

# < Differential Measurement >

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## Objective

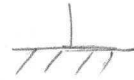
- Understand grounds and common-mode noise
- Differential amplifier
- Instrumentation amplifier

Grounds : a particular "conductor" in electric circuits.

## Symbols



Earth ground  
Mains ground



Chassis ground



PCB ground  
IC ground

Grounds can be categorized into two groups.

p.124

### ① Safety grounds (Ott. p.112)

- Not conducting currents during normal operations  
e.g.) Earth ground, Mains ground
- Role : protect people from electric shock.

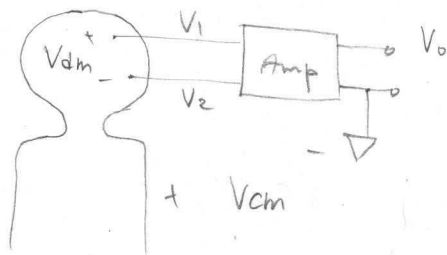
### ② Signal / power grounds (Ott. p.129)

- Conducting currents during normal operations  
e.g.) PCB ground plane
- Role :  $\begin{cases} \text{Voltage} : \text{common reference potential for node voltages.} \\ \text{Current} : \text{common low-impedance path for return currents.} \end{cases}$

## < Common-mode Voltage >

- When we connect two electric systems together, the potential difference between the two grounds is not always zero.

Example Bio-potential measurement



In general,  $V_{cm} \neq 0$

- Let's design a differential-input single-ended output amplifier

$V_1$ ,  $V_2$ , and  $V_o$  are referenced to  $\downarrow$

$$\begin{cases} \text{Common-mode voltage} : & V_{cm} \triangleq \frac{V_1 + V_2}{2} \\ \text{Differential voltage} : & V_{dm} \triangleq V_1 - V_2 \end{cases}$$

- In many cases,  $V_{cm}$  is unknown and unwanted, (e.g., EMI, ground noise, etc.)

and we want to reject it.

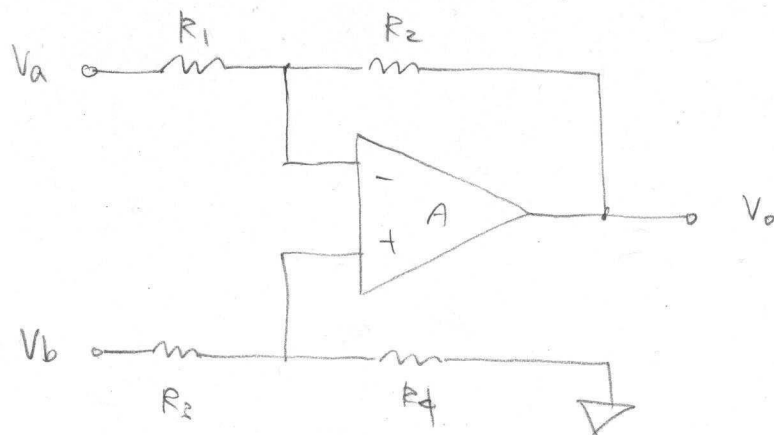
On the other hand,  $V_{dm}$  contains useful information (measurement) and we want to amplify it.

- Common-mode Rejection Ratio (CMRR) is the figure of merit

$$CMRR \triangleq \frac{A_{dm}}{A_{cm}}$$

# Differential (difference) Amplifier.

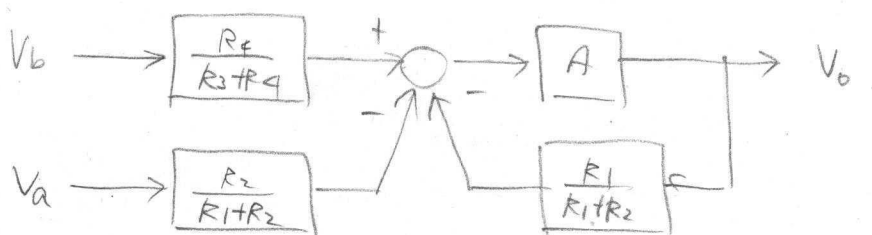
(Aptorger p. 149.)



