

# Laboratory 5: Fast Fourier Transform

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## 5.1 Introduction

The last laboratory covers the Discrete Fourier Transform (DFT). This laboratory will continue the discussion of the DFT and will introduce the Fast Fourier Transform (FFT). This lab will mainly contain the following contents:

- Continuation of DFT Analysis
- The Fast Fourier Transform Algorithm
- My explorations for FFT Algorithm

Through this lab, we could have a deeper understanding of principles of FFT algorithm. In order to show the advantages of FFT, we also involve the computational complexity to make a comparison between the conventional DFT calculated method (DFTsum in the previous lab) and FFT method.

## 5.2 Continuation of DFT Analysis

### 5.2.1 Shifting the Frequency Range

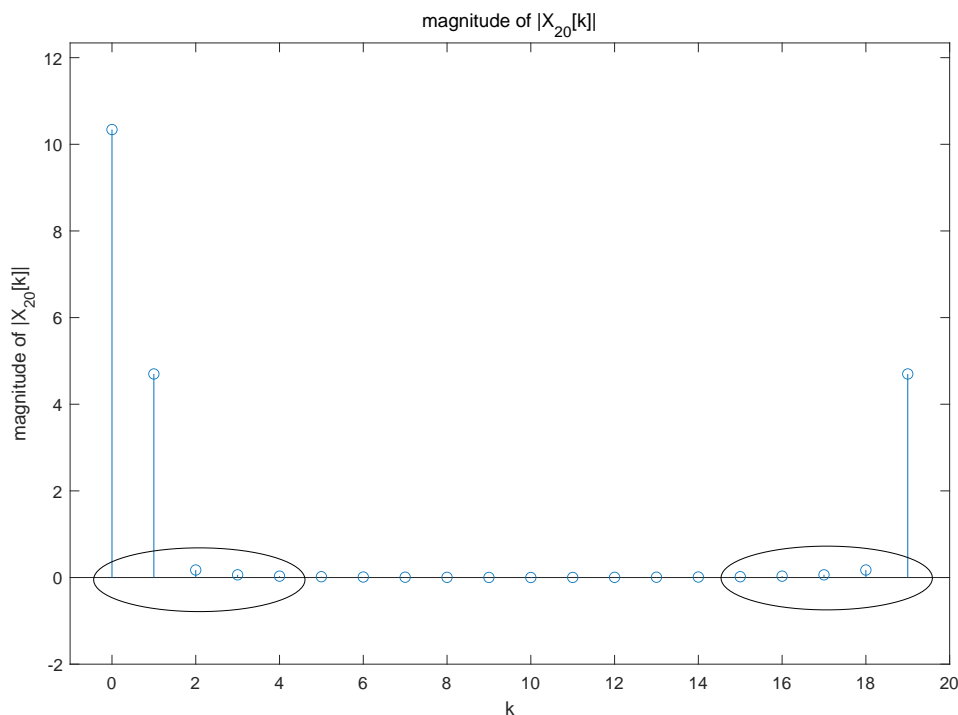
A. Plot of magnitude of  $|X_{20}[k]|$  versus  $k$  (DFTsum of hamming window with  $N = 20$ ).

- MATLAB codes:

q5\_2\_1.m

```
clear;
% part 1
N = 20;
k = 0:N-1;
x = hamming(20);
X20 = DFTsum(x);
figure(1);
stem(k, abs(X20)), xlabel('k'), ylabel('magnitude of |X_{20}[k]|'),
    title('magnitude of |X_{20}[k]|');
xlim([-1, 20]); ylim([-2, max(abs(X20))+2]);
% part 2
figure(2);
[X, w] = DTFTsamples(x);
figure(2);
stem(w, abs(X)), xlabel('\omega(rad/sample)'), ylabel('magnitude of
    DTFT samples of x[n]'), title('magnitude of DTFT samples of x[
    n]');
xlim([w(1)-0.1, w(N)+0.1]); ylim([-2, max(abs(X))+2]);
```

- Running Result:

Figure 1: Plot of magnitude of  $|X_{20}[k]|$  versus  $k$ 

- Analysis:
  - The regions of low frequency component is shown in the circles of above figures;
  - There are two disadvantages of this plot:
    1. The plot is against  $k$  rather than  $\omega$ .
    2. The arrangement of frequency samples goes against  $[0, 2\pi]$  rather than  $[-\pi, \pi]$ .

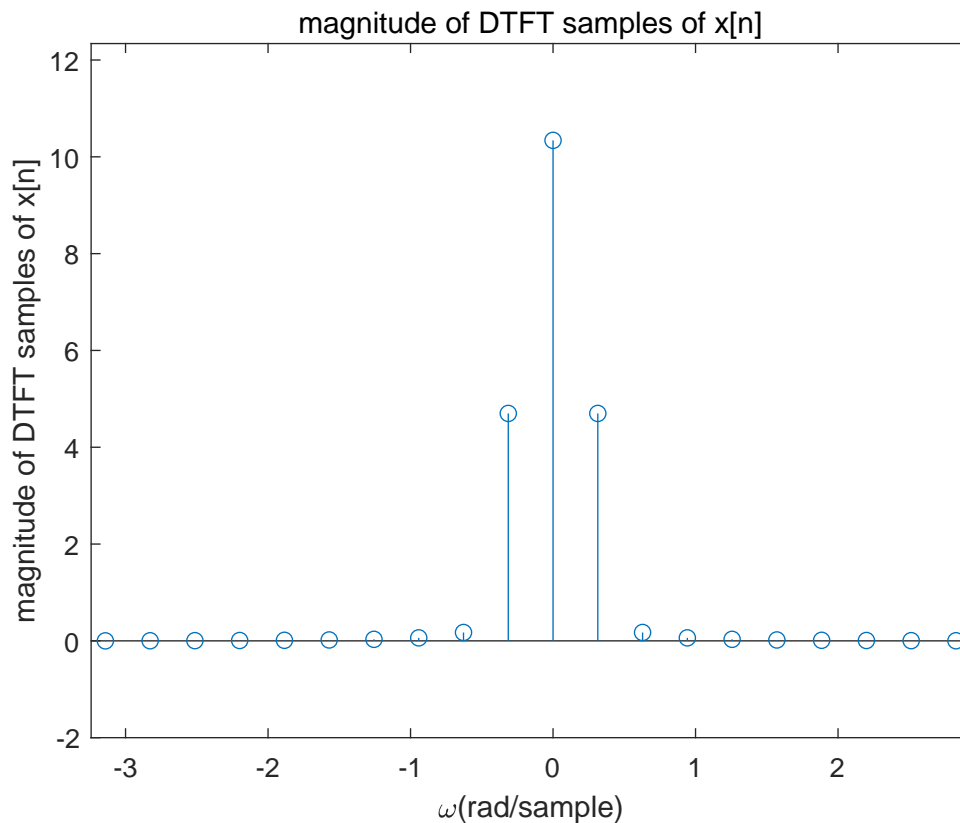
### B. Plot of DTFTsamples.

