

MECH468 : Modern Control Engineering

MECH509 : Controls

L27 : Discrete-time LQR

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Zoom lecture to be recorded and posted on Canvas



Course plan

Topics	CT	DT
Modeling	✓	✓
Stability	✓	✓
Controllability/observability	✓	✓
Realization	✓	✓
State feedback/observer	✓	✓
LQR/Kalman filter		→



Review & today's topic

- **Continuous-time** LQR optimal control
 - State feedback
 - Finite-horizon case
 - Time-varying gain
 - Matrix Riccati equation
 - Infinite-horizon case
 - Constant gain
 - Algebraic Riccati Equation (ARE)
- **Discrete-time** LQR optimal control

DT finite-horizon LQR optimal control

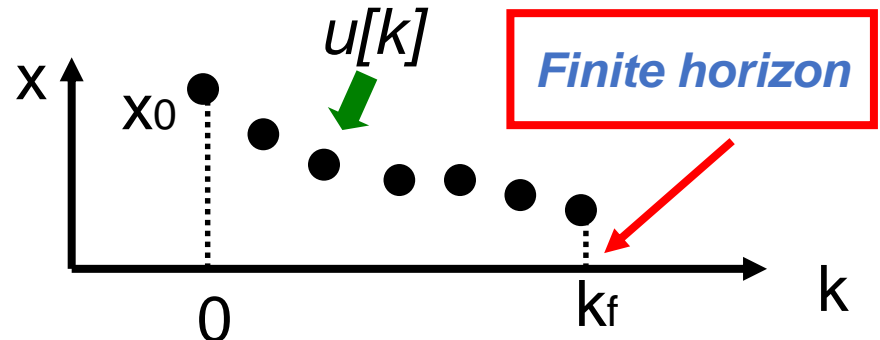
- Problem $\min_{u[\cdot]} J(u[\cdot])$ subj. to $\begin{cases} x[k+1] = Ax[k] + Bu[k] \\ x[0] = x_0 \text{ (given)} \end{cases}$

- J : **Quadratic** performance index (cost function)

$$J(u[\cdot]) := \sum_{k=0}^{k_f-1} \underbrace{[x^T[k]Qx[k]]}_{\text{For small state}} + \underbrace{[u^T[k]Ru[k]]}_{\text{For small input}} + \underbrace{[x^T[k_f]Sx[k_f]]}_{\text{For small final state}}$$

Design parameters

$$Q \geq 0, R > 0, S \geq 0$$



LQR optimal control law

- LQR optimal control is obtained as a **state feedback**

$$u[k] = - \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A x[k] \quad \boxed{\text{Linear}}$$

- $P[k]$: positive semidefinite solution to a **matrix Riccati difference equation** (A dual equation will appear in Kalman filter.)

$$\begin{cases} P[k] &= A^T P[k+1] A + Q \\ &\quad - A^T P[k+1] B \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A \\ P[k_f] &= S \end{cases}$$

- Optimal performance index $J(u) = x_0^T P[0] x_0$

(Proof given in the Appendix)



Example

$$\min_{u[\cdot]} \sum_{k=0}^1 (2x[k]^2 + u[k]^2) + x[2]^2$$

$$\text{subj. to } \begin{cases} x[k+1] = x[k] + u[k] \\ x(0) = x_0 \text{ (given)} \end{cases}$$



$$\begin{aligned} A &= 1, B = 1, Q = 2 \\ R &= 1, S = 1 \end{aligned}$$

$$\rightarrow \begin{cases} P[2] = 1 \\ P[1] = P[2] + 2 - \frac{P[2]^2}{1+P[2]} = \frac{5}{2} \\ P[0] = P[1] + 2 - \frac{P[1]^2}{1+P[1]} = \frac{17}{14} \end{cases}$$

*Compute these
off-line!*

$$\rightarrow \begin{cases} u[0] = -\frac{P[1]}{1+P[1]}x_0 \\ u[1] = -\frac{P[2]}{1+P[2]}\underbrace{(x[0] + u[0])}_{x[1]} \\ J(u[\cdot]) = \frac{17}{14}x_0^2 \end{cases}$$

DT infinite-horizon LQR optimal control

- Problem $\min_{u[\cdot]} J(u[\cdot])$ subj. to $\begin{cases} x[k+1] = Ax[k] + Bu[k] \\ x[0] = x_0 \text{ (given)} \end{cases}$

