

#### **Chapter 8**

UMBC ENEE610 Digital Signal Processing Course Notes © E F C LaBerge, 2009 All rights reserved

Lecture 1

# **Properties of Fourier Transform**

- **Duality:**  $X(f) = \mathcal{F}(x(t)) \Rightarrow x(f) = \mathcal{F}(X(-t))$  and  $x(-f) = \mathcal{F}(X(t))$
- Linearity:  $z(t) = ax(t) + by(t) \Rightarrow Y(f) = aX(f) + bY(f)$
- Time Shift:  $\mathcal{F}\left(x(t-t_0)\right) = e^{-j2\pi ft_0} \mathcal{F}\left(x(t)\right)$  Scaling: For  $a \neq 0 \in \mathbb{R}$ ,  $\mathcal{F}\left(x(at)\right) = \frac{1}{|a|} X\left(\frac{f}{a}\right)$
- Modulation:  $\mathcal{F}\left(x(t)e^{j2\pi f_0t}\right) = X(f-f_0)$
- Conjugation:  $\mathcal{F}(x^*(t)) = X^*(-f)$
- Parseval:  $\int_0^\infty x(t)y^*(t)dt = \int_0^\infty X(f)Y^*(f)df = \frac{1}{2\pi}\int_0^\infty X(\omega)Y^*(\omega)d\omega$
- Rayleigh  $\int_{0}^{\infty} \left| x(t) \right|^{2} dt = \int_{0}^{\infty} \left| X(f) \right|^{2} df = \frac{1}{2\pi} \int_{0}^{\infty} X(\omega) Y^{*}(\omega) d\omega$

## **Advanced Properties of Fourier Transform**

- Integration:  $\mathcal{F}\left(\int_{-\infty}^{t} x(t) dt\right) = \frac{X(f)}{j2\pi f} + \frac{1}{2}X(0)\delta(f)$
- Differentiation:  $\mathcal{F}\left(\frac{d}{dt}x(t)\right) = j2\pi fX(f)$
- Moments:  $\int_{-\infty}^{\infty} t^n x(t) dt = \left(\frac{j}{2\pi}\right)^n \frac{d^n}{df^n} X(f)$
- Real Signals: The Fourier transform of a real signal is EVEN in magnitude and ODD in phase

UMBC ENEE610 Digital Signal Processing Course Notes © E F C LaBerge, 2009 All rights reserved. Lecture 1 1-3

### The convolution theorem (very important!)

• The output of a LTI system with transfer function H(f)

$$Y(f) = X(f)H(f) \qquad y(t) = \int_{-\infty}^{\infty} Y(f)e^{j2\pi ft}df = \int_{-\infty}^{\infty} X(f)H(f)e^{j2\pi ft}df$$

$$Write H(f) = \int_{-\infty}^{\infty} h(\tau)e^{-j2\pi f\tau}d\tau$$

$$y(t) = \int_{-\infty}^{\infty} X(f)\left(\int_{-\infty}^{\infty} h(\tau)e^{-j2\pi f\tau}d\tau\right)e^{j2\pi ft}df$$

$$= \int_{-\infty}^{\infty} \left(\int_{-\infty}^{\infty} X(f)h(\tau)e^{j2\pi f(t-\tau)}df\right)d\tau$$

$$= \int_{-\infty}^{\infty} h(\tau)\left(\int_{-\infty}^{\infty} X(f)e^{j2\pi f(t-\tau)}df\right)d\tau$$

$$= \int_{-\infty}^{\infty} h(\tau)x(t-\tau)d\tau = \int_{-\infty}^{\infty} h(t-\tau)x(\tau)d\tau$$

$$\frac{1}{\sqrt{2\pi}} \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau$$

$$\frac{1}{\sqrt{2\pi}} \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau$$

$$\frac{1}{\sqrt{2\pi}} \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau + \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau$$

$$\frac{1}{\sqrt{2\pi}} \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau + \int_{-\infty}^{2\pi} h(\tau)x(\tau)d\tau$$

2

### Summary

- We can decompose a periodic signal into the weighted sum of complex exponentials,...
- ...or, equivalently, to the weighted sum of sines and cosines.
- We write the weighted sum as a Fourier Series
- The frequency-domain representation consists of a series of harmonically-related terms, with separation equal to the period of the signal,
- The coefficients have units of volts (or amps)
- We can decompose a non-periodic signal into the weighted integral of complex exponentials,...
- ...or, equivalently, to the weighted integral of sines and cosines
- The frequency-domain representation has a continuous spectrum with units of volts/Hz (or amps/Hz).

UMBC ENEE610 Digital Signal Processing
Course Notes © E F C LaBerge, 2009 All rights reserved.

Lecture 1 1-5

## Examples (done on board)

$$p(t) = \begin{cases} 0 & t < -\frac{\tau}{2} \\ 1 & -\frac{\tau}{2} \le t < \frac{\tau}{2} \\ 0 & t \ge \frac{\tau}{2} \end{cases}$$

$$v(t) = \begin{cases} 0 & t < -\tau \\ t + \tau & -\tau \le t < 0 \\ t - t & 0 \le t < \tau \\ 0 & t \ge \tau \end{cases}$$

g(t) = Gibbs phenomenon, p(t), perfect rectangular filter

UMBC ENEE610 Digital Signal Processing
Course Notes © E F C LaBerge, 2009 All rights reserved.

Lecture 1