Difficult Things

- There are four things that happen that complicate the process
 - **1)** $M \ge N$
 - 2) Repeated real roots of the form $(s-p_{\nu})^m$
 - 3) Complex roots
 - 4) Exponentials in H(s)

UMBC CMPE323 Signals and Systems
© EFCLaBerge, S. J. Kim 2014 All rights reserved
Portions © Addison-Wesley Publishing, Inc

1

Let's Digress just a little

• What is the Laplace Transform of $\delta(t)$?

$$\Delta(s) = \int_{-\infty}^{\infty} \delta(\tau) e^{-s\tau} d\tau = e^{-s\tau} \Big|_{\tau=0} = 1 \quad (!!)$$

- So the Laplace Transform of a delta function is a constant!...
- ...and vice-versa
- So $X_1(s) = 3 \Rightarrow 3\delta(t)$

$$X_2(s) = \frac{-19.5(s - 1.487)}{(s + 5)(s + 2)} = -19.5\left(\frac{R(-5)}{s + 5} + \frac{R(-2)}{s + 2}\right) \quad s > -2$$

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

2

And a more complicated case

$$H(s) = \frac{A}{s-r} + \frac{C+jD}{s-(\sigma+j\omega)} + \frac{C-jD}{s-(\sigma-j\omega)}, \text{ Re}[s] > \max(r,\sigma)$$

Causal!

$$h(t) = Ae^{rt}u(t) + (C + jD)e^{(\sigma + j\omega)t}u(t) + (C - jD)e^{(\sigma - j\omega)t}u(t)$$

$$= Ae^{rt}u(t) + Ce^{\sigma t}(e^{j\omega t} + e^{-j\omega t})u(t) + jDe^{\sigma t}(e^{j\omega t} - e^{-j\omega t})u(t)$$

$$= Ae^{rt}u(t) + 2Ce^{\sigma t}\cos(\omega t)u(t) + 2De^{\sigma t}\sin(\omega t)u(t)$$

 $= \left(Ae^{rt} + e^{\sigma t} \left(E\cos\omega t + F\sin\omega t\right)\right) u(t)$

= exponential term + second order solution from CMPE306!

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc -

Laplace Transforms III: Properties of the Laplace Transform

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

Summary

- LCCDE → LT: Put in standard form, fine the a, b coefficients, write the rational form, factor & find residuals, convert 1st and 2nd order terms to time domain based on RoC.
- $X(s) \rightarrow x(t)$: Use the synthesis equation

$$x(t) = \frac{1}{2\pi j} \int_{-\infty}^{\infty} X(s)e^{st} ds = \frac{1}{2\pi j} \int_{\sigma_{-i\infty}}^{\sigma_{+j\infty}} X(s)e^{st} ds$$

- ...which can be complicated because of the integration in the complex plane
- ...but there are properties that make things simpler

UMBC CMPE323 Signals and Systems
© EFCLaBerge, S. J. Kim 2014 All rights reserved
Portions © Addison-Wesley Publishing, Inc

_

Key properties

• Linearity: Of course it is linear!

$$\mathcal{L}(Ax_1(t) + Bx_2(t)) = AX_1(s) + BX_2(s), R' \supset R_1 \cap R_2$$

Time Shift

$$\mathcal{L}(x(t-t_0)) = e^{-st_0}X(s), R' = R$$

Modulation

$$\mathcal{L}(e^{s_0 t} x(t)) = X(s - s_0), \quad R' = R + \text{Re}[s_0]$$

Scaling (remember time scaling!)

$$\mathcal{L}(x(at)) = \frac{1}{|a|} X\left(\frac{s}{a}\right), R' = aR$$

http://lpsa.swarthmore.edu/LaplaceZTable/LaplacePropTable.html

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

6

More Properties

Differentiation in the time domain

$$x(t) = \frac{1}{2\pi j} \int_{-\infty}^{\infty} X(s)e^{st} ds = \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} X(s)e^{st} ds$$

$$\frac{dx}{dt} = \frac{d}{dt} \left[\frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} X(s)e^{st} ds \right] = \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} \frac{d}{dt} \left[X(s)e^{st} \right] ds$$

$$= \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} sX(s)e^{st} ds$$
Therefore $\mathcal{L}\left(\frac{dx}{dt}\right) = sX(s)$ $R' \supset R$

UMBC CMPE323 Signals and Systems
© EFCLaBerge, S. J. Kim 2014 All rights reserved

Differentiation in the transform domain

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

$$\frac{dX}{ds} = \frac{d}{ds} \left[\int_{-\infty}^{\infty} x(t)e^{-st} dt \right] = \int_{-\infty}^{\infty} x(t)\frac{d}{ds} \left[e^{-st} \right] dt = \int_{-\infty}^{\infty} (-t)x(t)e^{-st} dt$$

Therefore
$$\mathcal{L}((-t)x(t)) = \frac{dX}{ds}$$
, $R' = R$

• Integration in the time domain
$$\int_{-\infty}^{t} x(\tau) d\tau = \int_{-\infty}^{t} \frac{1}{2\pi j} \int_{\sigma-\infty}^{\sigma+j\infty} X(s) e^{st} ds = \frac{1}{2\pi j} \int_{\sigma-\infty}^{\sigma+j\infty} X(s) \int_{-\infty}^{t} e^{st} ds$$

$$= \int_{\sigma-\infty}^{\sigma+j\infty} \frac{1}{s} X(s) e^{st} ds$$

Therefore
$$\mathcal{L}\left(\int_{-\infty}^{t} x(\tau) d\tau\right) = \frac{1}{s}X(s), R' \supset R \cap \text{Re}[s] > 0$$



$$\lim_{s \to \infty} (sF(s)) = f(0^+)$$

Final Value Theorem

$$\lim_{s \to 0} \left(sF(s) \right) = \lim_{t \to \infty} f(t)$$

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

The Convolution Theorem

- One of the primary application of the Laplace (and later Fourier) Domain(s) is the Convolution Theorem
- What is the LT of the convolution?

$$\mathcal{L}(x * h) = \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau \right) e^{-st} dt$$

$$= \int_{-\infty}^{\infty} x(\tau) \left(\int_{-\infty}^{\infty} h(t-\tau)e^{-st} dt \right) d\tau = \int_{-\infty}^{\infty} x(\tau) \left(e^{-s\tau} H(s) \right) d\tau$$
time shift property!

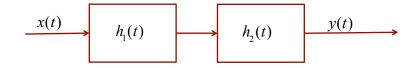
$$=H(s)\int_{-\infty}^{\infty}x(\tau)e^{-s\tau}\,d\tau=H(s)X(s)=X(s)H(s) \quad (!!)$$

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

10

So what?

• Remember our concatenation of systems?



$$y(t) = x * h = x * (h_1 * h_2)$$

$$H(s) = \mathcal{L}(h_1 * h_2) = H_1(s)H_2(s)$$

$$Y(s) = \mathcal{L}(x * h) = X(s)H(s) = X(s)H_1(s)H_2(s)$$

- A convolution in the time domain is equivalent to a multiplication in the frequency domain...
- ...and vice versa

UMBC CMPE323 Signals and Systems © EFCLaBerge, S. J. Kim 2014 All rights reserved Portions © Addison-Wesley Publishing, Inc

1.