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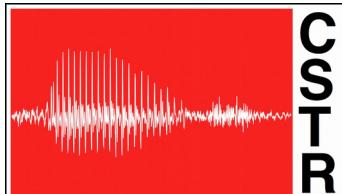
SpeechWave



Deep Scattering Spectrum (DSS) and its Applications in ASR

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Deep Scattering Spectrum

Joakim Andén, *Member, IEEE*, and Stéphane Mallat, *Fellow, IEEE*





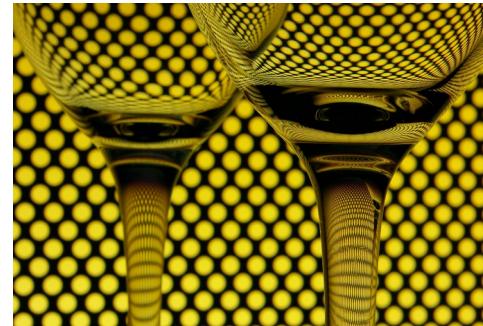
Outline

- Deep Scattering Spectrum (DSS)
- IBM+JHU, ICASSP 2014
- IBM+JHU, INTERSPEECH 2014
- KCL+CSTR, INTERSPEECH 2020
- Wrap-up



Goal: Construct a representation ...

- ... preserves info while remains *invariant* and *stable* to variabilities within class, for example ...
 - Stable to (small) deformation, e.g. time warping
 - Invariant to geometric transformations, e.g. translation, scale



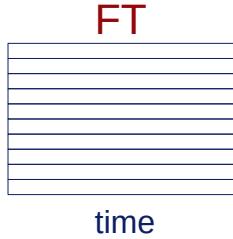
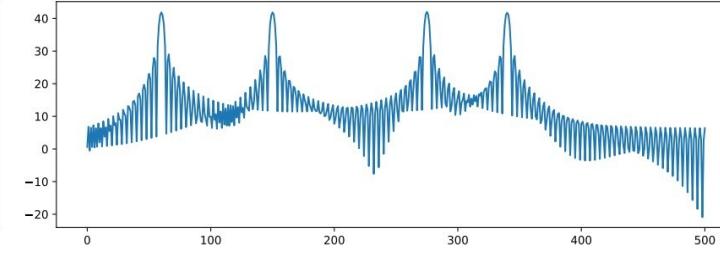
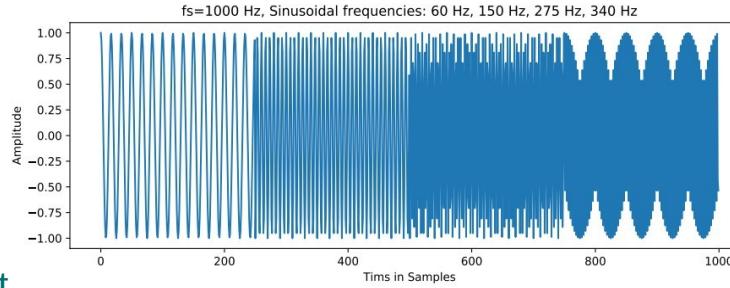
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4	4	4	4	4	4	4	4	4	4	4
5	5	5	5	5	5	5	5	5	5	5
6	6	6	6	6	6	6	6	6	6	6
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Detour

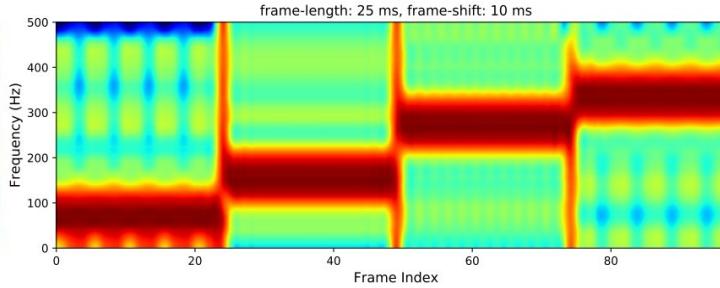
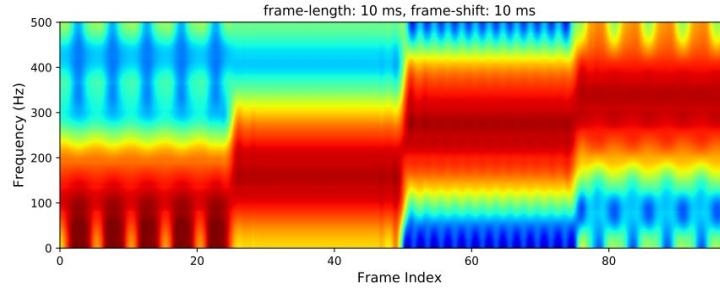
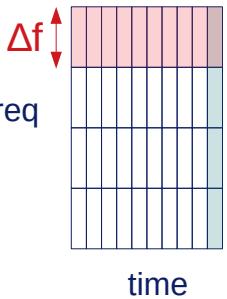
- Time-Frequency Analysis
- Wavelet Transform
- Amplitude Demodulation
- Time-warping Deformation
- Lipschitz Stability



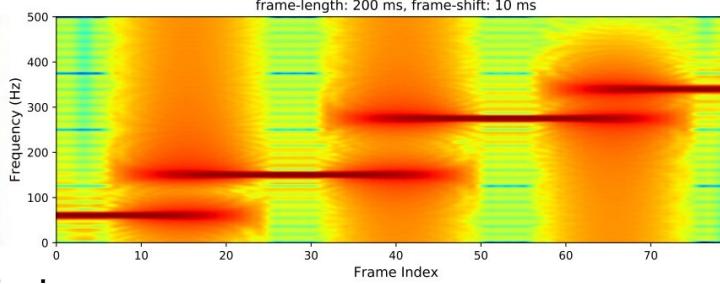
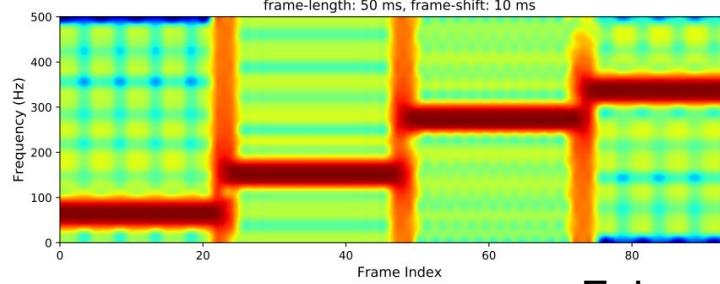
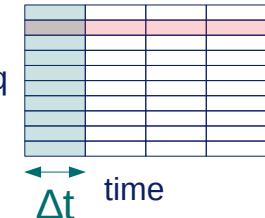
Time-Frequency Analysis (TFA)



STFT, small Δt



STFT, small Δt



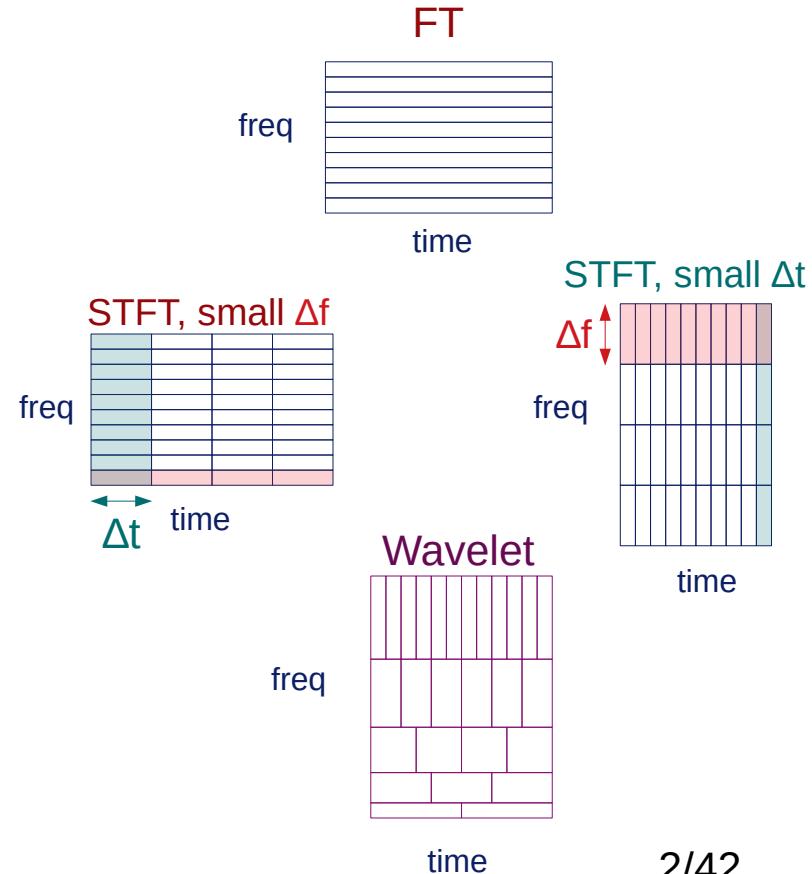
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Time-Frequency Analysis (TFA)

$$\text{FT: } X(\omega) = \int x(t) e^{-j\omega t} dt$$

$$\text{STFT: } X(t, \omega) = \int x(t') w(t' - t) \overset{\text{window}}{e^{-j\omega t'}} dt'$$

$$\text{Wavelet: } X(a, b) = \frac{1}{\sqrt{|a|}} \int x(t) \psi^*\left(\frac{t-b}{a}\right) dt$$



