

University of British Columbia  
Department of Mechanical Engineering

MECH468 Modern Control Engineering  
MECH522 Foundations in Control Engineering  
Final exam

Examiner: Dr. Ryoze Nagamune  
April 11 (Wednesday), 2018, noon-2:30pm

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Last name, First name

Name:

Student #:

Signature:

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**Exam policies**

- Allowed: Lecture note print-outs, hand-written notes, homework, books.
- Not-allowed: PC, calculators, mobile phones.
- Write all your answers on the provided exam booklet.
- Motivate your answers properly. (No chance to defend your answers orally.)
- 100 points in total. Mark will be scaled later.

**Before you start ...**

- Use washroom before the exam.
- Turn off your mobile phone.
- No eating.
- Questions are NOT allowed.

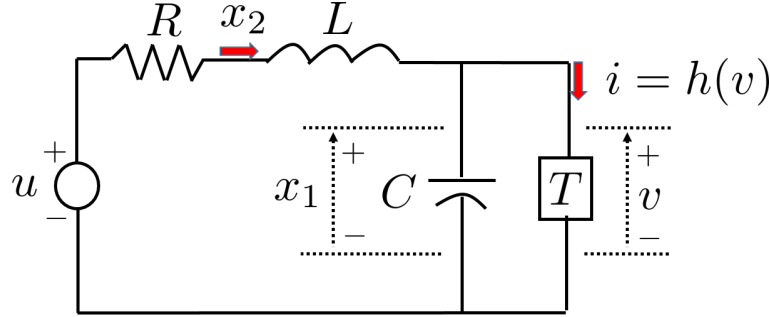
**If you finish early ...**

- If you would like to leave the room **before 2:20pm, raise your hand with this booklet**, and **wait at your seat** until the invigilator comes to you and collects your exam booklet.

**To be filled in by the instructor/marker**

Problem #	Mark	Full mark
1		20
2		20
3		20
4		20
5		20
Total		100

1. Consider the electric circuit depicted below. Here, the notations  $R$ ,  $L$  and  $C$  respectively denote the resistance, inductance and capacitance, and  $u$  is the voltage source. An electrical element  $T$  has the characteristic  $i = h(v)$ , where  $i$  is the current through  $T$  and  $v$  is the voltage across  $T$ , and  $h$  is a nonlinear function which is differentiable with respect to  $v$  (i.e.,  $h'(v)$  exists).



- (a) Let  $x_1$  be the voltage across the capacitance, and  $x_2$  be the current through the inductance. Prove that the state equation for this system is described as follows. (10pt)

$$\begin{aligned}\dot{x}_1(t) &= -\frac{1}{C}h(x_1(t)) + \frac{1}{C}x_2(t) \\ \dot{x}_2(t) &= -\frac{1}{L}x_1(t) - \frac{R}{L}x_2(t) + \frac{1}{L}u(t)\end{aligned}$$

- (b) Linearize the state equation above around the operating point  $(x_1, x_2, u) = (x_{10}, x_{20}, u_0)$ . (6pt)
  - (c) Express  $x_{20}$  and  $u_0$  as functions of  $x_{10}$ . (4pt)
2. Obtain minimal realizations of the following transfer functions. After obtaining minimal realizations, check if the realization is indeed minimal. (10pt-each)

- (a)  $G_1(s) = \begin{bmatrix} \frac{s^2 + 3s + 2}{s^2 + 2s + 1} \\ \frac{1}{s + 2} \end{bmatrix}$

- (b)  $G_2(s) = \begin{bmatrix} \frac{1}{s} & \frac{4}{s} \\ \frac{2}{s} & \frac{8}{s} \end{bmatrix}$

3. For the following state-space model, design an observer-based state-feedback controller. For the controller design, select the pole locations so that (20pt)
  - state estimation error converges to zero in about 0.4 second, and
  - (2%) settling time for initial condition excitation becomes about 1 second.

$$\begin{aligned}\dot{x}(t) &= \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t) \\ y(t) &= \begin{bmatrix} 0 & 1 \end{bmatrix} x(t)\end{aligned}$$

4. Determine whether each statement is True or False.

- If your answer is ‘True’, provide an explanation to support your answer.
- If your answer is ‘False’, provide a counter-example with two states, with an explanation, to support your answer. In counter-examples, use non-zero  $B$ -matrix and non-zero  $C$ -matrix.

One example is given below.

(10pt-each)

**Example** If a linear time-invariant system is stable, then it is controllable.

**Answer** False

**Counter-example**

$$\dot{x} = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

**Explanation** This system is stable because the eigenvalues of  $A$ -matrix are  $-1$  and  $-1$ , both of which are in the open left-half plane. However, it is not controllable because the rank of the controllability matrix  $\mathcal{C} = \begin{bmatrix} 1 & -1 \\ 1 & -1 \end{bmatrix}$  is one, i.e.,  $\mathcal{C}$  is not full rank.

- (a) If a linear time-invariant system is unstable, it is not controllable.
- (b) If a linear time-invariant system is detectable, it is observable.

5. Consider the following continuous-time infinite-horizon linear quadratic regulator (LQR) problem, where  $\alpha$  is a positive constant.

$$\min_{u(\cdot)} \int_0^\infty \{ \alpha x_2(t)^2 + u_1(t)^2 + u_2(t)^2 \} dt$$

$$\text{subject to } \begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} u_1(t) \\ u_2(t) \end{bmatrix}$$

- (a) Design the LQR control law. (10pt)
- (b) Prove that the designed LQR control law stabilizes closed-loop system for any  $\alpha > 0$ . (5pt)
- (c) For the closed-loop system with the designed LQR control law and  $\alpha = 3$ , draw the state trajectory in  $(x_1, x_2)$ -plane when the initial state is  $x(0) = (1, 1)$ . (**Hint:** The state trajectory must converge to  $(0, 0)$  (i.e., origin of the  $(x_1, x_2)$ -plane because the closed-loop system is stable.) (5pt)

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