

# Lecture 19

Monday, April 4, 2022 8:29 AM

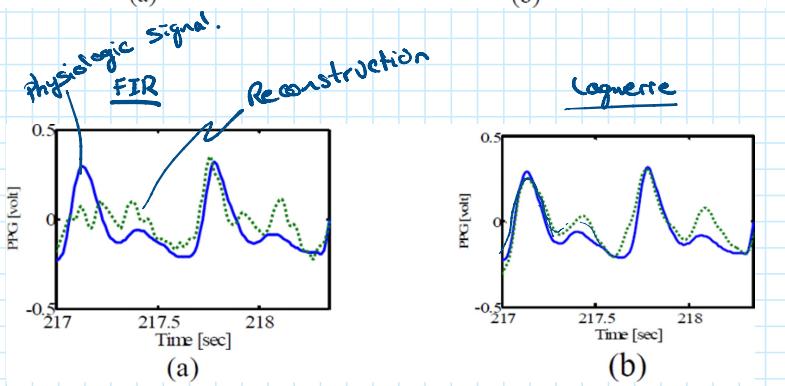
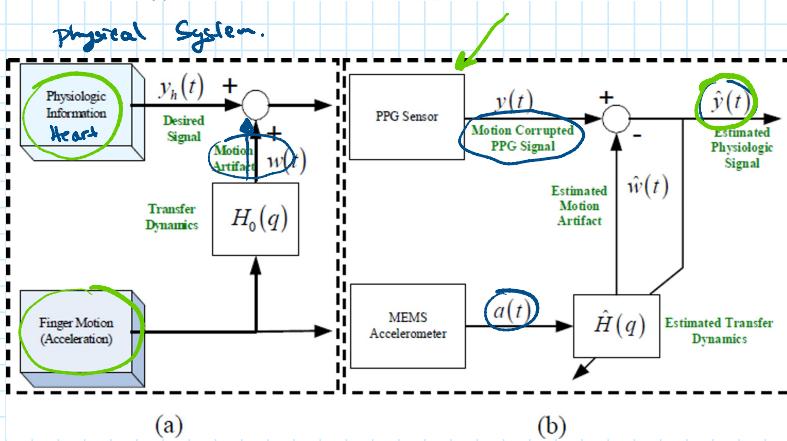
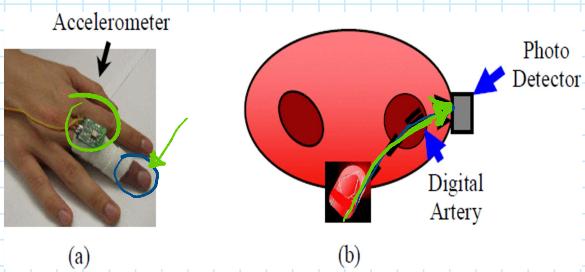
Admin

- (1) HW 64 - due Next Monday.
- (2) Term projects.

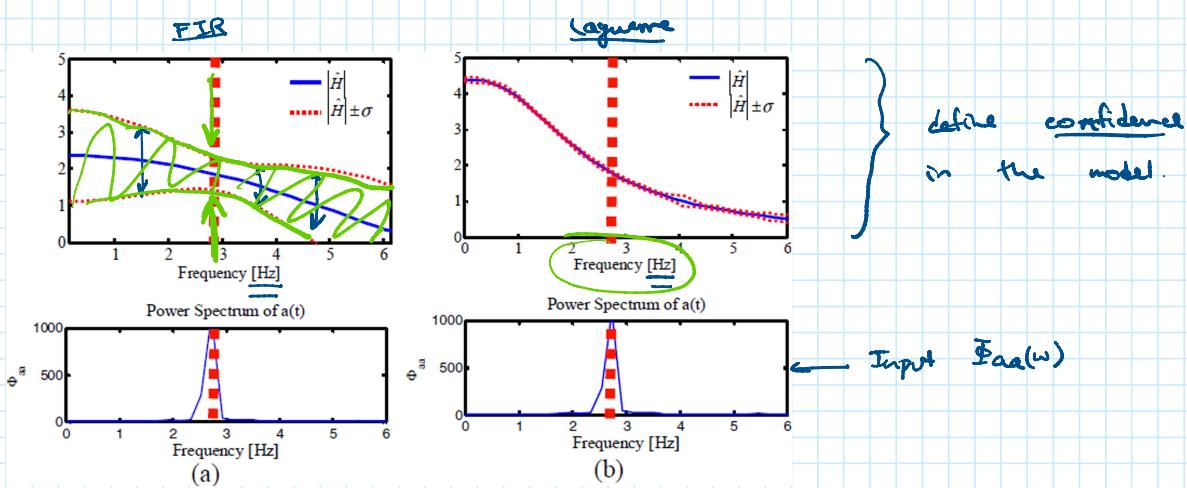
## System Identification Theory: Frequency Domain Analysis

- Suggested Reading: Ljung - chap.3, 4
- Question: what are we doing in this class?
  - Better understanding of a physical system
  - Quantitative parameterization. ← Assumed some model structure.