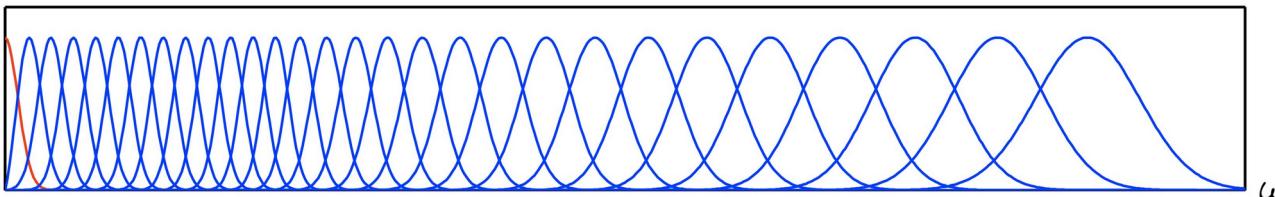
Wavelet Transform

$$X(a, b) = \frac{1}{\sqrt{|a|}} \int x(t) \psi^*\left(\frac{t-b}{a}\right) dt$$

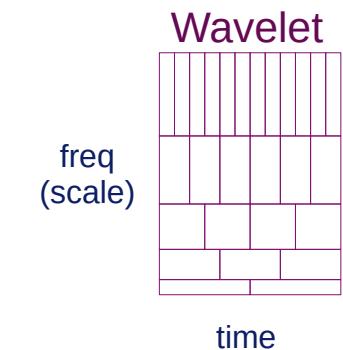
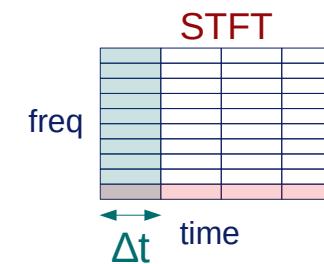
Mother wavelet
 shift
 scale

Example:
 (Complex) Morlet Wavelet

$$\psi(t) \propto \exp\left(-\frac{t^2}{2\sigma^2}\right) \exp(j2\pi f_c t)$$



Const-Q $\leftrightarrow \sigma f_c = \text{Cte}$



$$\text{scale} \propto \frac{1}{\text{frequency}}$$

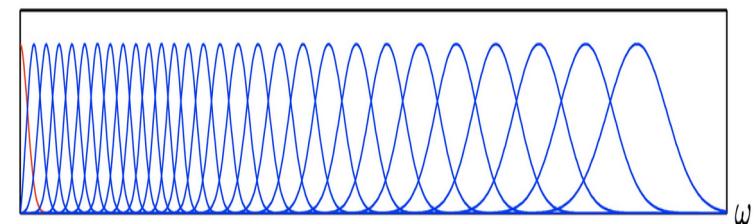
Wavelet Transform

- Wavelet is a filterbank, defined in time domain
- Conv. with each filter (ψ_λ) returns subband signal, $x_\lambda(t)$
- $x_\lambda(t)$ is complex; $| \cdot | \rightarrow$ extract envelop
 - Assume $x_\lambda(t)$ is an *analytic signal*

$$x_\lambda(t) = | x(t) * \psi_\lambda(t) |$$

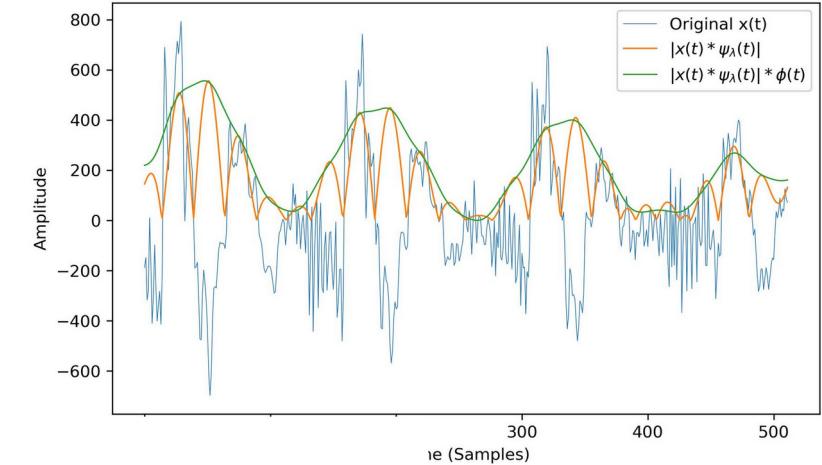
$$x_{\text{analytic}}(t) = x(t) + j\mathcal{H}\{x(t)\}$$

$| x_{\text{analytic}}(t) | = \text{Envelope of } x(t)$



Amplitude Demodulation

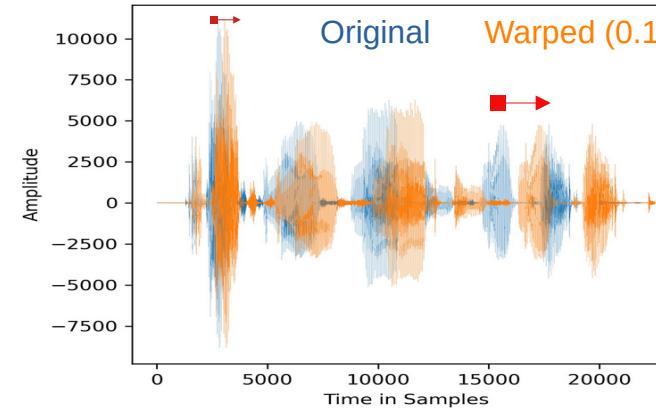
$$|x(t) * \psi_\lambda(t)| * \phi(t)$$



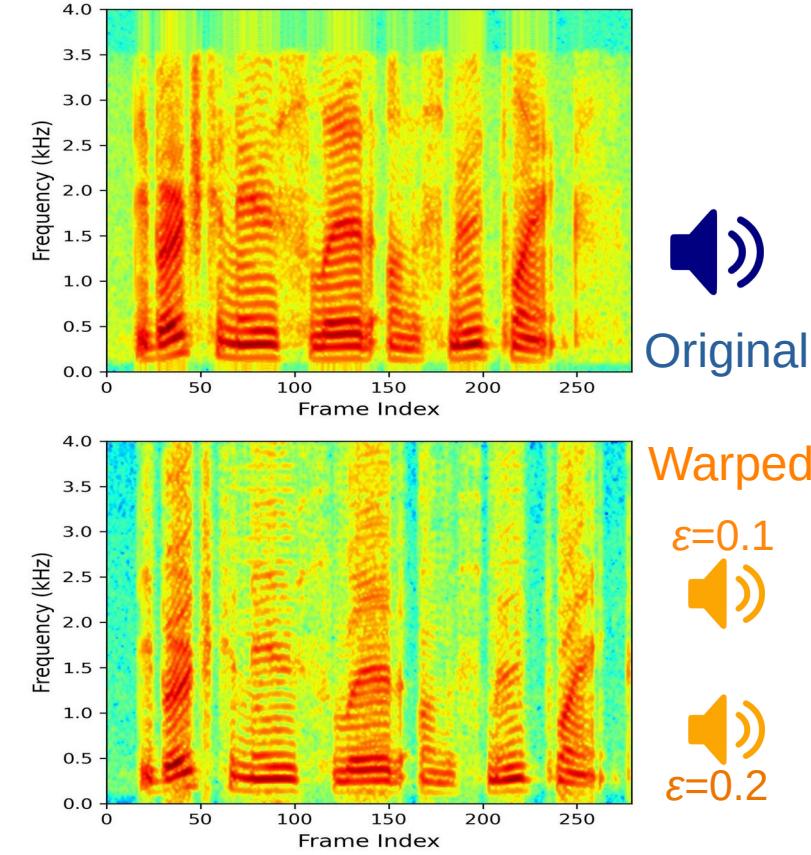
- $\phi(t)$: Low-pass filtering
- $| . | * \phi(t)$: Extract Envelop (amplitude demodulation)
- $x(t) * \psi_\lambda(t)$: Extract subband signal
- $|x(t) * \psi_\lambda(t)| * \phi(t)$: Extract envelop of subband signal

Time-warping Deformation (TWD)

- Variable **time-shift**
 - Definition: $x(t) = x_t(t - \tau(t))$
 - Example: $\tau(t) = \varepsilon t$



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Lipschitz Stability

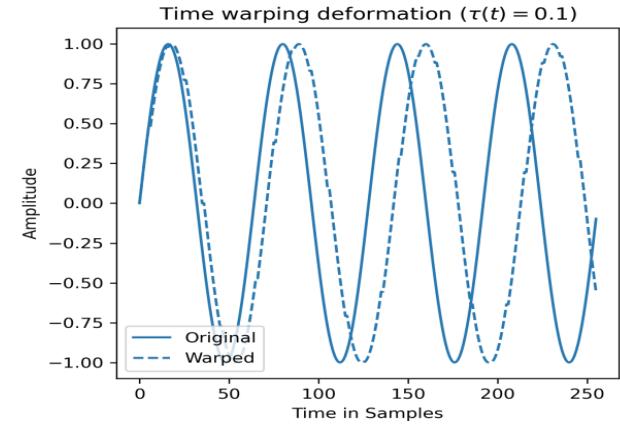
- Stability: small deformation in $x \implies$ small change in $\Phi(x)$
 - Deformation size measured by $\text{Sup}_t |\nabla \tau(t)|$
 - Change size \rightarrow Euclidean distance
- $\Phi(x)$ is Lipschitz stable to deformation $x_\tau(t)$ if a $C > 0$ exists s.t.

$$\forall \tau, \|\Phi(x) - \Phi(x_\tau)\| \leq C \sup_t |\nabla \tau(t)| \|x\|$$

- The lower the C , the higher the stability

Spectrogram ($|X(t, \omega)|$)

