

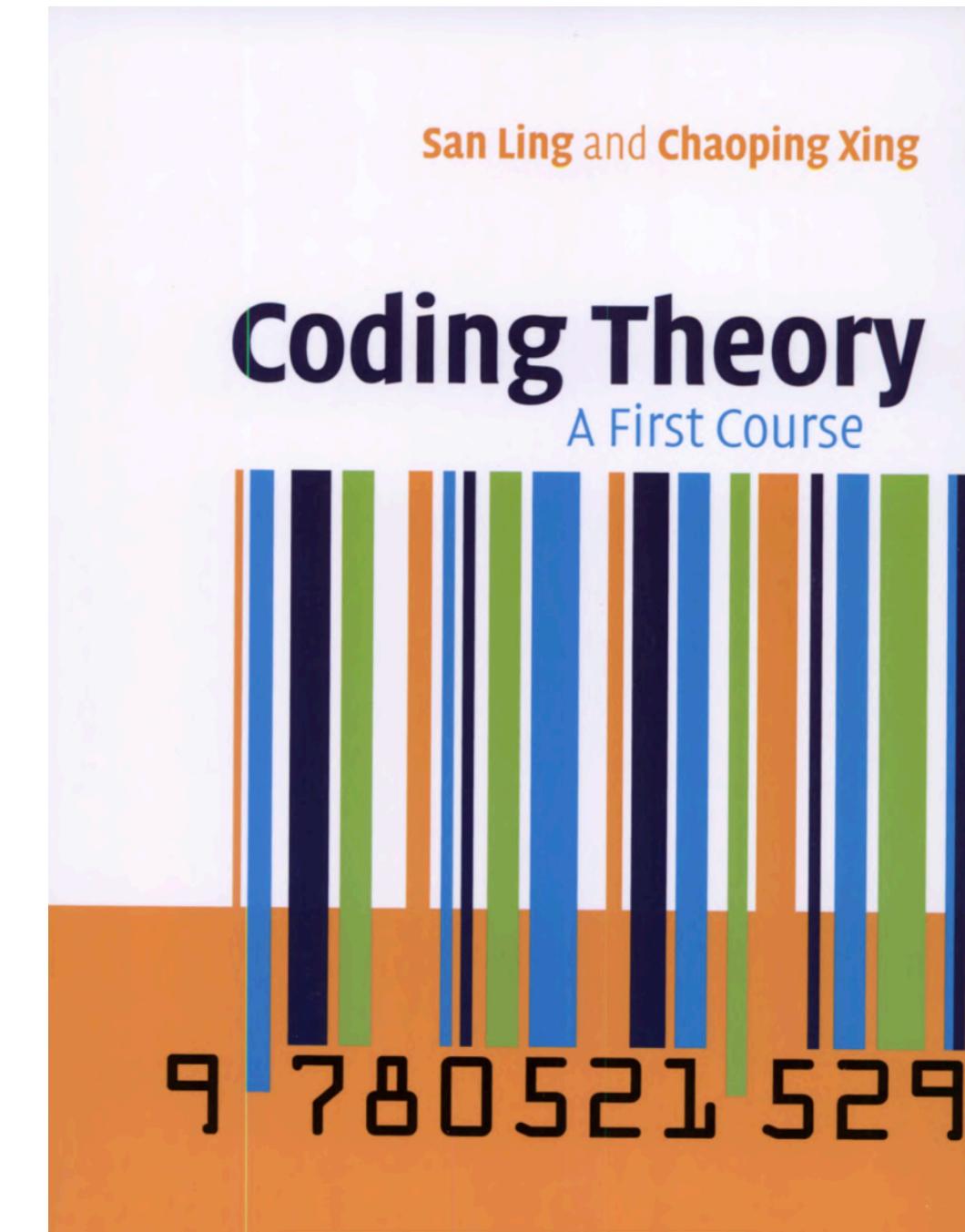
Coding Theory

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Contents

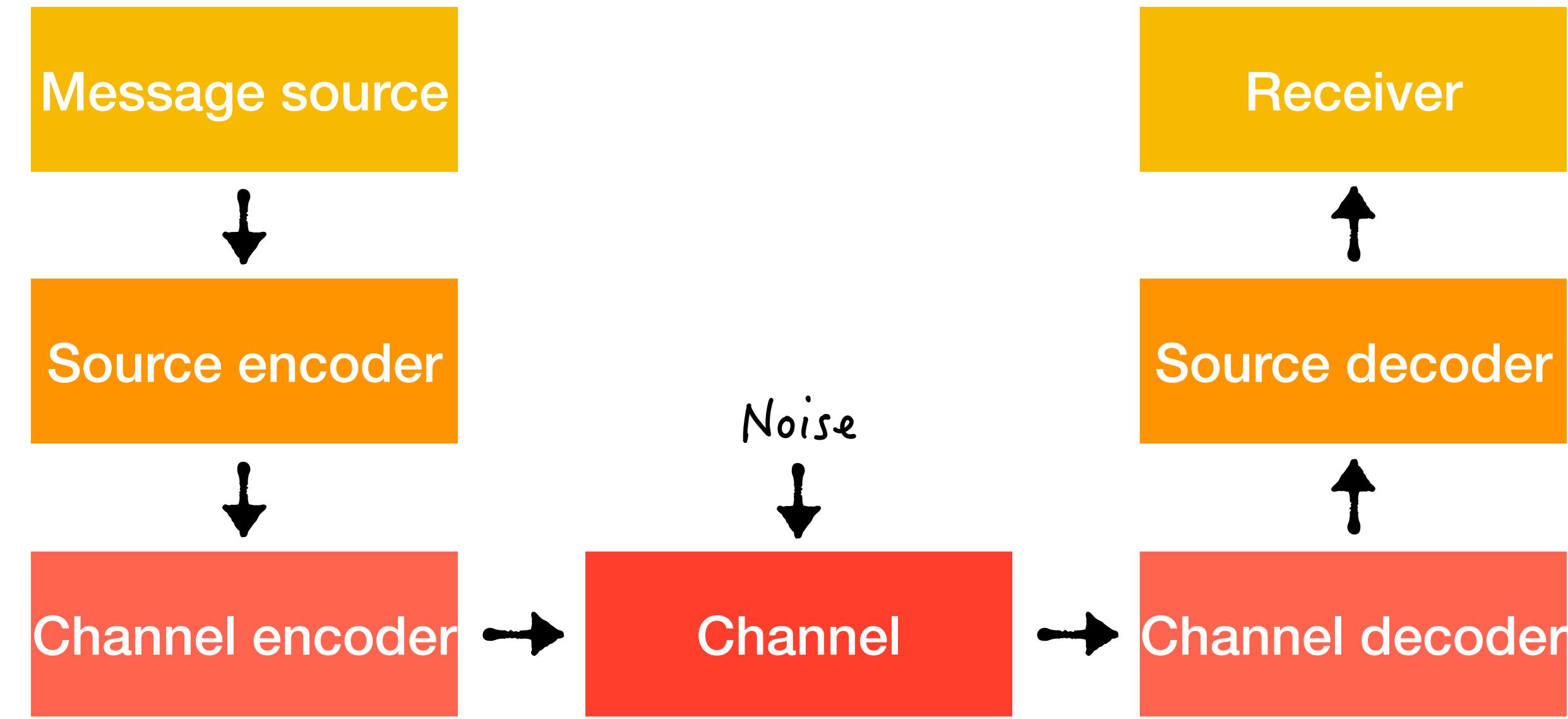
- Introduction
- Error detection, correction and decoding
- Linear codes



T. M. Thompson- From Error Correcting Codes Through Sphere Packings, to Simple Groups:
1.The origin of error correcting codes



Introduction



Source Coding:

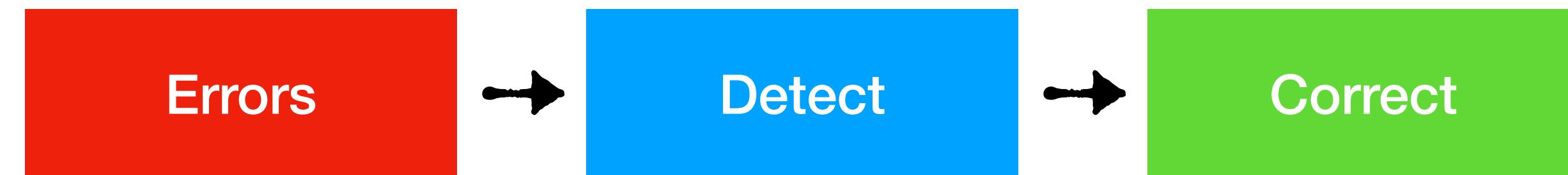
Changing the message source to a suitable code

Channel Coding:

Encoding the message by introducing some form of redundancy

Introduction

Example: repetition code $\rightarrow r+1$ times for detecting r errors



The goal of channel coding:

- (1) fast encoding of messages;
- (2) easy transmission of encoded messages;
- (3) fast decoding of received messages;
- (4) maximum transfer of information per unit time;
- (5) maximal detection or correction capability.