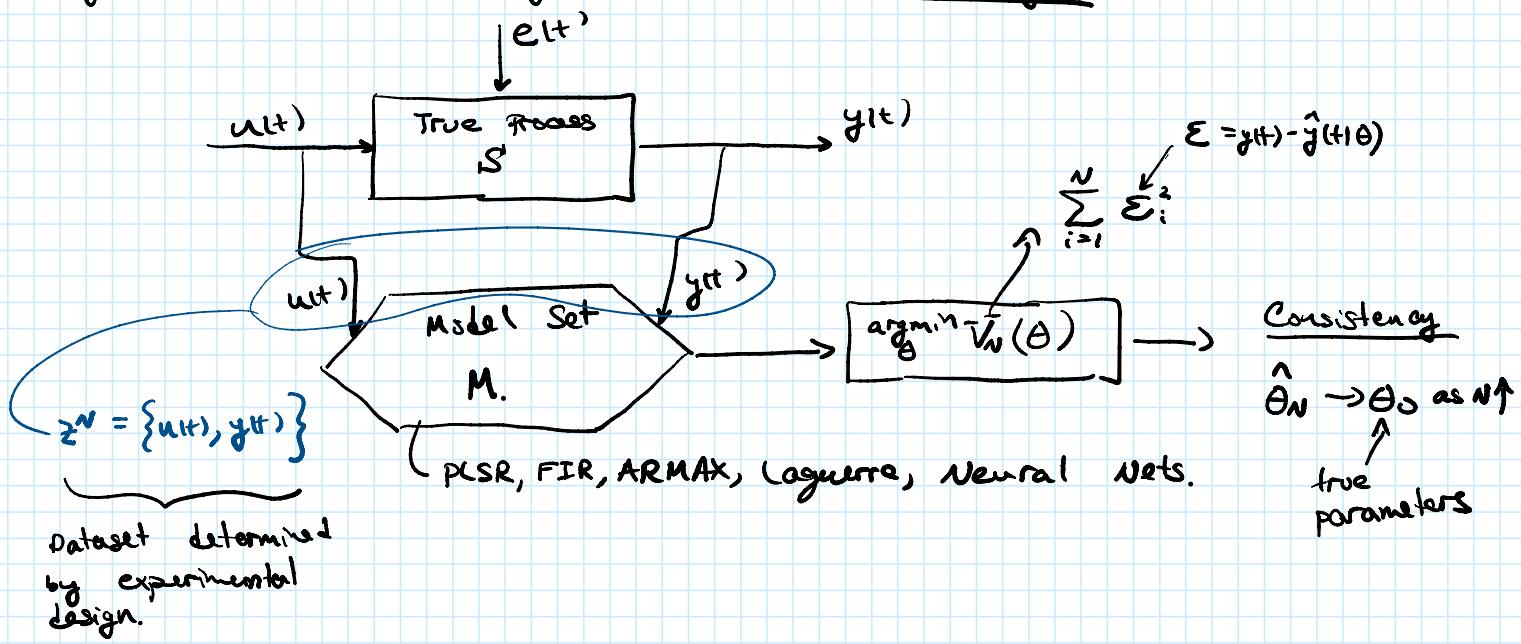
Example:



Laguerre is better than  
#FIR



### High-level view of System Identification Theory



### Key Questions

Q 1: Is a given dataset informative enough to determine a model from a given model set, M?

→ Does  $Z^N$  contain sufficient information to distinguish if one or two models in  $M$  is better?

Q 2: Is minimization of  $V_W(\theta)$  good enough to obtain true (undbiased) model?

→ what if true model is not in model set?

→ how is model fitting influenced by noise characteristics?

Q 3: How accurate is the model?

Q3: How accurate is the model?

- How much variance is expected?
- How much data is needed?

How do we design experiment

Key Results:

- Informative dataset of persistent excitation condition.
- Consistent (unbiased) estimate.
- Signal-to-noise ratio.
- Asymptotic variance.
- Input design: pseudo random binary signal.
- System order estimate: Model selection process.

Mathematical Tools

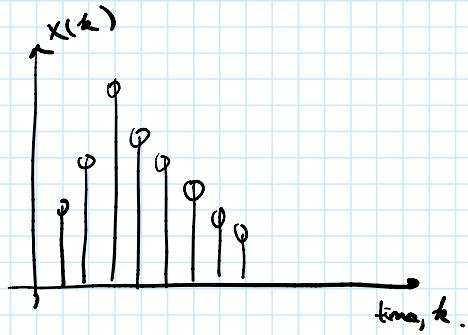
- Discrete Fourier Transform, spectral analysis
- Central limit theorems
- Random processes: stationary process & ergodicity.

