

CSE2209: Digital Electronics and Pulse Techniques

Course Conducted By:

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Prove that, for square wave (symmetric) input and $T \ll 2RC$

$$\text{i) } V_1 = \frac{V}{2} \left(1 + \frac{T}{4RC}\right)$$

$$\text{ii) } V_1' = \frac{V}{2} \left(1 - \frac{T}{4RC}\right)$$

i) We know,

$$V_1 = \frac{V}{\left(1 + e^{-\frac{T}{2RC}}\right)}$$

$$\therefore V_1 = \frac{V}{1 + \left(1 - \frac{T}{2RC} + \frac{T^2}{2! (2RC)^2} - \dots \dots\right)} \quad [e^{-x} = 1 - x + \frac{x^2}{2!} - \dots \dots]$$

$$= \frac{V}{1 + 1 - \frac{T}{2RC}} \quad [\text{Neglecting higher terms since } T \ll 2RC]$$

$$= \frac{V}{2 \left(1 - \frac{T}{4RC}\right)}$$

$$= \frac{V}{2} \left\{ 1 + \frac{T}{4RC} + \frac{T^2}{(4RC)^2} + \dots \dots \right\} \quad [(1 - x)^{-1} = 1 + x + x^2 + \dots \dots]$$

$$\therefore V_1 = \frac{V}{2} \left(1 + \frac{T}{4RC}\right) \quad [\text{Neglecting higher terms since } T \ll 2RC]$$

[Proved]

ii) We know, $V_1' = \frac{V}{\left(1 + e^{\frac{T}{2RC}}\right)}$

$$\therefore V_1' = \frac{V}{1 + \left(1 + \frac{T}{2RC} + \frac{T^2}{2! (2RC)^2} + \dots \dots\right)} \quad \left[e^x = 1 + x + \frac{x^2}{2!} + \dots \dots\right]$$

$$= \frac{V}{1 + 1 + \frac{T}{2RC}} \quad \text{[Neglecting higher terms since } T \ll 2RC]$$

$$= \frac{V}{2 \left(1 + \frac{T}{4RC}\right)} = \frac{V}{2} \left(1 + \frac{T}{4RC}\right)^{-1}$$

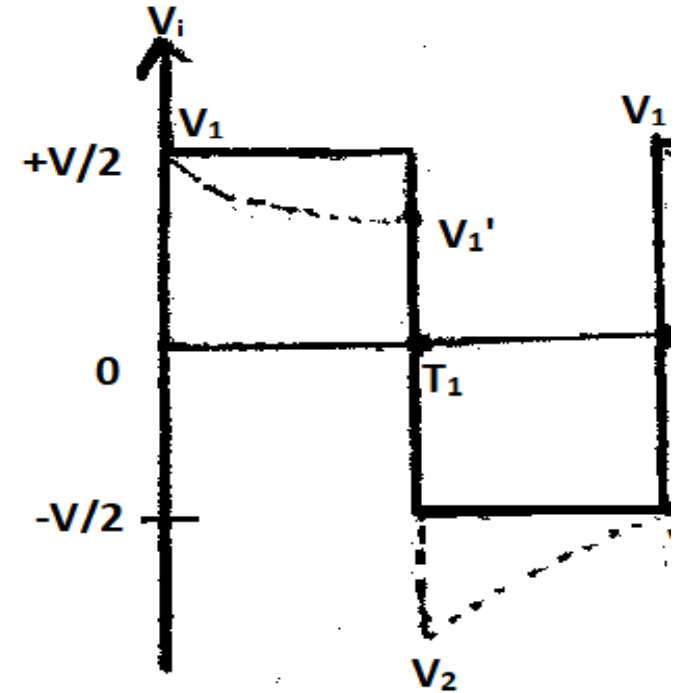
$$= \frac{V}{2} \left\{ 1 - \frac{T}{4RC} + \frac{T^2}{(4RC)^2} + \dots \dots \right\} \quad \left[(1 + x)^{-1} = 1 - x + x^2 - \dots \dots\right]$$

$$\therefore V_1' = \frac{V}{2} \left(1 - \frac{T}{4RC}\right) \quad \text{[Neglecting higher terms since } T \ll 2RC] \quad \text{[Proved]}$$