- J : **Quadratic** performance index (cost function)

$$J(u[\cdot]) := \sum_{k=0}^{\infty} \left[\underbrace{x^T[k]Qx[k]}_{\text{For small state}} + \underbrace{u^T[k]Ru[k]}_{\text{For small input}} \right]$$

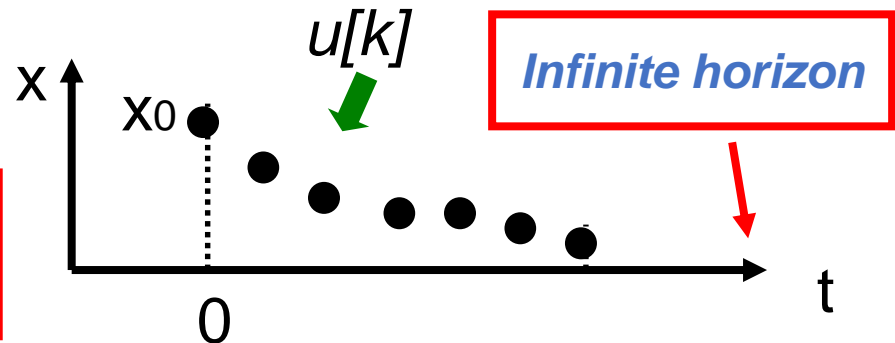
For small state

For small input

Design parameters

$$Q \geq 0, R > 0$$

**Assumptions: (A,B) controllable
& (A,Q) observable**



LQR optimal control law

- LQR optimal control is obtained as a **state feedback**

$$u[k] = - \left[R + B^T P B \right]^{-1} B^T P A x[k] \quad \boxed{\text{Linear}}$$

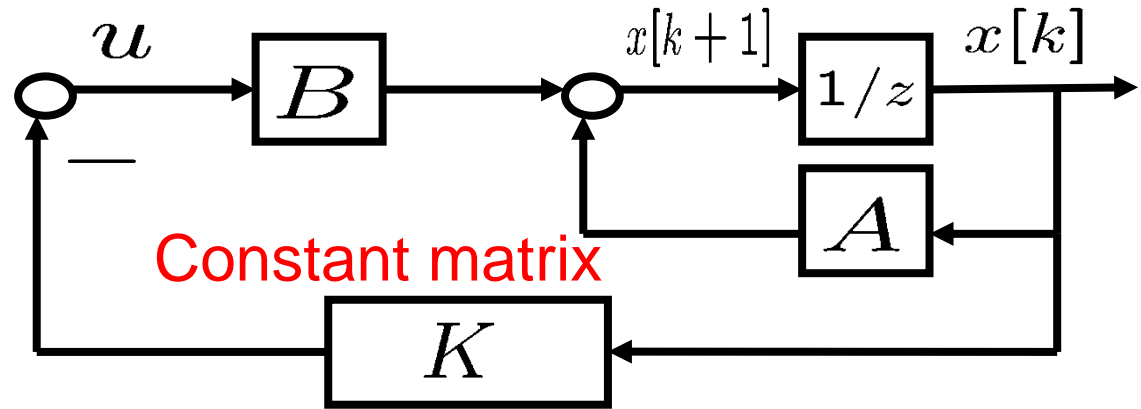
- P : unique positive definite solution to a **discrete algebraic Riccati equation (DARE)**

$$A^T P A - P + Q - A^T P B \left[R + B^T P B \right]^{-1} B^T P A = 0$$

- CL system is stable, i.e., $|\text{eig}(A-BK)| < 1$
- Optimal performance index $J(u) = x_0^T P x_0$

LQR optimal control law (cont'd)

- Block diagram



$$K = \left[R + B^T P B \right]^{-1} B^T P A$$

- DARE is obtained by setting $P[k]=P$ in matrix Riccati difference equation for finite-horizon LQR problem.

$$P[k] = A^T \left(P[k+1] A + Q - A^T P[k+1] B \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A \right)$$

How to solve DARE

$$A^T P A - P + Q - A^T P B [R + B^T P B]^{-1} B^T P A = 0$$

1. Numerically in Matlab.

$$[P, \Lambda_-, K] = \text{dare}(A, B, Q, R)$$

Closed-loop eigenvalues

Controller gain

2. Brute force (Next slide)
3. Method with Hamiltonian matrix (not covered)

Note: The **uniqueness** of the positive definite solution to DARE is guaranteed by the assumptions “ (A, B) is controllable and (A, Q) is observable.”