Frequency domain analysis

- Discrete Fourier Transform of power spectrum.

Discrete Fourier Transform of a sampled data system:  $x(k)$

$$X(\omega) = \sum_{k=0}^{\infty} x(k) e^{-i\omega k}$$



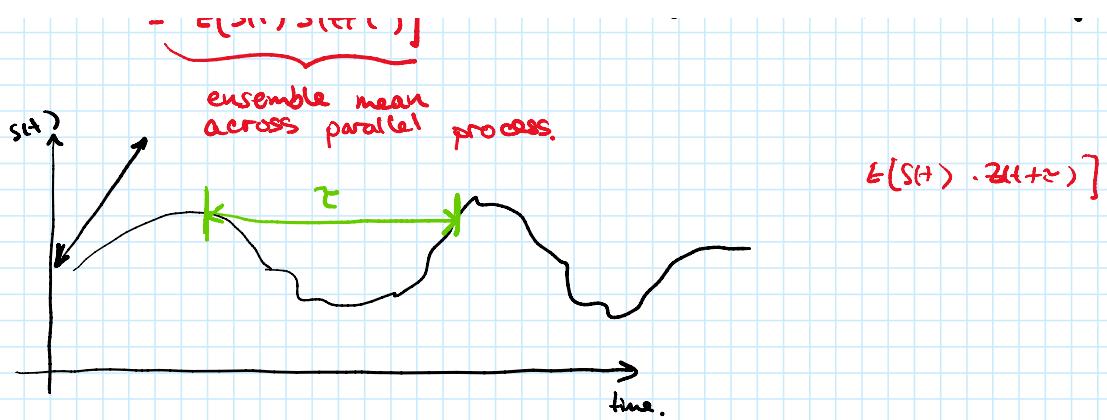
- Inverse transform:

$$x(k) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{i\omega k} d\omega$$

Power Spectrum

- Consider a deterministic, bounded  $\{s(t)\}$  for which

$$\begin{aligned} R_s(\tau) &\approx \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{t=1}^N (s(t) s(t+\tau)) \quad \leftarrow \text{covariance funcn.} \\ &= \underbrace{E[s(t) s(t+\tau)]}_{\text{ensemble mean}} \quad \leftarrow \text{if equal, the process is ergodic.} \end{aligned}$$



- The Fourier power spectrum of  $\{S(t)\}$  is defined as the auto covariance function,  $R_S(\tau)$ :

$$I_S(\omega) = \sum_{\tau=-\infty}^{\infty} R_S(\tau) e^{i\omega\tau}$$

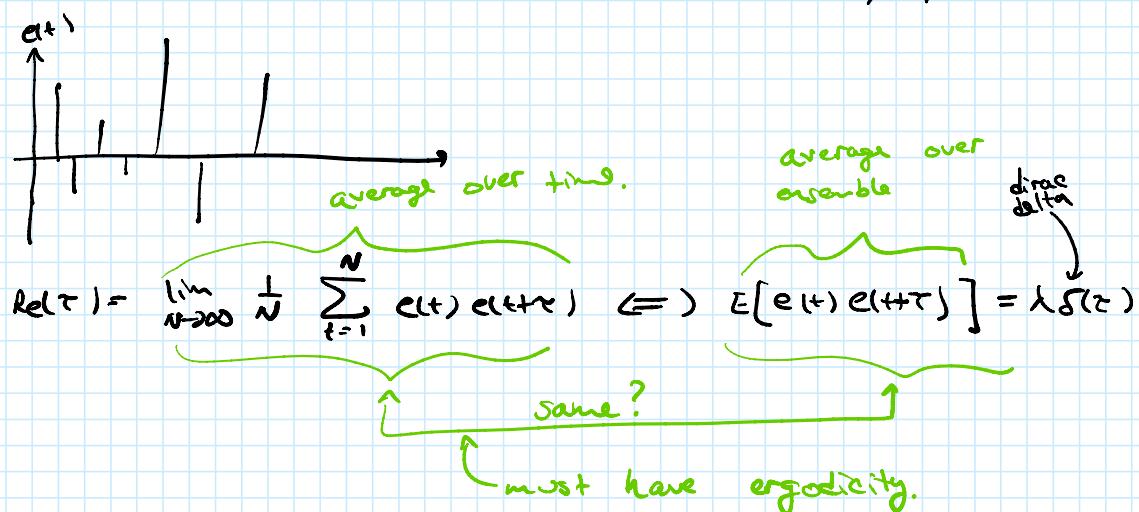
$\omega$  frequency

- Inverse

$$R_S(\tau) = \frac{1}{2\pi} \int_{-\pi}^{\pi} I_S(\omega) e^{i\omega\tau} d\omega$$

