

1. Consider the following continuous-time system:

$$\begin{cases} \dot{x}(t) = \begin{bmatrix} -1 & a \\ 1 & -2 \end{bmatrix} x(t) + \begin{bmatrix} b \\ 1 \end{bmatrix} u(t), \\ y(t) = \begin{bmatrix} c & 1 \end{bmatrix} x(t), \end{cases}$$

where  $a, b, c$  are constants.

- (a) Obtain the condition for asymptotic stability in terms of  $a$ . (2pt)

Write your answer here.

Eigenvalues of A-matrix are computed as

$$\det(\lambda I - A) = \det \begin{bmatrix} \lambda+1 & -a \\ -1 & \lambda+2 \end{bmatrix}$$

$$= (\lambda+1)(\lambda+2) - a$$

$$= \lambda^2 + 3\lambda + 2 - a = 0$$

The condition for  $\text{Re}[\lambda] < 0$  is that all coefficients have the same sign, because this is 2-nd order equation.

Thus  $2-a > 0$ , i.e.  $a < 2$

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- (b) Obtain the condition for controllability in terms of  $a$  and  $b$ . (2pt)  
(c) Obtain the condition for observability in terms of  $a$  and  $c$ . (2pt)

Write your answer here.

(b) Controllability matrix

$$C = \begin{bmatrix} b & -b+a \\ 1 & b-2 \end{bmatrix}$$

For  $C$  to be full rank, the condition is  $\det C \neq 0$

$$\det C = b(b-2) - (a-b) \neq 0$$

$$b^2 - b - a \neq 0$$

(c) Observability matrix

$$O = \begin{bmatrix} c & 1 \\ -c+1 & ac-2 \end{bmatrix}$$

For  $O$  to be full rank, the condition is  $\det O \neq 0$

$$\det O = c(ac-2) - (1-c) \neq 0$$

$$ac^2 - c - 1 \neq 0$$

(d) Consider the case when  $a = 0$  and  $b = c = 1$ , i.e.,

$$\begin{cases} \dot{x}(t) = \begin{bmatrix} -1 & 0 \\ 1 & -2 \end{bmatrix} x(t) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(t), \\ y(t) = [1 \ 1] x(t). \end{cases} \quad (1)$$

- i. Obtain the Kalman decomposition. (2pt)
- ii. Write explicitly which state is controllable / uncontrollable and observable / unobservable. (1pt)
- iii. Verify that the "controllable-and-observable part" ( $A_{co}, B_{co}, C_{co}$ ) is actually controllable and observable. (1pt)

Write your answer here.

$$(i) \quad C = \begin{bmatrix} 1 & -1 \\ 1 & -1 \end{bmatrix} \Rightarrow \text{rank } C = 1 < 2 \Rightarrow \text{NOT controllable}$$

$$O = \begin{bmatrix} 1 & 1 \\ 0 & -2 \end{bmatrix} \Rightarrow \text{rank } O = 2 \Rightarrow \text{observable}$$

$$\text{Im } C = \text{span} \left\{ \begin{bmatrix} 1 \\ 1 \end{bmatrix} \right\}$$

$$T^{-1} = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} \xrightarrow{T_c \ T_o} \Rightarrow T = \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix}$$

$$TAT^{-1} = \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix} \begin{bmatrix} -1 & 0 \\ 1 & -2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 2 & -2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 0 & -2 \end{bmatrix}$$

$$TB = \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$CT^{-1} = \begin{bmatrix} 1 & 0 \\ 2 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix} = \begin{bmatrix} 2 & 1 \end{bmatrix}$$

$$\begin{cases} \begin{bmatrix} \dot{z}_{co} \\ \dot{z}_{\bar{co}} \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} z_{co} \\ z_{\bar{co}} \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \end{cases}$$

$$y = \begin{bmatrix} 2 & 1 \end{bmatrix} \begin{bmatrix} z_{co} \\ z_{\bar{co}} \end{bmatrix}_4 \quad \begin{cases} z_{co} : \text{controllable \& observable} \\ z_{\bar{co}} : \text{uncontrollable \& observable} \end{cases} \quad (ii)$$

(iii)  $(A_{co}, B_{co}, C_{co}) = (-1, 1, 2)$

$C = 1 \neq 0 \therefore \text{controllable}$

$O = 2 \neq 0 \therefore \text{observable}$

- (e) For the system (1) in question (d), compute the  $A_d$ -matrix ( $A$ -matrix of a discrete-time system) of the discretized system (by zero-order-hold) with sampling period  $T = 1$ . (2pt)

Write your answer here.

