

General Information

Prof. Antonella Ferrara

<https://scholar.google.com/citations?user=r5JuMskAAAAJ&hl=en>

Course Teaching Material:

KIRO UNIPV <https://elearning.unipv.it/>

- [504462 - PROCESS CONTROL 2025-26 - PROF.SSA FERRARA ANTONELLA](#)

Lecture Time-table:

<http://www-3.unipv.it/ingserv/orario2526/1sem/insegnamenti/ProCont.html>

Exams:

<https://studentionline.unipv.it/esse3/Home.do>

<https://kirotesting.unipv.it/>



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Introduction

- Program of the course:

Advanced SISO control schemes:

Pre-filters and parallel compensators, two degrees of freedom control schemes, compensation of measurable disturbances, systems with delays and Smith Predictor, Padé approximation, decoupling in the frequency domain, control of open loop unstable systems.

Advanced MIMO control schemes:

Decoupling based control schemes, decentralized control, relative gain array.

PID controllers:

Features and properties. Rules for the empirical calibration. Wind-up and anti wind-up schemes.

Digital control:

Discrete-time systems. The concept of equilibrium for discrete-time systems. Stability of linear time-invariant discrete-time systems. Jury test. Digital control schemes. Zeta transform and its properties. Transfer functions in the z domain. Sampling and aliasing. Choice of the sampling time. Zero-order-Hold. Discretization of continuous-time controllers. Bilinear transformation, Euler and Tustin methods.

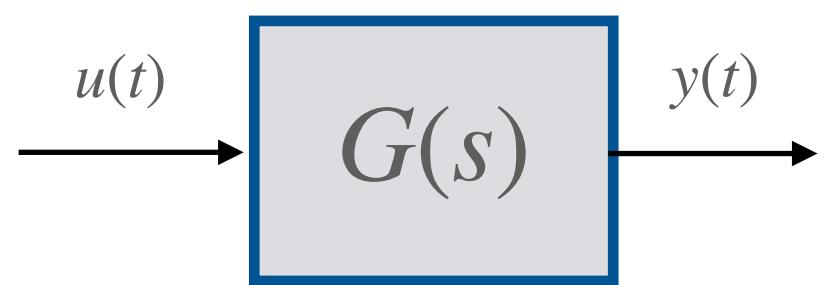


Introduction

- Some of the figures in these slides, kindly provided by McGraw-Hill, are those of the Textbook:



Discrete-time Systems



$$u(t) \in \mathbb{R}^m$$

$$y(t) \in \mathbb{R}^p$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



Discrete-time Systems



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



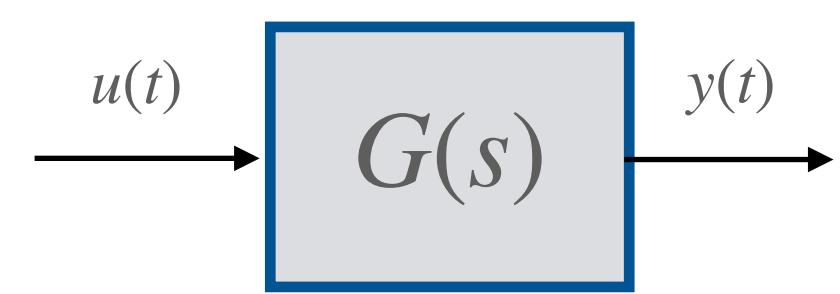
$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System



Discrete-time Systems



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

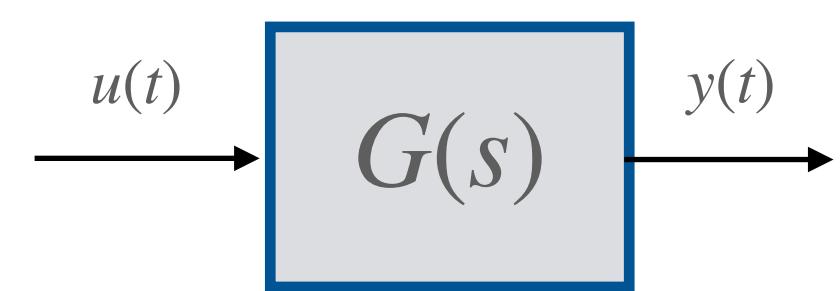
$$\boxed{\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}}$$



Discrete-time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

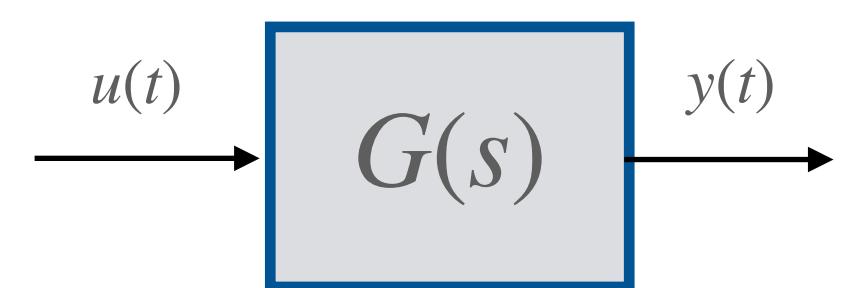
$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$



