

# Unit - II

## 2.16 Resonance

**Dr.Santhosh.T.K.**

# Syllabus

## UNIT – II

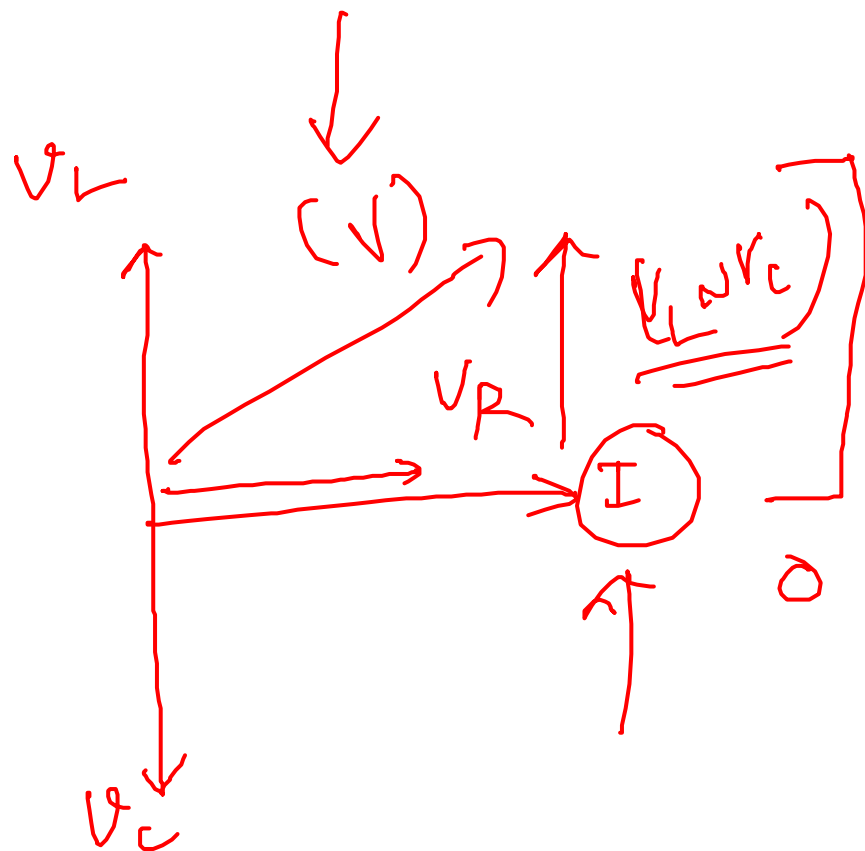
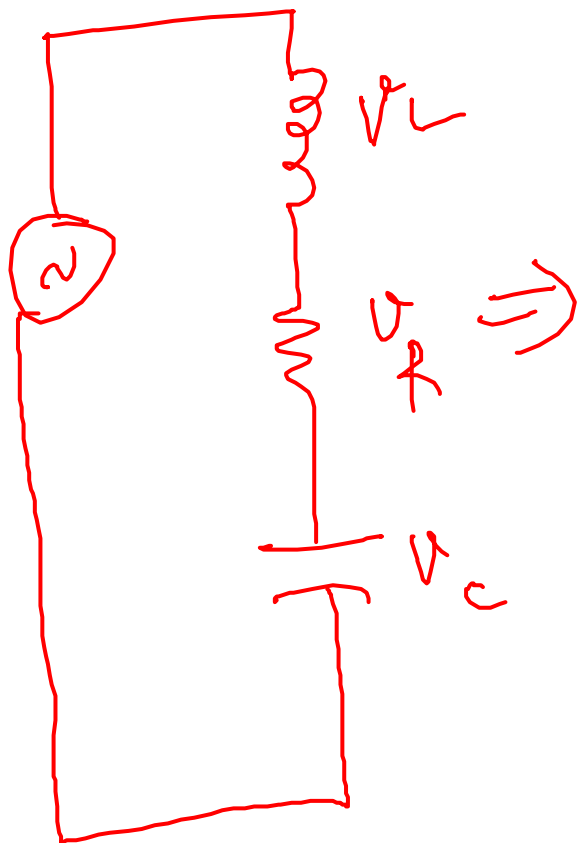
**14 Periods**

**DC Circuit Analysis:** Voltage source and current sources, ideal and practical, Kirchhoff's laws and applications to network solutions using mesh analysis, - Simplifications of networks using series- parallel, Star/Delta transformation, DC circuits-Current-voltage relations of electric network by mathematical equations to analyse the network (Superposition theorem, Thevenin's theorem, Maximum Power Transfer theorem), Transient analysis of R-L, R-C and R-L-C Circuits.

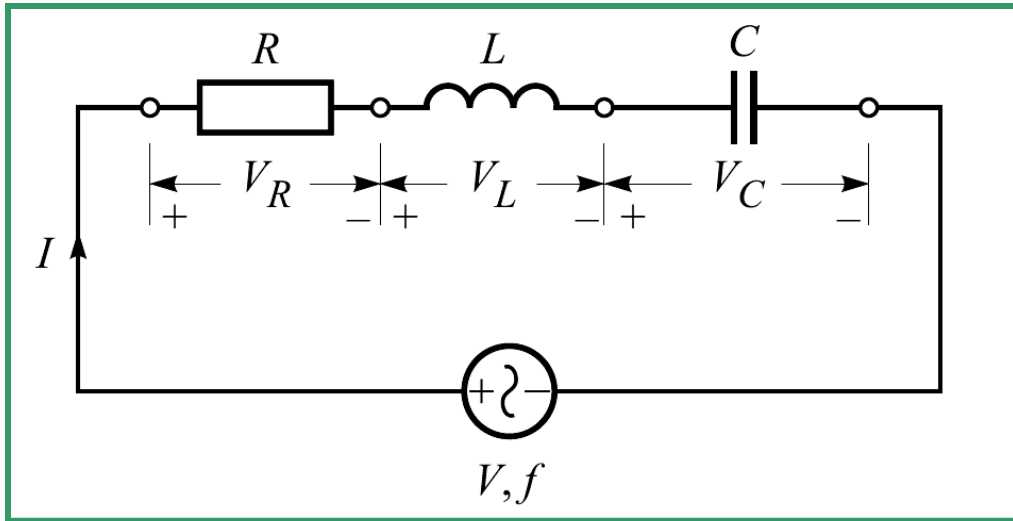
**AC Steady-state Analysis:** AC waveform definitions - Form factor - Peak factor - study of R-L - R-C -RLC series circuit - R-L-C parallel circuit - phasor representation in polar and rectangular form - concept of impedance - admittance - active - reactive - apparent and complex power - power factor, Resonance in R-L-C circuits - 3 phase balanced AC Circuits