Error detection, correction and decoding

Definitions:

- (1) Code Alphabet: $A = \{a_1, \dots, a_q\} \rightarrow$ finite field F_q of order q
- (2) Code Symbols: a_1, \dots, a_q
- (3) q -ary word: $W = w_1 \dots w_n : w_i \in A$
- (4) q -ary block code: non empty set C of q -ary words having the same length n .
- (5) Code word: $c \in C$
- (6) Size of C : $|C|$
- (7) Information rate of a code C of length n : $(\log_q |C|)/n$
- (8) (n, M) -code: A code of length n and size M

+	0	1	x	0	1
0	0	1	0	0	0
1	1	0	1	0	1

Addition and multiplication tables for \mathbb{Z}_2

Error detection, correction and decoding

Communication channel:

consists of a finite channel alphabet $A = \{a_1, \dots, a_q\}$ as well as a set of forward channel probabilities

$P(a_j \text{ received} | a_i \text{ sent})$, satisfying:

$$\sum_{j=1}^q P(a_j \text{ received} | a_i \text{ sent}) = 1 \text{ for all } i.$$

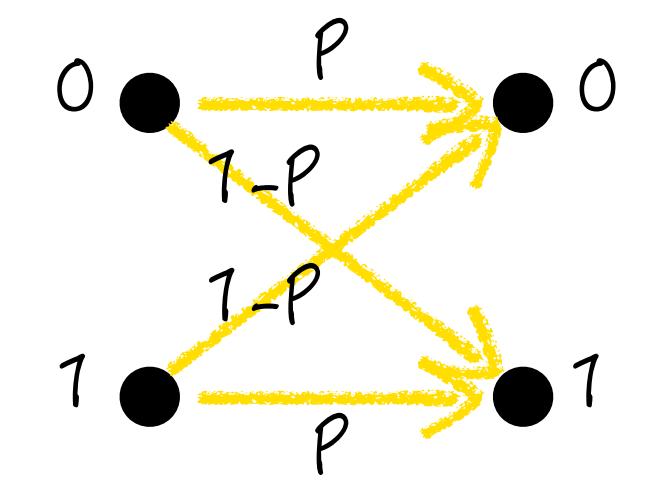
Memoryless:

if $c = c_1c_2\dots c_n$ and $x = x_1x_2\dots x_n$ are words of length n , then $P(x \text{ received} | c \text{ sent}) = \prod_{i=1}^n P(x_i \text{ received} | c_i \text{ sent})$

q -ary symmetric channel:

a memoryless channel which has a channel alphabet of size q such that

- (i) each symbol transmitted has the same probability $p (< 1/2)$ of being received in error;
- (ii) if a symbol is received in error, then each of the $q - 1$ possible errors is equally likely.



Binary symmetric channel
(BSC)

Error detection, correction and decoding

Maximum Likelihood decoding (MLD):

$$P(x \text{ received} | c_x \text{ sent}) = \max_{c \in C} P(x \text{ received} | c \text{ sent})$$

(i) Complete maximum likelihood decoding (CMLD):
If a word x is received, find the most likely codeword transmitted.
If there are more than one such codewords, select one of them arbitrarily.

(ii) Incomplete maximum likelihood decoding (IMLD):
If a word x is received, find the most likely codeword transmitted.
If there are more than one such codewords, request a retransmission.

Error detection, correction and decoding

Hamming distance:

If $x = x_1x_2\dots x_n$ and $y = y_1y_2\dots y_n$ then $d(x, y) = d(x_1, y_1) + \dots + d(x_n, y_n)$

$$\left\{ \begin{array}{ll} d(x_i, y_i) = 1 & \text{if } x_i \neq y_i \\ d(x_i, y_i) = 0 & \text{if } x_i = y_i \end{array} \right.$$

Nearest neighbour/minimum distance decoding (NND-MDD):

$$d(x, c_x) = \min_{c \in C} d(x, c)$$

(i) Complete nearest neighbour decoding (CNND)

(ii) Incomplete nearest neighbour decoding (INND)

Theorem 2.4.1 For a BSC with crossover probability $p < 1/2$, the maximum likelihood decoding rule is the same as the nearest neighbour decoding rule.