Discrete-time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

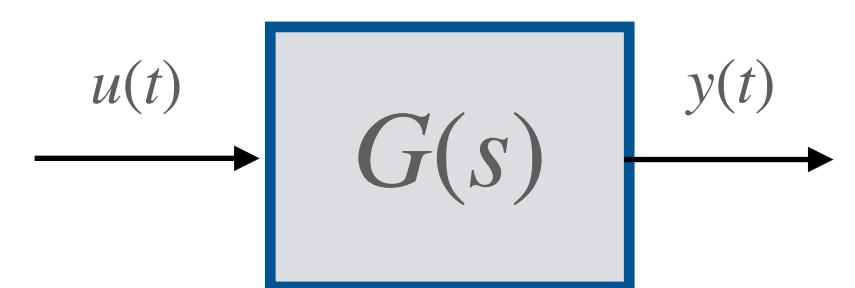
$$\begin{aligned} x(1) &= f(x_0, u(0)), \\ x(2) &= f(x(1), u(1)), \\ &\vdots \\ x(i) &= f(x(i-1), u(i-1)) \end{aligned}$$



Discrete-time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

$$\begin{aligned} x(1) &= f(x_0, u(0)), \\ x(2) &= f(x(1), u(1)), \\ &\vdots \\ x(i) &= f(x(i-1), u(i-1)) \end{aligned}$$

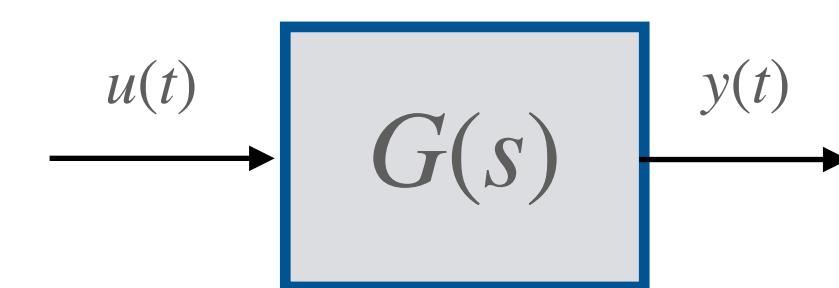
$$y(k) = g(x(k), u(k)), \quad k \geq 0$$



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

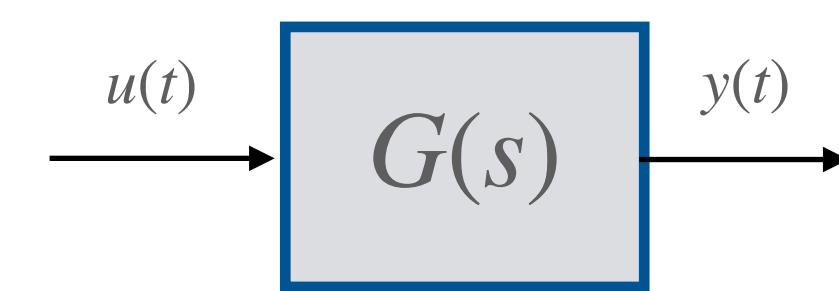
MIMO



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R} \\ y(t) &\in \mathbb{R} \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