# Introduction

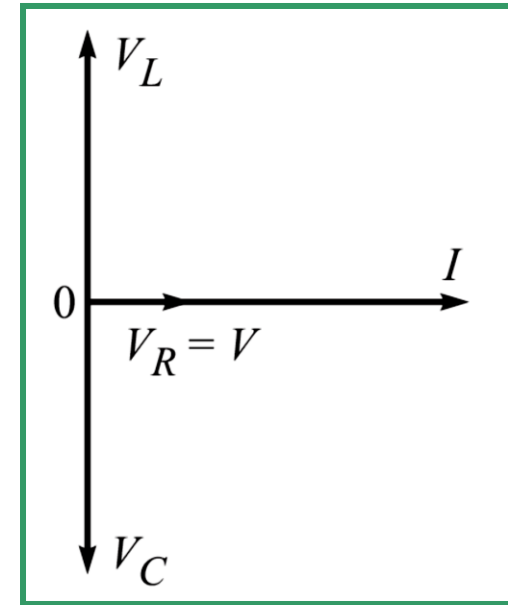
- **Resonance** is the condition that exists in ac circuits when the input current is in phase with the input voltage.
- When in resonance, the ac circuit is **purely resistive** and draws power at **unity power factor**.



# Series Resonant Circuit



(a) The circuit.



(b) Phasor diagram at resonance.

**The resonance in a series *RLC* circuit requires that**

$$X_{L0} - X_{C0} = 0 \quad \text{or} \quad X_{L0} = X_{C0} \quad \text{or} \quad \omega_0 L = \frac{1}{\omega_0 C}$$

**The frequency of resonance,**

$$\omega_0 = \frac{1}{\sqrt{LC}} \quad \text{or} \quad f_0 = \frac{1}{2\pi\sqrt{LC}}$$

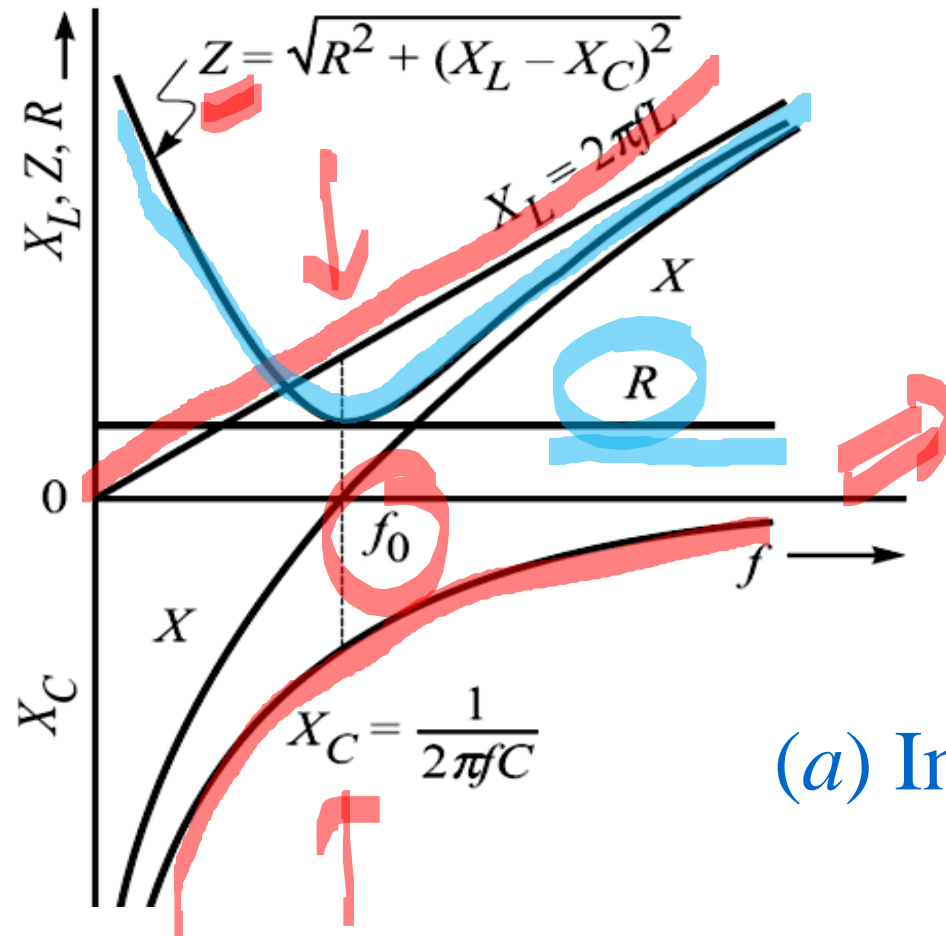
**The impedance of the circuit assumes a minimum value given as  $Z_0 = R + j0 = R$**