Proof. Let C denote the code in use and let \mathbf{x} denote the received word (of length n). For any vector \mathbf{c} of length n , and for any $0 \leq i \leq n$,

$$d(\mathbf{x}, \mathbf{c}) = i \Leftrightarrow \mathcal{P}(\mathbf{x} \text{ received} \mid \mathbf{c} \text{ sent}) = p^i(1-p)^{n-i}.$$

Since $p < 1/2$, it follows that

$$p^0(1-p)^n > p^1(1-p)^{n-1} > p^2(1-p)^{n-2} > \dots > p^n(1-p)^0.$$

By definition, the maximum likelihood decoding rule decodes \mathbf{x} to $\mathbf{c} \in C$ such that $\mathcal{P}(\mathbf{x} \text{ received} \mid \mathbf{c} \text{ sent})$ is the largest, i.e., such that $d(\mathbf{x}, \mathbf{c})$ is the smallest (or seeks retransmission if incomplete decoding is in use and \mathbf{c} is not unique). Hence, it is the same as the nearest neighbour decoding rule.

Error detection, correction and decoding

Definitions:

- (1) Distance of C : $d(C) = \min\{d(x, y) : x, y \in C, x \neq y\}$
- (2) (n, M, d) -code: A code of length n and size M and distance d
- (3) u -error-detecting: codeword incurs at least one but at most u errors $\Leftrightarrow d(C) \geq u + 1$
- (4) v -error-correcting: minimum distance (incomplete) decoding is able to correct v or fewer errors $\Leftrightarrow d(C) \geq 2v + 1$

Linear Codes

Let F_q be the **finite field** of order q . A nonempty set V , together with some (vector) addition $+$ and scalar multiplication by elements of F_q , is a **vector space** (or **linear space**) over F_q if it satisfies all of the following conditions. For all $u, v, w \in V$ and for all $\lambda, \mu \in F_q$:

- (i) $u+v \in V$;
- (ii) $(u+v)+w = u+(v+w)$;
- (iii) there is an element $0 \in V$ with the property $0+v = v = v+0$ for all $v \in V$;
- (iv) for each $u \in V$ there is an element of V , called $-u$, such that $u+(-u) = 0 = (-u)+u$;
- (v) $u+v=v+u$;
- (vi) $\lambda v \in V$;
- (vii) $\lambda(u+v)=\lambda u+\lambda v$, $(\lambda+\mu)u=\lambda u+\mu u$;
- (viii) $(\lambda\mu)u=\lambda(\mu u)$;
- (ix) if 1 is the multiplicative identity of F_q , then $1u=u$.

Linear Codes

$$C^\perp = \{x \in F_q^n \mid \langle x, y \rangle = 0 \text{ for all } y \in C\}$$

A linear code C of length n over F_q is a subspace of F_q^n

- (1) Dual code of C : C^\perp , the orthogonal complement of the subspace C of F_q^n .
- (2) Dimension of C : the dimension of C as a vector space over F_q . $\rightarrow \dim(C) = \log_q |C|$. ($|C| = q^{\dim(C)}$)
 $\rightarrow \dim(C) + \dim(C^\perp) = n$

Example (1):

$$C = \{0000, 1010, 0101, 1111\}$$

$$\dim(C) = \log_2 |C| = \log_2 4 = 2$$

$$C^\perp = \{0000, 1010, 0101, 1111\}$$

$$\dim(C^\perp) = \log_2 |C| = \log_2 4 = 2$$

Example (2):

$$C = \{000, 001, 002, 010, 020, 011, 012, 021, 022\}$$

$$\dim(C) = \log_3 |C| = \log_3 9 = 2$$

$$C^\perp = \{000, 100, 200\}$$

$$\dim(C^\perp) = \log_3 |C| = \log_3 3 = 1$$

Linear Codes

Hamming weight: $\text{wt}(x) = d(x, 0) = \sum_{i=1}^n \text{wt}(x_i)$

$$\left\{ \begin{array}{l} \text{wt}(x_i) = 1 \text{ if } x \neq 0 \\ \text{wt}(x_i) = 0 \text{ if } x = 0 \end{array} \right.$$