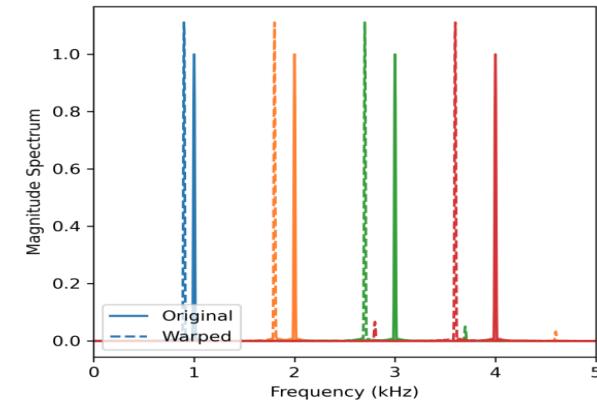
- Invariant to time-shift (c) $[:-]$
- Unstable to TWD $[:-|]$
 - Larger $\omega \rightarrow$ Larger $\Delta\omega \Rightarrow$ No C!



$$x_c(t) = x(t - c) \xrightarrow{\mathcal{F}} e^{-j\omega_k n_0} X(\omega) \xrightarrow{|.|} |X(\omega)|$$

$$x_\tau(t) = x(t - \tau(t)) = x(t - \epsilon t)$$

$$x_\tau(t) \xrightarrow{\mathcal{F}} X_\tau(\omega) = \frac{1}{1 - \epsilon} X\left(\frac{\omega}{1 - \epsilon}\right) \xrightarrow{|.|} \approx |X_\tau(\omega)|$$



Mel-Spectrogram

- $H(\omega; \lambda_i)$: frequency response of i^{th} filter (λ_i = centre frequency)
- Role: frequency **averaging** + subsampling \approx avg pooling
 - Makes $Mx(t; \lambda_i)$ Lipschitz stability (unlike $|X(t, \omega)|$) [:-)]
 - Brings about irreversible information loss [:-(|

$$\begin{aligned} Mx(t, \lambda_i) &= \int_{\omega} |X(t, \omega)|^2 |H(\omega; \lambda_i)|^2 d\omega \\ &= \int_{t'} |x(t, t') * h(t'; \lambda_i)|^2 dt' \end{aligned}$$



Plancherel
Theorem

Mel-Spectrogram

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$$\begin{aligned} Mx(t, \lambda_i) &= \int_{\omega} |X(t, \omega)|^2 |H(\omega; \lambda_i)|^2 d\omega \\ &\approx |x(t) * h(t; \lambda_i)|^2 * \phi^2(t) \end{aligned}$$



Proof in
Appendix A

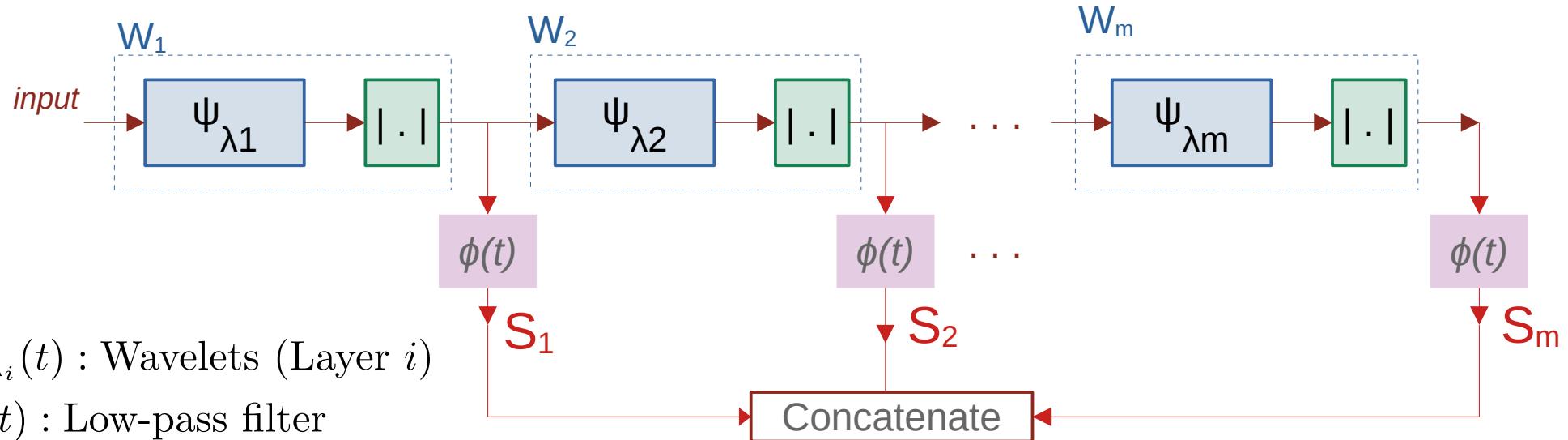


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Scattering Transform (1)

- A cascade of Wavelet transforms (linear) and modulus (non-linear)



Scattering Transform (2)

- A cascade of Wavelet ($\Psi_\lambda(t)$) transforms and modulus ($| \cdot |$)

0th order $S_0 x(t) = x(t) * \phi(t)$

1st order $S_1 x(t, \lambda_1) = |x(t) * \psi_{\lambda_1}(t)| * \phi(t)$

2nd order $S_2 x(t, \lambda_1, \lambda_2) = || x(t) * \psi_{\lambda_1}(t) | * \psi_{\lambda_2}(t) | * \phi(t)$

.

⋮

.

Mth order $S_m x(t, \lambda_1, \dots, \lambda_m) = | \dots |x(t) * \psi_{\lambda_1}(t)| * \dots | * \psi_{\lambda_m}(t)| * \phi(t)$



Scattering Transform (3)

- A cascade of Wavelet ($\psi_\lambda(t)$) transforms and modulus
- $\phi(t)$: low-pass filter \rightarrow averaging \rightarrow **stability** + **information loss**

0th order $S_0 x(t) = x(t) * \phi(t)$

1st order $S_1 x(t, \lambda_1) = |x(t) * \psi_{\lambda_1}(t)| * \phi(t)$

2nd order $S_2 x(t, \lambda_1, \lambda_2) = || x(t) * \psi_{\lambda_1}(t) | * \psi_{\lambda_2}(t) | * \phi(t)$

⋮

Mth order $S_m x(t, \lambda_1, \dots, \lambda_m) = | \dots | x(t) * \psi_{\lambda_1}(t) | * \dots | * \psi_{\lambda_m}(t) | * \phi(t)$



Role of Scattering Coefficients

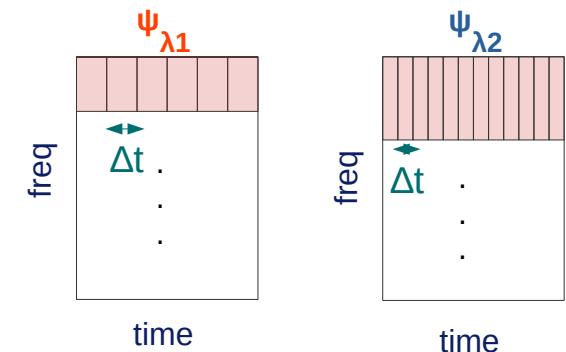
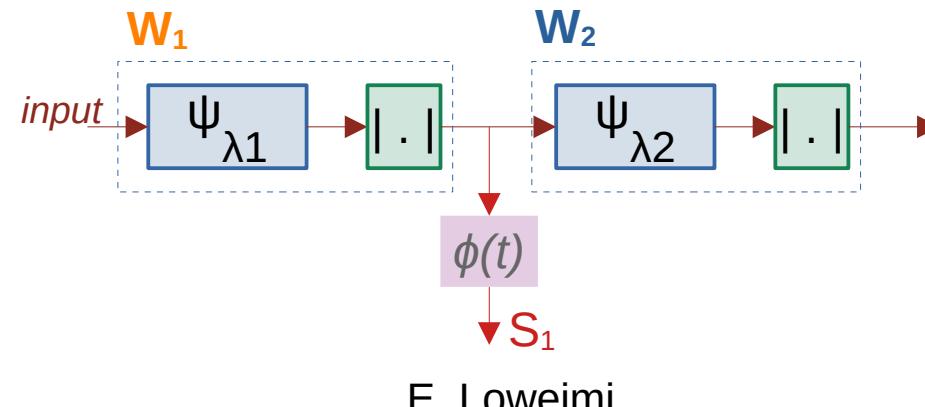
- First order scattering coef. (S_1) \equiv filterbank energies
- S_m aims at compensating for lost info in S_{m-1}
- Information loss ... due to low-pass filtering ...
 - Fast temporal transients (high freq.) info, e.g. attack, is lost!

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- S_m aims at compensating for lost info in S_{m-1}
- Information loss ... due to low-pass filtering ...
 - Fast temporal transients (high freq.) info, e.g. attack, is lost!
- **Solution:** Another transform with a higher *time resolution*
 - ... should better localise the transients in time

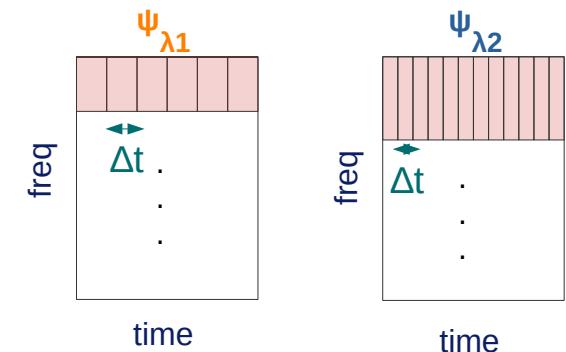
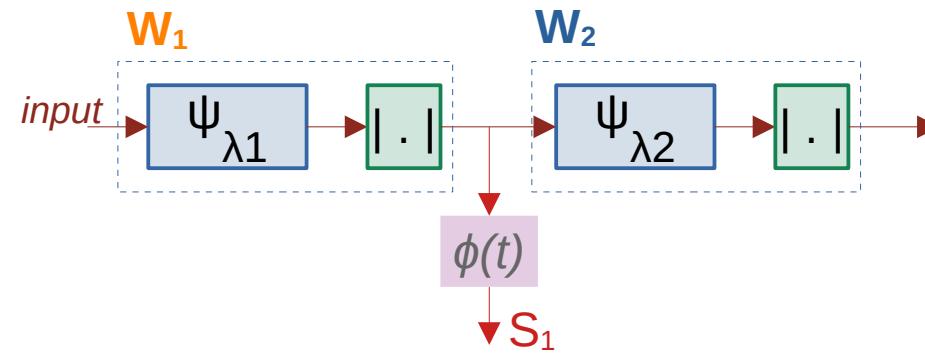
Role of high order Scattering Coef.