**The current has a maximum value given by**

$$I_0 = \frac{V}{Z_0} = \frac{V}{R}$$

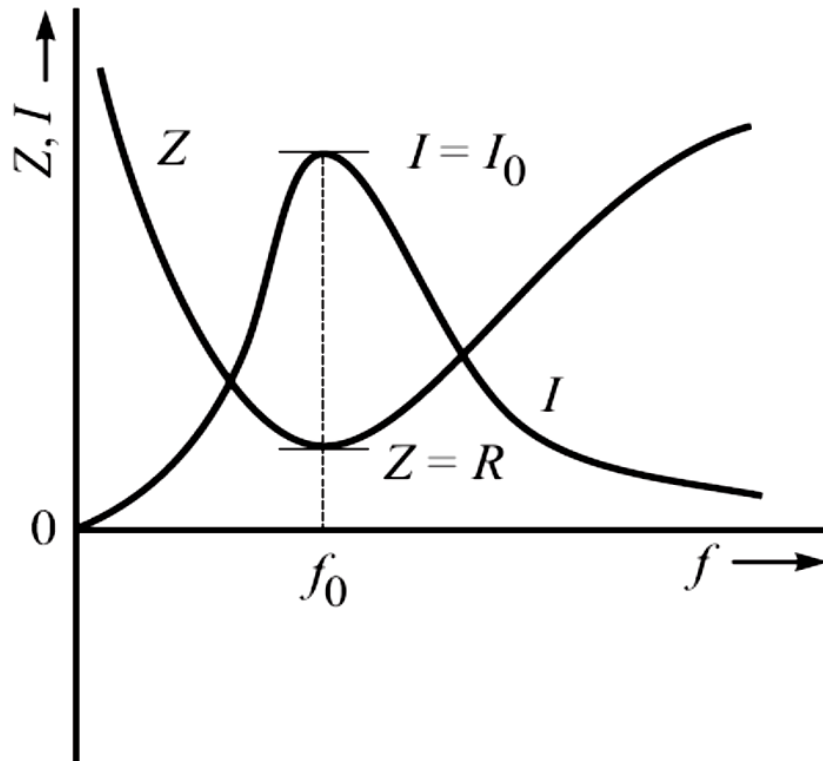
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# Effect of Variation of Frequency

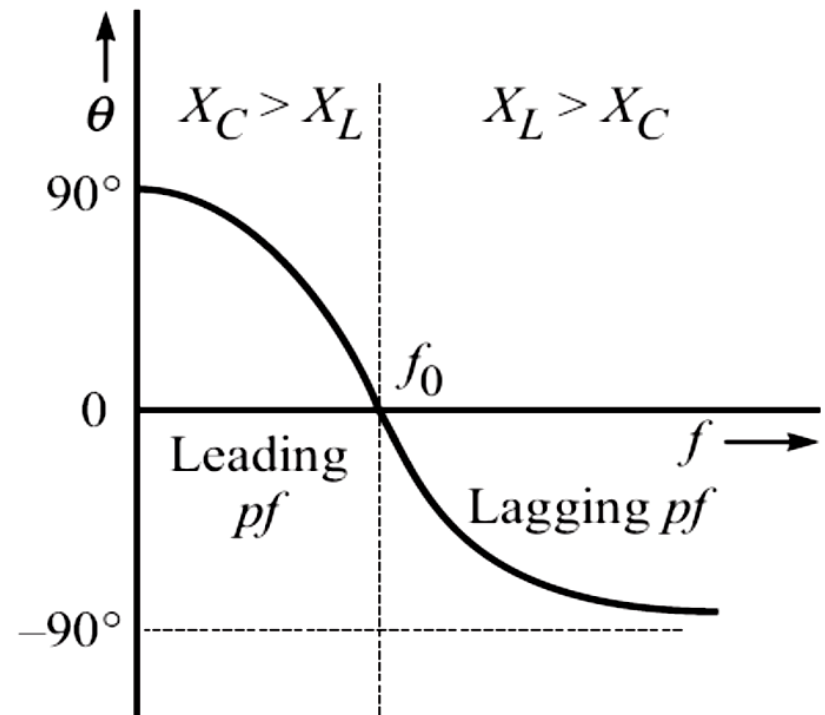


(a) Impedance.

Next



(b) Current.



(c) Power factor angle.

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- At  $f_0$ , the circuit behaves as *purely resistive*.

- Below  $f_0$ ,

$X$  has negative values

(i.e., the circuit is capacitive).

- Above  $f_0$ ,

$X$  has positive values

(i.e., the circuit is inductive).

- Since the current becomes maximum at resonant frequency, the series *RLC* resonant circuit is also called an *acceptor circuit*.

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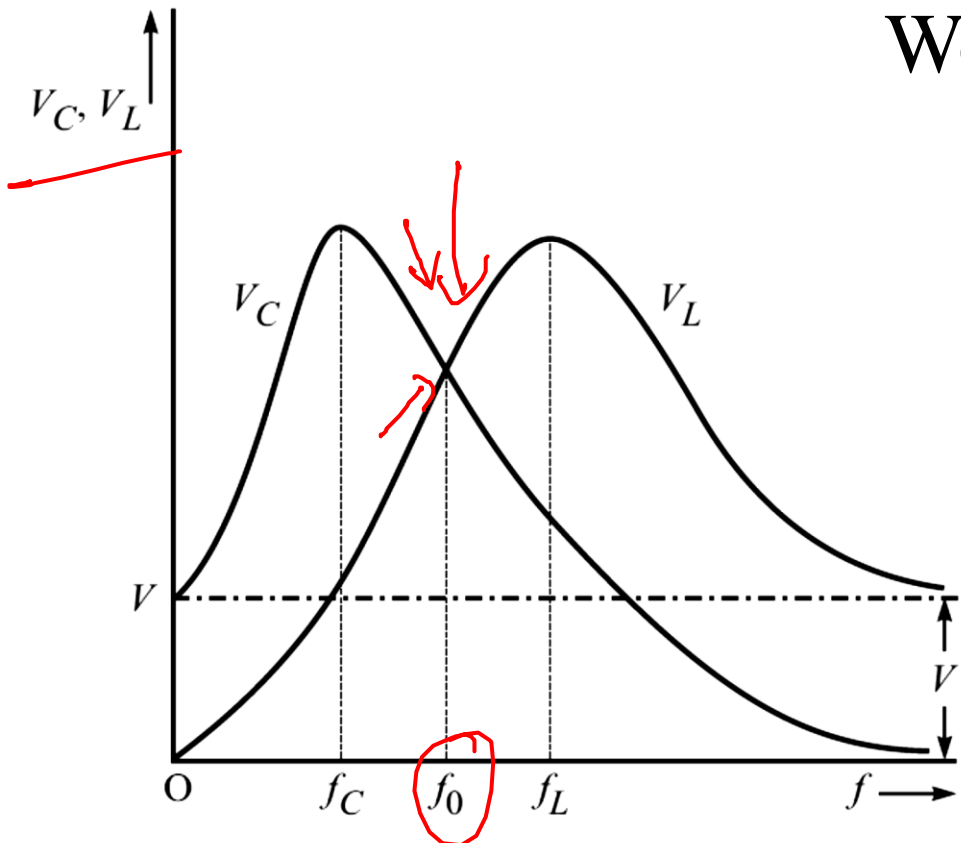
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# Variation of Voltage across $C$ and $L$ with Frequency