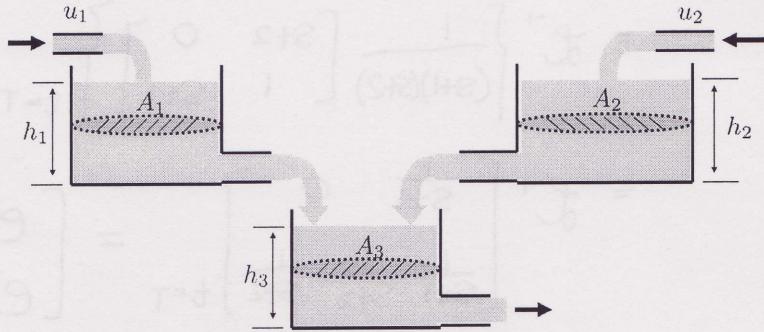
$$\begin{aligned}
 A_d = e^{AT} &= \mathcal{L}^{-1} \left\{ (sI - A)^{-1} \right\}_{t=T} \\
 &= \mathcal{L}^{-1} \left\{ \begin{bmatrix} s+1 & 0 \\ -1 & s+2 \end{bmatrix}^{-1} \right\}_{t=T} \\
 &= \mathcal{L}^{-1} \left\{ \frac{1}{(s+1)(s+2)} \begin{bmatrix} s+2 & 0 \\ 1 & s+1 \end{bmatrix} \right\}_{t=T} \\
 &= \mathcal{L}^{-1} \begin{bmatrix} \frac{1}{s+1} & 0 \\ \frac{1}{s+1} - \frac{1}{s+2} & \frac{1}{s+2} \end{bmatrix}_{t=T} = \begin{bmatrix} e^{-T} & 0 \\ e^{-T} - e^{-2T} & e^{-2T} \end{bmatrix}
 \end{aligned}$$

$$T=1: \quad A_d = \begin{bmatrix} e^{-1} & 0 \\ e^{-1} - e^{-2} & e^{-2} \end{bmatrix}$$

2. Consider a three-water-tank system in the figure below. Here,  $A_i$ ,  $i = 1, 2, 3$ , are tank section areas,  $h_i$ ,  $i = 1, 2, 3$ , are the water heights of the tanks, and  $u_i$ ,  $i = 1, 2$ , are input flow rates  $u_1$  and  $u_2$ . The nonlinear state equation of this system is assumed to be expressed as

$$\begin{aligned}\dot{h}_1(t) &= \frac{1}{\rho A_1} \left( -K\sqrt{h_1(t)} + u_1(t) \right), \\ \dot{h}_2(t) &= \frac{1}{\rho A_2} \left( -K\sqrt{h_2(t)} + u_2(t) \right), \\ \dot{h}_3(t) &= \frac{1}{\rho A_3} \left( -K\sqrt{h_3(t)} + K\sqrt{h_1(t)} + K\sqrt{h_2(t)} \right),\end{aligned}$$

where  $\rho$  and  $K$  are given positive constants.



We would like to linearize the nonlinear state equation around the situation when we maintain the water heights at  $h_1(t) = h_{10}$  and  $h_2(t) = h_{20}$ , where  $h_{10}$  and  $h_{20}$  are given positive constant heights.

- (a) Obtain the corresponding constant input flow rates  $u_1(t) = u_{10}$  and  $u_2(t) = u_{20}$  in terms of given constants  $h_{10}$  and  $h_{20}$ . (2pt)
- (b) Obtain the corresponding constant water height  $h_3(t) = h_{30}$  in terms of given constants  $h_{10}$  and  $h_{20}$ . (2pt)
- (c) Derive a linearized state equation  $\delta\dot{h}(t) = A\delta h(t) + B\delta u(t)$  around the equilibrium point  $(h_1, h_2, h_3) = (h_{10}, h_{20}, h_{30})$  and  $(u_1, u_2) = (u_{10}, u_{20})$ . To answer this question, you do not need to use solutions obtained in (a) and (b); just use  $(h_{10}, h_{20}, h_{30})$  and  $(u_{10}, u_{20})$ . (2pt)

Write your answer here.