### White (Gaussian) Noise

- We define  $\{e(t)\}$  as a sequence of indep. random variables:
  - zero mean.
  - Covariance:  $E[e(t)e(t+\tau)] = \begin{cases} \lambda, \tau=0 \\ 0, \tau \neq 0 \end{cases}$

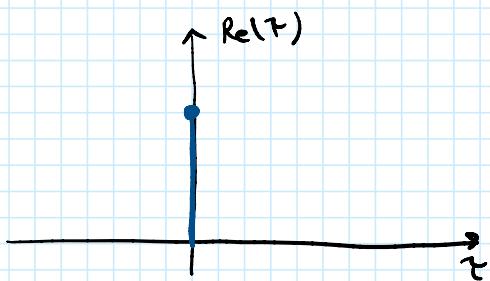


- The left side is a time average
- The right side is an ensemble
- If they are same, process is called ergodic.
- We assume ergodicity for most processes.

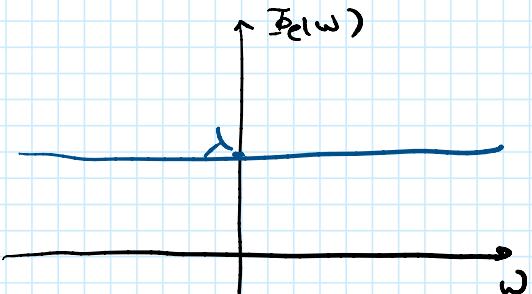
- We assume ergodicity for most processes.
- For the above, power spectrum is given by:

$$\mathbb{E}_e(\omega) = \sum_{\tau=-\infty}^{\infty} R_e(\tau) e^{i\tau\omega} = \sum_{\tau=-\infty}^{\infty} \lambda \delta(\tau) e^{i\tau\omega} = \lambda = \mathbb{E}_e(\omega)$$

Auto-covariance



Power Spectrum



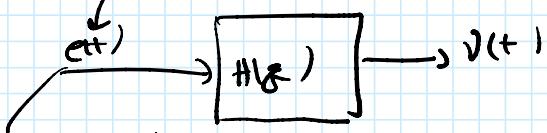
- A correlated (colored) signal can be created by white noise going through a dynamic process.

Theorem:

Let  $H(z)$  be a transfer function of a BIBO process with a white noise input  $e(t)$  w/ variance  $\lambda$ :

stationary.

→ bounded input - bounded output.



$$E[e(t)e(t+\tau)] = \begin{cases} \lambda, & \tau=0 \\ 0, & \tau \neq 0 \end{cases}$$

$$v(t) = H(z) e(t)$$

say  
 $H(z) = b_0 z^{-1} + b_1 z^{-2} + \dots$

The power spectrum of  $v(t)$  is given by:

$$\mathbb{E}_v(\omega) = \lambda |H(e^{i\omega})|^2$$

Proof.

$$\text{Proof } v(k) = \sum_{l=0}^{\infty} h(l) e(k-l)$$

$$= v(k-\tau) = \sum_{m=0}^{\infty} h(m) e(k-\tau-m)$$

combining these two

$$(15) \quad R_v(\tau) = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N v(k)v(k-\tau)$$

$$= \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N \sum_{l=0}^{\infty} h(l)e(k-l) \sum_{m=0}^{\infty} h(m)e(k-\tau-m)$$

$$R_v(\tau) = \sum_{l=0}^{\infty} \sum_{m=0}^{\infty} h(l)h(m) \underbrace{\lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N e(k-l)e(k-\tau-m)}_{\lambda \delta(l-\tau-m)}$$

$$\therefore R_v(\tau) = \lambda \sum_{l=\max(0,\tau)}^{\infty} h(l)h(l-\tau) \quad : \quad h(l)=0 \quad \text{for } l < 0$$

$$\lambda \delta(l-\tau-m) = \begin{cases} \lambda & l = \tau + m \\ 0 & \text{otherwise} \end{cases}$$

The power spectrum is then given by

$$(16) \quad \Phi_v(\omega) = \sum_{\tau=-\infty}^{\infty} R_v(\tau) e^{-i\omega\tau}$$

$$= \lambda \sum_{\tau=-\infty}^{\infty} \sum_{l=\max(0,\tau)}^{\infty} h(l) e^{-i\omega l} h(l-\tau) e^{i\omega(l-\tau)}$$

$$\text{Replacing } l-\tau \text{ by } s$$

$$= \lambda \sum_{l=0}^{\infty} h(l) e^{-i\omega l} \sum_{s=0}^{\infty} h(s) e^{i\omega s}$$

$$= \lambda H(e^{i\omega}) \cdot H(e^{-i\omega}) = \lambda |H(e^{i\omega})|^2$$