We can show that

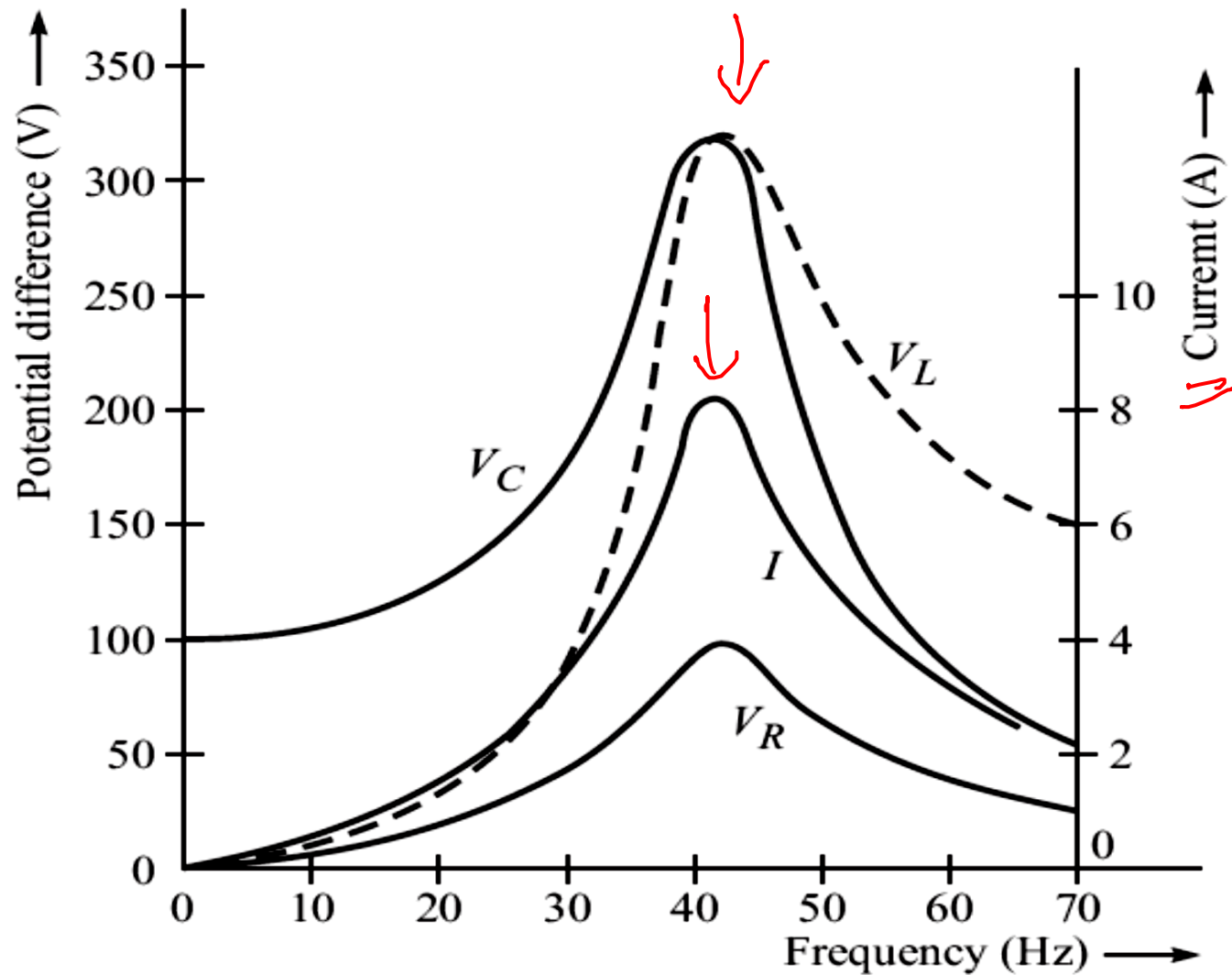
$$f_C = \frac{1}{2\pi} \sqrt{\frac{1}{LC} - \frac{R^2}{2L^2}}$$

$$f_L = \frac{1}{2\pi \sqrt{LC - (R^2 C^2 / 2)}}$$

**(a) When  $R$  has appreciable value.**

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(b) When  $R$  is very small.



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# Quality Factor ( $Q$ )

The quality of a resonant circuit to accept current (and power) at the resonant frequency to the exclusion of other frequencies is measured by its *quality factor* ( $Q$  factor), defined below.

$$Q = \frac{2\pi (\text{Maximum energy stored in } L \text{ or } C \text{ per cycle})}{\text{Energy dissipated per cycle}}$$
$$Q = \frac{2\pi [\frac{1}{2} L (I_m)^2]}{I_0^2 R T_0} = \frac{2\pi [\frac{1}{2} L (\sqrt{2} I_0)^2]}{I_0^2 R (2\pi / \omega_0)} = \frac{\omega_0 L}{R}$$

$\omega = 2\pi f$   
 $f = \frac{\omega}{2\pi}$   
 $T = \frac{1}{f} = \frac{2\pi}{\omega}$

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Since  $\omega_0 L = 1 / \omega_0 C$ ,  $Q = \frac{1}{\omega_0 CR}$

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By putting,  $\omega_0 = 1 / \sqrt{LC}$  we get another form for  $Q$  :

→  $Q = \frac{1}{\omega_0 CR} = \frac{\sqrt{LC}}{CR} = \frac{1}{R} \sqrt{\frac{L}{C}}$  ←

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- A capacitor usually has no losses.
- The  $Q$  of a series inductor-capacitor circuit is the same as the  $Q$  of the coil used.
- In fact,  $Q$  of the coil is used as a figure of merit for the coil.