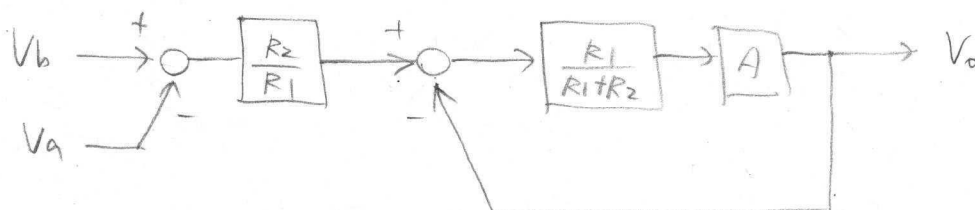
$$V_+ = \frac{R_4}{R_3 + R_4} \cdot V_b$$

$$V_o = A(V_+ - V_-)$$

$$V_- = \frac{R_2}{R_1 + R_2} V_a + \frac{R_1}{R_1 + R_2} V_o$$



$$\text{If } \frac{R_4}{R_3} = \frac{R_2}{R_1} \Rightarrow \frac{R_4}{R_3 + R_4} = \frac{R_2}{R_1 + R_2}$$



Below the bandwidth.

$$V_o = \frac{R_2}{R_1} (V_b - V_a)$$

$$\text{Let } V_{cm} = \frac{1}{2} (V_b + V_a) \Rightarrow V_o = \underbrace{\frac{R_2}{R_1}}_{A_{dm}} V_{dm} + \underbrace{0}_{A_{cm}} \cdot V_{cm}$$

$$V_{dm} = V_b - V_a$$

$$CMRR \approx \frac{A_{dm}}{A_{cm}} \rightarrow \infty$$

In practice,  $A_{cm} \neq \infty$  and therefore CMRR is finite.

CMRR is mainly limited by the resistor tolerance.  $\left( CMRR \geq \frac{R_2/R_1 + 1}{4\epsilon} \right)$

There exist monolithic diff-amps with laser-trimmed resistors.

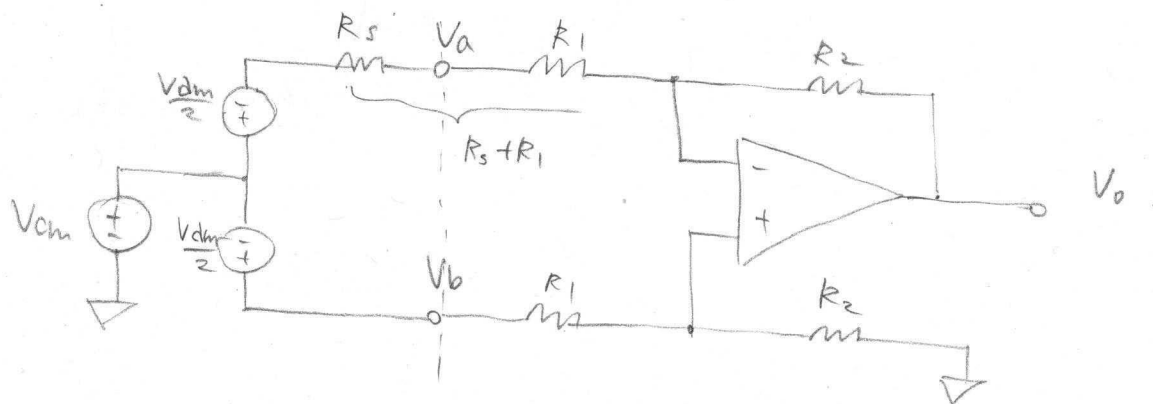
Particularly useful when  $\left\{ \begin{array}{l} \text{the source impedance is small, and} \\ \text{the common-mode voltage is large.} \end{array} \right.$

(  $V_{cm} > V_s$  possible. e.g. AC motor current measurement )

Limitations

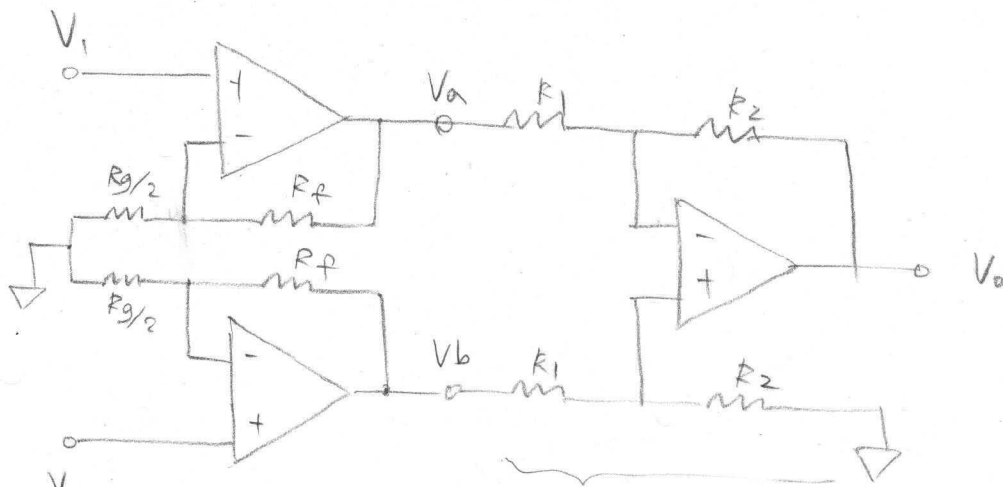
- Finite input impedance : even if the resistors are matched well,

the diff amp cannot reject  $V_{cm}$  well if the source impedance is not balanced.



