Incorrect Filter

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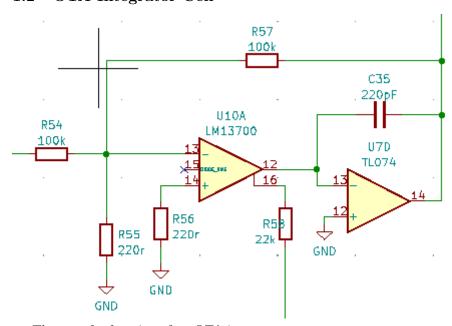
1 Introduction

Analysis of the original, incorrect design of the Ouroboros filter. Heavily based on the Serge VCFQ, but also with influence from the Mutable Instruments Blades. Much of this is pulled verbatim from the SSM2164 SVF Analysis by Emilie Gillet and then just somewhat modified.

1.1 Notations

- R_i is the value of the resistor through which the input signal is fed to the
- R_f is the value of the resistor through which the output signal of the integrator is fed back to the input.
- R_s is the value of the resistor from the OTA inputs to ground.
- C is the value of the integrators' capacitors.
- R_g is the value of the resistor through which the HP and LP outputs are fed back into the input.
- R_q is the value of the resistor through which the attenuated BP output is fed back into the input.
- v_{cv} is the cutoff frequency control voltage.
- v_q is the resonance control voltage.
- $v_i(s)$ is the input voltage.
- $v_{hp}(s)$, $v_{bp}(s)$ and $v_{lp}(s)$ are the high-pass, band-pass and low-pass output voltages.

1.2 OTA Integrator Cell



The transfer function of an OTA is:

$$i_o = g_m(v_+ - v_-)$$

 $i_o = 19.2i_{cv}(v_+ - v_-)$

If R_f is equal to R_i Kirchoff in v_- gives:

$$\begin{split} \frac{1}{R_i}(v_i(s) - v_-(s)) + \frac{1}{R_i}(v_{out}(s) - v_-(s)) &= \frac{1}{R_s}v_-(s) \\ R_sv_i(s) - R_sv_-(s) + R_sv_{out}(s) - R_sv_-(s) &= R_iv_-(s) \\ R_sv_i(s) + R_sv_{out}(s) - R_sv_-(s) - R_sv_-(s) &= R_iv_-(s) \\ R_sv_i(s) + R_sv_{out}(s) - 2R_sv_-(s) &= R_iv_-(s) \\ R_sv_i(s) + R_sv_{out}(s) &= R_iv_-(s) + 2R_sv_-(s) \\ R_s(v_i(s) + v_{out}(s)) &= v_-(s)(R_i + 2R_s) \\ \frac{R_s}{R_i + 2R_s}(v_i(s) + v_{out}(s)) &= v_-(s) \end{split}$$

The current i_c at the output if the OTA is:

$$\begin{split} i_c(s) &= g_m(v_+(s) - v_-(s)) \\ &= 19.2 i_{cv} \big(0 - \frac{R_s}{R_i + 2R_s} (v_i(s) + v_{out}(s)) \big) \\ &= -19.2 i_{cv} \frac{R_s}{R_i + 2R_s} (v_i(s) + v_{out}(s)) \end{split}$$

The voltage out of the OpAmp:

$$\begin{split} v_{out}(s) &= \frac{-i_c(s)}{Cs} \\ v_{out}(s) &= -(-19.2i_{cv}\frac{R_s}{Cs(R_i + 2R_s)}(v_i(s) + v_{out}(s))) \\ v_{out}(s) &= 19.2i_{cv}\frac{R_s}{Cs(R_i + 2R_s)}(v_i(s) + v_{out}(s)) \\ v_{out}(s)Cs(R_i + 2R_s) &= 19.2i_{cv}R_s(v_i(s) + v_{out}(s)) \\ v_{out}(s)Cs(R_i + 2R_s) &= v_i(s)19.2i_{cv}R_s + v_{out}(s)19.2i_{cv}R_s \\ v_{out}(s)Cs(R_i + 2R_s) - v_{out}(s)19.2i_{cv}R_s &= v_i(s)19.2i_{cv}R_s \\ v_{out}(s)(Cs(R_i + 2R_s) - 19.2i_{cv}R_s) &= v_i(s)19.2i_{cv}R_s \\ v_{out}(s) &= v_i(s)\frac{19.2i_{cv}R_s}{Cs(R_i + 2R_s) - 19.2i_{cv}R_s} \\ v_{out}(s) &= v_i(s)\frac{1}{(R_i + 2R_s)}Cs\frac{1}{19.2i_{cv}} - 1 \end{split}$$