Percentage Tilt:

$$\begin{aligned}
 \% \text{tilt} &= \frac{V_1 - V_1'}{\frac{V}{2}} \times 100\% \\
 &= \frac{\frac{V}{2} \left(1 + \frac{T}{4RC}\right) - \frac{V}{2} \left(1 - \frac{T}{4RC}\right)}{\frac{V}{2}} \times 100\% \\
 &= \left(1 + \frac{T}{4RC} - 1 + \frac{T}{4RC}\right) \times 100\% \\
 &= \frac{T}{2RC} \times 100\% \\
 &= \frac{T \cdot \pi}{2\pi RC} \times 100\% \quad \therefore \% \text{tilt} = T\pi f_1 \times 100\%
 \end{aligned}$$

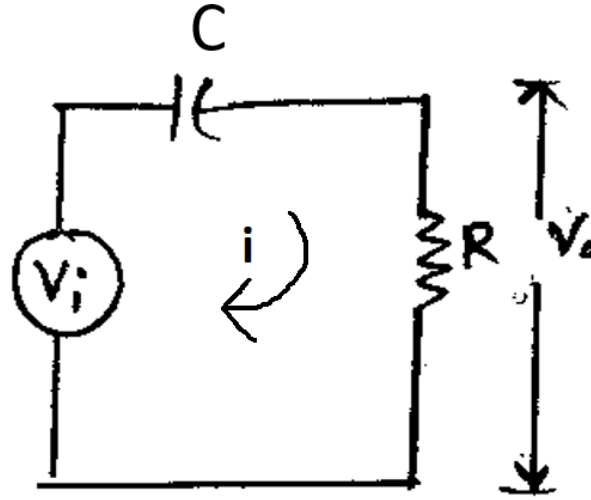
$$[\because \frac{1}{2\pi RC} = f_1 \text{ (Linear frequency)}]$$



Prove that, for High Pass RC circuit amplification $|A| = 0.707$

$$Z_C = \frac{1}{j\omega C}$$

$$Z_C = -\frac{j}{\omega C} \quad [\because \frac{1}{j} = -j]$$



$$i = \frac{V_i}{Z_C + R}$$

$$V_o = iR = \frac{V_i}{Z_C + R} R$$

Here, j is imaginary coefficient, $j = \sqrt{-1}$

ω = angular frequency = $2\pi f$

$$A = \frac{V_0}{V_i} = \frac{V_i \cdot R}{Z_C + R} \cdot \frac{1}{V_i} \quad [V_0 = \frac{V_i}{Z_C + R} R]$$

$$= \frac{R}{Z_C + R}$$

$$= \frac{1}{\frac{Z_C}{R} + 1}$$

$$= \frac{1}{\frac{-j}{\omega R C} + 1} = \frac{1}{1 - \frac{j}{\omega R C}}$$

$$A = \frac{1}{1 - \frac{j}{wRc}}$$

$$= \frac{\left(1 + \frac{j}{wRc}\right)}{\left(1 - \frac{j}{wRc}\right) \left(1 + \frac{j}{wRc}\right)}$$

$$= \frac{\left(1 + \frac{j}{wRc}\right)}{1^2 - \left(\frac{j}{wRc}\right)^2}$$

$$= \frac{\left(1 + \frac{j}{wRc}\right)}{1 + \frac{1}{w^2 R^2 c^2}}$$

$$A = \frac{1}{1 + \frac{1}{w^2 R^2 c^2}} + j \frac{\frac{1}{w R c}}{1 + \frac{1}{w^2 R^2 c^2}} \quad \left[\begin{array}{l} r = a + ib \\ \therefore |r| = \sqrt{a^2 + b^2} \end{array} \right]$$

$$\therefore |A| = \sqrt{\left(\frac{1}{1 + \frac{1}{w^2 R^2 c^2}} \right)^2 + \left(\frac{\frac{1}{w R c}}{1 + \frac{1}{w^2 R^2 c^2}} \right)^2} = \sqrt{\frac{1 + \frac{1}{w^2 R^2 c^2}}{\left(1 + \frac{1}{w^2 R^2 c^2} \right)^2}}$$

$$= \sqrt{\frac{1}{1 + \frac{1}{w^2 R^2 c^2}}}$$

$$\frac{1}{wRc} = \frac{1}{2\pi f R c} = \frac{1}{f} \cdot \frac{1}{2\pi R c} = \frac{1}{f} \cdot f_1 \quad [\because \frac{1}{2\pi R c} = f_1]$$

$$\therefore |A| = \frac{1}{\sqrt{1 + \left(\frac{f_1}{f}\right)^2}} = \frac{f_1}{f}$$

