an appropriate criterion:

$$MSE = E\left\{ \left(x(n) - \hat{x}(n) \right)^2 \right\}$$



$$E = \sum_{k=k_0}^{k_1} (x(k) - \hat{x}(k))^2 = \sum_{k=k_0}^{k_1} \left(x(k) + \sum_{l=1}^p a_l x(k-l) \right)^2$$

where k_0 and k_1 are chosen indices (more on that later)



This expression is again quadratic with respect to the coefficients. We estimate the global minimum by zero-ing the gradient:

$$\frac{\partial E}{\partial a_j} = 2 \sum_{k=k_0}^{k_1} x(k-j) \left(x(k) + \sum_{l=1}^{p} a_l x(k-l) \right) = 0 \quad \text{for } j = 1, ..., p$$

This yields the equation set:

$$\sum_{l=1}^{p} a_l \phi(j,l) = -\phi(j,0) \quad \text{for } j = 1, \dots, p$$
 with:
$$\phi(j,l) = \sum_{l=1}^{k_1} x(k-j)x(k-l)$$

And, the system can be written in matrix form:

$$\Phi \boldsymbol{a} = -\boldsymbol{\psi}$$
$$[\Phi]_{j,l} = \phi(j,l)$$
$$[\psi]_j = \phi(j,0)$$



The choice of bounds k_0 , k_1

Two main approaches exist:

• "Autocorrelation":
$$k_0 = 0$$
, $k_1 = N + p - 1$

• "Covariance":
$$k_0 = p, k_1 = N - 1$$

(the designations are historical)

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Autocorrelation approach $(k_0 = 0, k_1 = N + p - 1)$

- It is equivalent to considering the sample as part of a signal s of infinite duration: ... 0 0 0 0 x_0 x_1 ... x_{N-2} x_{N-1} 0 0 0 ...
- We can in fact consider k_0 and k_1 as expanding to $\pm \infty$, since the prediction error is zero for k < 0 and k > N 1 + p. So we can write:

$$E = \sum_{k=-\infty}^{+\infty} \left(x(k) + \sum_{l=1}^{p} a_l x(k-l) \right)^2 \qquad \phi(j,l) = \sum_{k=-\infty}^{+\infty} x(k) x(k+j-l)$$

• Taking us back to Y.W.: $\sum_{l=1}^{p} a_l R_{ss}(j-l) = -R_{ss}(l)$





Advantages:

- In this case the matrix Φ is Hermitian Toeplitz, and one can use Levinson-Durbin to solve the system. Additionally, one gets a stable model.
- Fully equivalent to solving the Y.W. equations using the biased sample autocorrelation estimates.

Disadvantages:

Boundary effects: prediction error will be larger for k=0,...,p-1 (trying to predict signal samples using null samples), and for k=N,...,N+p-1 (predict null samples with signal samples), which will then affect the parameter estimation.



AR models: linear prediction

Covariance approach $(k_0 = p, k_1 = N - 1)$

- It includes in the estimation only the prediction errors that are "fully supported", i.e. errors in predicting signal samples using signal samples.
- We then use:

$$\phi(j, l) = \sum_{k=p}^{N-1} x(k - j)x(k - l)$$



Advantages:

- The matrix Φ is symmetric and positive-definite.
- The AR estimation is more accurate than with the autocorrelation approach.

Disadvantages:

- The matrix Φ is not Toeplitz, and Levinson-Durbin cannot be applied.
- The AR model is not guaranteed to be stable.







AR models: linear prediction

Matrix formulation

- Provides an equivalent way to pursue the solution to the MSE criterion.
- For example for the covariance method, we would write:

 $e = x - \hat{x}$

$$\begin{bmatrix} e(p) \\ e(p+1) \\ \vdots \\ e(N-1) \end{bmatrix} = \begin{bmatrix} x(p) \\ x(p+1) \\ \vdots \\ x(N-1) \end{bmatrix} - \begin{bmatrix} x(p-1) & x(p-2) & \cdots & x(0) \\ x(p) & x(p-1) & \cdots & x(1) \\ \vdots & \vdots & \ddots & \vdots \\ x(N-2) & x(N-3) & \cdots & x(N-p-1) \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \\ \vdots \\ a_p \end{bmatrix}$$

$$e \qquad C \qquad Xa$$

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AR models: linear prediction

• We then minimize e in the least-squares sense:

$$||e||^2 = (Xa + c)^T (Xa + c) = a^T X^T Xa + 2c^T Xa + c^T c$$

To minimize, we search for the solution that nulls the gradient:

$$\frac{\partial \|\boldsymbol{e}\|^2}{\partial \boldsymbol{a}} = 0 \iff 2X^T X \boldsymbol{a} + 2X^T \boldsymbol{c} = \boldsymbol{0}$$

$$\Leftrightarrow \mathbf{a} = -(X^T X)^{-1} X^T \mathbf{c}$$

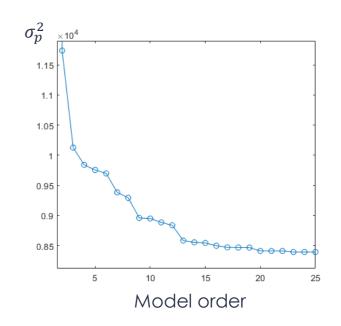
If X^TX is invertible.





How to choose the model order p?

- The larger the better? Not in practice:
 - Limited estimation accuracy of the autocorrelation due to limited sample size;
 - Computational resources are not infinite either.
- An empirical rule of thumb suggests that, with N samples, the order should not be larger than N/3.
- The rate of decrease of the excitation variance σ_p^2 with respect to p may help setting the stop point.