Example of DT LQR

$$\min_{u[\cdot]} \sum_{k=0}^{\infty} \left[\underbrace{2x[k]^2}_{Q=2} + \underbrace{u[k]^2}_{R=1} \right] \quad \text{subj. to } x[k+1] = x[k] + u[k]$$

- DARE $P = A^T P A + Q - A^T P B (R + B^T P B)^{-1} B^T P A \quad \Rightarrow \quad P = 1 + \sqrt{3}$
- Control gain $K = (R + B^T P B)^{-1} B^T P A = \frac{1 + \sqrt{3}}{2 + \sqrt{3}}$
- Closed-loop A-matrix $A_{cl} = \frac{1}{2 + \sqrt{3}}$

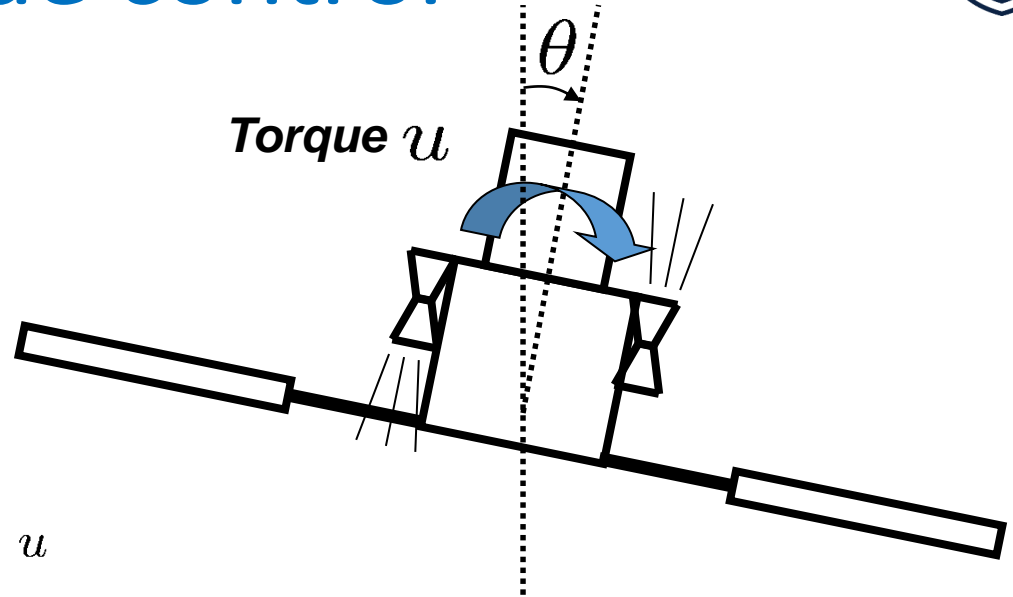
Satellite attitude control

- After normalization,

$$\ddot{\theta} = u$$

- SS model $x := [\theta, \dot{\theta}]^T$

$$\begin{cases} \dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \\ y = \begin{bmatrix} 1 & 0 \end{bmatrix} x \end{cases}$$