- MATLAB codes for DTFTsamples.m:

#### DTFTsamples.m

```
function [X,w] = DTFTsamples(x)
N = length(x);
k = 0:N-1;
w = 2*pi*k/N;
w(w>=pi)=w(w>=pi)-2*pi;% shift the range from [0,2*pi] to [-pi,pi]
w = fftshift(w);
X = fftshift(DFTsum(x));
display(w); display(X);
end
```

- Plot of DTFTsamples of hamming window with  $N = 20$ .

Figure 2: Plot of DTFT samples of hamming window with  $N = 20$ 

- Analysis:
  - By applying the DTFT samples ( $\omega = 2\pi k/N$ ), the plot is against  $\omega$  now.
  - By applying the algorithm of `fftshift()` for both  $\omega$  and DTFT samples, we could achieve the arrangement of frequency goes against  $[-\pi, \pi]$ .

### 5.2.2 Zero Padding

- Finite duration signal considered in this section:

$$x[n] = \begin{cases} \sin(0.1\pi n), & 0 \leq n \leq 49 \\ 0, & \text{otherwise} \end{cases}$$

$\text{length}(x[n]) = N$

- MATLAB codes for this section:

**q5\_2\_2.m**

```
clear;
x = sin(0.1*pi*[0:49]);
N = input("Please specify a range N:\n");
display(N);
x = [x zeros(1,N-50)]; % zero padding;
display(x)
[X,w] = DTFTsamples(x);
stem(w,abs(X)), xlabel('\omega(rad/sample)'), ylabel('magnitude of
DTFT samples of x[n]'), title('magnitude of DTFT samples of x[
n]');
```

- Plot of DTFT samples when  $N = 50$  (No zero paddings):

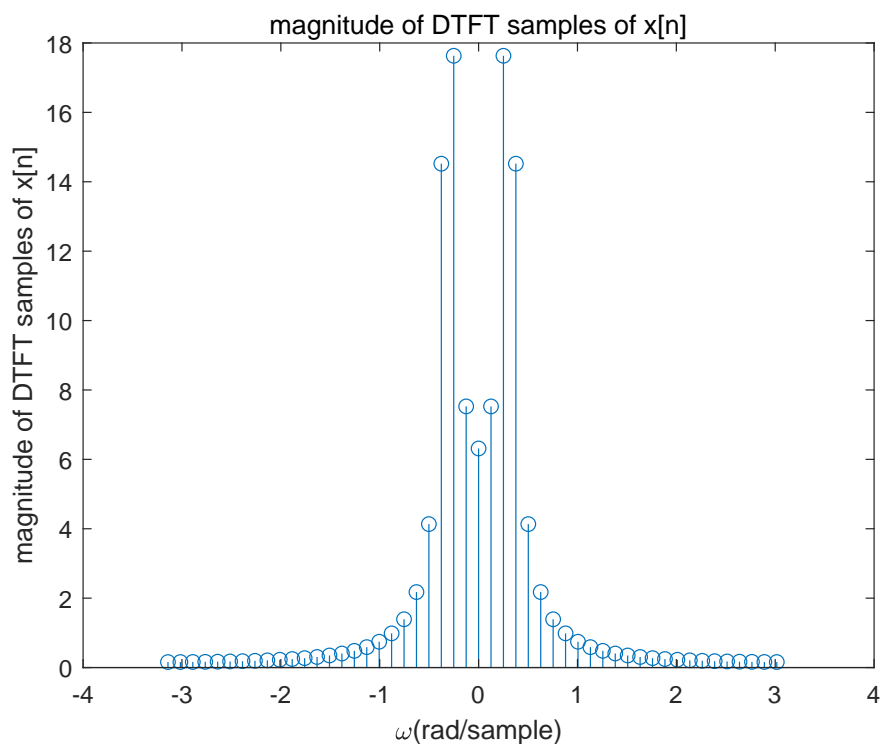


Figure 3: Plot of DTFT samples when  $N = 50$

- Plot of DTFT samples when  $N = 200$ :