# < Buffered Diff-amp >

Let's place a buffer stage to increase the input impedance



$$A_{cm} \neq 0, \quad A_{dm} \approx \frac{R_2}{R_1}$$

$$V_a = \frac{R_f + R_{g/2}}{R_{g/2}} V_1$$

$$V_b = \frac{R_f + R_{g/2}}{R_{g/2}} V_2$$

$$\therefore V_o = \underbrace{A_{dm} \left( \frac{R_f + R_{g/2}}{R_{g/2}} \right)}_{A_{dm}'} (V_2 - V_1) + \underbrace{A_{cm} \left( \frac{R_f + R_{g/2}}{R_{g/2}} \right)}_{A_{cm}'} \left( \frac{V_2 + V_1}{2} \right)$$

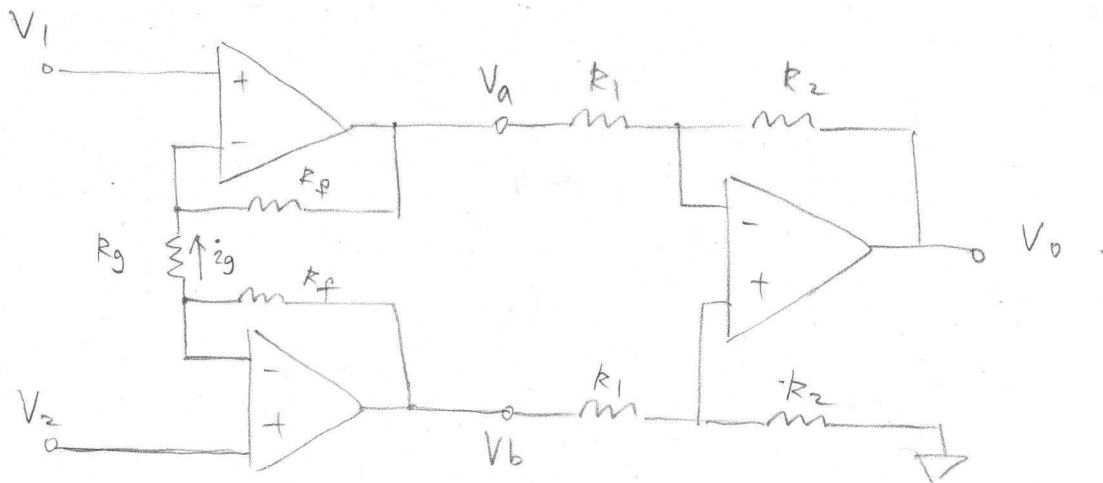
$$CMRR' = \frac{A_{dm}'}{A_{cm}'} = \frac{A_{dm}}{A_{cm}}$$

No improvement in CMRR

$$A_{dm}' = A_{dm} \left( \frac{2R_f + R_g}{R_g} \right) \quad \therefore \text{can be increased by decreasing } R_g \text{ (gain resistor)}$$

# < Instrumentation Amplifier (In-amp) >

Let the gain resistor  $R_g$  float!



$$i_g = \frac{V_2 - V_1}{R_g}$$

$$V_a = V_1 - R_f i_g = V_1 - \frac{R_f}{R_g} (V_2 - V_1) = \left( \frac{R_g + R_f}{R_g} \right) V_1 - \frac{R_f}{R_g} V_2$$

$$V_b = V_2 + R_f i_g = V_2 + \frac{R_f}{R_g} (V_2 - V_1) = -\frac{R_f}{R_g} V_1 + \left( \frac{R_g + R_f}{R_g} \right) V_2$$

$$\Rightarrow V_0 = A_{dm} (V_b - V_a) + A_{cm} \left( \frac{V_b + V_a}{2} \right)$$

$$= \underbrace{A_{dm} \left( \frac{R_g + 2R_f}{R_g} \right)}_{A_{dm}'} (V_2 - V_1) + \underbrace{A_{cm} \left( \frac{R_g}{R_g} \right)}_{A_{cm}'} \left( \frac{V_2 + V_1}{2} \right)$$

$$CMRR' = \frac{A_{dm}'}{A_{cm}'} = \frac{A_{dm}}{A_{cm}} \left( \frac{R_g + 2R_f}{R_g} \right) \quad \text{Improved by } \frac{R_g + 2R_f}{R_g}$$

This is because the buffer stage only amplifies  $V_{dm}$  and doesn't amplify  $V_{cm}$ .

- There exist monolithic in-amps with laser-trimmed resistors ( $R_1$ ,  $R_2$ , and  $R_f$ ). Typically  $R_1 = R_2$ .
- One external resistor  $R_g$  is needed for gain tuning.
- As opposed to diff-amps,  $V_{cm}$  cannot exceed the tail voltages.
- Extremely useful building blocks for measurement & instrumentation.