- Ψ_{λ_2} should have a smaller Δt than Ψ_{λ_1}
 - Ψ_{λ_2} 's filters should be narrower in time domain
 - wider in frequency domain



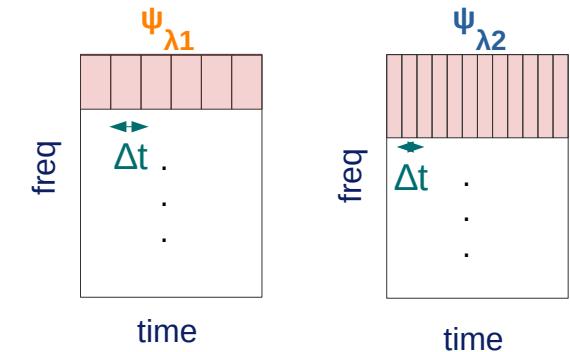
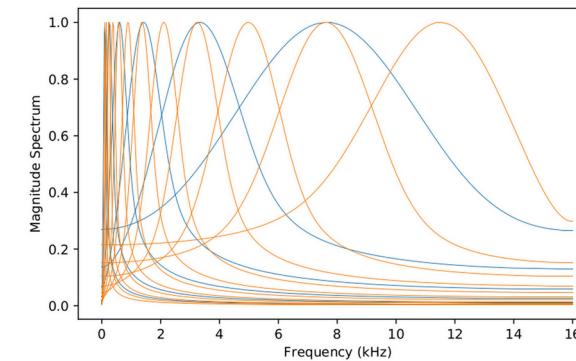
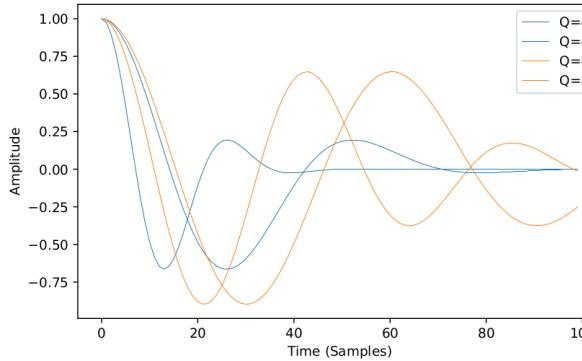
Role of high order Scattering Coef.

- Ψ_{λ_2} should have a smaller Δt than Ψ_{λ_1}
 - Ψ_{λ_2} 's filters should be narrower in time domain (wider in Hz)
- Ψ_λ is in a constant-Q filterbank ($Q \equiv \text{knob}$)
 - $Q_1 > Q_2$ or $Q_1 < Q_2$?



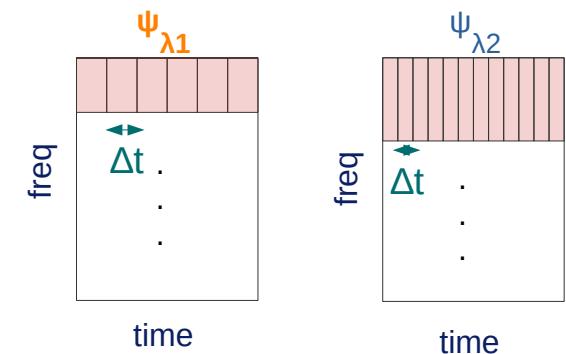
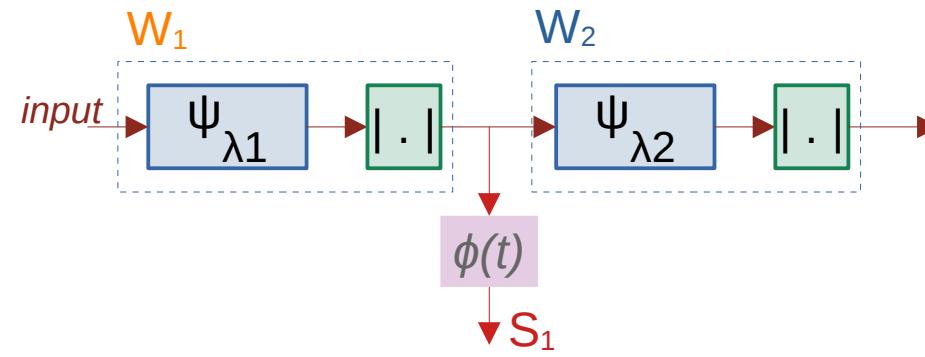
Role of high order Scattering Coef.

- Ψ_{λ_2} should have a smaller Δt than Ψ_{λ_1}
 - Ψ_{λ_2} 's filters should be narrower in time domain
- Smaller $Q \rightarrow$ filters wider in freq domain \rightarrow narrower in time



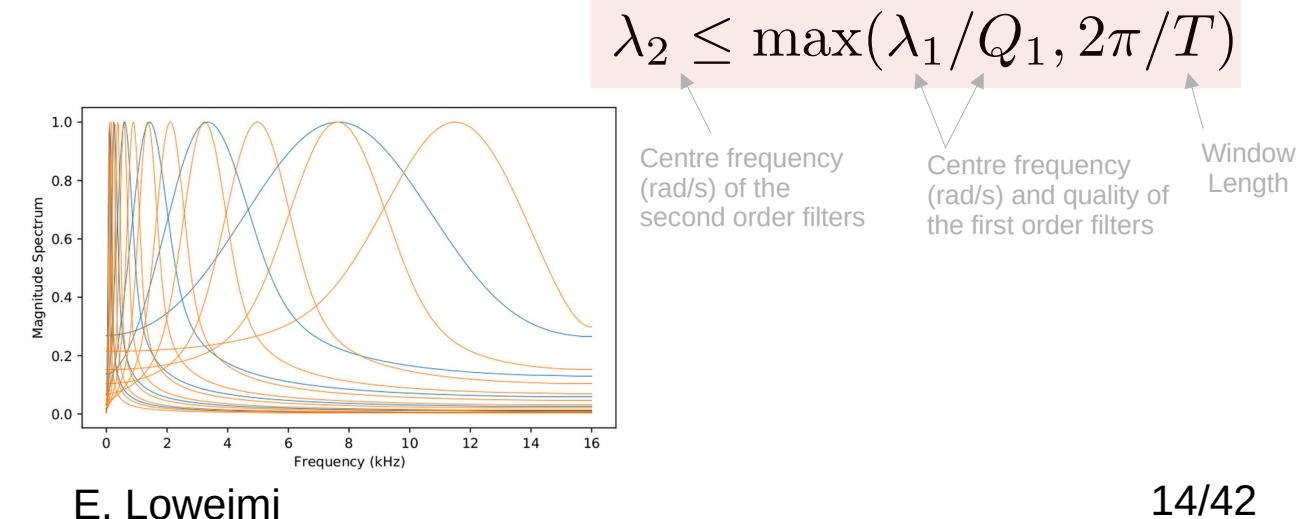
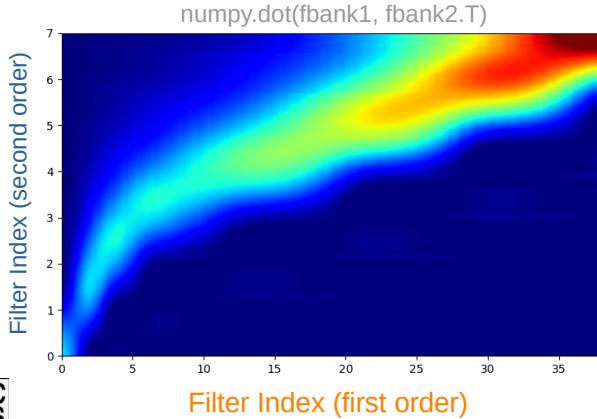
Role of high order Scattering Coef.