Next

- Coils with  $Q < 10$  are described as *low- $Q$  coils*.
- Coils with  $Q > 10$  are described as *high- $Q$  coils*.
- Coils having  $Q$  as high as 200-300 are used in electronic circuits.

Next



# Example 1

- A series combination of a resistance of  $4\ \Omega$ , an inductance of  $0.5\ \text{H}$  and a variable capacitance is connected across a  $100\text{-V}$ ,  $50\text{-Hz}$  supply.

Calculate

- (*a*) the capacitance to give resonance,
- (*b*) the voltage across the inductance and the capacitance, and
- (*c*) the  $Q$  factor of the circuit.

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## Solution :

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(a) For resonance,

$$X_{L0} = X_{C0} \quad \text{or} \quad 2\pi f_0 L = 1 / 2\pi f_0 C$$

$$\Rightarrow C = \frac{1}{(2\pi f_0)^2 L} = \frac{1}{(2\pi \times 50)^2 \times 0.5} = \mathbf{20.3 \mu F}$$

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(b) At resonance,  $I_0 = \frac{V}{R} = \frac{100}{4} = 25 \text{ A}$

$$\therefore V_L = I_0 X_L = 25 \times (2\pi \times 50 \times 0.5) = \mathbf{3925 \text{ V}}$$

$$\therefore V_C = V_L = \mathbf{3925 \text{ V}}$$

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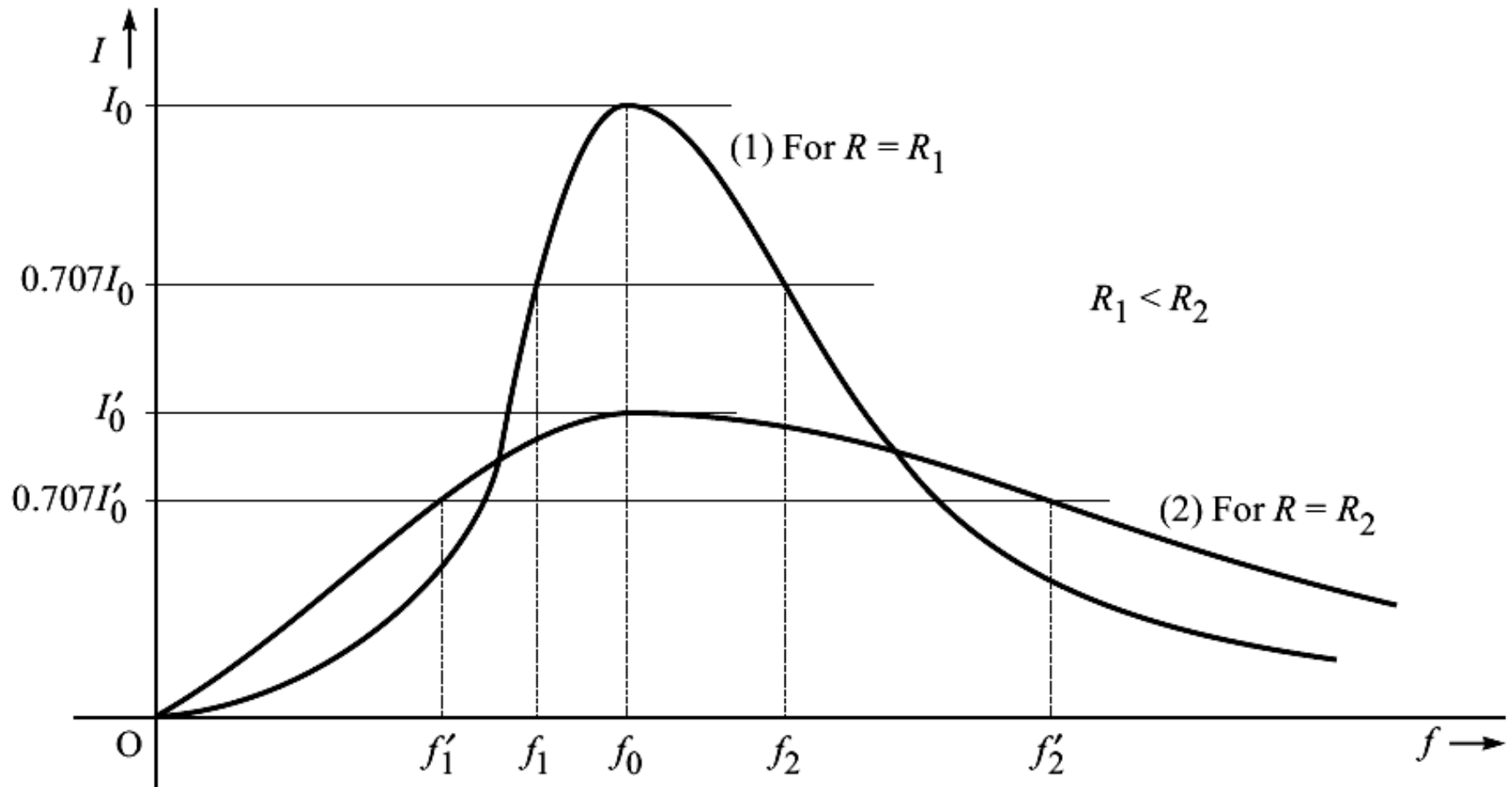
(c)  $Q = \frac{\omega_0 L}{R} = \frac{2\pi f_0 L}{R} = \frac{2\pi \times 50 \times 0.5}{4} = \mathbf{39.25}$

Next





# Resonance Curve



Next

- The frequencies  $f_1$  and  $f_2$  are often called *lower* and *upper cutoff frequencies*.
- At these frequencies, the current reduces to  $0.707I_0$ .
- These frequencies are also called *half-power frequencies*.
- The lower the value of  $R$ , the sharper is the resonance curve.
- We say that such a resonant circuit has high selectivity.
- For larger values of  $R$ , not only the peak value of current falls, but even the response curve becomes less sharp (i.e., *low selectivity*).