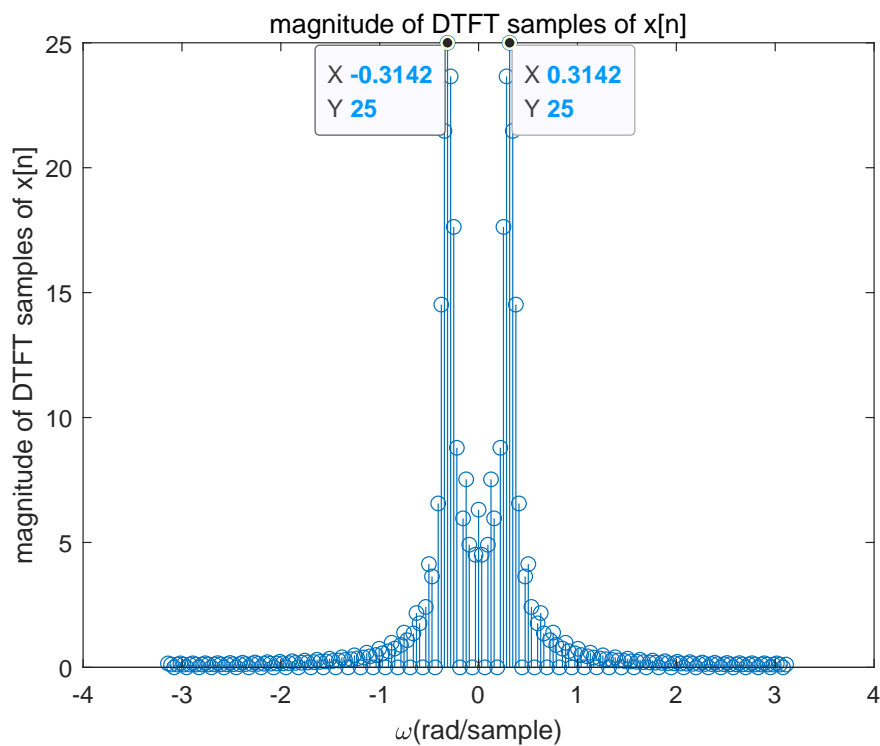


Figure 4: Plot of DTFT samples when  $N = 200$

- Analysis: Evidently, graph of case  $N = 200$  looks more like DTFT than case  $N = 50$ .
- Proof (Why is case  $N = 200$  closer to DTFT?):

– Shape of truncated signal  $x[n] = \sin(0.1\pi n), 0 \leq n \leq 49$  's DTFT:

Evidently,  $x[n]$  is a truncated signal truncated by sine function  $x'[n] = \sin(0.1\pi n), -\infty < n < \infty$  with window function  $w[n] = 1, 0 \leq n \leq 49$ .

It has been shown in the previous laboratory that truncated signal's DTFT has below forms:

$$\begin{aligned} X(e^{j\omega}) &= Ft\{x[n] \times w[n]\} \\ &= X'(e^{j\omega}) \otimes W(e^{j\omega}) \\ &= (\pi\delta(\omega - 0.1\pi) + \pi\delta(\omega + 0.1\pi)) \otimes W(e^{j\omega}) \\ &= \pi W(e^{j(\omega-0.1\pi)}) + \pi W(e^{j(\omega+0.1\pi)}) \end{aligned}$$

Where,  $W(e^{j\omega})$  is a sinc wave such that:

$$\begin{aligned} W(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} w[n]e^{-j\omega n} = \sum_{n=0}^{N-1} e^{-j\omega n} \\ &= \begin{cases} \frac{1-e^{-j\omega N}}{1-e^{-j\omega}}, & \text{for } \omega \neq 0, \pm 2\pi, \dots \\ N, & \text{for } \omega = 0, \pm 2\pi, \dots \end{cases} \\ W(e^{j\omega}) &= \frac{e^{-j\omega N/2} e^{j\omega N/2} - e^{-j\omega N/2}}{e^{-j\omega/2} e^{j\omega/2} - e^{-j\omega/2}} = e^{-j\omega(N-1)/2} \frac{\sin(\omega N/2)}{\sin(\omega/2)} \end{aligned}$$

$\therefore$  DTFT waveform of  $x[n]$  is composed of two sinc function centered at  $-0.1\pi$  and  $0.1\pi$  respectively.

– Already known the waveform of DTFT, let us pay attention to the Figure 3 and Figure 4:

Evidently, Figure 4 provides closer waveform to sinc waveforms. And the center of waveform shown in Figure 4 (0.3142 and -0.3142) just satisfy the previous deduction that two sinc waves centered at  $-0.1\pi$  and  $0.1\pi$ .

- Why these two plots look so differently?

– This is just because of effects of zero-paddings. The more padding zeros there are, the closer to DTFT waveform of DFT is.

– Proof: Just take a look into the equation of DTFT and DFT:

DTFT:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

DFT:

$$X_N[k] = \sum_{n=0}^{N-1} x_N[n]e^{j2\pi kn/N}$$

It could be observed that DFT is frequency sampling of DTFT at  $\omega = 2\pi k/N, k = 0, 1, \dots, N-1$ .

$\therefore$  Case  $N = 200$  provides smaller sampling interval than case  $N = 50$ .

$\therefore$  These two cases look different and case  $N = 200$  is closer to DTFT.

## 5.3 The Fast Fourier Transform Algorithm

### 5.3.1 Implementation of Divide-and-Conquer DFT

- Diagram of Figure dcDFT. (Only consider  $N/2$  points dcDFT)

