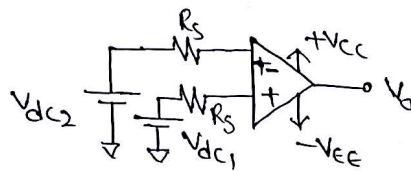


\* 1. DC characteristics of op-Amp:-

① \* Input off set Voltage ( $V_{io}$ ):-

\* The input off-set voltage is the voltage that must be applied between the two terminals of the op-amp to Null the op.

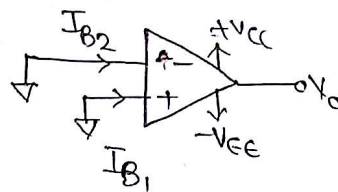
\* For IC741 the max value of  $V_{io}$  is 6mV



$$* V_{io} = V_{dc1} - V_{dc2}$$

② \* Input off set current ( $I_{io}$ ):-

\* The difference between the currents into the Non-inverting & inverting terminals.



$$* I_{io} = I_{B1} - I_{B2}$$

\*  $I_{io}$  for IC741 is 200nAmps.

③ \* Input Bias current ( $I_B$ ):-

\* Average of the currents that flow

into inverting & non-inverting terminals  $I_B = \frac{I_{B1} + I_{B2}}{2}$

### ① Thermal drift:-

\* Off-set Voltage, Off-set Current, Bias current changes with the temperature, it is called thermal drift.

### \* AC- Characteristics of op-Amp:-

#### \* ① Gain-Bandwidth product:-

\* It is B.W of op-Amp when voltage gain is 1.

\* For IC741 it is 1MHz.

#### \* ② Large Scale voltage gain:-

\* The voltage gain of an op-Amp is given as

$$A = \frac{V_o}{V_{id}}$$

\* If  $R_L > 2k\Omega$ , &  $V_o = \pm 10V$  then  $A = 2 \times 10^5$ .

#### \* ③ Slew rate:-

\* Defined as max rate of change of op voltage per unit time.

\* Expressed as V/ $\mu$ sec.

$$* SR = \frac{dV_o}{dt} \text{ V}/\mu\text{sec}$$

\*  $SR = 0.5 \text{ V}/\mu\text{sec}$  for IC741.

#### \* ④ Supply voltage rejection ratio:-

\* Change in op-Amp's  $V_{os}$  off-set voltage with the variations in supply voltage.

$$* SVRR = \frac{\Delta V_{io}}{\Delta V}$$

$$* \text{For } 741 \text{ IC } SVRR = 150 \text{ mV/V.}$$

\*5. Differential input Resistance ( $R_i$ )  $\rightarrow$

\* Equivalent Resistance that is measured at either inverting or non-inverting terminal with the other terminal at ground.

$$* \text{For IC741 it is } 2 \text{ M}\Omega.$$

\*6. Input Capacitance ( $C_i$ )  $\rightarrow$

\* Equivalent Capacitance that is measured either at inverting or non-inverting terminal with the other terminal connected to ground.

$$* \text{For IC741 it is } 1.4 \text{ pF}$$

\*7. Transient Response  $\rightarrow$

\* It is based on Rise time & overshoot.

$$* \text{For IC741 Rise-time} = 0.3 \mu\text{s}$$

$$\text{Overshoot} = 5\%$$

\*8. Output Resistance ( $R_o$ )  $\rightarrow$

\* The Equivalent resistance that is measured at the output terminal of an op-Amp.

$$* \text{IC741 } R_o = 75 \Omega.$$

### \* Supply Current $\rightarrow$

\* The current that is drawn by the op-Amp from the Supply Voltage

$$* I_S = 2.8 \text{ mA for IC741.}$$

### \* Power Consumption $\rightarrow (P_c)$

\*  $P_c$  is the Quiescent power that must be Consumed by the op-Amp to operate properly.

$$* \text{IC741 } P_c = 85 \text{ mW.}$$

### \*2. \* Instrumentation Amplifier $\rightarrow$

\* In a number of Consumer & Industrial applications there is a need to measure & Control physical quantities such as Control of the temperature, measuring of light intensity, water flow etc.

\* These are measured with the help of transducers.

\* Transducer is a device that converts one form of Energy into another.

\* The o/p of the transducer should be amplified so that it can drive the indicator or display system.

\* This function is performed by an Instrumentation amplifier.