# Bandwidth ( $BW$ ) in Terms of Circuit Parameters

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$$BW = f_2 - f_1 = \left[ \frac{R}{4\pi L} + \frac{1}{4\pi} \sqrt{\left(\frac{R}{L}\right)^2 + \frac{4}{LC}} \right] - \left[ -\frac{R}{4\pi L} + \frac{1}{4\pi} \sqrt{\left(\frac{R}{L}\right)^2 + \frac{4}{LC}} \right]$$

$$BW = \frac{R}{2\pi L}$$

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# Bandwidth ( $BW$ ) in Terms of $Q$ and $f_0$

We know that

$$Q = \frac{\omega_0 L}{R} \Rightarrow \frac{R}{L} = \frac{\omega_0}{Q} = \frac{2\pi f_0}{Q}$$

$$\therefore BW = \frac{R}{2\pi L} = \frac{1}{2\pi} \left( \frac{R}{L} \right) = \frac{1}{2\pi} \left( \frac{2\pi f_0}{Q} \right)$$

or  $BW = \frac{f_0}{Q}$

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## Example 2

- A series ac circuit has a resonance frequency of 150 kHz and a bandwidth of 75 kHz. Determine its half-power frequencies.

**Solution :** Let us calculate the  $Q$  of the circuit,

Click



$$\text{Since } BW = \frac{f_0}{Q}, \text{ we have } Q = \frac{f_0}{BW} = \frac{150 \text{ kHz}}{75 \text{ kHz}} = 2$$

Hence, we cannot use the approximate relations.  
Using the exact relations, and working in kHz,

Next



$$75 = f_2 - f_1 \quad \text{and} \quad 150 = \sqrt{f_2 f_1}$$

Eliminating  $f_2$  between the two equations, we get

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$$f_1^2 + 75f_1 - 22500 = 0$$

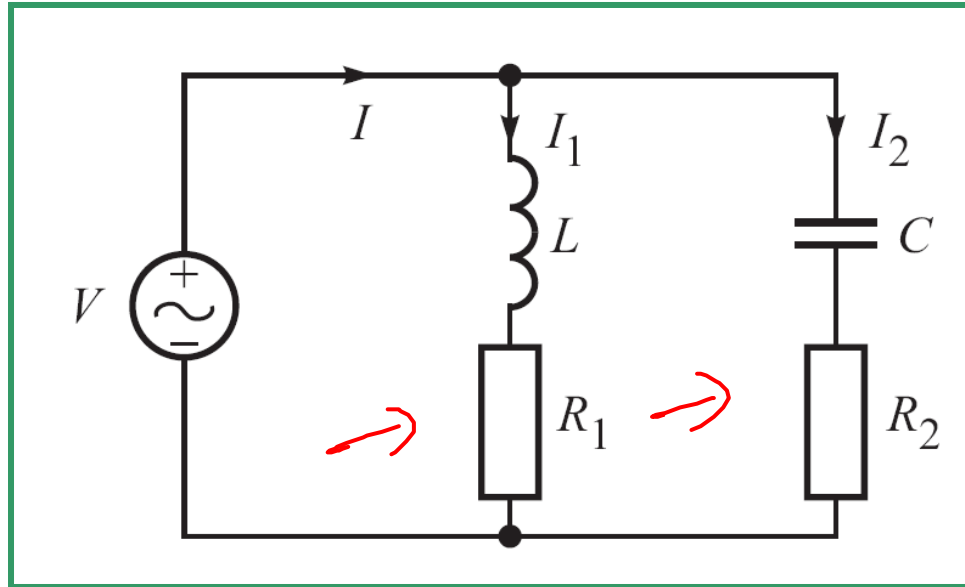
$$\Rightarrow f_1 = 117.1 \text{ kHz or } -192.1 \text{ kHz}$$

- Ignoring the negative value, we have  $f_1 = \underline{117.1 \text{ kHz}}$ .
- Hence,  $f_2 = 75 + f_1 = \underline{192.1 \text{ kHz}}$ .

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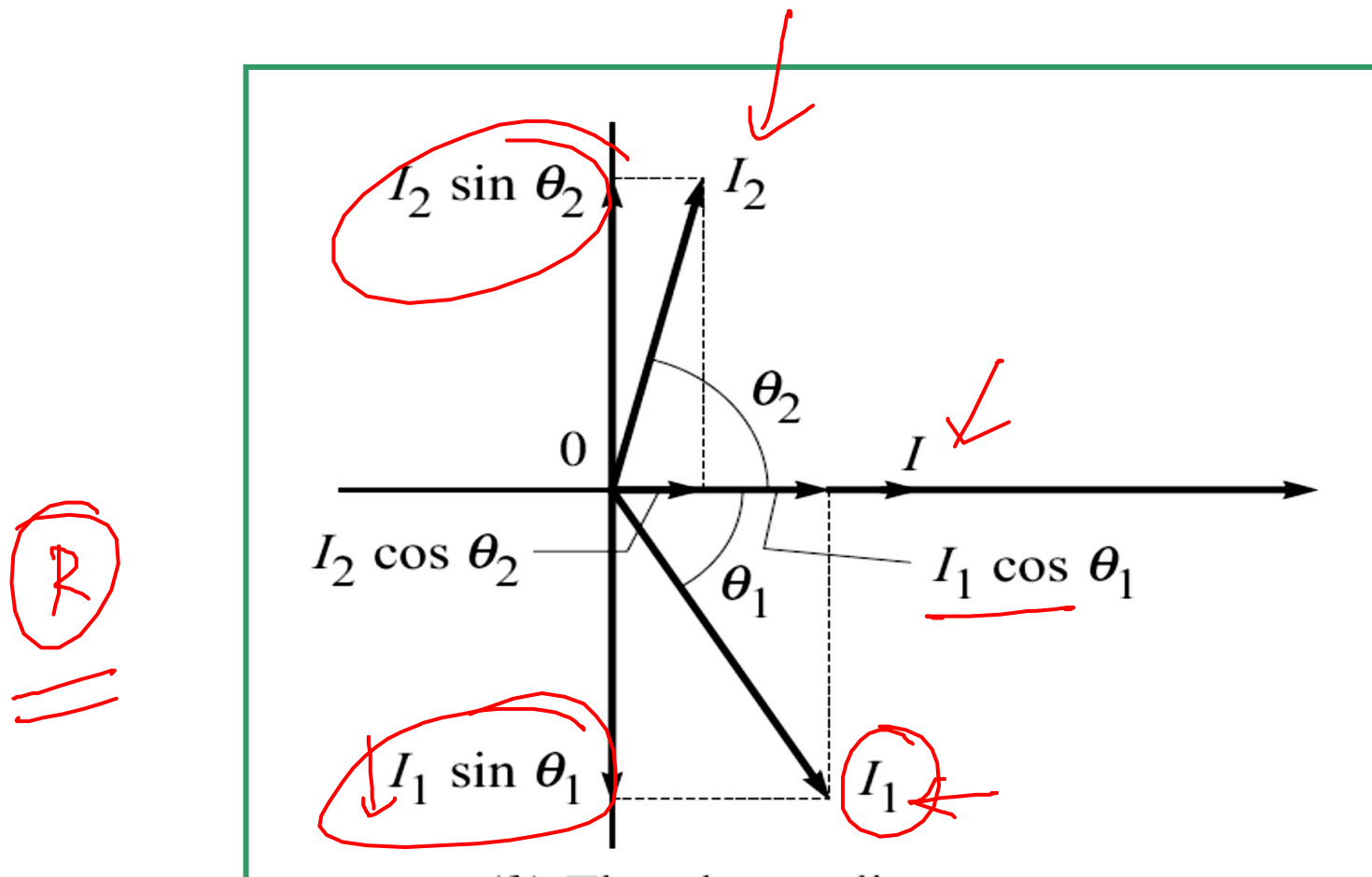