So the transfer function is:

$$\alpha(s) = \frac{1}{\frac{(R_i + 2R_s)}{R_s} Cs \frac{1}{19.2i_{cv}} - 1} \tag{1}$$

This should work as a single filter stage in itself, with the cutoff frequency being the point at which the outure is 3dB lower than the input, or $\frac{1}{\sqrt{2}} \approx 0.707$

$$0.707 = \frac{1}{\frac{(R_i + 2R_s)}{R_s}} C2\pi f \frac{1}{19.2i_{cv}} - 1$$

$$0.707 \frac{(R_i + 2R_s)}{R_s} C2\pi f \frac{1}{19.2i_{cv}} - 0.707 = 1$$

$$0.707 \frac{(R_i + 2R_s)}{R_s} C2\pi f \frac{1}{19.2i_{cv}} = 1.707$$

$$\frac{(R_i + 2R_s)}{R_s} C2\pi f \frac{1}{19.2i_{cv}} = \frac{19.2i_{cv}1.707}{0.707}$$

$$f \frac{(R_i + 2R_s)}{R_s} C2\pi = 19.2i_{cv}2.414$$

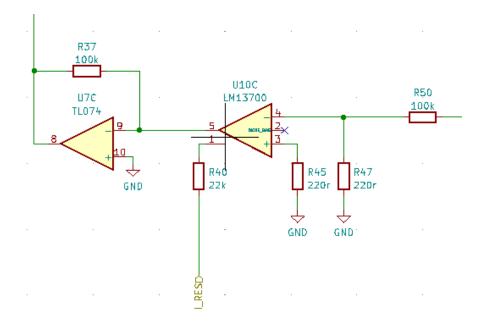
$$f = \frac{19.2i_{cv}2.414}{2\pi \frac{(R_i + 2R_s)}{R_s} C}$$

$$f = \frac{19.2i_{cv}1.207}{\pi \frac{(R_i + 2R_s)}{R_s} C}$$

$$f = \frac{23.18i_{cv}}{\pi \frac{(R_i + 2R_s)}{R_s} C}$$

1.3 Feedback Control

The gain of the feedback circuit is:



$$\begin{split} v_{-} &= v_{i} \frac{R_{s}}{R_{i} + R_{s}} \\ i_{o} &= 19.2 i_{res} (0 - v_{-}) \\ i_{o} &= -19.2 i_{res} v_{i} \frac{R_{s}}{R_{i} + R_{s}} \\ i_{o} + \frac{v_{o}}{R_{f}} &= 0 \\ i_{o} &= -\frac{v_{o}}{R_{f}} \\ -i_{o} R_{f} &= v_{o} \\ v_{o} &= -i_{o} R_{f} \\ v_{o} &= 19.2 i_{res} v_{i} \frac{R_{f} R_{s}}{R_{i} + R_{s}} \\ v_{o} &= 19.2 i_{res} v_{i} \frac{R_{f} R_{s}}{R_{i} + R_{s}} \\ \beta &= 19.2 i_{res} \frac{R_{f} R_{s}}{R_{i} + R_{s}} \end{split}$$

1.4 Filter

$$\begin{split} \frac{v_i - v_-}{R_i} + \frac{v_{hp} - v_-}{R_f} + \frac{v_{lp} - v_-}{R_f} + \frac{v_{bp}\beta - v_-}{R_q} &= i_- = 0 \\ \frac{v_i}{R_i} + \frac{v_{hp}}{R_f} + \frac{v_{lp}}{R_f} + \frac{v_{bp}\beta}{R_q} &= 0 \end{split}$$

It is important to note that $v_{bp}(s) = v_{hp}(s)\alpha(s)$, and $v_{lp}(s) = v_{hp}(s)\alpha(s^2)$.