### \* Features $\rightarrow$

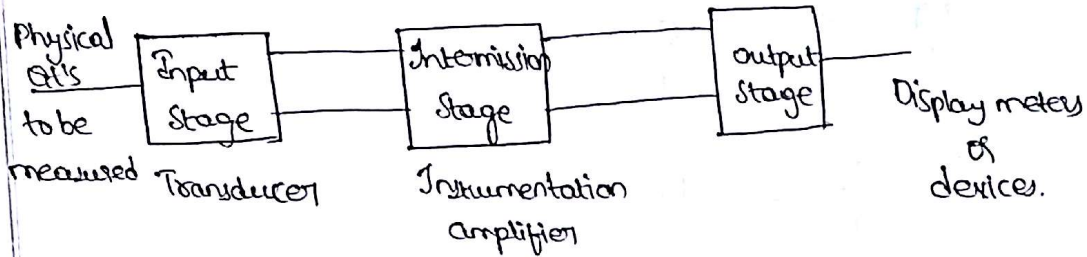
\*1. High gain accuracy.

\*2. High CMRR.

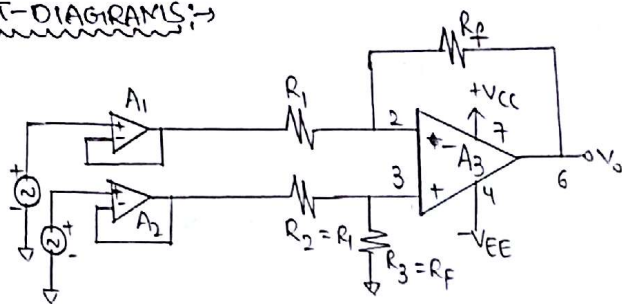
\*3. High gain stability with low temperature Co-efficient.

\*4. low DC ~~off-set~~ Voltage

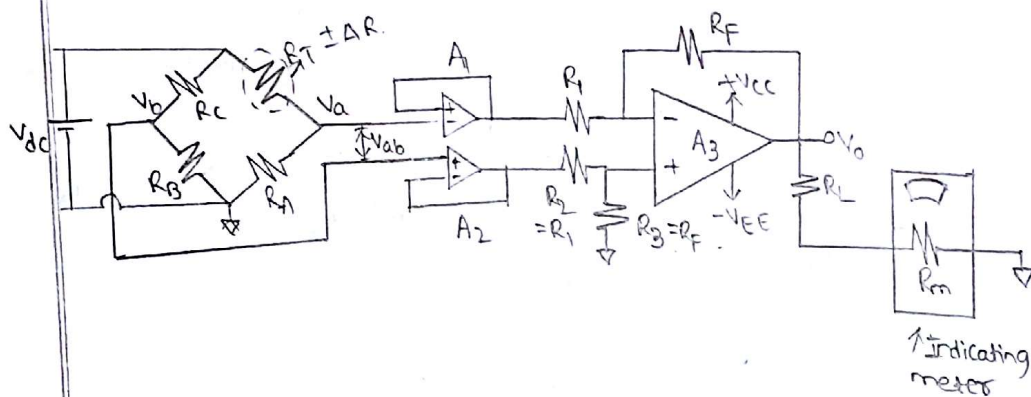
\*5. low  $\phi$ p Impedance.



CIRCUIT-DIAGRAMS:-



\* Instrumentation amplifier using transducer Bridge:-



\* As shown in the above circuit as resistive transducer i.e., the resistance changes as a function of some physical Energy is connected in arm of bridge with a small circle around it.

\* where  $R_T$  is Resistance of the transducer &  $\Delta R$  is the change in the Resistance.

\* For a balance bridge no change in " $R_T$ " value then  $V_a = V_b$ .

$$* \frac{V_{dc} \cdot R_A}{R_A + R_T} = \frac{V_{dc} \cdot R_B}{R_B + R_C}$$

$$\Rightarrow \frac{R_A}{R_T} + 1 = \frac{R_B}{R_C} + 1$$

$$\Rightarrow \frac{R_A}{R_T} = \frac{R_B}{R_C}$$

\*  $V_A$  &  $V_B$  Voltages are given to op-amp ( $A_3$ ) then.

$$V_o = -\frac{R_F}{R_1} \cdot V_{ab} = -\frac{R_F}{R_1} [V_A - V_B] \rightarrow \textcircled{1}$$