-> Lemma: $d(x, y) = \text{wt}(x - y)$

The minimum (Hamming) weight: $\text{wt}(C) = \min_{0 \neq x \in C} \text{wt}(x)$

Theorem 4.3.8 *Let C be a linear code over \mathbf{F}_q . Then $d(C) = \text{wt}(C)$.*

Proof. Recall that for any words \mathbf{x}, \mathbf{y} we have $d(\mathbf{x}, \mathbf{y}) = \text{wt}(\mathbf{x} - \mathbf{y})$.

By definition, there exist $\mathbf{x}', \mathbf{y}' \in C$ such that $d(\mathbf{x}', \mathbf{y}') = d(C)$, so

$$d(C) = d(\mathbf{x}', \mathbf{y}') = \text{wt}(\mathbf{x}' - \mathbf{y}') \geq \text{wt}(C),$$

since $\mathbf{x}' - \mathbf{y}' \in C$.

Conversely, there is a $\mathbf{z} \in C \setminus \{\mathbf{0}\}$ such that $\text{wt}(C) = \text{wt}(\mathbf{z})$, so

$$\text{wt}(C) = \text{wt}(\mathbf{z}) = d(\mathbf{z}, \mathbf{0}) \geq d(C).$$

Linear Codes

Bases for linear codes:

Input: A nonempty subset S of F_q^n .

Output: A basis for $C = \langle S \rangle$, the linear code generated by S .

Algorithm 1:

Description: Form the matrix A whose rows are the words in S . Use elementary row operations to find an REF of A . Then the nonzero rows of the REF form a basis for C .

$$S = \{12101, 20110, 01122, 11010\}.$$

$$A = \begin{pmatrix} 12101 \\ 20110 \\ 01122 \\ 11010 \end{pmatrix} \rightarrow \begin{pmatrix} 12101 \\ 02211 \\ 01122 \\ 02212 \end{pmatrix} \rightarrow \begin{pmatrix} 12101 \\ 01122 \\ 00001 \\ 00000 \end{pmatrix}.$$

Linear Codes

Algorithm 2:

Description: Form the matrix A whose columns are the words in S. Use elementary row operations to put A in **REF** and locate the leading columns in the **REF**. Then the original columns of A corresponding to these leading columns form a basis for C.

$$S = \{11101, 10110, 01011, 11010\}.$$

$$A = \begin{pmatrix} 1101 \\ 1011 \\ 1100 \\ 0111 \\ 1010 \end{pmatrix} \rightarrow \begin{pmatrix} 1101 \\ 0110 \\ 0001 \\ 0111 \\ 0111 \end{pmatrix} \rightarrow \begin{pmatrix} 1101 \\ 0110 \\ 0001 \\ 0000 \\ 0000 \end{pmatrix}.$$

Linear Codes

Algorithm 3:

Description: Form the matrix A whose rows are the words in S . Use elementary row operations to place A in **RREF**. Let G be the $k \times n$ matrix consisting of all the nonzero rows of the **RREF**: $A \rightarrow \begin{pmatrix} G \\ 0 \end{pmatrix}$

$$A \rightarrow \begin{pmatrix} G \\ 0 \end{pmatrix}$$

The matrix G contains k leading columns. Permute the columns of G to form $G' = (I_k | X)$.

Form a matrix H as follows: $H' = (-X^T | I_{n-k})$.

Apply the inverse of the permutation applied to the columns of G to the columns of H' to form H . Then the rows of H **form a basis for C^\perp** .

$$G = \begin{pmatrix} 1 & 2 & 3 & 4 & 5 & 6 & 7 & 8 & 9 & 10 \\ 1 & 0 & 2 & 0 & 0 & 2 & 0 & 1 & 0 & 2 \\ 0 & 0 & 0 & 1 & 0 & 1 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 2 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 2 \end{pmatrix}$$

$$G' = (I_5 | X) = \begin{pmatrix} 1 & 4 & 5 & 7 & 9 & 2 & 3 & 6 & 8 & 10 \\ 1 & 0 & 0 & 0 & 0 & 0 & 2 & 2 & 1 & 2 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 2 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 2 \end{pmatrix}$$