- Requirements

- Small θ
- Small u

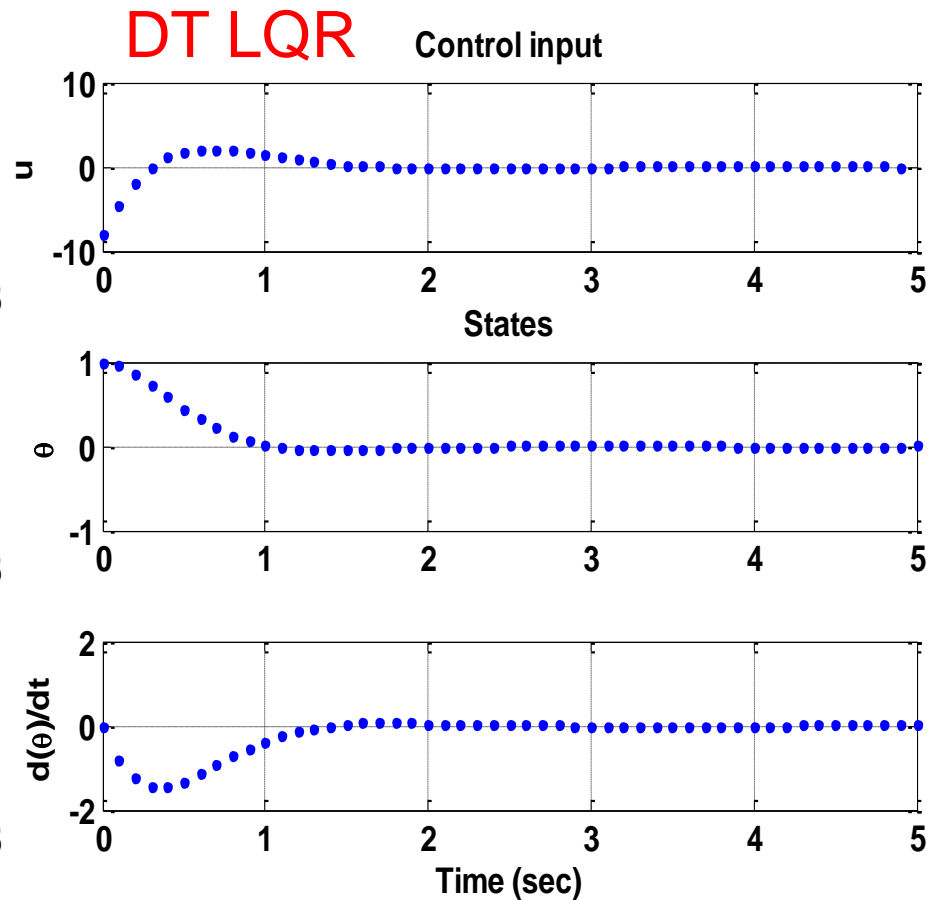
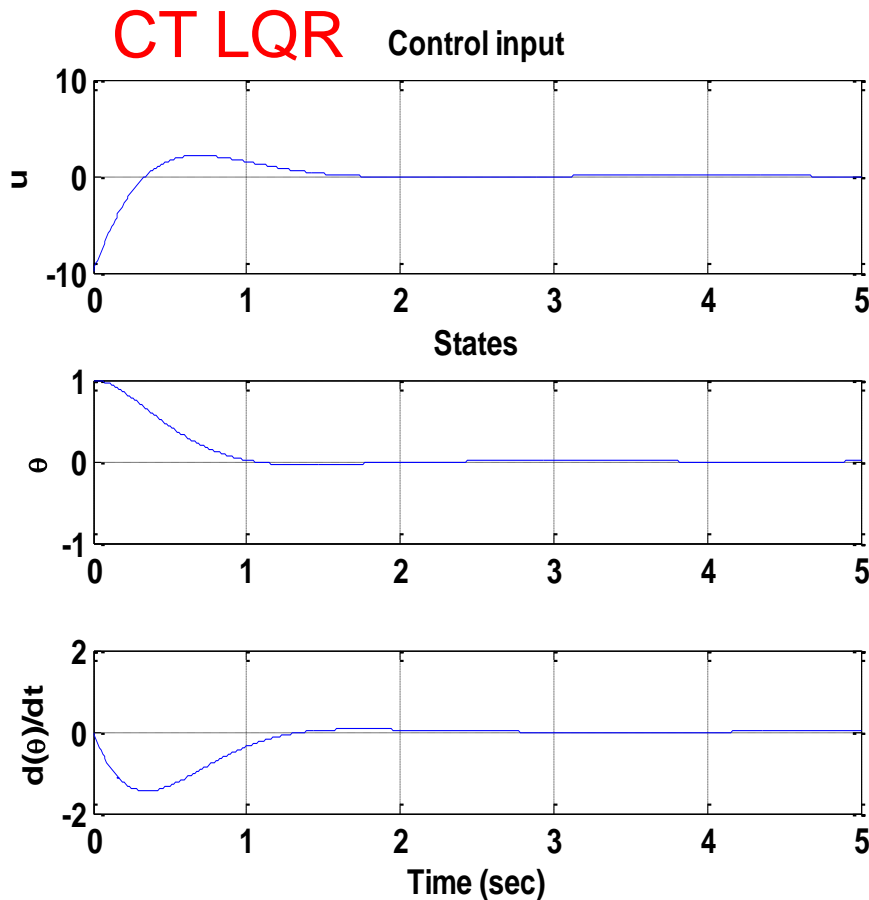


Design LQR controller for the discretized system.