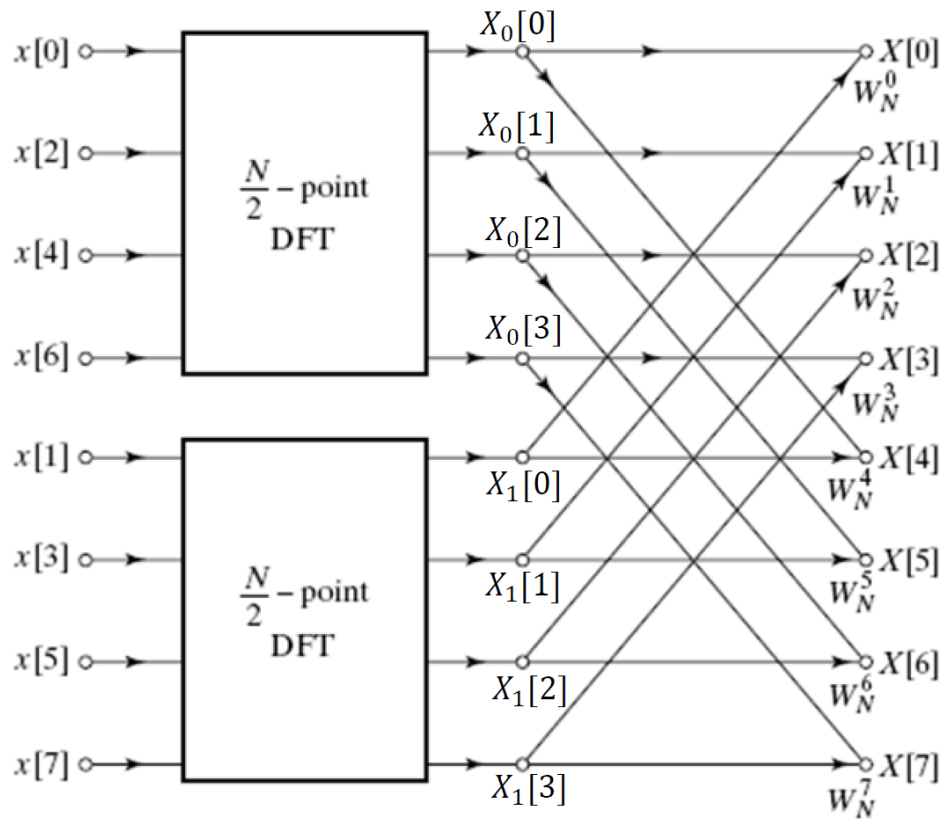


Figure 5: Diagram of dcDFT

Figure 5 could be described as below equation:

$$X[k] = X_0[\langle k \rangle_{N/2}] + W_N^k X_1[\langle k \rangle_{N/2}] \quad (\text{Strategy 1})$$

Where,  $k = 0, 1, \dots, N - 1$ .

- MATLAB code for **dcDFT.m**:

#### dcDFT.m

```
function X = dcDFT(x)
% Using divide and conquer method to calculate FFT
% 2 stages of FFT only, computational complexity: N+N^2/2
% Divide the input into even and odd parts
N = length(x); % Even length N
x0 = x(1:2:N); % Even part of x[n]: x[0], x[2], ..., x[N-2]
x1 = x(2:2:N); % Odd part of x[n]: x[1], x[3], ..., x[N-1]
X0 = DFTsum(x0); X0 = [X0 X0];
X1 = DFTsum(x1); X1 = [X1 X1];
X = X0 + exp(-j*2*pi/N*[0:1:N-1]).*X1;
end
```

- Verifications for dcDFT:

– Test functions:

- \*  $x_1[n] = \delta[n]$ , for  $N = 10$
- \*  $x_2[n] = 1$ , for  $N = 10$
- \*  $x_3[n] = e^{j2\pi n/N}$ , for  $N = 10$

– Test codes:

q5\_3\_1

```
clear;
n = 0:9; N = length(n);
x1 = (n==0);
x2 = ones(1,N);
x3 = exp(j*2*pi*n/N);
k = 0:9;
sgtitle('dcDFT and DFTsum')
subplot(321), stem(n,abs(dcDFT(x1))), xlabel('k'), ylabel('dcDFT(x1)'), title('magnitude of dcDFT(x1)');
subplot(322), stem(n,abs(DFTsum(x1))), xlabel('k'), ylabel('DFTsum(x1)'), title('magnitude of DFTsum(x1)');
subplot(323), stem(n,abs(dcDFT(x2))), xlabel('k'), ylabel('dcDFT(x2)'), title('magnitude of dcDFT(x2)');
subplot(324), stem(n,abs(DFTsum(x2))), xlabel('k'), ylabel('DFTsum(x2)'), title('magnitude of DFTsum(x2)');
subplot(325), stem(n,abs(dcDFT(x3))), xlabel('k'), ylabel('dcDFT(x3)'), title('magnitude of dcDFT(x3)');
subplot(326), stem(n,abs(DFTsum(x3))), xlabel('k'), ylabel('DFTsum(x3)'), title('magnitude of DFTsum(x3)');
```

– Test Result:

## dcDFT and DFTsum

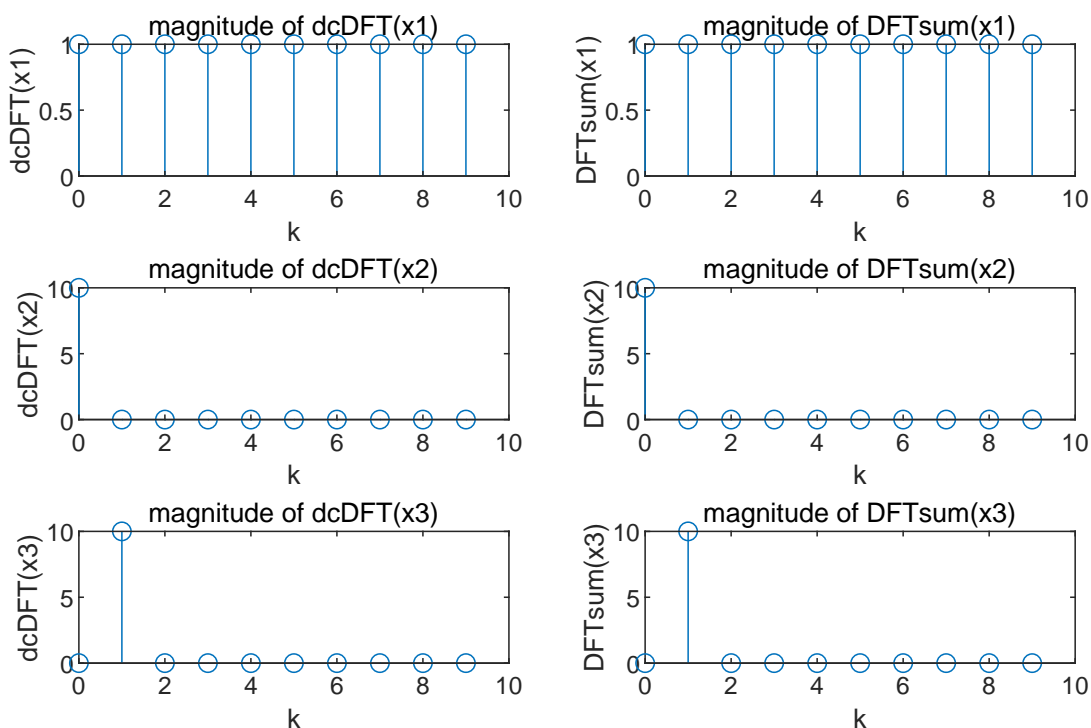


Figure 6: Verifications for dcDFT  
(On the left: dcDFT, on the right: DFTsum)

