University of British Columbia Department of Mechanical Engineering

MECH468 Modern Control Engineering MECH509 Controls Final exam

Examiner: Dr. Ryozo Nagamune April 26 (Monday), 2021, noon-2:30pm (PST)

Exam policies

- Allowed: Open-book. Any distributed material and any textbook. You can see course materials on your computer.
- Not-allowed: Matlab. Calculators. Web-browsing.
- Write all your answers on your own sheets.
- Motivate your answers properly. (No chance to defend your answers orally.)
- No questions are allowed.
- 100 points in total. (It will be scaled later.)

Before you start ...

- Turn off your mobile phone, or make it in the airplane mode.
- Make sure that your above-neck is fully captured by the webcam.

After you finish the exam ...

- You will get enough time to scan and upload your answer sheets on Canvas "Assignments" **AFTER** the exam writing time (150 minutes).
- If you finish the exam early, you can upload answer sheets and leave Zoom room.
- Make sure that you have uploaded all your answer sheets. You cannot add some sheets later even if you somehow missed uploading them.

Marking scheme

Question #	Expected duration	Full mark
Q1	about 30 min	20 %
Q2	about 30 min	20 %
Q3	about 45 min	30 %
Q4	about 45 min	30 %
Total	about 150 min	100 %

- 1. Select **only one** correct statement for the following sentences. There is no need to motivate your answers. (2pt each)
 - (a) If we place the poles of the closed-loop system with state-feedback at s = -0.04, -0.4, -4, it is roughly estimated that the 2\% settling time for the step response would be about:
 - i. 0.01 second.
 - ii. 0.1 second.
 - iii. 1 second.
 - iv. None of i, ii, iii is correct.
 - (b) A series connection $G_1(s)G_2(s)$ of the two systems $G_1(s) = \frac{s+1}{s+2}$ and $G_2(s) = \frac{s+2}{s+1}$ is:
 - i. asymptotically stable but not BIBO stable.
 - ii. BIBO stable but not asymptotically stable.
 - iii. both BIBO stable and asymptotically stable.
 - iv. None of i, ii, iii is correct.
 - (c) If a linear time-invariant system is:
 - i. asymptotically stable, then it is always observable.
 - ii. observable, then it is always stabilizable.
 - iii. stabilizable, then it is always detectable.
 - iv. None of i, ii, iii is correct.
 - (d) In infinite-horizon LQR control, if we want to regulate y = Cx + Du $(C \neq 0, D \neq 0)$ around zero, then the cost function to be minimized will be in the form of:

i.
$$\int_0^\infty (x^TQx + u^TRu)dt$$

ii.
$$\int_0^\infty x^TQxdt$$

iii.
$$\int_0^\infty u^TRudt$$

iv. None of i, ii, iii is correct.

ii.
$$\int_0^\infty x^T Qx dt$$

iii.
$$\int_{0}^{\infty} u^{T} Ru dt$$

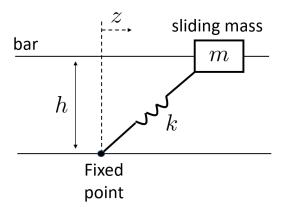
- (e) If a symmetric matrix $M = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 2 \\ 0 & 2 & 4 \end{bmatrix}$ is:
 - i. positive definite.
 - ii. positive semidefinite.
 - iii. negative definite.
 - iv. None of i, ii, iii is correct.