$$A_d = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix}, \quad B_d = \begin{bmatrix} T^2/2 \\ T \end{bmatrix}$$

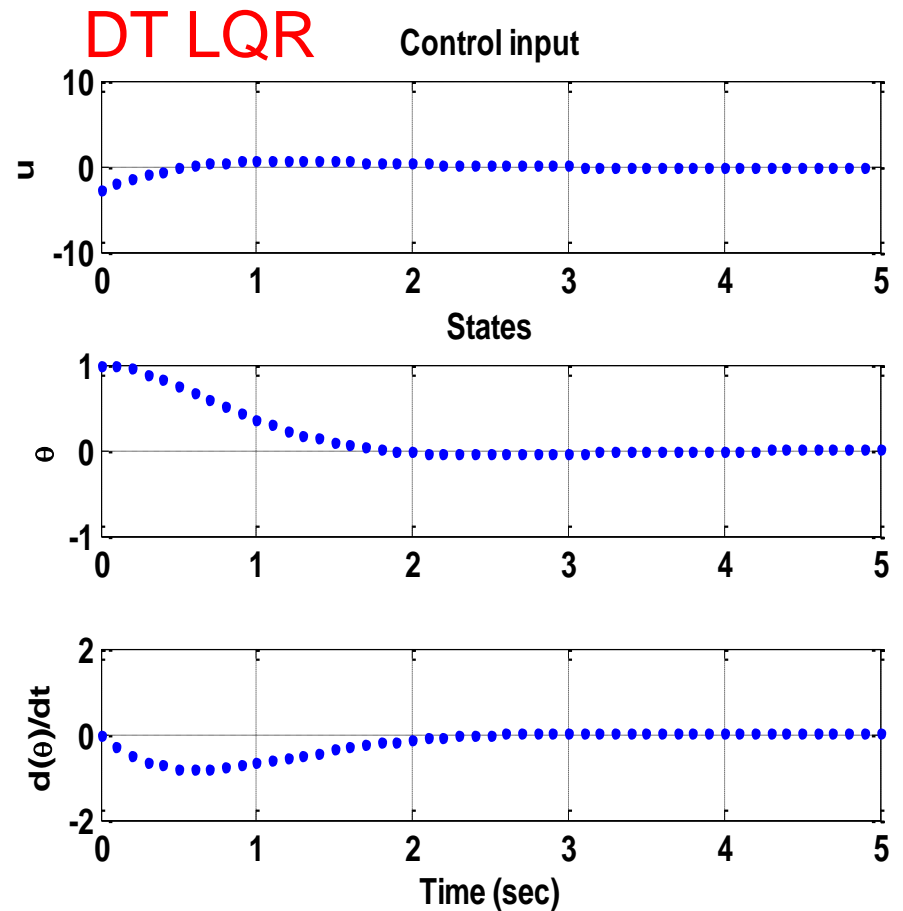
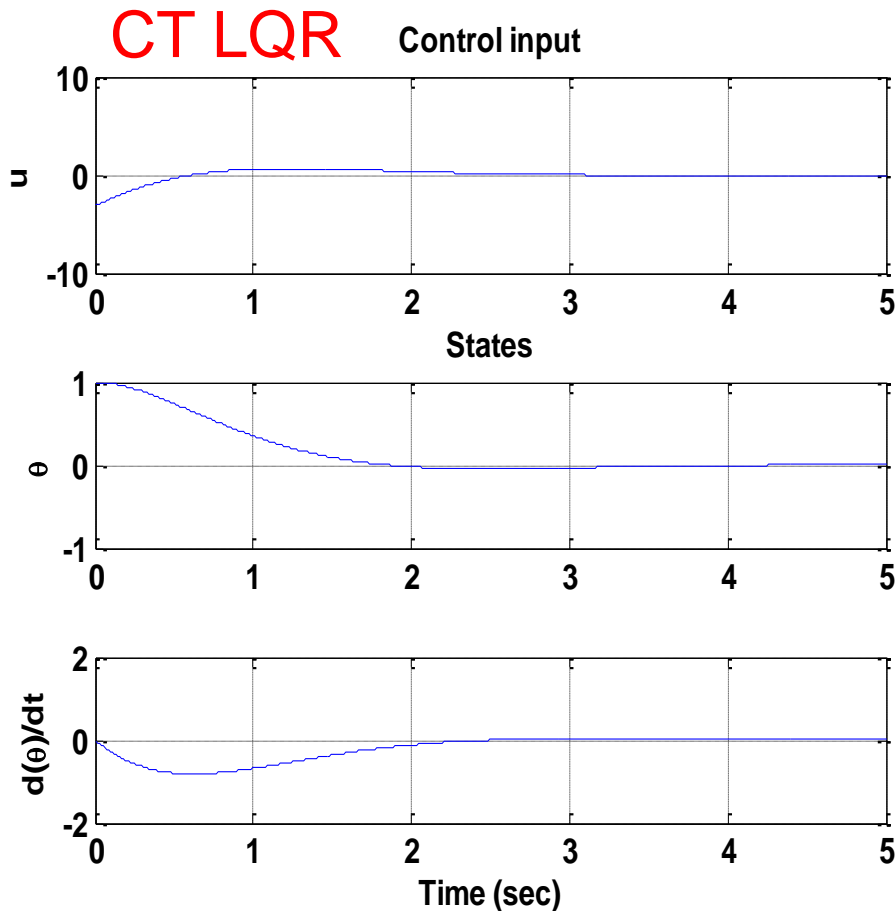
Satellite attitude control