(a)  $\dot{h}_1 = 0 \Rightarrow u_1(t) = K\sqrt{h_1(t)}$  Since  $h_1(t) = h_{10}$ ,  $u_{10} = K\sqrt{h_{10}}$   
 $\dot{h}_2 = 0 \Rightarrow u_2(t) = K\sqrt{h_2(t)}$  Since  $h_2(t) = h_{20}$ ,  $u_{20} = K\sqrt{h_{20}}$

(b)  $\dot{h}_3 = 0 \Rightarrow K\sqrt{h_3(t)} = K\sqrt{h_1(t)} + K\sqrt{h_2(t)}$   
With  $h_1(t) = h_{10}$ ,  $h_2(t) = h_{20}$ ,  $h_{30} = (\sqrt{h_{10}} + \sqrt{h_{20}})^2$

Write your answer here.

$$(c) \quad \delta h = A \delta h + B \delta u$$

$$\delta h := \begin{bmatrix} h_1 - h_{10} \\ h_2 - h_{20} \\ h_3 - h_{30} \end{bmatrix}, \quad \delta u := \begin{bmatrix} u_1 - u_{10} \\ u_2 - u_{20} \\ u_3 - u_{30} \end{bmatrix}$$

$$A = \frac{\partial f}{\partial h} \Bigg|_{\substack{h=h_0 \\ u=u_0}} = \begin{bmatrix} -\frac{k}{PA_1} \frac{1}{2\sqrt{h_{10}}} & 0 & 0 \\ 0 & -\frac{k}{PA_2} \frac{1}{2\sqrt{h_{20}}} & 0 \\ \frac{k}{PA_3} \frac{1}{2\sqrt{h_{10}}} & \frac{k}{PA_3} \frac{1}{2\sqrt{h_{20}}} & -\frac{k}{PA_3} \frac{1}{2\sqrt{h_{30}}} \end{bmatrix}$$

$$B = \frac{\partial f}{\partial u} \Bigg|_{\substack{h=h_0 \\ u=u_0}} = \begin{bmatrix} \cancel{y_{PA_1}} & 0 \\ 0 & \cancel{y_{PA_2}} \\ 0 & 0 \end{bmatrix}$$

3. Consider the following controllable discrete-time system:

$$x[k+1] = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} x[k] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[k],$$

Compute the minimum energy control  $u[k]$ ,  $k = 0, 1, 2$ , which transfers state vector from  $x[0] = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$  to  $x[3] = \begin{bmatrix} 6 \\ 6 \end{bmatrix}$ . (2pt)

— (End of Midterm Exam) —

Write your answer here.

Min. energy control

$$\begin{bmatrix} u[2] \\ u[1] \\ u[0] \end{bmatrix} = C_d[3]^T \left( C_d[3] C_d[3]^T \right)^{-1} \left( \underbrace{x[3]}_{\begin{bmatrix} 6 \\ 6 \end{bmatrix}} - \underbrace{A^3 x[0]}_{\begin{bmatrix} 0 \\ 0 \end{bmatrix}} \right)$$

$$C_d[3] = \begin{bmatrix} B & AB & A^2B \end{bmatrix} \quad = \begin{bmatrix} 0 & 1 \\ 1 & 1 \\ 2 & 1 \end{bmatrix} \frac{1}{6} \begin{bmatrix} 3 & -3 \\ -3 & 5 \end{bmatrix} \begin{bmatrix} 6 \\ 6 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & 1 & 2 \\ 1 & 1 & 1 \end{bmatrix} \quad = \begin{bmatrix} 0 & 1 \\ 1 & 1 \\ 2 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix}$$

$$(C_d[3] C_d[3]^T)^{-1} = \left( \begin{bmatrix} 0 & 1 & 2 \\ 1 & 1 & 1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 1 & 1 \\ 2 & 1 \end{bmatrix} \right)^{-1} \quad = \begin{bmatrix} 2 \\ 2 \\ 2 \end{bmatrix}$$

$$= \begin{bmatrix} 5 & 3 \\ 3 & 3 \end{bmatrix}^{-1} = \frac{1}{6} \begin{bmatrix} 3 & -3 \\ -3 & 5 \end{bmatrix} \quad //$$