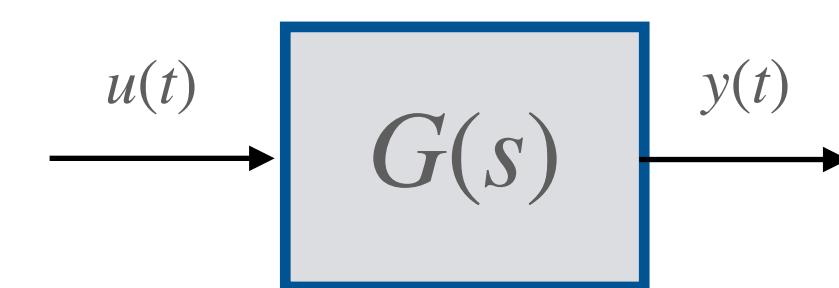
SISO



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

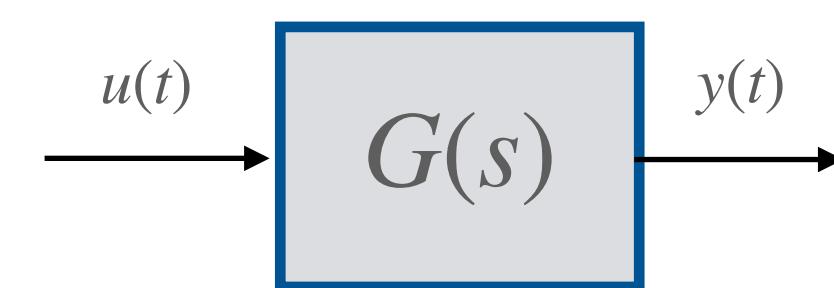
Proper



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), k) \end{aligned}$$

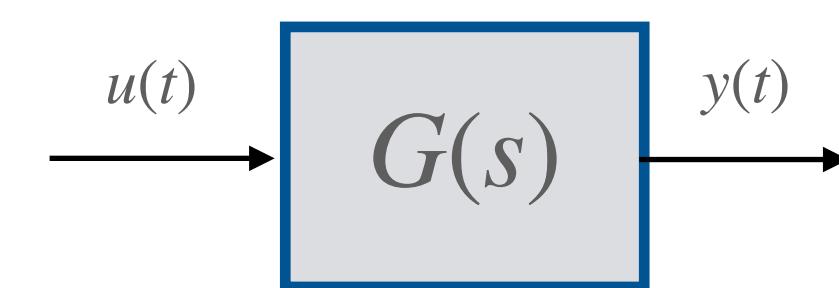
Strictly proper



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

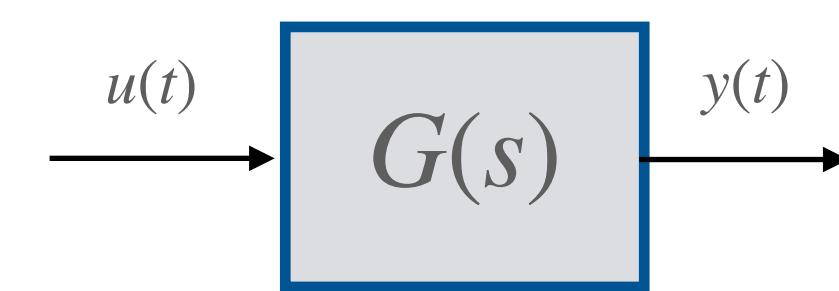
Time-varying



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$

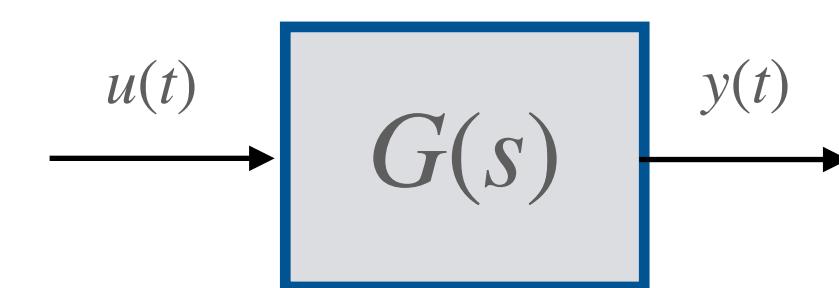
Time-invariant



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k), k) \\ y(k) &= g(x(k), u(k), k) \end{aligned}$$

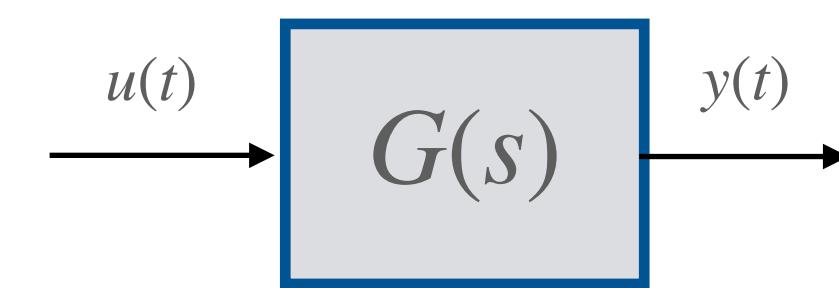
Nonlinear



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= A(k)x(k) + B(k)u(k) \\ y(k) &= C(k)x(k) + D(k)u(k) \end{aligned}$$

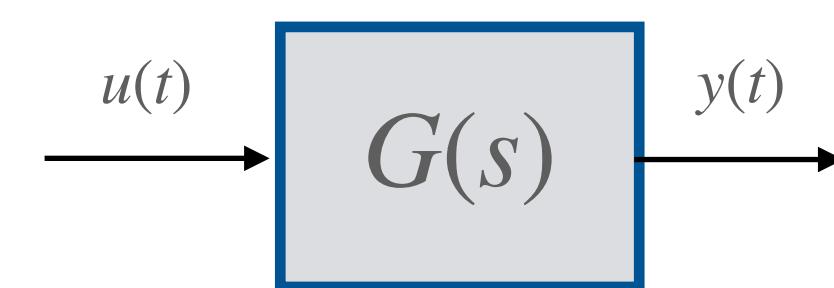
Linear Time-varying



Discrete-time Systems: Classification

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\boxed{\begin{aligned} x(k+1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}}$$