# Parallel Resonant Circuit



- The losses in inductor and capacitor are accounted for by equivalent resistances  $R_1$  and  $R_2$ .
- Resonant condition reaches when the reactive (or wattless) component of line current  $I$  reduces to zero.

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Under resonance condition, the reactive components of these two currents are equal in magnitude (but opposite in phase). That is,

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$$I_1 \sin \theta_1 = I_2 \sin \theta_2$$

$$\frac{V}{\sqrt{R_1^2 + (\omega_0 L)^2}} \times \frac{(\omega_0 L)}{\sqrt{R_1^2 + (\omega_0 L)^2}} = \frac{V}{\sqrt{R_2^2 + (1/\omega_0 C)^2}} \times \frac{(1/\omega_0 C)}{\sqrt{R_2^2 + (1/\omega_0 C)^2}}$$

or

$$\frac{(\omega_0 L)}{R_1^2 + (\omega_0 L)^2} = \frac{(1/\omega_0 C)}{R_2^2 + (1/\omega_0 C)^2}$$



or

$$\frac{\omega_0 L}{R_1^2 + \omega_0^2 L^2} = \frac{\omega_0 C}{R_2^2 \omega_0^2 C^2 + 1}$$



$$\Rightarrow f_0 = \frac{1}{2\pi\sqrt{LC}} \sqrt{\frac{R_1^2 - (L/C)}{R_2^2 - (L/C)}}$$



# Effect of Variation of Frequency

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$$\mathbf{Y}_1 = \frac{1}{R + j\omega L} = \frac{R}{R^2 + (\omega L)^2} - j \frac{\omega L}{R^2 + (\omega L)^2} = G - jB_L$$

$$\mathbf{Y}_2 = \frac{1}{(1/j\omega C)} = +j\omega C = +jB_C$$

The conductance and susceptance of the inductive branch,

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$$G = \frac{R}{R^2 + (\omega L)^2} \quad \text{and} \quad B_L = \frac{\omega L}{R^2 + (\omega L)^2}$$

For the capacitive branch,

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$$G = 0 \quad \text{and} \quad B_C = \omega C$$

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**Total admittance of the circuit,**

$$\mathbf{Y} = \mathbf{Y}_1 + \mathbf{Y}_2 = \frac{R}{R^2 + (\omega L)^2} - j \frac{\omega L}{R^2 + (\omega L)^2} + j\omega C = G + j(B_C - B_L)$$