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 - Ψ_{λ_2} 's filters should be narrower in time domain
- Ψ_λ is in a constant-Q filterbank ($Q \equiv \text{knob}$)
 - ✓ $Q_2 < Q_1$

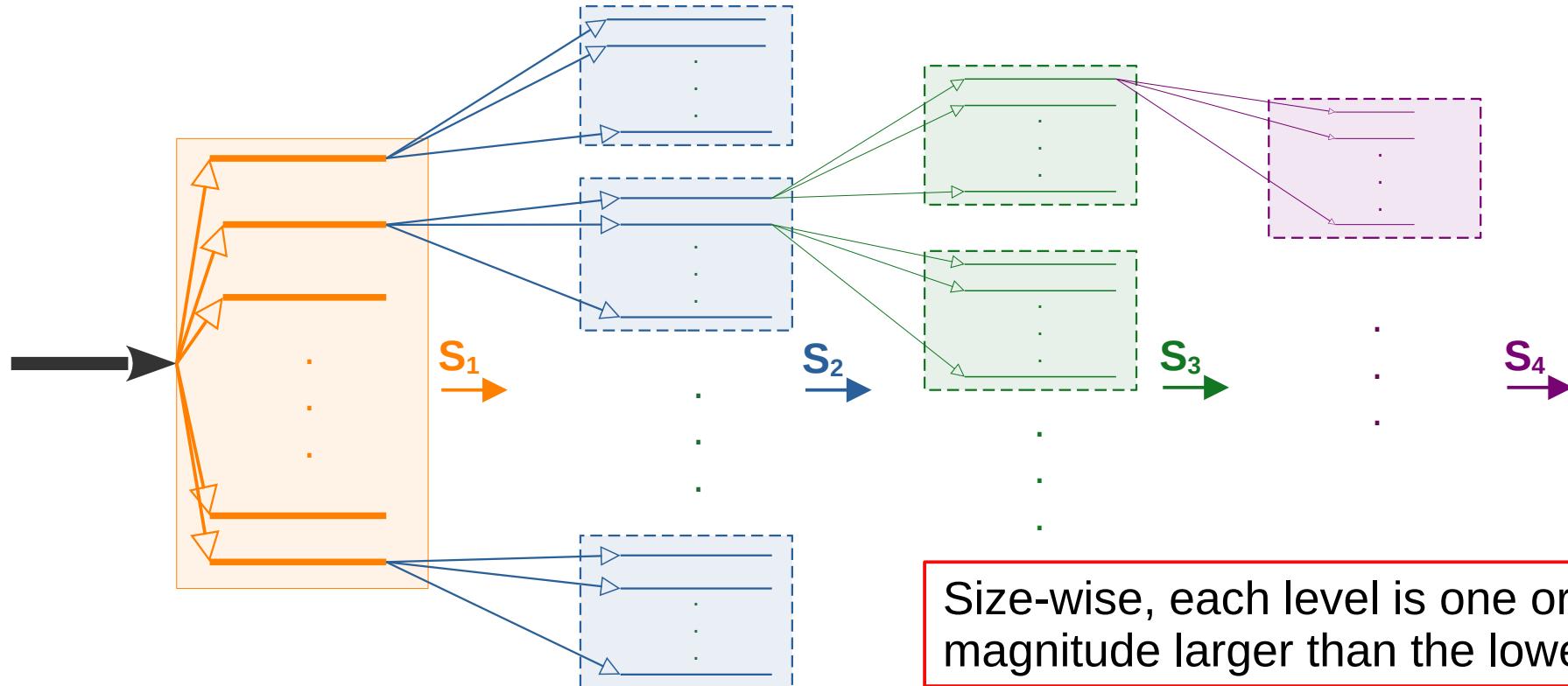


Sparsity of Higher Order Coef

- Higher order coef are sparse (mostly zero)
- Non-zero if Ψ_{λ_1} and Ψ_{λ_2} overlap
- Only compute *non-negligible* coefficients ...



Dimension of Scattering Coef.



Energy (?) of Scattering Coef.

- For 25ms signal decomposition ...
 - 94.5% of energy is in S_1 , $\sim 4.8\%$ in S_2
- By frame extension energy of high order Coef. increases
 - Not useful for speech, but may be music

Averaged $\| S_m x \|^2 / \| x \|^2$

T	$m = 0$	$m = 1$	$m = 2$	$m = 3$
23 ms	0.0%	94.5%	4.8%	0.2%
93 ms	0.0%	68.0%	29.0%	1.9%
370 ms	0.0%	34.9%	53.3%	11.6%
1.5 s	0.0%	27.7%	56.1%	24.7%

Normalising Scattering Coef.

- Normalise order m with order $m-1$
- Goal: improve invariability, e.g. to channel distortion

$$S_1(t, \lambda_1) = \frac{S_1(t, \lambda_1)}{S_0(t, \lambda_1) + \epsilon}$$

$$S_2(t, \lambda_1, \lambda_2) = \frac{S_2(t, \lambda_1, \lambda_2)}{S_1(t, \lambda_1) + \epsilon}$$

Silence
detection
threshold
(to avoid x/0)

$$S_m(t, \lambda_1, \dots, \lambda_m) = \frac{S_m(t, \lambda_1, \dots, \lambda_m)}{S_{m-l}(t, \lambda_1, \dots, \lambda_{m-1}) + \epsilon}$$

Normalising Scattering Coef.

- Normalise order m with order $m-1$
- Goal: improve invariability, e.g. to channel distortion

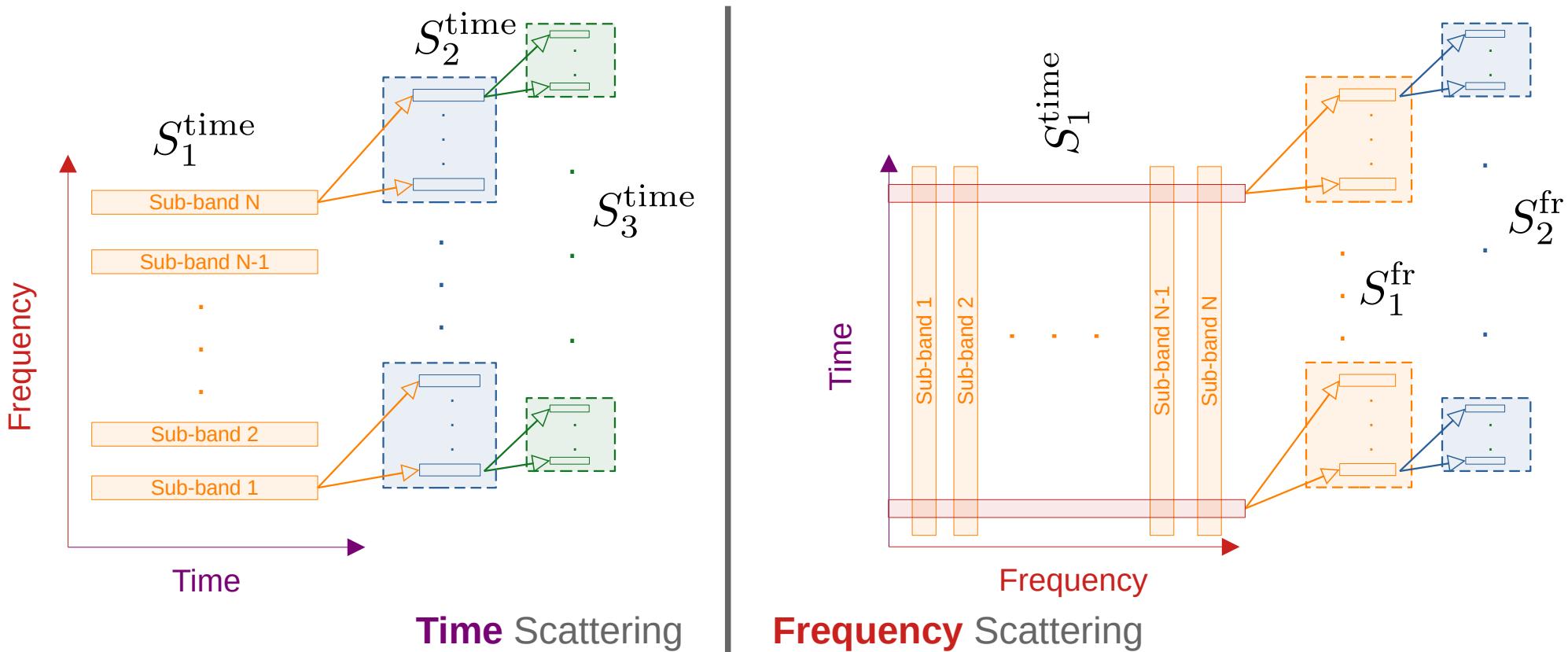
$$S_m(t, \lambda_1, \dots, \lambda_m) = \frac{S_m(t, \lambda_1, \dots, \lambda_m)}{S_{m-1}(t, \lambda_1, \dots, \lambda_{m-1}) + \epsilon}$$

$$h(t) * \psi_\lambda(t) \approx |H(\omega = \lambda)| \psi_\lambda(t)$$
$$| (x(t) * h(t)) * \psi_\lambda(t) | \approx |H(\omega = \lambda)| |x(t) * \psi_\lambda(t)|$$

Holds only when $H(\omega)$ is approximately constant over support of $\psi(\omega; \lambda)$

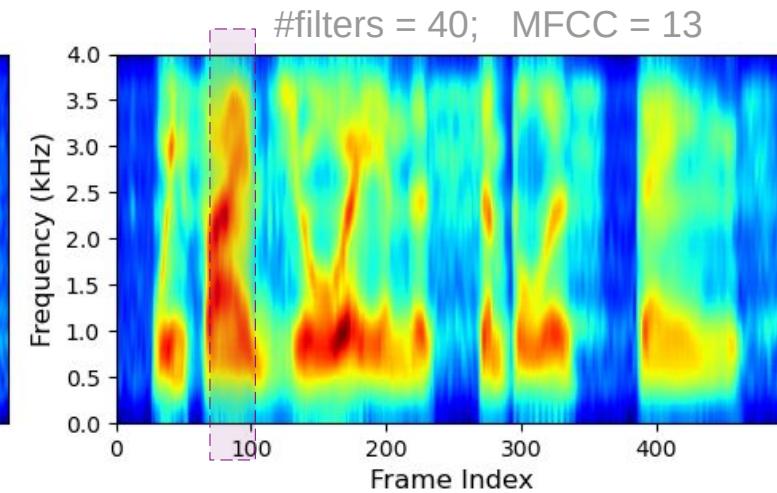
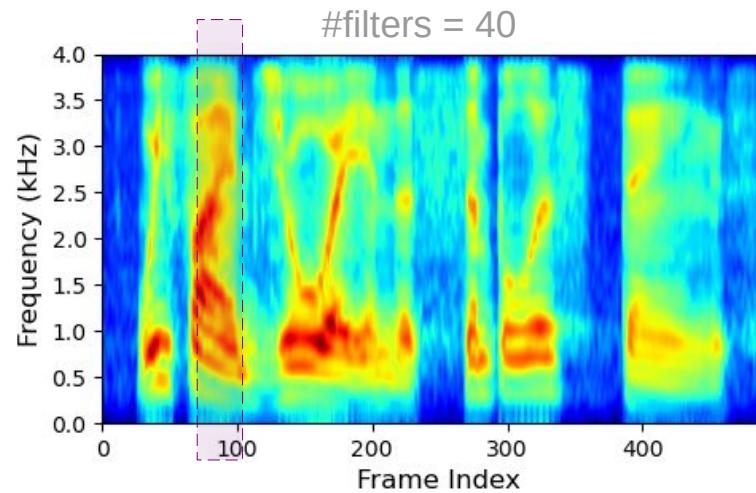