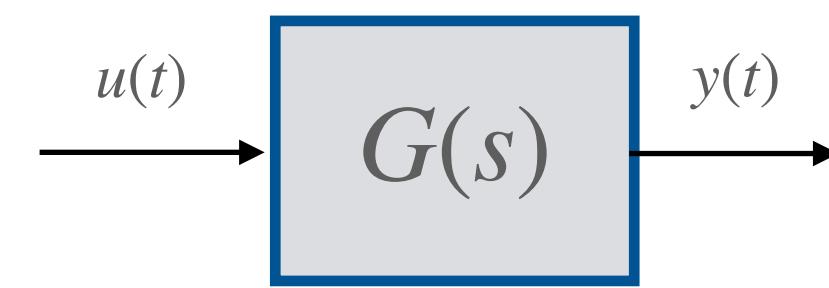
Linear Time-invariant



Discrete-time Systems: Equilibrium

Given:

$$u(k) = \bar{u}, \quad \forall k \geq 0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$

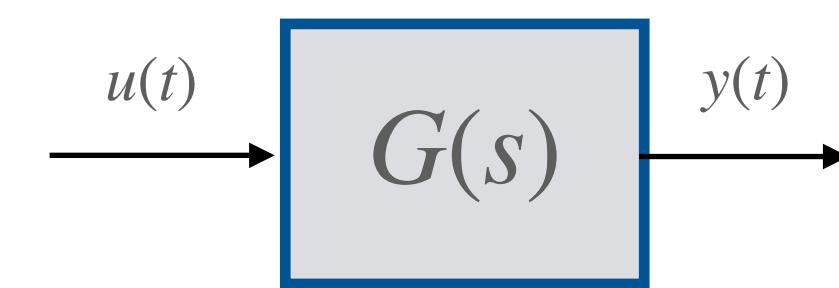
Time-invariant



Discrete-time Systems: Equilibrium

Given:

$$u(k) = \bar{u}, \quad \forall k \geq 0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$

Time-invariant

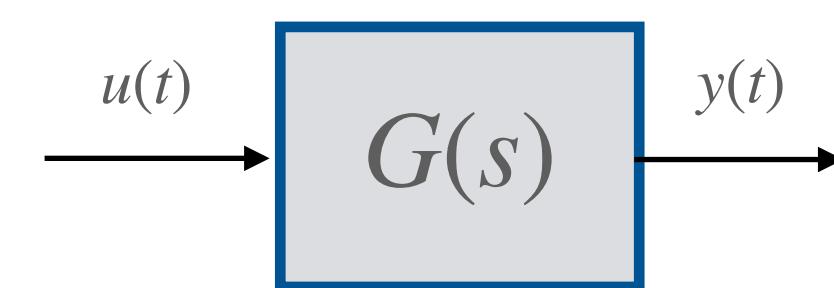
$$\dot{x}(t) = 0$$



Discrete-time Systems: Equilibrium

Given:

$$u(k) = \bar{u}, \quad \forall k \geq 0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$

Time-invariant

$$\dot{x}(t) = 0$$

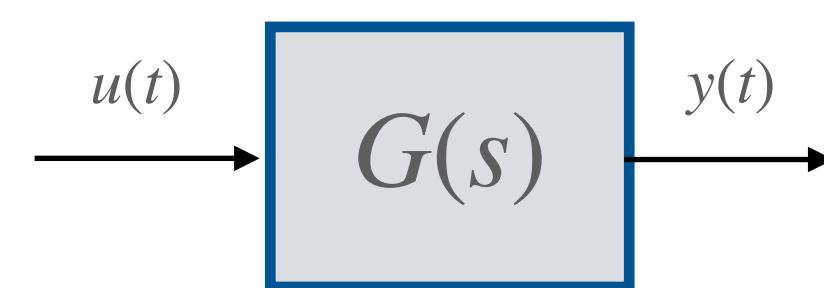
$$x(k+1) = x(k)$$



Discrete-time Systems: Equilibrium

Given:

$$u(k) = \bar{u}, \quad \forall k \geq 0$$



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

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Continuous-time System



$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$$t \in \mathbb{N}_{\{0\}}$$