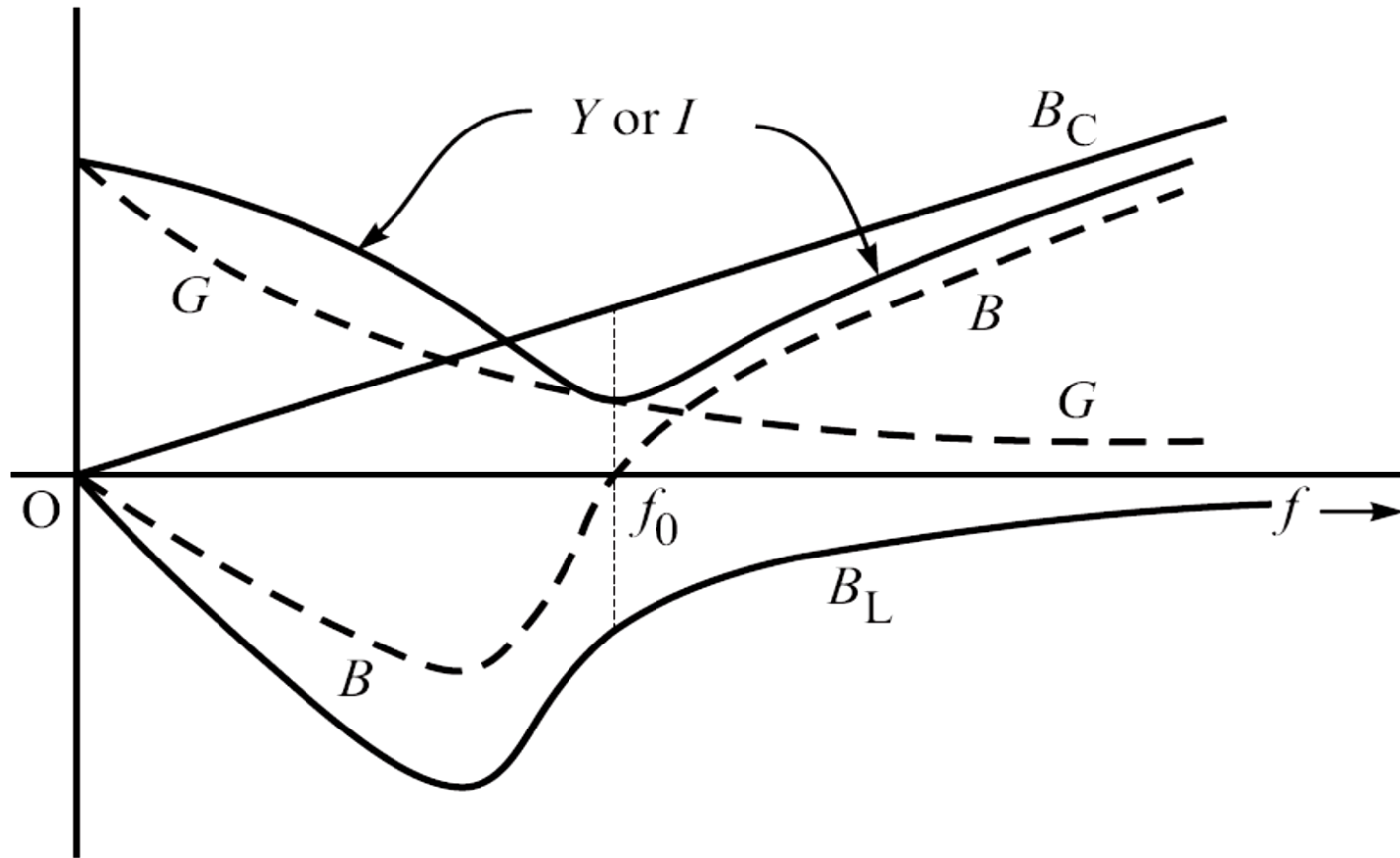
$$G = \frac{R}{R^2 + (\omega L)^2}$$

and

$$B_L = \frac{\omega L}{R^2 + (\omega L)^2}$$

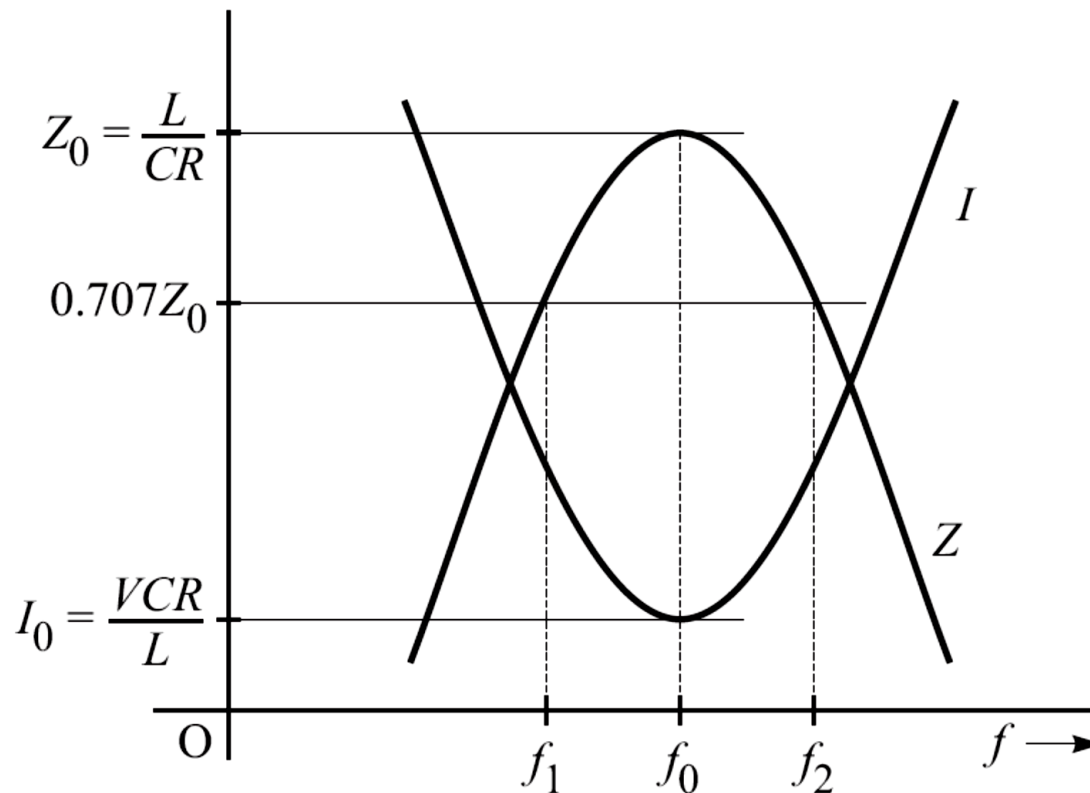
$$B_C = \omega C$$

**Next** 



(a) Admittance versus frequency.

Next



### (b) Current versus frequency.

At resonance frequency, the line current is seen to have minimum,

$$I_0 = \frac{V}{Z_0} = \frac{V}{L/CR} = \frac{VCR}{L}$$

Next

# Some Important Points

- Since the circuit rejects the current at resonance (i.e., it has minimum value), the parallel resonant circuit is also called **rejector circuit** or **anti-resonant circuit**.
- Since the circulating current between the two branches is many times the line current, the parallel resonant circuit is also called **current resonant circuit**.
- The circuit is also called a **tank circuit**.



# Review

## ■ Introduction.

## ■ Series Resonant Circuit.

- Effect of Variation of Frequency.
- Variation of Voltage across  $C$  and  $L$  with Frequency.
- Quality Factor ( $Q$ ).
- Voltage Magnification.
- Resonance Curve.

- Relation between  $f_0$ ,  $f_1$  and  $f_2$ .

- Bandwidth ( $BW$ ) in Terms of Circuit Parameters.

- Bandwidth ( $BW$ ) in Terms of  $Q$  and  $f_0$ .

## ■ Parallel Resonant Circuit.

- Practical Circuit.
- Resonance Curve.
- Effect of Variation of Frequency.
- Some Important Points.

Next