Frequency Scattering (1)



Frequency Scattering (2)

- Similar to freq. avg. by setting higher order MFCCs to 0
- Provides stability to *frequency transposition*



Frequency Scattering (2)

- Similar to freq. avg. by setting higher order MFCCs to 0
- Provides stability to *frequency transposition*
- Only the first-order is used, with small Q (e.g. Q=1)
- Filters are centred at ***quefrency*** λ

$$S^{\text{fr}} z(\gamma, \lambda_q) = |z(\gamma) * \psi_{\lambda_q}(\gamma)| * \phi^{\text{fr}}(\gamma)$$

$$\gamma = \log_2(\lambda)$$

Experimental Results

- Second order helps
 - Especially for music (Y?)
- Third order may slightly help
 - Costly because of dimension
- Freq. scattering helps

Representations	GTZAN	TIMIT
$\Delta\text{-MFCC}$ ($T = 23$ ms)	20.2 ± 5.4	18.5
$\Delta\text{-MFCC}$ ($T = 740$ ms)	18.0 ± 4.2	60.5
State of the art (excluding scattering)	9.4 ± 3.1 [8]	16.7 [43]
	$T = 740$ ms	$T = 32$ ms
Time Scat., $l = 1$	19.1 ± 4.5	19.0
Time Scat., $l = 2$	10.7 ± 3.1	17.3
Time Scat., $l = 3$	10.6 ± 2.5	18.1
Time & Freq. Scat., $l = 2$	9.3 ± 2.4	16.6
Adapt Q_1 , Time & Freq. Scat., $l = 2$	8.6 ± 2.2	15.9

* GTZAN: Music Genre Classification

* TIMIT: phone classification

* Classifier: SVM with Gaussian Kernel

* Adapt → multi-resolution: $Q=1, 8$

Sturm, 2012, “An Analysis of the GTZAN Music Genre Dataset”
 “... 5% ... exact duplicates, 10.8% is mislabelled ...”



Properties of Scattering Transform

- Similar to CNNs (hierarchical) but involves no learning
 - Learns a general (not task-specific) representation; interpretable
- Translation-invariant, stable to deformation, preserves info
- Some similarities to physiological models (cochlea, const-Q)
- Energy conservative and contractive mapping
- Has approximate and non-trivial inverse transformation
- Poorer frequency resolution than STFT



DEEP SCATTERING SPECTRUM WITH DEEP NEURAL NETWORKS

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This paper investigates ...

- Usefulness of ...
 - DSS for TIMIT phone recognition
 - Multi-resolution DSS
- Optimal architecture for ...
 - Processing S_1 and S_2 , simultaneously
 - Multi-resolution DSS

Experimental Setup

- Task: TIMIT phone recognition
- Baseline: 40-dim log-mel fbank + Δ + $\Delta\Delta$
- DNN: 2 x CNN (256 filters) → 3 x FC (1024)
- Output/Target: CI (147) and CD (2400)
- **MVN** for log-Mel and S_1 ; **MN** for S_2
 - Scatter transfer operator act like var-norm (?)
- **Delta** only for log-mel and S_1 ; not S_2 [not beneficial]

Experimental Results – TIMIT

- PERs of log-Mel and S_1 are **similar**
 - TIMIT, **0.3**, statistically significant?
- Using S_2 may help, but **NOT** consistently!
 - Why? Functionality overlap ...
 - Δ and S_2 ? $\Delta\Delta$ and S_3 ?
- ReLU and Regularisation help

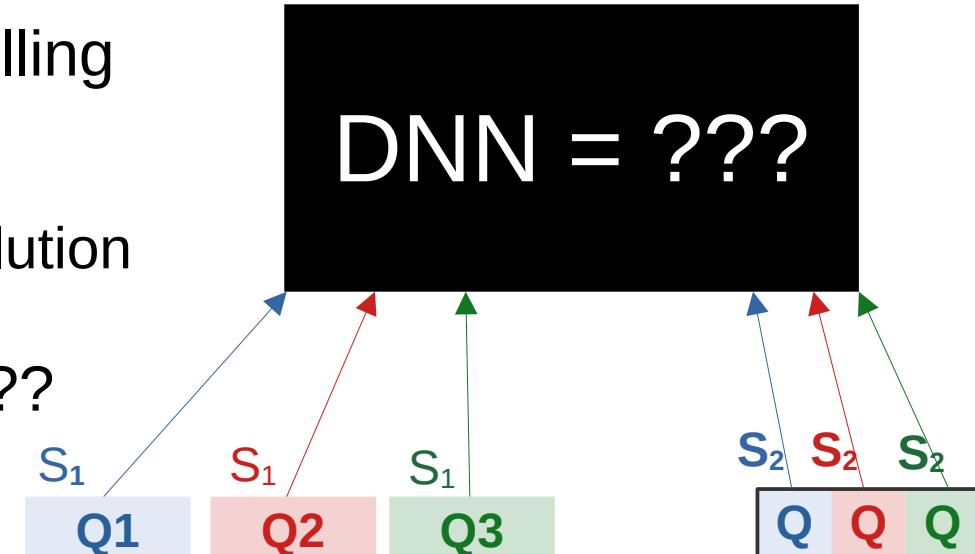
Feature	PER	
	CI	CD
<i>logmel + Δ + $\Delta\Delta$</i>	19.3	18.7
$S_1x(t, \lambda_1) + \Delta + \Delta\Delta$	↓ 19.0	18.7

Non-linearity	$S_1 + \Delta + \Delta\Delta$	$S_1 + \Delta + \Delta\Delta + S_2$
Sigmoid	21.3	20.9
ReLU	20.0	20.3
ReLU+regularization	19.0	18.8

- * CI: context-independent (147)
- * CD: context-dependent (2400)
- * Regularisation: MaxNorm and Dropout

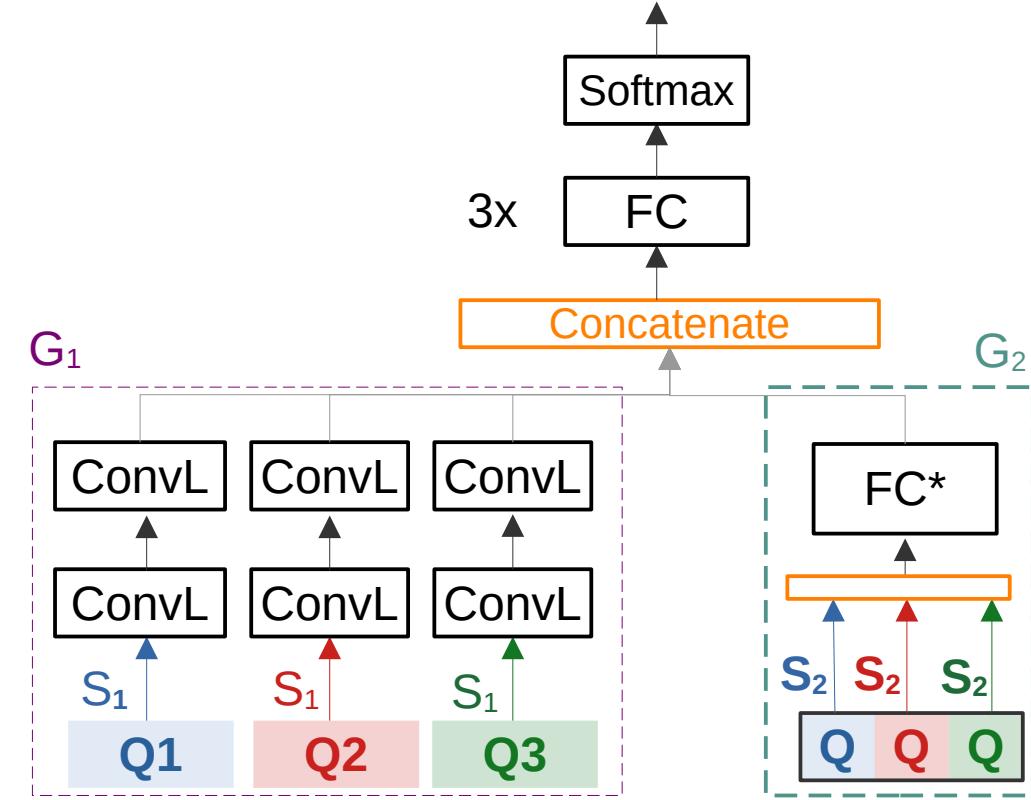
Multi-Resolution Approach

- Use multiple filterbanks with various Qs
 - ONLY for S_1 ; $S_2 \leftrightarrow$ always $Q=1$
- Advantage: complementary modelling
 - Small Q → better time resolution
 - Large Q → better frequency resolution
- Optimal architecture to combine???