Discrete-time System

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$

Time-invariant



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If the solution exists:

$$\begin{aligned} x(k) &= \bar{x}, \quad \forall k \geq 0 \\ \bar{y}(k) &= g(\bar{x}, \bar{u}) \end{aligned}$$

$$\dot{x}(t) = 0$$

$$x(k+1) = x(k)$$

Discrete-time Systems: Stability of Equilibrium

Given:

$$x(k_0) = x_0, \quad u(k) = \bar{u}$$

An equilibrium state is STABLE if:

$$\forall \varepsilon > 0 \quad \exists \quad \delta > 0 \quad \text{s.t.} \quad \|x(0) - \bar{x}\| < \delta$$

$$\implies \|x(k) - \bar{x}\| < \varepsilon, \quad \forall k \geq 0$$

$$\boxed{\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}}$$



Time-invariant

If the solution exists:

$$\boxed{\begin{aligned} x(k) &= \bar{x}, \quad \forall k \geq 0 \\ \bar{y}(k) &= g(\bar{x}, \bar{u}) \end{aligned}}$$



Discrete-time Systems: Stability of Equilibrium

Given:

$$x(k_0) = x_0, \quad u(k) = \bar{u}$$

An equilibrium state is AS. STABLE if:

$$\forall \varepsilon > 0 \quad \exists \quad \delta > 0 \quad \text{s.t.} \quad \|x(0) - \bar{x}\| < \delta$$

$$\implies \|x(k) - \bar{x}\| < \varepsilon, \quad \forall k \geq 0$$

and in addition:

$$\lim_{k \rightarrow \infty} \|x(k) - \bar{x}\| = 0$$

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$



Time-invariant

If the solution exists:

$$\begin{aligned} x(k) &= \bar{x}, \quad \forall k \geq 0 \\ \bar{y}(k) &= g(\bar{x}, \bar{u}) \end{aligned}$$



Discrete-time Systems: Stability of Equilibrium

Given:

$$x(k_0) = x_0, \quad u(k) = \bar{u}$$

An equilibrium state is UNSTABLE if:

it is not stable

$$\boxed{x(k + 1) = f(x(k), u(k)) \\ y(k) = g(x(k), u(k))}$$



Time-invariant

If the solution exists:

$$\boxed{x(k) = \bar{x}, \quad \forall k \geq 0 \\ \bar{y}(k) = g(\bar{x}, \bar{u})}$$



Discrete-time Systems: Stability of a Generic Motion of the State

An motion of the state $\tilde{x}(k)$ is STABLE if:

$$\forall \varepsilon > 0 \ \exists \ \delta > 0 \text{ s.t. } \|x(0) - \tilde{x}(0)\| < \delta$$

$$\implies \|x(k) - \tilde{x}(k)\| < \varepsilon, \quad \forall k \geq 0$$

Given:

$$x(k_0) = x_0, \quad u(k) = \bar{u}$$

$$x(k+1) = f(x(k), u(k))$$

$$y(k) = g(x(k), u(k))$$

An motion of the state $\tilde{x}(k)$ is AS. STABLE if:

$$\forall \varepsilon > 0 \ \exists \ \delta > 0 \text{ s.t. } \|x(0) - \tilde{x}(0)\| < \delta$$

$$\implies \|x(k) - \tilde{x}(k)\| < \varepsilon, \quad \forall k \geq 0 \quad \text{and in addition:}$$

$$\lim_{k \rightarrow \infty} \|x(k) - \tilde{x}(k)\| = 0$$

Time-invariant

If the solution exists:

$$x(k) = \bar{x}, \quad \forall k \geq 0$$

$$\bar{y}(k) = g(\bar{x}, \bar{u})$$



Discrete-time Systems: Stability of a Generic Motion of the State

Given:

$$x(k_0) = x_0, \quad u(k) = \bar{u}$$

An motion of the state $\tilde{x}(k)$ is UNSTABLE if:

it is not stable

$$\begin{aligned} x(k+1) &= f(x(k), u(k)) \\ y(k) &= g(x(k), u(k)) \end{aligned}$$



Time-invariant

If the solution exists:

$$\begin{aligned} x(k) &= \bar{x}, \quad \forall k \geq 0 \\ \bar{y}(k) &= g(\bar{x}, \bar{u}) \end{aligned}$$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$