- Analysis:

- Figure 6 shows that the design of dcDFT is successful.
- The number of multiplies that are required in this approach to computing an point DFT:(Consider a multiply to be one multiplication of real or complex numbers.)

$$N + 2 \times (N/2)^2 = N + N^2/2$$

Proof: According to above Strategy 1, it could be observed that each butterfly needs one multiplication, , which could lead to a stage of  $N$  points that needs  $N$  multiplications. And combining with the condition that  $N/2$ -points DFTsum operation needs  $(N/2)^2$  multiplications, the overall multiplications are  $N + 2 \times (N/2)^2 = N + N^2/2$ .

### 5.3.2 Recursive Divide and Conquer

#### A. 8-point FFT: FFT2,FFT4,FFT8

- Diagram of 8-point FFT:

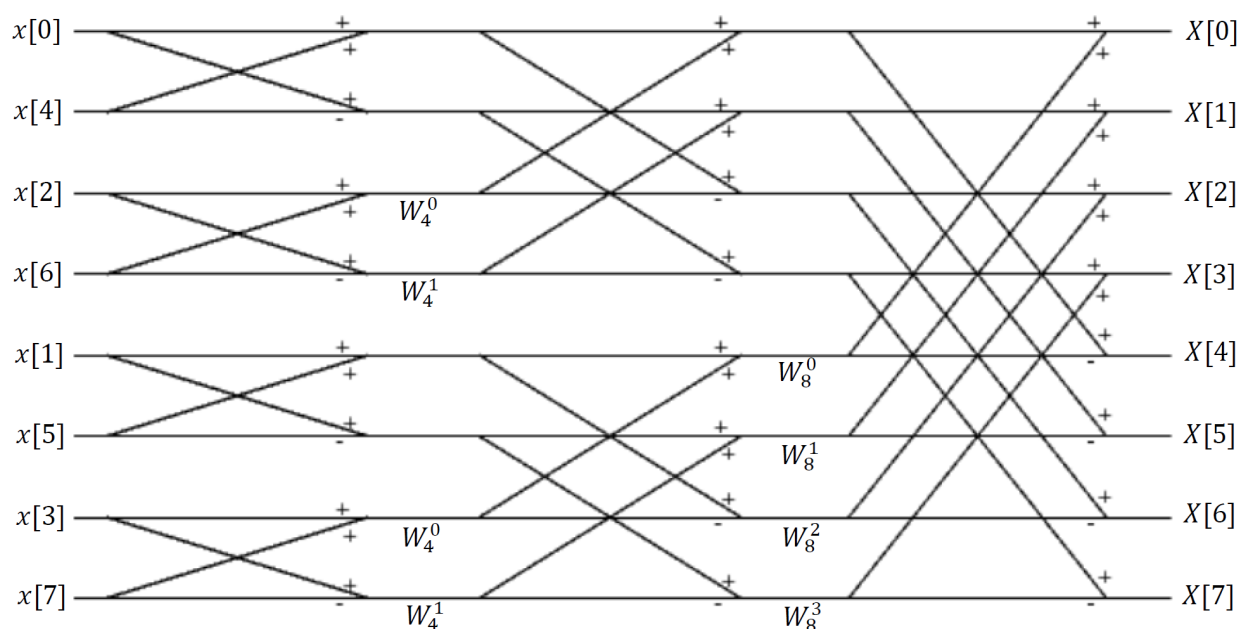


Figure 7: Diagram of 8-point FFT

- FFT design strategy in this section:

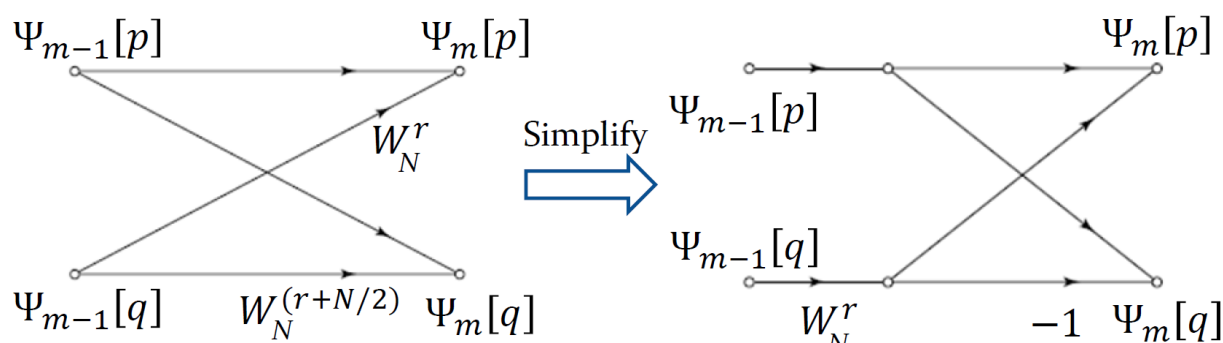


Figure 8: Simplified butterfly computation



In this section, we use the below strategy to simplify the calculation of each butterfly.

$$\begin{aligned}\Psi_m[p] &= \Psi_{m-1}[p] + W_N^r \Psi_{m-1}[q] \\ \Psi_m[q] &= \Psi_{m-1}[p] + W_N^{r+(N/2)} \Psi_{m-1}[q] = \Psi_{m-1}[p] - W_N^r \Psi_{m-1}[q]\end{aligned}\quad (\text{Strategy 2})$$

By applying such strategy, each butterfly just need one multiplication:  $W_N^r \Psi_{m-1}[q]$ . Compared with previous Strategy 1, Strategy 2 is much more efficient that each  $N$ -point stage just needs  $(N/2)$  multiplications.