\* If there is a change in resistance then

$$V_A = V_{dc} \cdot \frac{R_A}{R_A + R_T \pm \Delta R} ; V_B = V_{dc} \cdot \frac{R_B}{R_B + R_C}$$

$$\Rightarrow V_{ab} = V_A - V_B = V_{dc} \left[ \frac{R_A R_B + R_A R_C \mp R_B R_A - R_T R_B \mp \Delta R R_B}{(R_A + R_T \pm \Delta R)(R_B + R_C)} \right]$$

$$\text{If } R_A = R_B = R_C = R_T = R.$$

$$\Rightarrow V_{ab} = -V_{dc} \left[ \frac{\pm \Delta R R}{(2R \pm \Delta R)(2R)} \right]$$

$$\Rightarrow V_{ab} = -V_{dc} \left[ \frac{\pm \Delta R}{2(2R \pm \Delta R)} \right]$$

\* o/p voltage

$$V_o = -\frac{R_F}{R_1} \times -V_{dc} \left[ \frac{\pm \Delta R}{2(2R \pm \Delta R)} \right]$$

$$\Rightarrow V_o = \frac{R_F}{R_1} \times \frac{(\pm \Delta R)}{4R} \times V_{dc} \quad \{2R \pm \Delta R \approx 2R\}$$

$$\text{i.e., } V_o \propto \Delta R$$

usually we have  $V_o' = +V_{sat}$ , then the o/p of the integrator will be -ve going ramp (because  $V_o = -\frac{1}{R_C} \int V_{in} dt$ ).



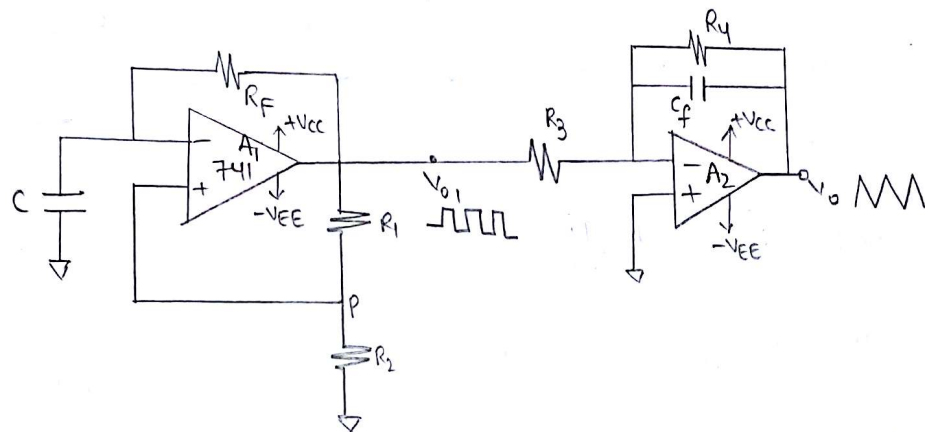
### ③ \* Triangular wave generator :-

\* It can be obtained using Astable m.v & an Integrator.

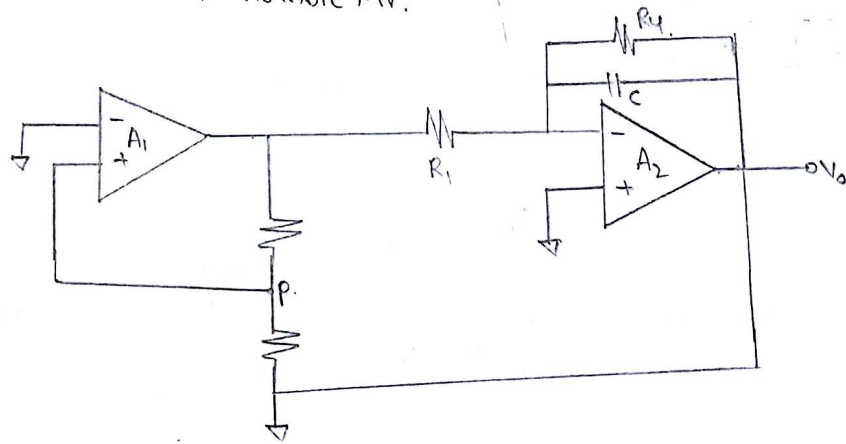
\* The triangular wave generator can also be designed using a Comparator & an integrator.