The stability property is referred to the system

Stability only depends on the eigenvalues of matrix A

$$\begin{aligned} x(k+1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}$$



Linear Time-invariant



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$

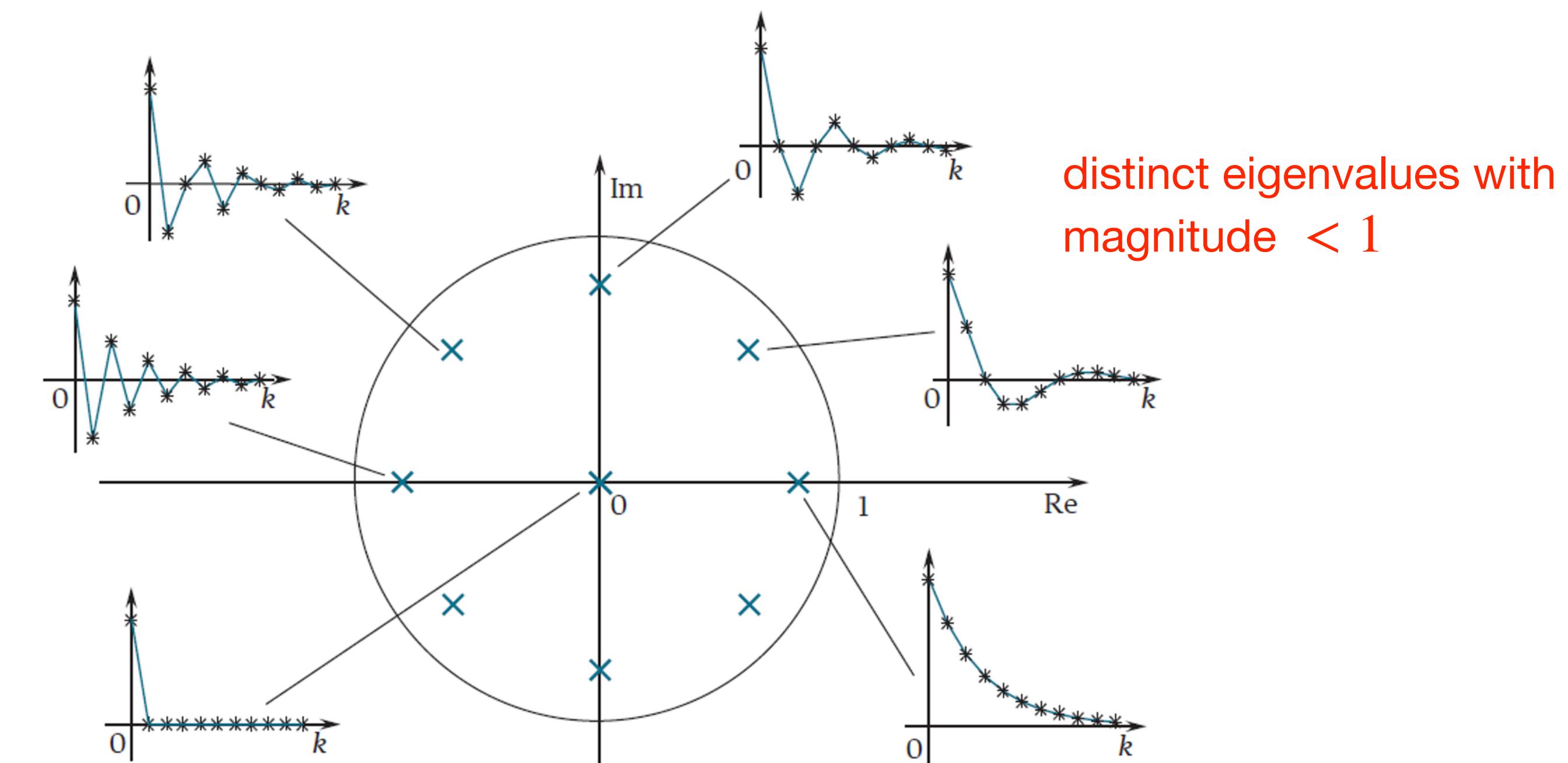
$$\begin{aligned} x(k + 1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}$$