- Three MATLAB functions:

#### FFT2.m

```
function X = FFT2(x)
% length(x) = 2
X(1) = x(1)+x(2);
X(2) = x(1)-x(2);
end
```

#### FFT4.m

```
function X = FFT4(x)
% length(x) = 4
r = 0:1;
X0 = FFT2([x(1) x(3)]);
X1 = FFT2([x(2) x(4)]);
temp = X1.*exp(-j*2*pi/4*r);
X = [X0+temp X0-temp];
end
```

#### FFT8.m

```
function X = FFT8(x)
% length(x) = 8
r = 0:3;
X0 = FFT4([x(1) x(3) x(5) x(7)]);
X1 = FFT4([x(2) x(4) x(6) x(8)]);
temp = X1.*exp(-j*2*pi/8*r);
X = [X0+temp X0-temp];
end
```

- Output of **FFT8** for the case  $x[n] = 1$ , for  $N = 8$ :

```
» FFT8(ones(1,8))
ans =
8 0 0 0 0 0 0 0
» fft(ones(1,8))==FFT8(ones(1,8))
ans =
1×8 logical 数组
1 1 1 1 1 1 1 1
```

- Other test examples:  $x[n] = \delta[n]$  and  $x[n] = e^{\frac{j2\pi n}{8}}$  for  $N = 8$ .

```

» FFT8([1 zeros(1,7)])
ans =
1 1 1 1 1 1 1
» fft([1 zeros(1,7)])==FFT8([1 zeros(1,7)])
ans =
1×8 logical 数组
1 1 1 1 1 1 1 1

```

```

» x=exp(j*2*pi/8*[0:1:7]);
» FFT8(x)
ans =
-0.0000 + 0.0000i 8.0000 - 0.0000i 0.0000 + 0.0000i 0.0000 + 0.0000i 0.0000 + 0.0000i 0.0000 + 0.0000i
0.0000 + 0.0000i 0.0000 + 0.0000i
» fft(x)
ans =
-0.0000 + 0.0000i 8.0000 - 0.0000i 0.0000 + 0.0000i 0.0000 + 0.0000i 0.0000 + 0.0000i -0.0000 + 0.0000i
0.0000 + 0.0000i 0.0000 + 0.0000i

```

- Comparison between results obtain from **DFTsum.m** and **FFT8.m**:

- $x_1[n] = \delta[n]$ , for  $N = 8$
- $x_2[n] = 1$ , for  $N = 8$
- $x_3[n] = e^{j2\pi n/N}$ , for  $N = 8$
- Test Codes:

#### q5\_3\_2a.m

```

clear
n=0:1:7;N = length(n);
x1 = (n==0);
x2 = ones(1,N);
x3 = exp(j*2*pi*n/N);
k = 0:7;
sgtitle('FFT8 and DFTsum')
subplot(321),stem(n,abs(FFT8(x1))),xlabel('k'),ylabel('FFT8(
x1)'),title('magnitude of FFT8(x1)');
subplot(322),stem(n,abs(DFTsum(x1))),xlabel('k'),ylabel('
DFTsum(x1)'),title('magnitude of DFTsum(x1)');
subplot(323),stem(n,abs(FFT8(x2))),xlabel('k'),ylabel('FFT8(
x2)'),title('magnitude of FFT8(x2)');
subplot(324),stem(n,abs(DFTsum(x2))),xlabel('k'),ylabel('
DFTsum(x2)'),title('magnitude of DFTsum(x2)');
subplot(325),stem(n,abs(FFT8(x3))),xlabel('k'),ylabel('FFT8(
x3)'),title('magnitude of FFT8(x3)');
subplot(326),stem(n,abs(DFTsum(x3))),xlabel('k'),ylabel('
DFTsum(x3)'),title('magnitude of DFTsum(x3)');

```

- Below result shows that the design of **FFT8.m** is right.