1.4.1 High-pass

$$\begin{split} \frac{v_{hp}(s)}{R_f} &= -\frac{v_i(s)}{R_i} - \frac{v_{lp}(s)}{R_f} - \frac{v_{bp}(s)\beta}{R_q} \\ \frac{v_{hp}(s)}{R_f} &= -\frac{v_i(s)}{R_i} - \frac{v_{hp}(s)\alpha(s^2)}{R_f} - \frac{v_{hp}(s)\alpha(s)\beta}{R_q} \\ H_{hp}(s) &= \frac{v_{hp}(s)}{v_i(s)} \\ \frac{v_i(s)}{R_i} &= -\frac{v_{hp}(s)}{R_f} - \frac{v_{hp}(s)\alpha(s^2)}{R_f} - \frac{v_{hp}(s)\alpha(s)\beta}{R_q} \\ \frac{v_i(s)}{R_i} &= -v_{hp}(s)(\frac{1}{R_f} + \frac{\alpha(s^2)}{R_f} + \frac{\alpha(s)\beta}{R_q}) \\ \frac{1}{R_i} &= -H_{hp}(s)(\frac{1}{R_f} + \frac{\alpha(s^2)}{R_f} + \frac{\alpha(s)\beta}{R_q}) \\ \frac{-1}{R_i} &= H_{hp}(s)(\frac{1}{R_f} + \frac{\alpha(s^2)}{R_f} + \frac{\alpha(s)\beta}{R_q}) \\ H_{hp}(s) &= \frac{-1/R_i}{\frac{1}{R_f} + \frac{\alpha(s^2)}{R_f} + \frac{\alpha(s)\beta}{R_q}} \\ &= \frac{-R_f/R_i}{1 + \frac{R_f\alpha(s)\beta}{R_q} + \alpha(s^2)} \end{split}$$

1.4.2 Low-Pass

$$\begin{split} H_{lp}(s) &= \frac{v_{lp}(s)}{v_i(s)} \\ &= \frac{v_{hp}(s)}{v_i(s)} \alpha^2(s) \\ &= \frac{-1}{\frac{1}{\alpha^2(s)} + \frac{R_f \beta}{R_q \alpha(s)} + 1} \\ &= \frac{-R_f / R_i}{\frac{1}{\alpha^2(s)} + \beta \frac{R_f}{R_q} \frac{1}{\alpha(s)} + 1} \\ &= \frac{-R_f / R_i}{\frac{1}{(\frac{1}{R_i + 2R_s} - C_s \frac{1}{10 \cdot 2i_{cv}} - 1)^2} + (19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) \frac{R_f}{R_q} \frac{1}{(\frac{1}{R_i + 2R_s} - C_s \frac{1}{10 \cdot 2i_{cv}} - 1)} + 1 \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}} - 1)^2 + (19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) \frac{R_f}{R_q} (\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}} - 1) + 1 \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - 2(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 1 + (19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s} \frac{R_f}{R_q}) (\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}} - 1) + 1 \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s} \frac{R_f}{R_q} \frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}} + 2}{-R_f / R_i} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s} \frac{R_f}{R_q} (\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}}) - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_i + R_s}) + 2}{-R_f / R_s} \\ &= \frac{-R_f / R_i}{(\frac{R_i + 2R_s}{R_s} C s \frac{1}{19 \cdot 2i_{cv}})^2 - (2 + \frac{R_f}{R_q})(19 \cdot 2i_{res} \frac{R_f R_s}{R_$$

At this point it seems clear that this doesn't fit the standard transfer function for a second order filter....