Linear Time-invariant

The stability property is referred to the system

Stability only depends on the eigenvalues of matrix A



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$

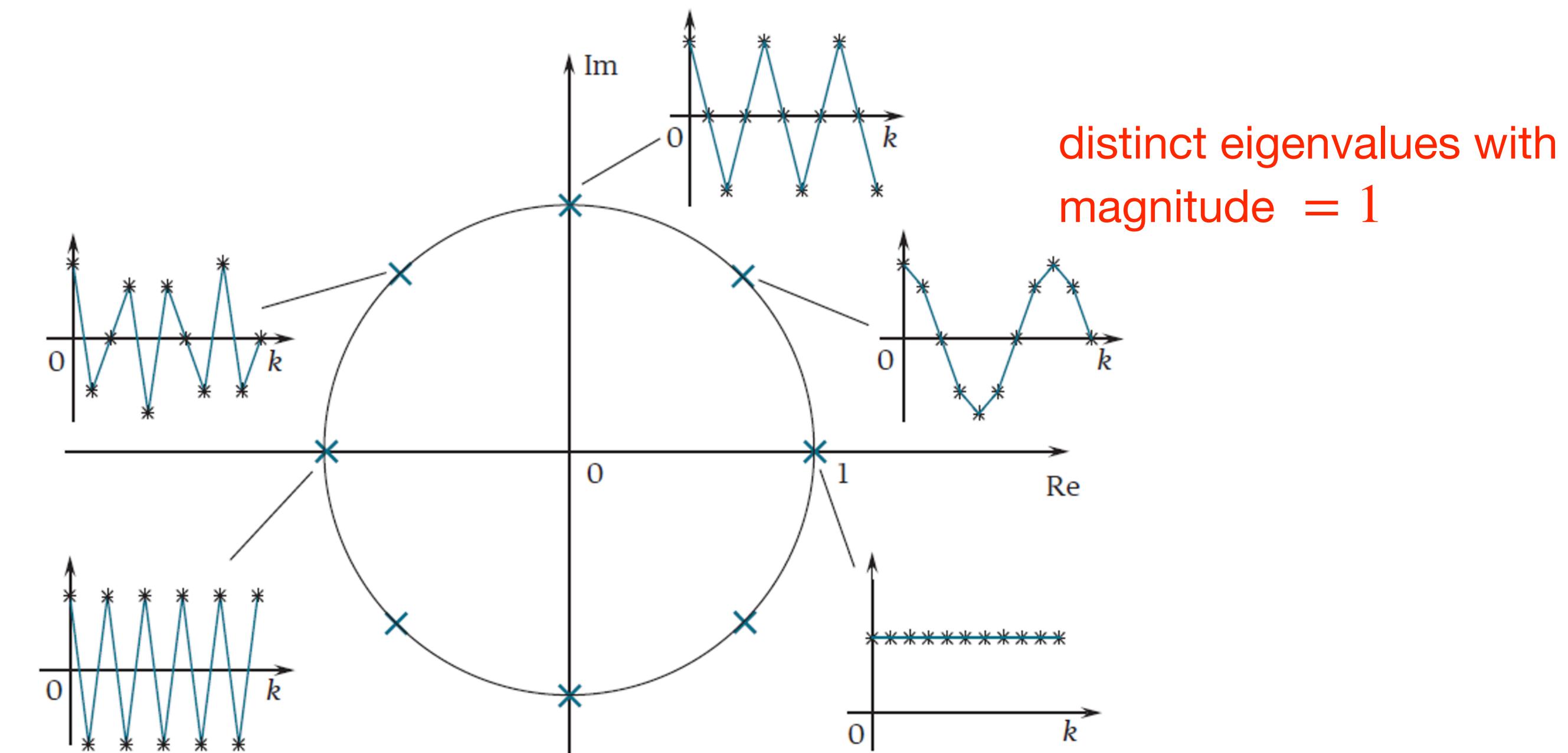
$$\begin{aligned} x(k + 1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}$$



Linear Time-invariant

The stability property is referred to the system

Stability only depends on the eigenvalues of matrix A



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$

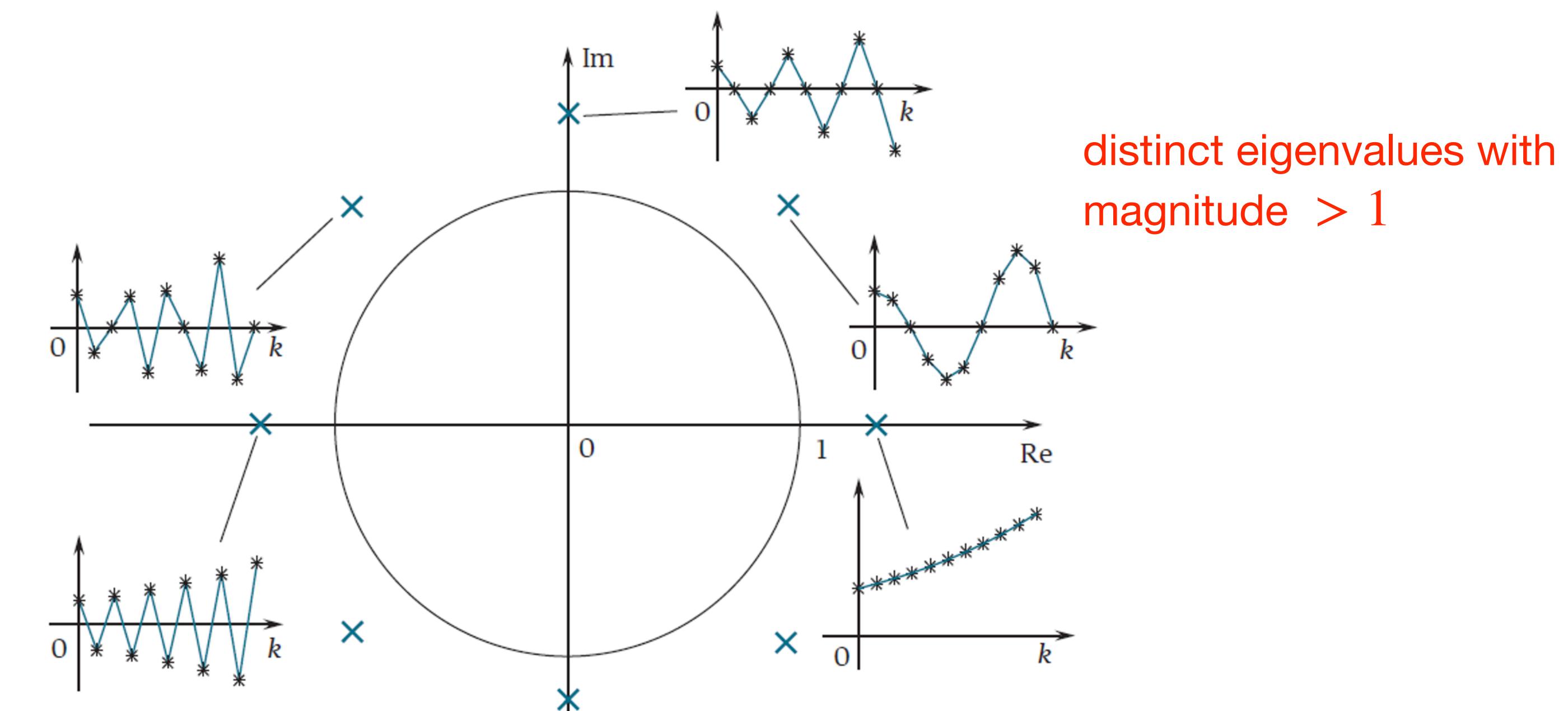
$$\begin{aligned} x(k + 1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}$$