## FFT8 and DFTsum

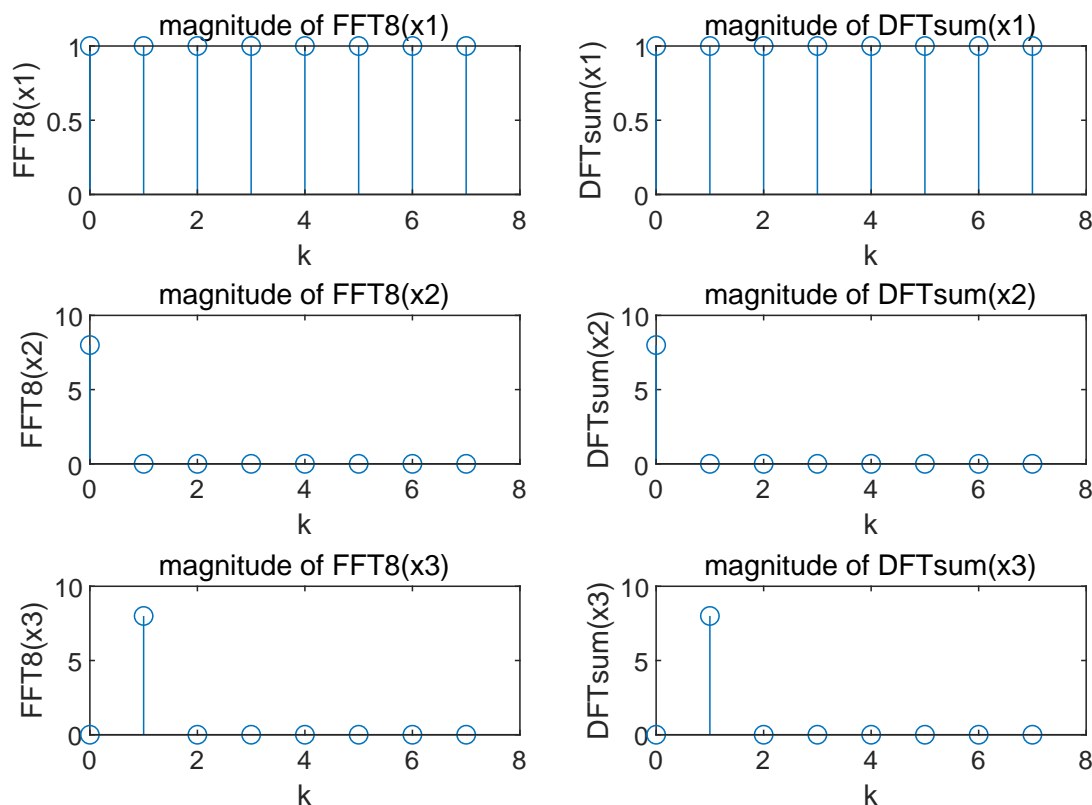


Figure 9: Comparison between results obtain from **DFTsum.m** and **FFT8.m**  
(On the right: FFT8, On the left: DFTsum)

• Analysis:

- The total number of multiplies by twiddle factors required for 8-point FFT (A multiply is a multiplication by a real or complex number.) :  $\frac{8}{2} \times \log_2 8 = 12$ . (Without considering  $W_N^0 = 1$  and  $W_N^{N/2} = -1$  in above Strategy 2 here.)
- For  $N = 2^p$  point FFT, the overall multiplications are  $\frac{N}{2} \log_2 N = p2^{p-1}$ . (Without considering  $W_N^0 = 1$  and  $W_N^{N/2} = -1$  in above Strategy 2 here.)

Compared to number of multiplies required for direct implementation ( $N^2 = 4^p$ ), when  $p = 10$ , FFT just needs  $10 \times 2^9 = 5120$  multiplications while direct method needs  $4^{10} = 1048576$  multiplications.

- Actually, the number of multiplications could be even lower if we consider  $W_N^0 = 1$  and  $W_N^{N/2} = -1$  in above Strategy 2. In brief, the number of multiplications could be expressed as  $\frac{N}{2} (\log_2 N - 2) + 1$ .

Proof:

$$\begin{aligned}
& \underbrace{\frac{N}{2} \log_2 N}_{\text{previous multiplications}} - \underbrace{\left( \frac{N}{2} + \frac{N}{4} + \cdots + 1 \right)}_{m=\log_2 N \text{ terms}} \\
&= \frac{N}{2} \log_2 N - \sum_{\text{stage } i=1}^{\log_2 N} \left( \frac{1}{2} \right)^{i-1} \left( \frac{N}{2} \right) \\
&= \frac{N}{2} \log_2 N - \frac{\frac{N}{2} \left( 1 - \left( \frac{1}{2} \right)^{\log_2 N} \right)}{1 - \frac{1}{2}} \\
&= \frac{N}{2} (\log_2 N - 2) + 1
\end{aligned}$$

Now consider the new computational complexity, when  $p = \log_2 N = 10$ , number of multiplications needed is  $2^{p-1}(p-2) + 1 = 4097$ . (My design doesn't consider this case because of MATLAB's feature of matrix computing, though it is feasible.)

## B. N-point FFT: `fft_stage.m`

- In this section, we need to write a recursive method to implement the general case of fft. And the idea of the recursive algorithm is shown below:

---

### Algorithm 1 Fast Fourier Transform `fft_stage(x)`

---

**Require:**  $N = 2^p$ ,  $p \in \mathbb{N}$

**Ensure:**  $N = \text{length}(x)$

```

if  $N = 2$  then
    do 2-pt DFT
else
    calling fft_stage to compute the  $(N/2)$ -point DFTs
end if

```

---

- MATLAB code for `fft_stage.m`:

#### `fft_stage.m`

```

function X = fft_stage(x)
N = length(x); %  $N=2^m$  in this function
if N == 2
    X(1) = x(1)+x(2);
    X(2) = x(1)-x(2);
else
    r = 0:1:N/2-1;
    xeven = x(1:2:N);
    xodd = x(2:2:N);
    X0 = fft_stage(xeven);
    X1 = fft_stage(xodd);
    temp = X1.*exp(-j*2*pi/N*r);
    X = [X0+temp X0-temp];
end
end

```

- Test three previous functions again:

- $x_1[n] = \delta[n]$ , for  $N = 8$
- $x_2[n] = 1$ , for  $N = 8$

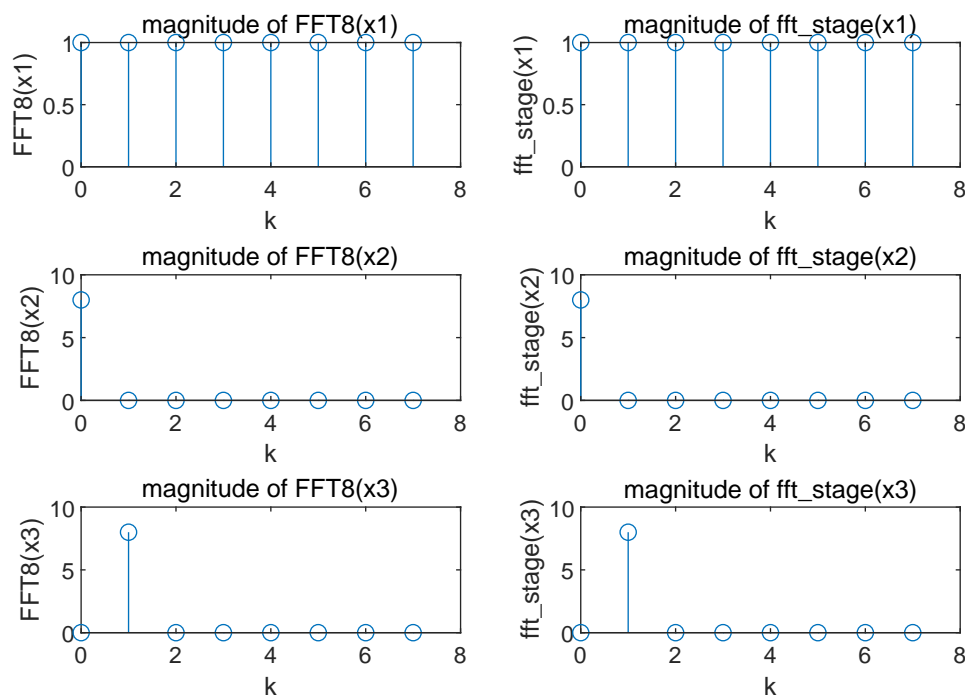
- $x_3[n] = e^{j2\pi n/N}$ , for  $N = 8$
- Test codes:

## q5\_3\_2b.m

```
clear
n=0:1:7;N = length(n);
x1 = (n==0);
x2 = ones(1,N);
x3 = exp(j*2*pi*n/N);
k = 0:7;
sgtitle('FFT8 and fft_stage')
subplot(321),stem(n,abs(FFT8(x1))),xlabel('k'),ylabel('FFT8(x1)'),title('magnitude of FFT8(x1)');
subplot(322),stem(n,abs(fft_stage(x1))),xlabel('k'),ylabel('fft_stage(x1)'),title('magnitude of fft_stage(x1)');
subplot(323),stem(n,abs(FFT8(x2))),xlabel('k'),ylabel('FFT8(x2)'),title('magnitude of FFT8(x2)');
subplot(324),stem(n,abs(fft_stage(x2))),xlabel('k'),ylabel('fft_stage(x2)'),title('magnitude of fft_stage(x2)');
subplot(325),stem(n,abs(FFT8(x3))),xlabel('k'),ylabel('FFT8(x3)'),title('magnitude of FFT8(x3)');
subplot(326),stem(n,abs(fft_stage(x3))),xlabel('k'),ylabel('fft_stage(x3)'),title('magnitude of fft_stage(x3)');
```

- Below result shows that the design of q5\_3\_2b.m is right.

## FFT8 and fft\_stage

Figure 10: Comparison between results obtain from `fft_stage.m` and `FFT8.m`

## 5.4 My explorations for FFT algorithm

- In the section 5.3.2, FFT algorithm of  $N = 2^p$  was proposed. But when it comes to the case of  $N \neq 2^p$ , what should we do?

Actually, function `fft()` in MATLAB just use DFTsum algorithm to calculate it directly. It is inefficient when  $N$  is a really large number. What we should is to choose the nearest power-of-two FFT algorithm for implementing a DFT by zero-padding a sequence into an  $N$ -point. The idea of zero-padding is shown below:

---

### Algorithm 2 Fast Fourier Transform `fft_stage(x)`

---

**Ensure:**  $N = \text{length}(x)$

```

if  $N = 2^p$  then
    if  $N = 2$  then
        do 2-pt DFT
    else
        calling fft_stage to compute the  $(N/2)$ -point DFTs
    end if
else
     $x = \text{zero-padding}(x)$ 
    fft_stage( $x$ )
end if

```

---

- After searching the Internet, I choose a famous zero-padding operation for OFDM and implement the MATLAB code below:

#### `fft_stagep.m`

```

function X = fft_stagep(x)
N = length(x);
if log2(N) == fix(log2(N)) %  $N = 2^m$ 
    if N == 2
        X(1) = x(1)+x(2);
        X(2) = x(1)-x(2);
    else
        r = 0:2:N-1;
        xeven = x(1:2:N);
        xodd = x(2:2:N);
        X0 = fft_stage(xeven);
        X1 = fft_stage(xodd);
        temp = X1.*exp(-j*2*pi/N*r);
        X = [X0+temp X0-temp];
    end
else %  $N \neq 2^m$ 
    Nzeros = 2^(fix(log2(N))+1)-N-1;
    x = [x(1:fix(N/2)) zeros(1,Nzeros) x(fix(N/2)+1:N)]; %
        insert null tones;
    x = [0 x]; % DC subcarrier
    X = fft_stagep(x);
end
end

```

## 5.5 Summary & Experience

- In this lab, we mainly discuss two things:
  - The relationship between the DFT and DTFT: DFT is the frequency sampling of DTFT.
  - Using the method of divide and conquer to recursively implement the Fast Fourier Transform (FFT), which could lead a change of number of multiplications from  $O(N^2)$  to  $O(N \log_2 N)$ .
- All these efforts makes me have a deeper understanding to the principle of DFT and FFT, especially the part of FFT algorithm design.
- For the further explorations for FFT algorithm, we can focus more on the zero-padding of FFT.

## Appendix

DFTsum.m

```
function X = DFTsum(x)
%  $X_N[k] = \sum_{n=0}^{N-1} x[n] e^{-j * 2 * \pi * k * n / N}$ 
%  $X(k) = \sum_{n=1}^N x(n) e^{-j * 2 * \pi * (k-1) * (n-1) / N}$ ,  $1 \leq k \leq N$ 
N = length(x);
X = zeros(1,N);
for k=1:N
    for n=1:N
        X(k) = X(k) + x(n) * exp(-j * 2 * pi * (k-1) * (n-1) / N);
    end
end
end
```