(Finite-horizon, $R=0.01$, $T=0.1$ sec)



Satellite attitude control

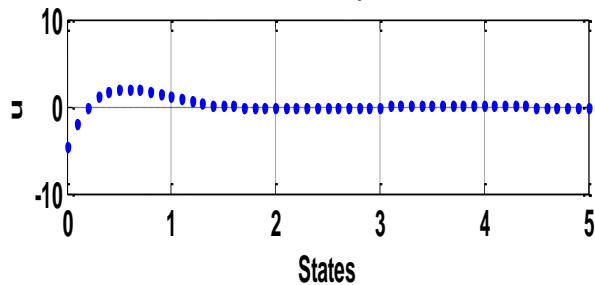
(Finite-horizon, $R=0.1$, $T=0.1$ sec)



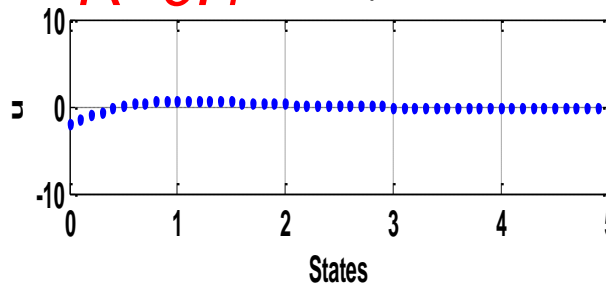


Satellite attitude control (Infinite-horizon, $T=0.1$ sec)