Architecture for Multi-Resolution

- Multi-resolution \equiv Various Qs
- Process S_1 with (2x) ConvL
- Process S_2 with FC*
 - S_2 is sparse + Limited local corr
 - Not optimal for ConvL
 - Too short filters



Multi-Resolution -- TIMIT -- CI

- Multi-resolution helps!
- Multi-resolution for S_1 (G_1) is more helpful than S_2
 - 0.6 vs 0.2
- Optimal width for FC* is 128

Feature Stream	PER
$S_1 + \Delta + \Delta\Delta$	19.0
$G_1 + \Delta + \Delta\Delta$	18.4
$S_1 + \Delta + \Delta\Delta + S_2$	18.8
$G_1 + \Delta + \Delta\Delta + G_2 + 1024$ HU	19.1
$G_1 + \Delta + \Delta\Delta + G_2 + 256$ HU	18.7
$G_1 + \Delta + \Delta\Delta + G_2 + 128$ HU	18.2
$G_1 + \Delta + \Delta\Delta + G_2 + 64$ HU	18.6

- * G_1 : multi-resolution S_1
- * G_2 : multi-resolution S_2
- * HU: #hidden units of FC*

Multi-Resolution -- TIMIT -- CD

- Using S_2 helps
 - PER: 18.7 → 17.9 [0.8]
 - For CI: 19.0 → 18.8 [0.2]
- Multi-Resolution helps
 - PER: 17.9 → 17.4 [0.5]
 - For CI: 18.8 → 18.2 [0.6]

Feature Stream	PER
$S_1 + \Delta + \Delta\Delta$	18.7
$S_1 + \Delta + \Delta\Delta + S_2$ 128 HU	17.9
$G_1 + \Delta + \Delta\Delta + G_2 + 128$ HU	<u>17.4</u>

- * G_1 : multi-resolution S_1
- * G_2 : multi-resolution S_2
- * HU: #hidden units of FC*



Deep Scattering Spectra with Deep Neural Networks for LVCSR Tasks

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14-18 September 2014, Singapore



This paper investigates ...

- LVCSR (BN: 50h; BN: 430h)
- Multi-resolution + frequency scattering effect
- Dimensionality reduction
- Speaker adaptation
- Sequence training

Experimental Results

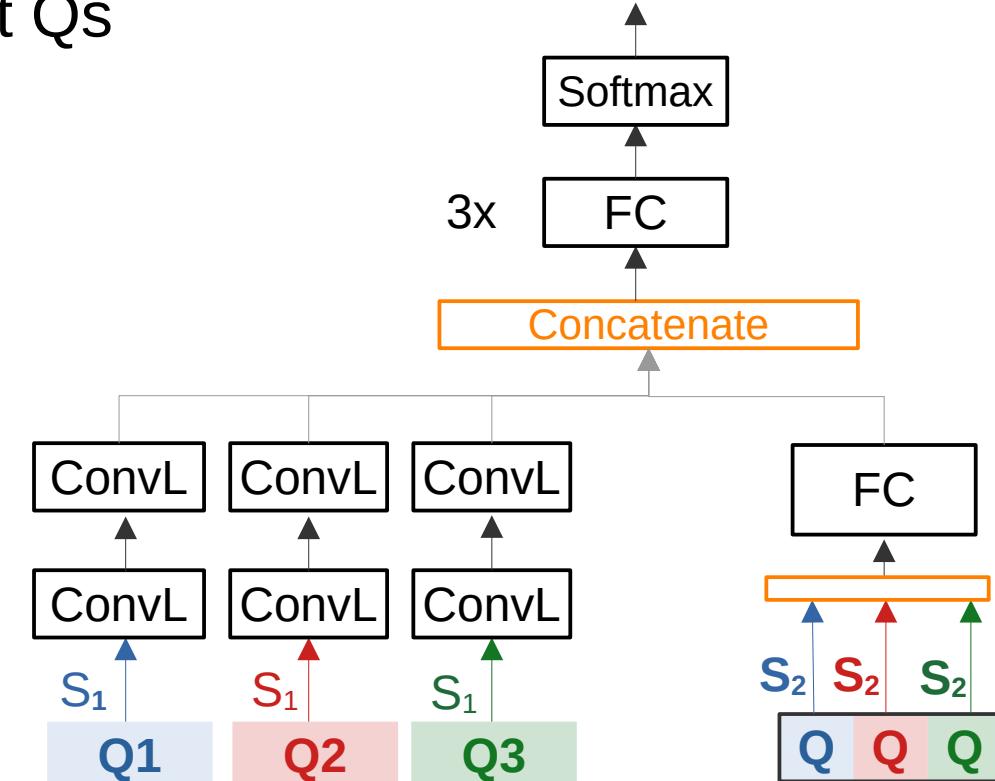
- $S_1(+S_2)$ is comparable to log-mel!
- S_2 slightly helps!
 - WER: 16.0 → 15.9
- Frequency scattering helps!
 - WER: 15.9 → 15.5
- Gain carries over to larger tasks

English Broadcast News, 50h

Feature	WER
log-mel baseline	15.9
S_1 , time	16.0
S_1+S_2 , time	15.9
S_1+S_2 , time+frequency	15.5

Multi-Resolution Approach

- Multiple filterbanks with different Qs
- Various Qs ONLY for S_1
 - For S_2 , always Q=1
- S_1 modelled by ConvL
- S_2 modelled by FC
 - S_2 is sparse; Limited local corr
 - Not optimal for ConvL



Experimental Results – Multi-Resolution

- $Q=8$ is optimal
 - Consistent with human system
- Multi-resolution helps
 - Best $Q=(8,13)$
- Time+Frequency scattering helps
 - Not if Q is too low!

Feature	WER Time Scat.	WER Time+Freq. Scat.
log-mel baseline	15.9	15.9
S_1+S_2 ($Q=1$)	20.5	25.0
S_1+S_2 ($Q=4$)	16.2	16.3
S_1+S_2 ($Q=8$)	15.9	15.5
S_1+S_2 ($Q=13$)	16.1	15.7
S_1+S_2 ($Q=1,8$)	15.7	15.5
S_1+S_2 ($Q=1,13$)	15.5	15.5
S_1+S_2 ($Q=4,13$)	15.6	15.1
S_1+S_2 ($Q=8,13$)	15.3	15.1
S_1+S_2 ($Q=1,4,13$)	15.7	-

Dimensionality Reduction of S_1 & S_2

- Dim. Reduction methods ...
 - $S_2 \rightarrow$ PCA & LDA
 - $S_1 \rightarrow$ Linear bottleneck
- Conclusion
 - Identical results with a smaller network

Feature	WER	Params
Baseline $S_1, tf + S_2, tf$ (Q=4,13)	15.1	26.5M
$S_1, tf + \text{pca128}(S_1, f, S_2)$	15.2	14.1M
$S_1, tf + \text{pca256}(S_1, f, S_2)$	15.2	15.5M
$S_1, tf + \text{lda128}(S_1f, S_2)$	15.1	14.1M

Feature	WER	Params
Baseline $S_1, tf + S_2, tf$ (Q=4,13)	15.1	26.5M
$S_1, tf + \text{lda128}(S_1f, S_2)$	15.1	14.1M
$S_1, tf, \text{bn=128} + \text{lda128}(S_1f, S_2)$	15.4	10.0M
$S_1, tf, \text{bn=256} + \text{lda128}(S_1f, S_2)$	15.1	10.8M

Speaker Adaptation

- VTLN helps!
 - ONLY for S_1 (S_2 unwarped)
- fMLLR & i-vector help!
 - Extra input stream to the FC
 - Do not obey locality
 - More effective than VTLN!
- Using 2xConv Layers help!

Feature	WER no VTLN	WER with VTLN
log-mel	15.9	15.4
S_1+S_2 , time+freq, Q=8	15.5	15.0
S_1+S_2 , time+freq, Q=4,13	15.1	14.7

Feature	WER
log-mel +fMLLR+ivectors	13.9
S_1+S_2 , time+freq, Q=4,13	13.4

Feature	WER
joint CNN/DNN	13.4
DNN	14.2

Experimental Results

- Sequence training (after CE) improves the results
- Gain carries over to larger data (50h → 430h)
- Comparing multiQ DSS with log-mel; is it fair?

English Broadcast News, 50h

Feature	WER
log-mel	12.5
S_1+S_2 , time+freq, Q=4,13	12.0

English Broadcast News, 430h

Feature	WER
log-mel	14.2
m1+m2, time+freq, mulitQ	13.2

What are m1 and m2?

“Log-Mel+MFCC” vs DSS

- S_1 and log-mel have identical WER!
- S_2 slightly helps ($15.4 \rightarrow 15.2$)
- Frequency scatter slightly helps ($15.2 \rightarrow 15.0$)
- Frequency scatter effect is similar to MFCC
- MultiQ “log-mel+MFCCs” match DSS with all bells & whistles!

Feature	WER
log-mel, Q=8	15.4
S_1 , time scatter, Q=8	15.4
$S_1 + S_2$ time scatter, Q=8	15.2
$S_1 + S_2$ time+freq scatter, Q=8	15.0
log-mel+mfcc, Q=8	15.0
$S_1 + S_2$ time+freq scatter, Q=4,13	14.7
log-mel + mfcc, Q=4,13	14.6



INTERSPEECH 2020

October 25–29, 2020, Shanghai, China



Deep Scattering Power Spectrum Features for Robust Speech Recognition

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INTERSPEECH 2020

OCTOBER 25-29/ SHANGHAI, CHINA
SHANGHAI INTERNATIONAL CONVENTION CENTER



This paper ...