\* The Comparator is used because it has lesser number of components than Astable m.v.

\*



USING ASTABLE MV.



USING COMPARATOR.

\* The voltage at "p" is compared with "0"

\* Initially we have  $V_{O1} = +V_{sat}$ , then the o/p of the integrator will be -ve going ramp {because  $V_{O2} = -\frac{1}{RC} \int V_{in} dt$ }

\* The circuit consists of Comparator "A<sub>1</sub>" & an integrator "A<sub>2</sub>", the Comparator "A<sub>1</sub>" compares the voltage at "P" with the zero volts at -ve terminal of "A<sub>1</sub>".

\* When the voltage at "P" goes slightly below or above zero volts the op at A<sub>1</sub> is -ve or +ve Sat values respectively.

\* Initially let op of "A<sub>1</sub>" is at +V<sub>sat</sub>, which acts as i/p to A<sub>2</sub>.

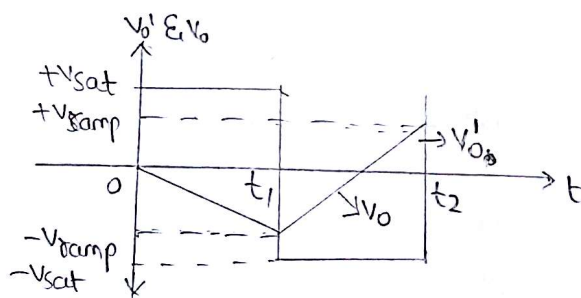
\* Then the op of A<sub>2</sub> is a -ve gng ramp & the voltage divider circuit R<sub>1</sub> & R<sub>2</sub> has +V<sub>sat</sub> of A<sub>1</sub> at one end & -V<sub>ramp</sub> A<sub>2</sub> at other end.

\* At certain point of -ve ramp, the point "P" goes below zero volts, then the op of A<sub>1</sub> will be -ve & op of A<sub>2</sub> will be a +ve gng ramp.

\* Hence, this cycle repeats & a Angular wave is generated.

\* Note:- Angular & Square wave freq's are same.

\* The amplitude of Angular wave decreases as frequency increases.



$$f = \frac{1}{4R_2R_1C_1}$$



\* at P we get  $I_1 = I_2$ .

$$\Rightarrow \frac{+V_{sat} - 0}{R_3} = \frac{0 - V_o}{R_2} = \frac{0 - (-V_{ramp})}{R_2}$$

$$\Rightarrow \frac{V_{sat}}{R_3} = \frac{V_{ramp}}{R_2}$$

$$\Rightarrow -V_{ramp} = -\frac{R_2}{R_3} \cdot V_{sat} \rightarrow (1)$$

Similarly for  $V_o' = -V_{sat}$  then  $V_o = +V_{ramp}$ .

$$\Rightarrow \frac{-V_{sat} - 0}{R_3} = \frac{0 - (+V_{ramp})}{R_2}$$

$$\Rightarrow V_{ramp} = \frac{R_2}{R_3} \cdot V_{sat} \rightarrow (2)$$

\* P-P Voltage is

$$V_o(p-p) = (1) - (2) = \frac{2R_2}{R_3} \cdot V_{sat} \rightarrow (3)$$

\* The time, the o/p swings from  $-V_{ramp}$  to  $+V_{ramp}$  is Equal to  $T/2$ .

\* The time period can be calculated by using Integrator op Equation

$$\Rightarrow V_o(p-p) = \frac{-1}{R_1 C_1} \int_0^{T/2} -V_{sat} \cdot dt \quad \left\{ \begin{array}{l} \text{at } V_o' = -V_{sat} \\ V_{o4} = V_o(p-p) \end{array} \right\}$$

$$\Rightarrow V_o(p-p) = \frac{V_{sat}}{R_1 C_1} \left[ \frac{T}{2} \right] \rightarrow (4)$$

\* From (3) & (4) we get.

$$\Rightarrow \frac{V_{sat}}{R_1 C_1} \cdot \frac{T}{2} = \frac{2R_2}{R_3} \cdot V_{sat}$$

$$\Rightarrow T = \frac{4R_2 R_1 C_1}{R_3}$$

$$\Rightarrow f = \frac{R_3}{4R_2 R_1 C_1}$$