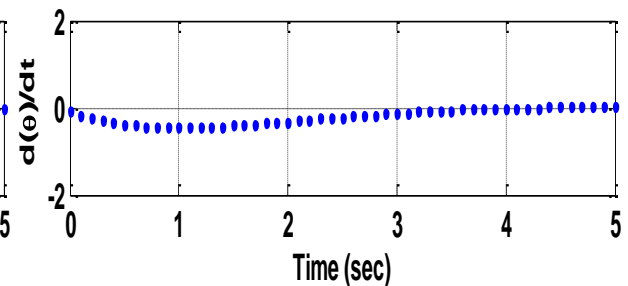
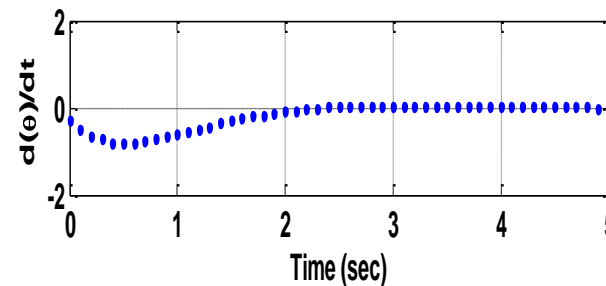
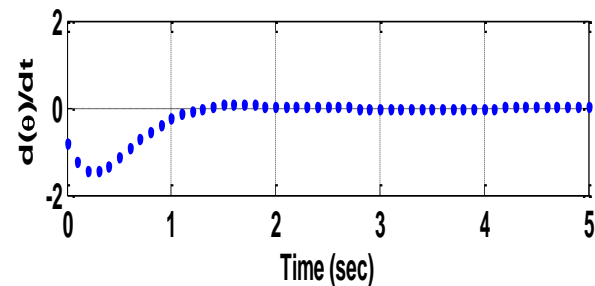
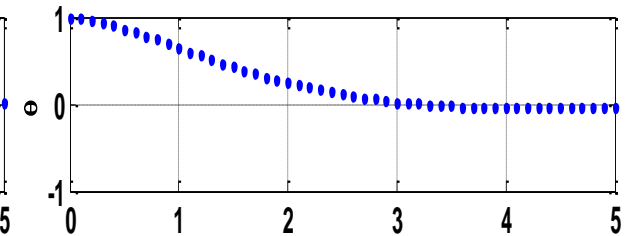
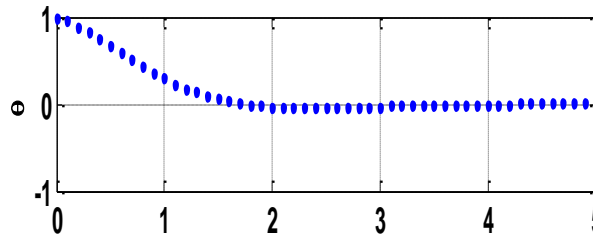
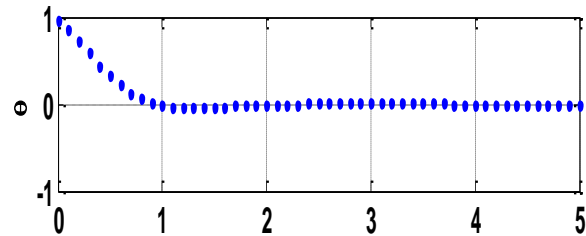
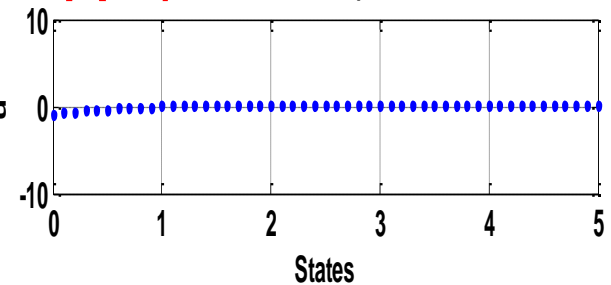
$R=0.01$ Control input



$R=0.1$ Control input



$R=1$ Control input





Summary

- Discrete-time LQR
 - Finite horizon: Matrix Riccati difference equation
 - Infinite horizon: Discrete Algebraic Riccati Equation (DARE)
- Next, Kalman filter
 - We will see later that discrete-time LQR is “dual” of Kalman filter.

Optimality of DT LQR control law (optional)

1. For any n-by-n symmetric $P[k]$ and $x[k]$ satisfying

$$x[k+1] = Ax[k] + Bu[k], x[k] = x_0$$

we have

$$\begin{aligned} & x^T[k_f]P[k_f]x[k_f] - x^T[0]P[0]x[0] \\ = & \sum_{k=0}^{k_f-1} \left[x^T[k+1]P[k+1] \underbrace{x[k+1]}_{Ax[k]+Bu[k]} - x^T[k]P[k]x[k] \right] \\ = & \sum_{k=0}^{k_f-1} \left[x^T[k] \left\{ A^T P[k+1]A - \textcircled{P[k]} \right\} x[k] \right. \\ & \left. + u^T[k]B^T P[k+1]Ax[k] + x^T[k]A^T P[k+1]Bu[k] \right. \\ & \left. + u^T[k]B^T P[k+1]Bu[k] \right] \end{aligned}$$

→ See the next slide

Optimality of DT LQR control law

2. Select a special $P[k]$ satisfying

$$\begin{cases} P[k] = A^T P[k+1]A + Q \\ \quad - A^T P[k+1]B [R + B^T P[k+1]B]^{-1} B^T P[k+1]A \\ P[k_f] = S \end{cases}$$

Then,

$$\begin{aligned} 0 &= -x^T[k_f]Sx[k_f] + x^T[0]P[0]x[0] \\ &\quad + \sum_{k=0}^{k_f-1} \left[x^T[k] \left\{ -Q + A^T P[k+1]B [R + B^T P[k+1]B]^{-1} B^T P[k+1]A \right\} x[k] \right. \\ &\quad \left. + u^T[k]B^T P[k+1]Ax[k] + x^T[k]A^T P[k+1]Bu[k] \right. \\ &\quad \left. + u^T[k]B^T P[k+1]Bu[k] \right] \end{aligned}$$