- (f) McMillan degree of the transfer matrix $G(s) = \frac{1}{s+1} \begin{bmatrix} 1 & 1 & 1 \\ 1 & 1 & 1 \end{bmatrix}$ is:
 - i. 1
 - ii. 2
 - iii. 3
 - iv. None of i, ii, iii is correct.
- (g) The solution to a state equation $\dot{x}(t) = Ax(t)$ with condition x(1) = 2 is:
 - i. $x(t) = 2e^{At}$
 - ii. $x(t) = 2e^{A(t-1)}$
 - iii. $x(t) = 2e^{A(t+1)}$
 - iv. None of i, ii, iii is correct.
- (h) A system $\dot{x}(t) = \begin{bmatrix} \alpha & 1 \\ -1 & \alpha \end{bmatrix} x(t)$ with a constant α is:
 - i. asymptotically stable for any α .
 - ii. marginally stable for any α .
 - iii. unstable for any α .
 - iv. None of i, ii, iii is correct.
- (i) In the state-feedback controller design using pole-placement technique, if the overshoot of the step response for a closed-loop system needs be reduced, then we move the closed-loop poles to the locations:
 - i. far from the real axis of the complex plane.
 - ii. close to the real axis of the complex plane.
 - iii. close to the imaginary axis of the complex plane.
 - iv. None of i, ii, iii is correct.
- (j) For a discrete-time system $x[k+1] = \begin{bmatrix} 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} x(t) + \begin{bmatrix} 1 \\ 0 \\ 1 \end{bmatrix} u[k]$, it is possible to transfer the state from $x[0] = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$ to $x[k_f] = \begin{bmatrix} 1 \\ 1 \\ 0 \end{bmatrix}$ for:
 - i. any k_f satisfying $k_f \geq 1$.
 - ii. any k_f satisfying $k_f \geq 2$.
 - iii. any k_f satisfying $k_f \geq 3$.
 - iv. None of i, ii, iii is correct.

2. Consider a mechanical system consisting of a mass m sliding on a horizontal bar and connected to a spring with the spring constant k as shown in the figure below. When the spring is not stretched, its angle with the horizontal is equal to 45 degrees. When a force f is applied to the mass, the horizontal displacement z of the system evolves according to the following equation:

$$m\ddot{z} = k \left(\frac{\sqrt{2}h}{\sqrt{z^2 + h^2}} - 1 \right) z + f,$$

where h is the distance shown in the figure.



- (a) Obtain the nonlinear state equation (output equation is NOT necessary) for the system when the input is the force f, by introducing the state vector $x := [z, \dot{z}]^T$. (5pt)
- (b) Obtain all equilibrium states x_{eq} of the system for the zero input $f_{eq} = 0$. For each equilibrium, draw a figure which illustrates the mass location and the spring angle are physically. (10pt)
- (c) Compute linearizations of the state equation around each equilibrium point. (5pt)

3. Consider to estimate a 1-dimensional position x [m] of a mass which stays at one location, using multiple position sensor measurements:

$$y_i = x + n_i, \quad i = 1, 2, 3,$$

where n_i is the noise for *i*-th measurement. The variances of the measurement noises n_i are assumed to be

$$E\left\{n_1^2\right\} = 1, \quad E\left\{n_2^2\right\} = \frac{1}{2}, \quad E\left\{n_3^2\right\} = \frac{1}{4},$$

and they are not correlated, i.e.,

$$E\{n_1n_2\} = E\{n_2n_3\} = E\{n_3n_1\} = 0.$$

- (a) First, using y_1 , y_2 and y_3 , and using the **batch weighted least-squares estimation**, obtain the position estimate \hat{x}_b and its error variance $P_b := E\{(x \hat{x}_b)^2\}$. (10pt)
- (b) Next, in (b) and (c), we will demonstrate that x_b and P_b in (a) can be obtained by the recursive least-squares estimation. For this purpose, first, using only y_1 and y_2 , and using the **batch weighted least-squares estimation**, obtain the position estimate \hat{x}_o and its error variance $P_o := E\{(x \hat{x}_o)^2\}$. (In the question (b), the subscript 'o' means that the estimate is based on "old" measurements.) (10pt)
- (c) Finally, using (\hat{x}_o, P_o) in (b) and y_3 , and using the **recursive weighted** least-squares estimation, obtain the position estimate \hat{x}_n and its error variance $P_n := E\{(x \hat{x}_n)^2\}$. (In (c), the subscript 'n' means that the estimate is updated based on "new" measurements.) (10pt)
- 4. Consider a system $\dot{x}(t) = Ax(t) + Bu(t)$ with the following matrices:

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & -1 & 0 \\ 1 & 0 & 1 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix}.$$

- (a) Determine whether this system is controllable.
- (b) Without finding a stabilizing state-feedback gain K, prove that this system is stabilizable. (10pt)

Hint: Although the coordinate transformation matrix T is 4-by-4 in this question, if you select T^{-1} in a 'clever' way, you can hand-calculate its inverse T.

(10pt)

(c) By selecting the closed-loop poles in the stable region by yourself, compute a matrix K such that the eigenvalues of A-BK are placed at your selected closed-loop pole locations. (10pt)