$$H' = \begin{pmatrix} 1 & 4 & 5 & 7 & 9 & 2 & 3 & 6 & 8 & 10 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 1 & 2 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 2 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 1 & 2 & 0 & 2 & 1 & 0 & 0 & 0 & 0 & 1 \end{pmatrix}$$

$$H = \begin{pmatrix} 1 & 2 & 3 & 4 & 5 & 6 & 7 & 8 & 9 & 10 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 2 & 0 & 1 & 0 & 0 & 0 & 0 \\ 2 & 0 & 0 & 0 & 1 & 0 & 0 & 1 & 0 & 0 \\ 1 & 0 & 0 & 2 & 0 & 0 & 2 & 0 & 1 & 1 \end{pmatrix}$$

Linear Codes

Generator matrix: Matrix G whose rows form a basis for C . \rightarrow Standard: $G = (I_k | X)$

Parity-check matrix: Matrix H is a generator matrix for the dual code C^\perp . \rightarrow Standard: $H = (Y | I_{n-k})$

Let C be a linear code and let H be a parity-check matrix for C . Then the following statements are equivalent:

- (i) C has distance d ;
- (ii) any $d - 1$ columns of H are linearly independent and H has d columns that are linearly dependent.

\rightarrow If $G = (I_k | X)$ is the standard form generator matrix of an $[n,k]$ -code C , then a parity-check matrix for C is $H = (-X^T | I_{n-k})$.

Linear Codes

Two (n, M) -codes over F_q are **equivalent** if one can be obtained from the other by a combination of operations of the following types:

- (i) permutation of the n digits of the codewords;
- (ii) multiplication of the symbols appearing in a fixed position by a nonzero scalar.

-> Any linear code C is equivalent to a linear code C' with a generator matrix in standard form.

Example (1):

Let $q = 2$ and $n = 4$. Choosing to rearrange the bits in the order 2, 4, 1, 3, we see that the code

$C = \{0000, 0101, 0010, 0111\}$ is equivalent to the code
 $C' = \{0000, 1100, 0001, 1101\}$.

Example (2):

Let $q = 3$ and $n = 3$. Consider the ternary code
 $C = \{000, 011, 022\}$.

Permuting the first and second positions,
followed by multiplying the third position by 2,
we obtain the equivalent code

$C = \{000, 102, 201\}$.

Linear Codes

Encoding with a linear code:

Let C be an $[n, k, d]$ -linear code over the finite field F_q . Each codeword of C can represent one piece of information, so C can represent q^k distinct pieces of information. Once a basis $\{r_1, \dots, r_k\}$ is fixed for C , each codeword v , or, equivalently, each of the q^k pieces of information, can be uniquely written as a linear combination, $v = u_1r_1 + \dots + u_kr_k = uG$. (G is the generator matrix of C whose i th row is the vector r_i in the chosen basis.)

-> The process of representing the elements u of F_q^k as codewords $v = uG$ in C is called **encoding**.

Example (1):

Let C be the binary $[5, 3]$ -linear code with the generator matrix

$$G = \begin{pmatrix} 10110 \\ 01011 \\ 00101 \end{pmatrix} \rightarrow \text{for } u = 101: v = uG = (101) \begin{pmatrix} 10110 \\ 01011 \\ 00101 \end{pmatrix} = 10011$$

Linear Codes

If an $[n, k, d]$ -linear code C has a generator matrix G in standard form, $G = (I_k | X)$, then it is trivial to recover the message u from the codeword $v = uG \rightarrow v = uG = u(I|X) = (u, uX)$;

Message digits: the first k digits in the codeword $v = uG$ give the message u .

Check digits: The remaining $n - k$ digits!

The check digits represent the **redundancy** which has been added to the message for protection against noise.



Linear Codes

Coset: (of C) determined by u to be the set: $u + C = C + u = \{v + u : v \in C\}$.

Coset leader: A word of the least (Hamming) weight in a coset.