Optimality of DT LQR control law

3. By adding the cost function below to both sides

$$J(u[\cdot]) := \sum_{k=0}^{k_f-1} \left[x^T[k] Q x[k] + u^T[k] R u[k] \right] dt + x^T[k_f] S x[k_f]$$

we have

$$\begin{aligned} J(u[\cdot]) = & x_0^T P[0] x_0 \\ & + \sum_{k=0}^{k_f-1} x^T[k] \left\{ A^T P[k+1] B \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A \right\} x[k] \\ & + \sum_{k=0}^{k_f-1} u^T[k] B^T P[k+1] A x[k] + x^T[k] A^T P[k+1] B u[k] \\ & + \sum_{k=0}^{k_f-1} u^T[k] \left\{ R + B^T P[k+1] B \right\} u[k] \end{aligned}$$

Optimality of DT LQR control law

3. (cont'd) By **completion of square**

$$\begin{aligned}
 J(u[\cdot]) &= x_0^T P[0] x_0 \\
 &+ \sum_{k=0}^{k_f-1} \left[u[k] + \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A x[k] \right]^T \\
 &\quad \times \left[R + B^T P[k+1] B \right] \\
 &\quad \times \left[u[k] + \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A x[k] \right]
 \end{aligned}$$

4. Since $R > 0$, the function J achieves its minimum when

$$u[k] = - \left[R + B^T P[k+1] B \right]^{-1} B^T P[k+1] A x[k], \quad k = 0, 1, \dots, k_f-1$$

DT LQR with an integrator

- Problem

$$\min_{u[\cdot]} J(u[\cdot]) \text{ subj. to } \begin{cases} x[k+1] &= Ax[k] + Bu[k], \quad x[0] = 0 \\ y[k] &= Cx[k] \end{cases}$$

- J : Quadratic performance index (cost function)

$$J(u[\cdot]) := \sum_{k=0}^{\infty} \underbrace{(r - y[k])^T Q (r - y[k])}_{\text{For small deviation from reference } r} + \underbrace{(u[k+1] - u[k])^T R (u[k+1] - u[k])}_{\text{For small input rate of change}}$$

For small deviation
from reference r

For small input
rate of change

- Design parameters $Q \geq 0, R > 0$

Reduction to standard DT LQR

- Consider an auxiliary state vector: $\tilde{x}[k] := \begin{bmatrix} x[k+1] - x[k] \\ e[k] \end{bmatrix}$
- Then,

$$\tilde{x}[k+1] = \begin{bmatrix} x[k+2] - x[k+1] \\ e[k+1] \end{bmatrix} = \underbrace{\begin{bmatrix} A & 0 \\ -C & I \end{bmatrix}}_{\tilde{A}} \tilde{x}[k] + \underbrace{\begin{bmatrix} B \\ 0 \end{bmatrix}}_{\tilde{B}} \underbrace{(u[k+1] - u[k])}_{\Delta u[k]}$$

$$J(u[\cdot]) := \sum_{k=0}^{\infty} \tilde{x}[k]^T \underbrace{\begin{bmatrix} 0 \\ I \end{bmatrix}^T Q \begin{bmatrix} 0 & I \end{bmatrix}}_{\tilde{Q}} \tilde{x}[k] + \Delta u[k]^T R \Delta u[k]$$

- Problem is reduced to a standard LQR problem:

$$\min_{u[\cdot]} \sum_{k=0}^{\infty} \tilde{x}[k]^T \tilde{Q} \tilde{x}[k] + \Delta u[k]^T R \Delta u[k] \text{ subj. to } \tilde{x}[k+1] = \tilde{A} \tilde{x}[k] + \tilde{B} \Delta u[k]$$

Reduction to DT LQR (cont'd)

- LQR optimal control is

$$\Delta u[k] = -(R + \tilde{B}^T \tilde{P} \tilde{B})^{-1} \tilde{B}^T \tilde{P} \tilde{A} \tilde{x}[k] = -K_x(x[k+1] - x[k]) - K_e e[k]$$

$$\rightarrow u[k] = \sum_{k=0}^{k-1} \Delta u[k] = -K_x(x[k] - \underbrace{x[0]}_0) - K_e \sum_{k=0}^{k-1} e[k]$$