Selection criteria

Whiteness of the excitation term:

Once the model is estimated, one can compute the excitation term:

$$\varepsilon(n) = x(n) + \sum_{l=1}^{p} a_l x(n-l)$$

• ε is supposed to be a white noise process. A whiteness test can therefore be applied: if it fails the test, the order can be increased.





Selection criteria

Several statistical criteria have been proposed as well:

• Final prediction error (FPE):

$$FPE(p) = \sigma_p^2 \left(\frac{N+p+1}{N-p-1} \right)$$
Decreases Increases with Ωp with Ωp

Akaike information criterion (AIC): $AIC(p) = N \ln(\sigma_p^2) + 2(p+1)$



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Selection criteria

Several statistical criteria have been proposed as well:

• Minimum description length (MDL): $MDL(p) = N \ln(\sigma_p^2) + (p+1) \ln(N)$

Decreases Increases

with $\hat{\mathbf{p}}$ with $\hat{\mathbf{p}}$

- MDL is probably the best criterion, especially with short signals.
- For purposes of spectral estimation, it is recommended to choose larger orders than what is given by the criteria.



AR models: sinusoids

- A sinusoid is a rather "organized" signal
- Sinusoids can be represented by an AR(2) process with no excitation:

$$x(n) = \sin(2\pi f n) = 2\cos(2\pi f)x(n-1) - x(n-2)$$
Using:
$$\sin(a+b) + \sin(a-b) = 2\sin(a)\cos(b)$$
with $a = 2\pi f(n-1)$ and $b = 2\pi f$

The denominator of the transfer function $H(z) = \frac{1}{1 + \sum_{k=1}^{p} a(k)z^{-k}}$ is then:

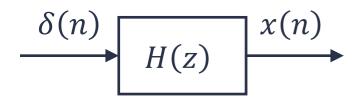
$$1 - 2\cos(2\pi f)z^{-1} + z^{-2}$$

Which is equal to: $(1 - e^{+j2\pi f}z^{-1})(1 - e^{-j2\pi f}z^{-1})$



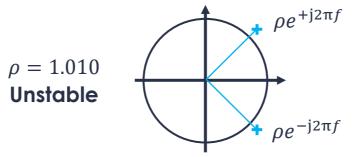


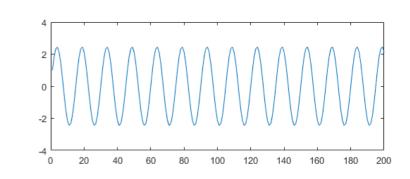
AR models: sinusoids

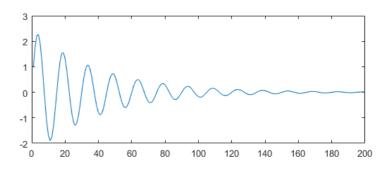


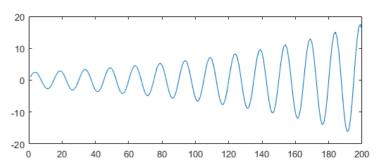
- As the poles get closer to the unit circle, the signal approaches a stationary sinusoid;
- It can be considered to become more "organized" in this sense;
- Analogous observations can be done for sums of sinusoids.

Neutrally stable $\rho = 0.975$ Stable $\rho = 0.975$ Stable















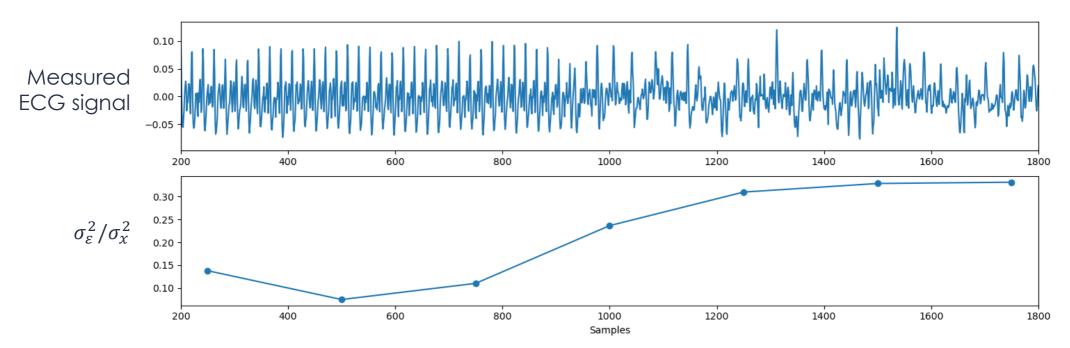
Signal structure/organization

$$x(n) = -\sum_{l=1}^{p} a_l x(n-l) + \varepsilon(n)$$

Variance σ_x^2 contains everything

Variance σ_{ε}^2 contains the random contribution

- Once the model is estimated, both σ_{x}^{2} and σ_{ε}^{2} are known;
- The ratio $\sigma_{\varepsilon}^2/\sigma_{x}^2$ can inform on how "disorganized" or "unstructured" the signal is;
- It also depends on how "noisy" the measurement is.









- M. H. Hayes, Statistical Digital Signal Processing and Modeling, Wiley, 1996.
- R. Shiavi, Introduction to Applied Statistical Signal Analysis, 2nd Ed., Academic Press 1999.
- S. M. Kay and S. L. Marple, Spectrum analysis A modern perspective, Proc. IEEE, vol. 69, no. 11, Nov. 1981, pp. 1380-1419.