Example (1):

Let $q = 2$ and $C = \{000, 101, 010, 111\}$. Then

$$C + 000 = \{\textcolor{red}{000}, 101, 010, 111\},$$

$$C + 001 = \{\textcolor{red}{001}, 100, 011, 110\},$$

$$C + 010 = \{\textcolor{red}{010}, 111, 000, 101\},$$

$$C + 011 = \{011, 110, \textcolor{red}{001}, 100\},$$

$$C + 100 = \{100, \textcolor{blue}{001}, 110, 011\},$$

$$C + 101 = \{101, 000, 111, \textcolor{blue}{010}\},$$

$$C + 110 = \{110, 011, \textcolor{blue}{100}, 001\},$$

$$C + 111 = \{111, \textcolor{blue}{010}, 101, 000\}.$$

Theorem 4.8.4 Let C be an $[n, k, d]$ -linear code over the finite field \mathbf{F}_q . Then,

- (i) every vector of \mathbf{F}_q^n is contained in some coset of C ;
- (ii) for all $\mathbf{u} \in \mathbf{F}_q^n$, $|C + \mathbf{u}| = |C| = q^k$;
- (iii) for all $\mathbf{u}, \mathbf{v} \in \mathbf{F}_q^n$, $\mathbf{u} \in C + \mathbf{v}$ implies that $C + \mathbf{u} = C + \mathbf{v}$;
- (iv) two cosets are either identical or they have empty intersection;
- (v) there are q^{n-k} different cosets of C ;
- (vi) for all $\mathbf{u}, \mathbf{v} \in \mathbf{F}_q^n$, $\mathbf{u} - \mathbf{v} \in C$ if and only if \mathbf{u} and \mathbf{v} are in the same coset.

Linear Codes

Assume the codeword v is transmitted and the word w is received,

Error pattern (Error string): $e = w - v \in w + C$.

Then $w - e = v \in C$, so the error pattern e and the received word w are in the same coset.

Nearest neighbour decoding:

Since error patterns of small weight are the most likely to occur, nearest neighbour decoding works for a linear code C in the following manner. Upon receiving the word w , we choose a word e of least weight in the coset $w + C$ and conclude that $v = w - e$ was the codeword transmitted.

Example (1):

Let $q = 2$ and $C = \{0000, 1011, 0101, 1110\}$.

Decode the following received words:

(i) $w = 1101$; (ii) $w = 1111$.

First, we write down the standard array of C

$C + 0000 = \{0000, 1011, 0101, 1110\}$,

$C + 0001 = \{0001, 1010, 0100, 1111\} \rightarrow$ (ii) $e = 0001$ or $0100 \rightarrow$ complete or incomplete ?!

$C + 0010 = \{0010, 1001, 0111, 1100\}$,

$C + 1000 = \{1000, 0011, 1101, 0110\} \rightarrow$ (i) $e = 1000 \rightarrow w - e = 0101$

Linear Codes

Syndrome decoding: $\forall w \in F_q^n : S(w) = wH^T \in F_q^{n-k}$

- (i) $S(u+v) = S(u) + S(v)$;
- (ii) $S(u) = 0$ if and only if u is a codeword in C ;
- (iii) $S(u) = S(v)$ if and only if u and v are in the same coset of C .

Syndrome look-up table (Standard Decoding Array (SDA).): table which matches each coset leader with its syndrome.

-> Steps to construct a syndrome look-up table assuming complete nearest neighbour decoding

Step 1: List all the cosets for the code, choose from each coset a word of least weight as coset leader u .

Step 2: Find a parity-check matrix H for the code and, for each coset leader u , calculate its syndrome $S(u) = uh^T$.

Coset leader u	Syndrome $S(u)$
0000	00
0001	01
0010	10
1000	11

$$C = \{0000, 1011, 0101, 1110\}.$$

Linear Codes

Decoding procedure for syndrome decoding:

Step 1: For the received word w , compute the syndrome $S(w)$.

Step 2: Find the coset leader u next to the syndrome $S(w) = S(u)$ in the syndrome look-up table.

Step 3: Decode w as $v = w - u$.

Example (1): Let $q = 2$ and let $C = \{0000, 1011, 0101, 1110\}$. Use the syndrome look-up table below to decode (i) $w = 1101$; (ii) $w = 1111$.

(i) $w = 1101$. The syndrome is $S(w) = wH^T = 11$. From Table, we see that the coset leader is 1000.

Hence, $1101 + 1000 = 0101$ was a most likely codeword sent.

(ii) $w = 1111$. The syndrome is $S(w) = wH^T = 01$. From Table, we see that the coset leader is 0001.

Hence, $1111 + 0001 = 1110$ was a most likely codeword sent.

Coset leader u	Syndrome $S(u)$
0000	00
0001	01
0010	10
1000	11



Thank You

For your Attention