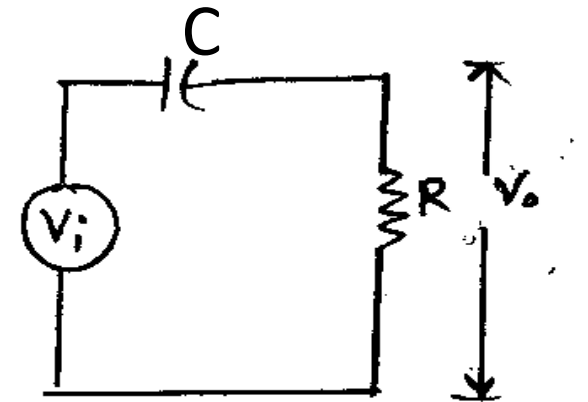
If $f_1 = f$, then $|A| = \frac{1}{\sqrt{1 + 1}} = \frac{1}{\sqrt{2}} = 0.707$

[Proved]

High Pass RC Circuit as Differentiator

When time constant (T) is too small i.e. R and C are very small in comparison with the time required for i/p signal to make an appreciable change, then the circuit acts like a differentiator. This name arises from the fact that under this circumstances, the voltage drop across R will be very small compared to the drop across C .

Hence, we may consider that the total i/p V_i appears across C



$$V_i(t) = V_c(t) + V_R(t)$$

$$V_i(t) \approx V_c(t) \quad [\because R \text{ is too small}]$$

$$i(t) = \frac{V_R(t)}{R}$$

Again,

$$i_c(t) = C \frac{d}{dt} V_c(t) \quad [\text{As, } V_c = \frac{Q}{C} \text{ and } i = \frac{Q}{t}]$$

$$\frac{V_R(t)}{R} = C \frac{d}{dt} V_c(t)$$

$$V_R(t) = RC \frac{d}{dt} V_c(t)$$

$$V_o(t) = RC \frac{d}{dt} V_i(t) \quad [V_i(t) \approx V_c(t)]$$

$$V_o(t) \propto \frac{d}{dt} V_i(t)$$

So, the output is proportional to the derivative of the input signal.

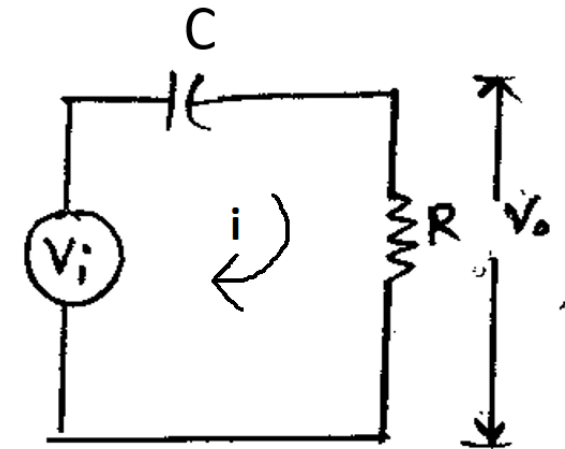
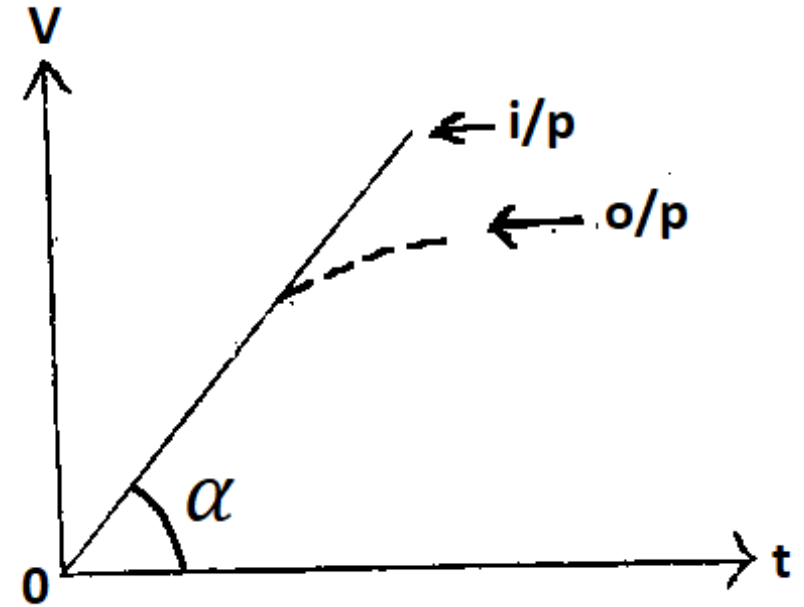
Ramp Input:

$$V_i(t) = \begin{cases} \alpha t, & \text{if } t \geq 0 \\ 0, & \text{Otherwise} \end{cases}$$

$$V_i(t) = \alpha t$$

$$\mathcal{L}\{V_i(t)\} = \alpha \mathcal{L}\{t\}$$

$$\therefore V_i(S) = \alpha \frac{1}{S^2} \quad \left[\mathcal{L}\{t\} = \frac{1}{S^2} \right]$$



We know, for High Pass RC Circuit

$$V_0(S) = \left[\frac{1}{1 + \frac{1}{SRC}} \right] V_i(S) = \left[\frac{1}{1 + \frac{1}{SRC}} \right] * \frac{\alpha}{S^2}$$

$$= \frac{\alpha}{S^2 \left(1 + \frac{1}{SRC} \right)}$$

$$\therefore V_0(S) = \frac{\alpha}{S \left(S + \frac{1}{RC} \right)}$$

$$V_0(S) = \frac{\alpha}{S(S + \frac{1}{RC})}$$

Do all the partial fraction steps yourself

$$\frac{px+q}{(x-a)(x-b)}, a \neq b \quad \frac{A}{x-a} + \frac{B}{x-b}$$

Using this

Now using partial fraction,

$$V_0(S) = \alpha RC \left(\frac{1}{S} - \frac{1}{S + \frac{1}{RC}} \right)$$

$$V_0(S) = \alpha RC \left(\frac{1}{S} - \frac{1}{S + \frac{1}{RC}} \right)$$

Applying Inverse Laplace transformation

$$V_0(t) = \alpha RC \cdot \left(1 - e^{-\frac{t}{RC}} \right) \quad \left[\mathcal{L} \left\{ \frac{1}{S + a} \right\} = e^{-at} \right]$$

Now we have to find the transmission error for Ramp input.

Transmission Error

Now, if $\frac{t}{RC} \ll 1$, then $[e^{-x} = 1 - x + \frac{x^2}{2!} - \dots \dots]$

$$\begin{aligned} V_0(t) &= \alpha RC \left\{ 1 - \left(1 - \frac{t}{RC} + \frac{t^2}{2! (RC)^2} - \dots \dots \right) \right\} \\ &= \alpha RC \left(\frac{t}{RC} - \frac{t^2}{2 RC^2} \right) \quad [\text{Neglecting higher terms}] \\ &= \alpha \left(t - \frac{t^2}{2 RC} \right) \end{aligned}$$

Now transmission error at time $t = T$

$$e_i(T) = \frac{V_i(T) - V_0(T)}{V_i(T)}$$

$$= \frac{\alpha T - \alpha \left(T - \frac{T^2}{2RC} \right)}{\alpha T}$$

$$= 1 - \left(1 - \frac{T}{2RC} \right)$$

$$\therefore e_i(T) = \frac{T}{2RC}$$

Again, if $\frac{t}{RC} \gg 1$, then $e^{-\frac{t}{RC}} = e^{-\frac{\infty}{RC}} = 0$

$$V_0(t) = \alpha RC \cdot (1 - 0) = \alpha RC$$

Now transmission error at time $t = T$

$$e_i(T) = \frac{V_i(T) - V_0(T)}{V_i(T)} = \frac{\alpha T - \alpha RC}{\alpha T}$$

$$\therefore e_i(T) = 1 - \frac{RC}{T}$$