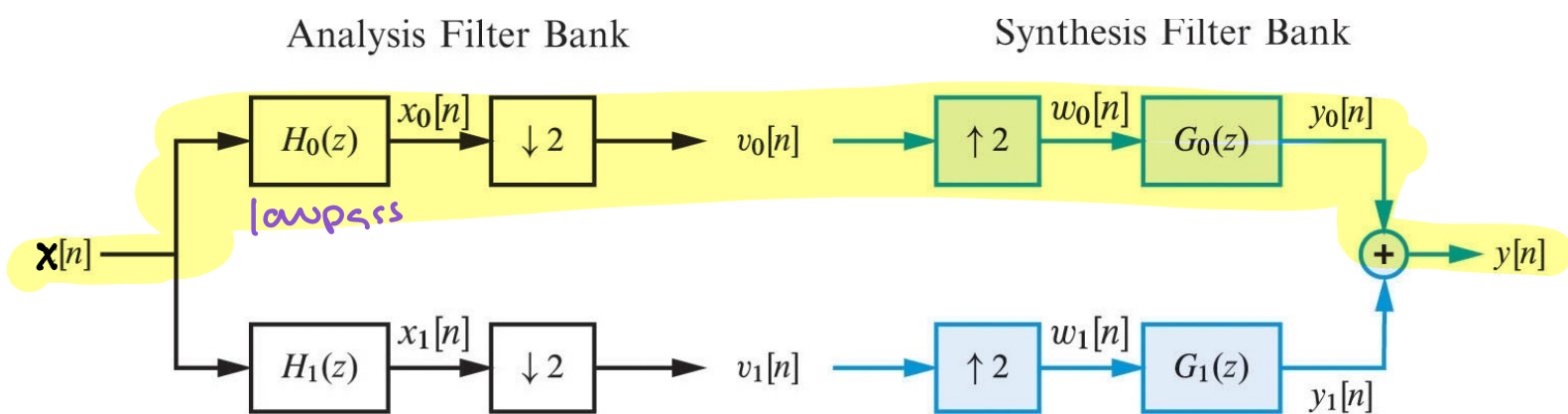


Chap 12.4.3



We want to design a perfect reconstruction two band filter bank

We need to design

$$H_0(z), H_1(z), G_0(z), G_1(z)$$

Frequency response

$$H_0(e^{j\omega}), H_1(e^{j\omega}), G_0(e^{j\omega}), G_1(e^{j\omega})$$

Impulse response

$$h_0[n], h_1[n], g_0[n], g_1[n]$$

We can choose to design either FIR or IIR filters. Because we don't want to have to worry about stability, we will choose a design method where all the filters are FIR

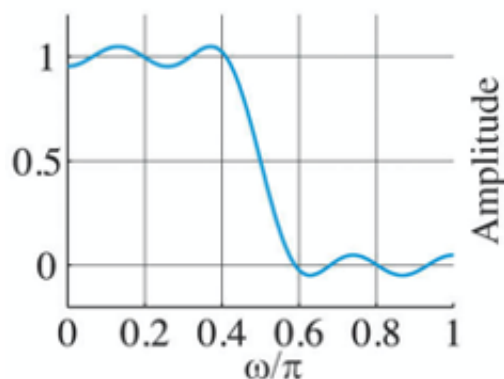
We will use design conjugate quadrature filters (CQFs)

$H(e^{j\omega})$ and $G(e^{j\omega})$ must be odd-order filters !!

Recall $R_o(z) \equiv z^{n_D} \underbrace{H_o(z)}_{\substack{\text{M order filter,} \\ \text{M is odd}}} \underbrace{G_o(z)}_{\substack{\text{M order filter,} \\ \text{M is odd}}}$

$R_o(z)$ has order $2M$

$R_o(e^{j\omega})$ is a half band filter, meaning that it is a lowpass filter whose cutoff frequency is 0.5π



Algorithm

1. Design a lowpass zero-phase half-band FIR filter $R_0(z)$ of order $2M$, where the number M *must* be an *odd* integer
2. If the minimum value δ_{\min} of the real and even function $R_0(e^{j\omega})$ is negative, form a nonnegative function as

$$R_+(e^{j\omega}) = R_0(e^{j\omega}) + |\delta_{\min}| \geq 0$$

This is equivalent to adding the value $|\delta_{\min}|$ to the sample $r_0[0]$, that is,

$$r_+[n] = r_0[n] + |\delta_{\min}| \delta[n]$$

3. Scale $R_+(z)$ so that the frequency response is equal to $1/2$ at $\omega = \pi/2$,

$$R(z) = \frac{1/2}{1/2 + |\delta_{\min}|} R_+(z)$$

4. Determine the minimum-phase filter $H(z)$ by solving the spectral factorization problem $R(z) = H(z)H(z^{-1})$

This minimum-phase filter $H(z)$ can then be used to find $H_0(z)$, $H_1(z)$, $G_0(z)$, $G_1(z)$ with the following mapping

$$\begin{aligned} h_0[n] = h[n] & \xleftrightarrow{\text{DTFT}} H_0(e^{j\omega}) = H(e^{j\omega}), \\ h_1[n] = (-1)^n h[M-n] & \xleftrightarrow{\text{DTFT}} H_1(e^{j\omega}) = -H(e^{-j\omega})e^{-j\omega M}, \\ g_0[n] = 2h[M-n] & \xleftrightarrow{\text{DTFT}} G_0(e^{j\omega}) = 2H(e^{-j\omega})e^{-j\omega M}, \\ g_1[n] = -2(-1)^n h[n] & \xleftrightarrow{\text{DTFT}} G_1(e^{j\omega}) = -2H(e^{-j\omega}). \end{aligned}$$

How to do spectral factorization:

We want to find $H(z)$ such that

$$R(z) = H(z)H(z^{-1})$$

If $H(z)$ has a zero at z_k , $H(z^{-1})$ has a zero at $\frac{1}{z_k^*} \leftarrow$ complex conjugate

We only work with FIR filters so we don't have to worry about poles

1) Find the poles and zeros of $R(z)$

2) Keep all the zeros inside the unit circle and only one zero from each pair of zeros on the unit circle

Then you can use the function `zp2tf` (with gain $k=1$) to get the system function $H(z)$