- Investigates usefulness of DSS (S_1 and S_2) for robustness ASR
- Replaces modulus with squared modulus non-linearity
- Comparison with similar architectures

Replace Modulus with Squared Modulus (1)

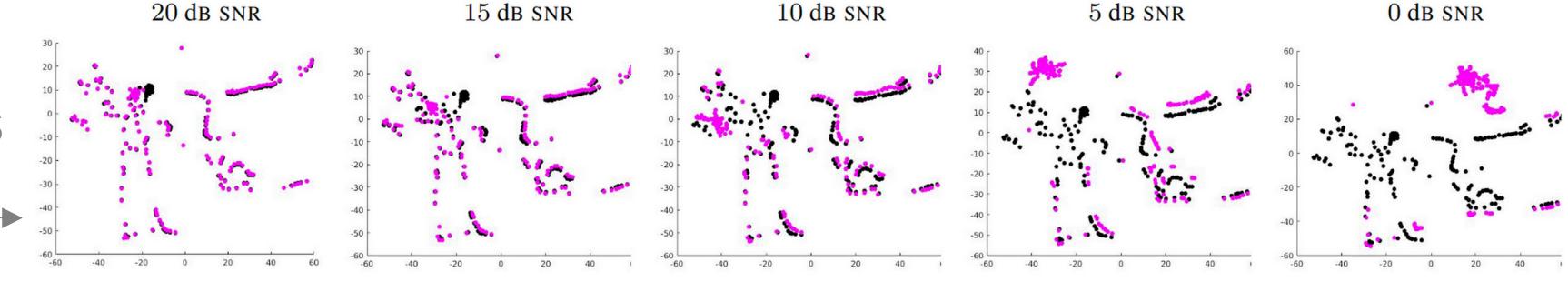
- Amplifies strong coefficients
 - may improve **robustness** + better speech/noise **separation**
- Amplifies *sparsity*

$$\hat{S}_1(t, \lambda_1) = |x(t) * \psi_{\lambda_1}(t)|^2 * \phi(t)$$

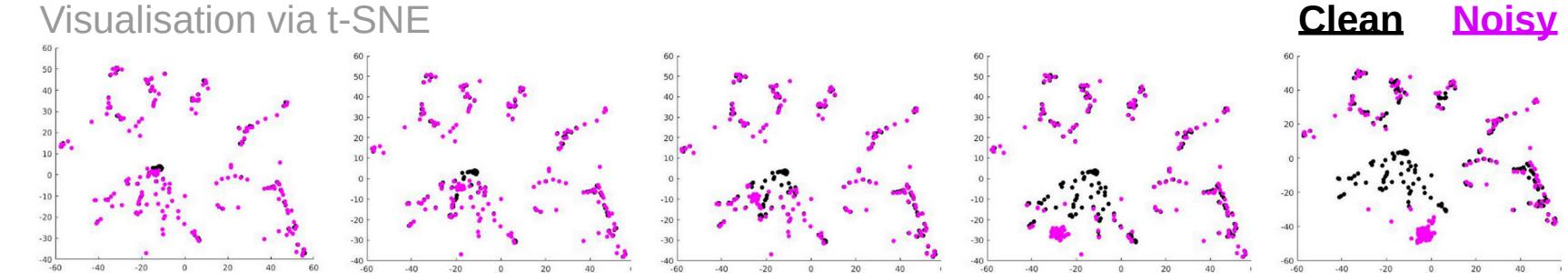
$$\hat{S}_2(t, \lambda_1, \lambda_2) = \left| |x * \psi_{\lambda_1}(t)|^2 * \psi_{\lambda_2}(t) \right|^2 * \phi(t)$$

Replace Modulus with Squared Modulus (2)

Standard DSS
 $(| \cdot |)$



Proposed
 $(| \cdot |^2)$



Visualisation via
t-SNE

$| \cdot |^2$ is less variant to the additive noise (Guassian) at different SNRs.

Experimental Results – Aurora-4

Clean

Very good WER
for this task!

FEATURES	A ₁	B _{2–7}	C ₈	D _{9–14}	AVG _{1–14}
DSPS ₁	2.76	13.83	7.74	17.90	14.35
DSPS ₁ + DSPS ₂	2.58	11.14	6.89	14.33	11.59
DSS ₁ [5]	2.62	14.72	7.89	19.07	15.23
DSS ₁ + DSS ₂ [5]	2.61	11.95	7.33	15.33	12.40
FBANK ₄₀ [4]	2.65	13.75	7.96	16.89	13.89
FBANK ₆₀ [4]	2.54	13.06	8.33	17.08	13.69
FBANK ₈₀ [4]	2.69	12.04	8.03	16.19	12.86
FBANK ₁₀₀ [4]	2.52	12.60	7.60	16.52	13.20

- * Squared modulus → Helps! → 0.9, 0.8% abs
- * Second-order features → Helps! → 2.8, 2.8% abs

Experimental Results – Aurora-4

Multi (1)

FEATURES	A ₁	B _{2–7}	C ₈	D _{9–14}	AVG _{1–14}
DSPS ₁	2.97	5.88	6.71	15.96	10.05
DSPS ₁ + DSPS ₂	2.73	5.20	4.73	14.15	8.83 ↓
DSS ₁ [5]	2.99	5.69	6.56	15.95	9.96
DSS ₁ + DSS ₂ [5]	2.86	5.45	6.11	15.08	9.44 ↓
FBANK ₄₀ [4]	3.06	6.08	7.10	16.09	10.23
FBANK ₆₀ [4]	2.90	5.72	6.46	15.65	9.83
FBANK ₈₀ [4]	2.88	5.58	5.92	15.22	9.55
FBANK ₁₀₀ [4]	2.69	5.33	5.74	15.26	9.43

* Squared modulus [S₁] → WER → Slight WER increase

* Second-order features → Helps! → 1.2, 0.5% abs

Experimental Results – Aurora-4

Multi (2)

- Multi-Resolution is useful but should not be overdone!
 - $Q = \{1, 4, 8, 13\}$ is the worst!
 - Best multi-resolution results
 $\rightarrow Q = \{4, 13\}$
- Comparable results with other complicated DNNs

ARCHITECTURE	CNN DEPTH	AVG _{1–14}
DSPS ₁ + DSPS ₂ (MULTI-RESOLUTION SCATTERING)		
$Q = \{8\}$	3	8.83
$Q = \{1, 4, 13\}$	3	8.76
$Q = \{1, 4, 8, 13\}$	3	8.94
$Q = \{4, 13\}$	3	8.64
FBANK BASELINES		
FMLLR + MLP	-	10.21
VD6CNN [23]	6	10.34
VD10CNN [23]	10	8.81
M-OCT CNN [24]	15	8.31

Wrap-up

- Deep scattering spectrum (DSS) is a cascade of wavelet (linear) and modulus (non-linear) transforms
- Advantages: translation invariant, Lipschitz stable & preserves information
- First-order coefficients are similar to filterbank features
- [Novelty] Higher-order aims at recovering lost info in lower level; sparse
 - Usually only first (S_1) and second (S_2) orders are used
- DSS has similar hierarchical structure to CNNs but involves no learning
- Frequency scattering and multi-resolution time scattering are helpful
- Performance on ASR task: comparable to classic features + marginal gain
- Suggestions: learn S_1 via parametric CNNs, use CNN+group for S_2



That's It!

- Thanks for Your Attention!
- Q/A

SpeechWave



- Appendix A: Proof of $Mx(t, \lambda_i) = \int_{\omega} |X(t, \omega)|^2 |H(\omega; \lambda_i)|^2 d\omega$
- Appendix B: DSS vs ... $\approx |x(t) * h(t; \lambda_i)|^2 * \phi^2(t)$



Appendix A: Proof

STFT

$$\chi(t, \omega) = \int_{-\infty}^{\infty} \chi(u) \phi(u-t) e^{-j\omega u} du$$

Frame index
of windowed cosine

$M_X(t, \lambda) = \frac{1}{2\pi} \int_{-\infty}^{\infty} |\chi(t, \omega)|^2 |\psi_\lambda(\omega)|^2 d\omega$

 $= \frac{1}{2\pi} \int_{-\infty}^{\infty} |\chi(t, \omega) \psi_\lambda(\omega)|^2 d\omega$

Planckel's Finsler

or Pascual's theorem

$$= \int_{-\infty}^{\infty} \underbrace{\chi(u) * \psi_\lambda(u)}_{\text{time adapt. window}} \overbrace{|\phi(u-t) \psi_\lambda(u-t)|^2}^{\text{N: parameter of wavelet } \psi_\lambda(\omega)} du$$

N: centre freq

* Hcde: Convolution

$\begin{aligned} &= \int_{-\infty}^{\infty} \left| \chi(v) \phi(v-t) \psi_\lambda(v-t) \right|^2 du \\ &\quad \approx \int_{-\infty}^{\infty} \left| \chi(v) \phi(v-t) \psi_\lambda(v-t) \right|^2 du \\ &\quad \approx \int_{-\infty}^{\infty} \left| \chi(v) \psi_\lambda(v-t) \right|^2 \left| \phi(v-t) \right|^2 du \\ &\quad \approx \int_{-\infty}^{\infty} \underbrace{\chi(v) \psi_\lambda(v-t)}_{\text{2.5ms}} \underbrace{\left| \phi(v-t) \right|^2}_{\substack{\text{phi in time} \\ \text{much shorter than} \\ \psi, if } \lambda \gg 400} du \\ &= \chi(u) * \psi_\lambda(u) \end{aligned}$

because $\phi(t) = \phi(-t)$
 ϕ is symmetric

$M_X(t, \lambda) \approx \underbrace{\chi(t) * \psi_\lambda(t)}_{\substack{\text{constant envelope} \\ \text{amplitude modulation}}} \underbrace{\left| \phi(t) \right|^2}_{\substack{\text{constant amplitude} \\ \text{single}}$

assuming $\chi(t, \lambda) = \chi(t) \psi_\lambda(t)$

Appendix B: DSS vs Modulation Spectrum

Speech Communication 25 (1998) 117-132

Robust Speech Recognition Using the Modulation Spectrogram

Brian Kingsbury, Nelson Morgan and Steven Greenberg

