In real-world system, we may need to convert a signal from one sampling rate to another (for example, store 10 instead of 20 paints to save memory)

Computation of a sequence $X_0[n] \equiv X_c(nT_0)$ from the known Sequence $X[n] = X_c(nT)$ for $T_0 \neq T$ without reconstructing $X_c(t)$ is called resampling or sampling rate change

T= original sampling period
To= new sampling period

3 cases to consider

1) To = DT, D is an integer (To=3T)

New sampling period is longer ord integer multiple

New sampling frequency is slower

= This is known as downsampling or

sampling rate compression

2)
$$T_0 = \frac{T}{I}$$
, I is an integer $(T_0 = \frac{T}{5})$

New sampling period is shorter

New sampling Prepuency is faster

= This is known as interpolation

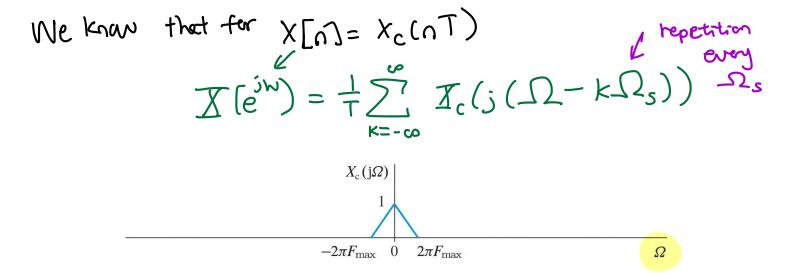
3)
$$T_0 = T(\frac{D}{I})$$
, Dad I are integer, $\frac{D}{I}$ is a rational number

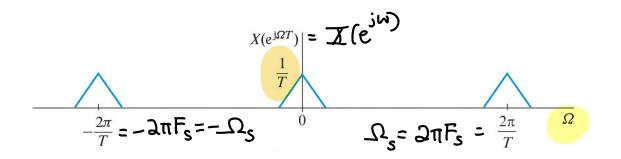
Case 1, sampling frequency decrease by an integer factor $T_o = DT$

 $X[n] = X_c(nT)$ original signal sampled at T period $X_D[n] = X_c(nDT)$

This means $X_D[n] = X[nD]$. This is called dawnsampling. We will see that after times, just dawnsampling is not enough.

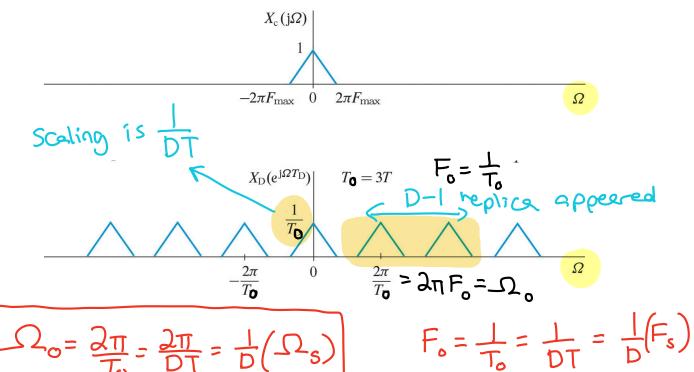
We went to understand how the down sampling affects the frequency domain





And for
$$X_D[n] = X_c(nDT)$$

$$X_D(e^{iw}) = \frac{1}{T_0} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_0)) = \frac{1}{DT} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_0))$$



However, things get a bit confusing if we want to plot the DTFT with a normalized frequency axes. Which sampling frequency do we normalize by Fs or Fo?

normalized frequency normalized by Fs $f = \frac{E}{E_c}$

normalized frequency normalized by Fo

$$f = F_o = \frac{F}{L(F_s)} = \frac{DE}{F_s}$$

Let us choose to define the normalized frequency as

$$W = 2\pi f = 2\pi \left(\frac{F}{F_0}\right) = \Omega T_0$$

Then

$$\frac{1}{N_{D}}(e^{iw}) = \frac{1}{D} \frac{D-1}{N_{D}} \times \frac{(w-2\pi m)}{D}$$

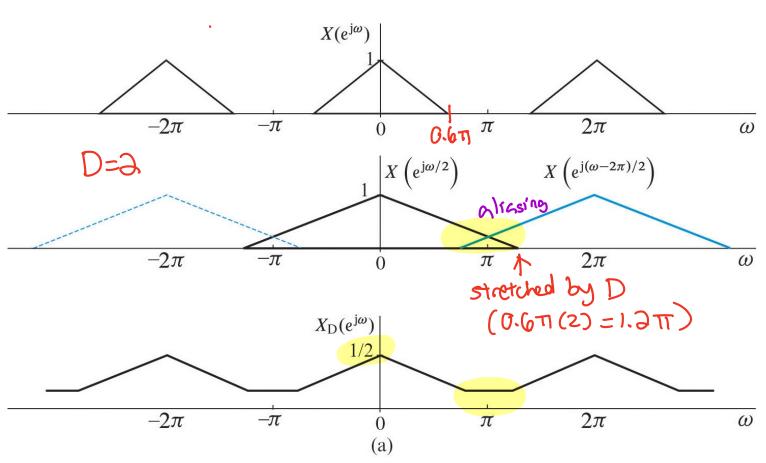
$$\frac{1}{N_{D}}(e^{iw}) = \frac{1}{D} \frac{(w-2\pi m)}{N_{D}}$$

How are the 2 DTFT related

$$\underline{X}_{D}(e^{j\pi}) = \underline{J} \underbrace{\sum_{m=0}^{D-1} \underline{X}(e^{j(\underline{\pi}-2\underline{\pi}m)})}$$

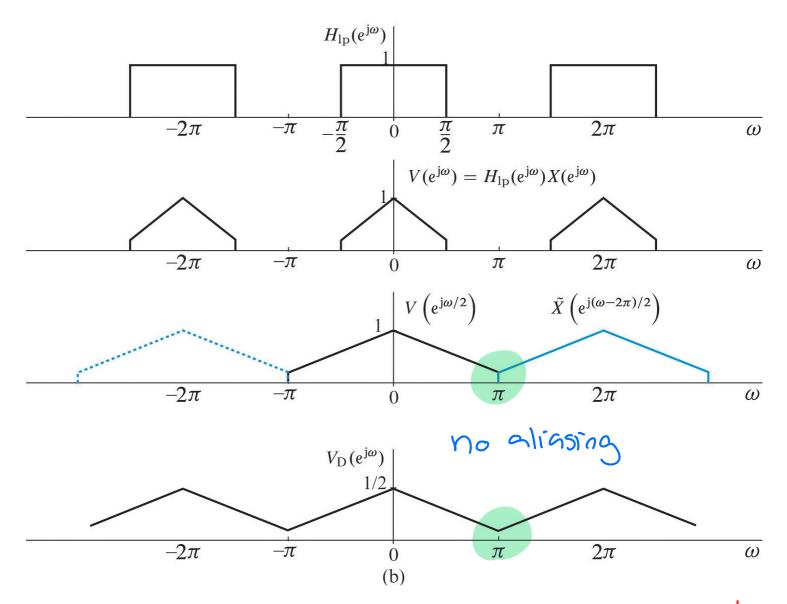
one way to think of this is that
$$X_0(e^{i\omega})$$
 Stretches $X(e^{i\omega})$

Because of the "stretching" and replicants, integer factor D has to be chosen corefully to avoid aliasing



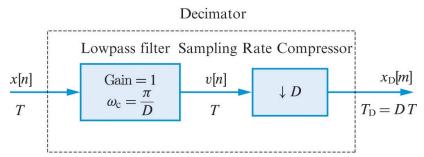
To avoid alising the maximum frequency component of x[n] has to be $w \leq \frac{\pi}{D}$

If this is not true, the XInT need to be passed through a low pass filter



In reality, to avoid aliasing, we need to lawpass filter XInI before dawnsambing.

lawpass fitter + downsampling = decimations



Of cause in practice, ue con not use an ideal

laupass filter. Therefore, unlike downsampling, decimater output XD[m] do not perfectly reproduce the valves of X[n] When do you known down sampling is Sufficient? avoided if Aliasing can be I(ein) = 0, WH S/W/ET)

Theximum frequency

component of XED then $W_s = \frac{2\pi}{D} \ge 2W_H$ $W_H = 0.6\pi$ In the previous example $2\pi \geq a(0.6\pi) = 1.2\pi$ If D=2, we see that $T\neq 1.2T$. Therefore, my downsampling of previous example will introduce aliasing! we reed to decimate!