Linear Time-invariant

The stability property is referred to the system

Stability only depends on the eigenvalues of matrix A



distinct eigenvalues with
magnitude > 1



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Given:

$$x(k_0) = x_0, \quad u(k) \geq k_0$$

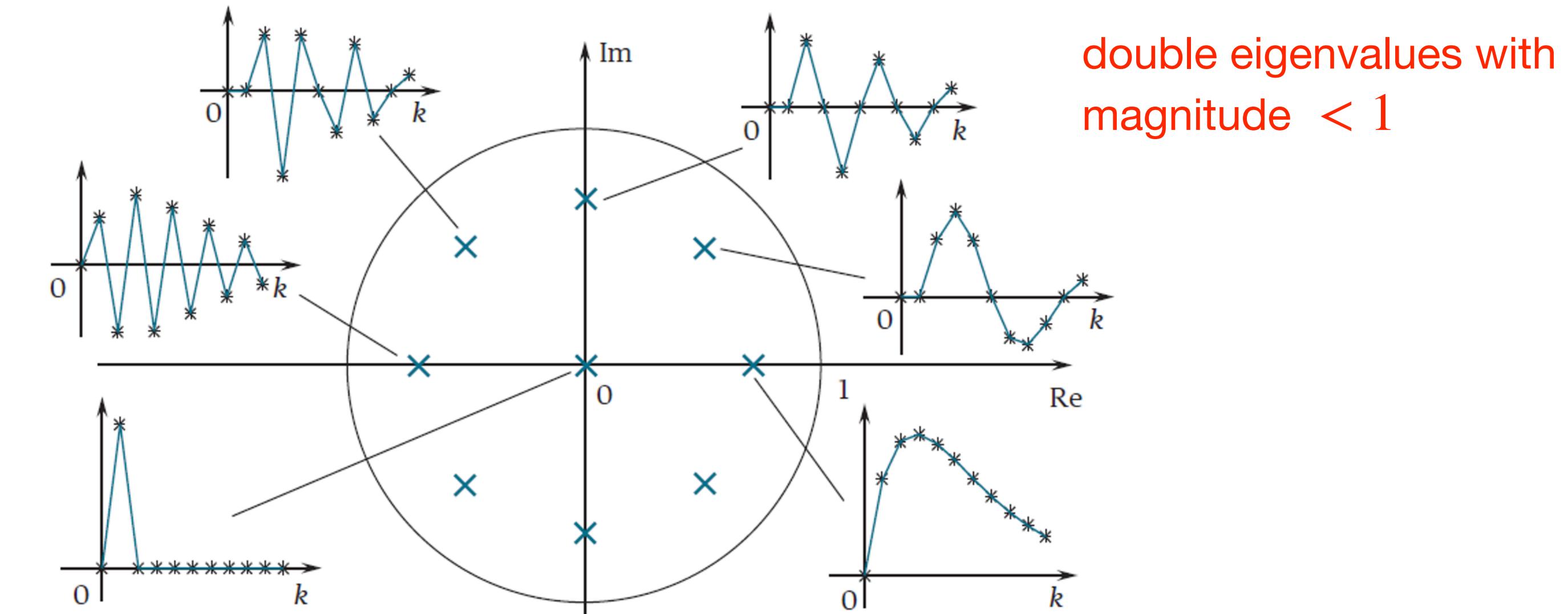
$$\begin{aligned} x(k + 1) &= Ax(k) + Bu(k) \\ y(k) &= Cx(k) + Du(k) \end{aligned}$$



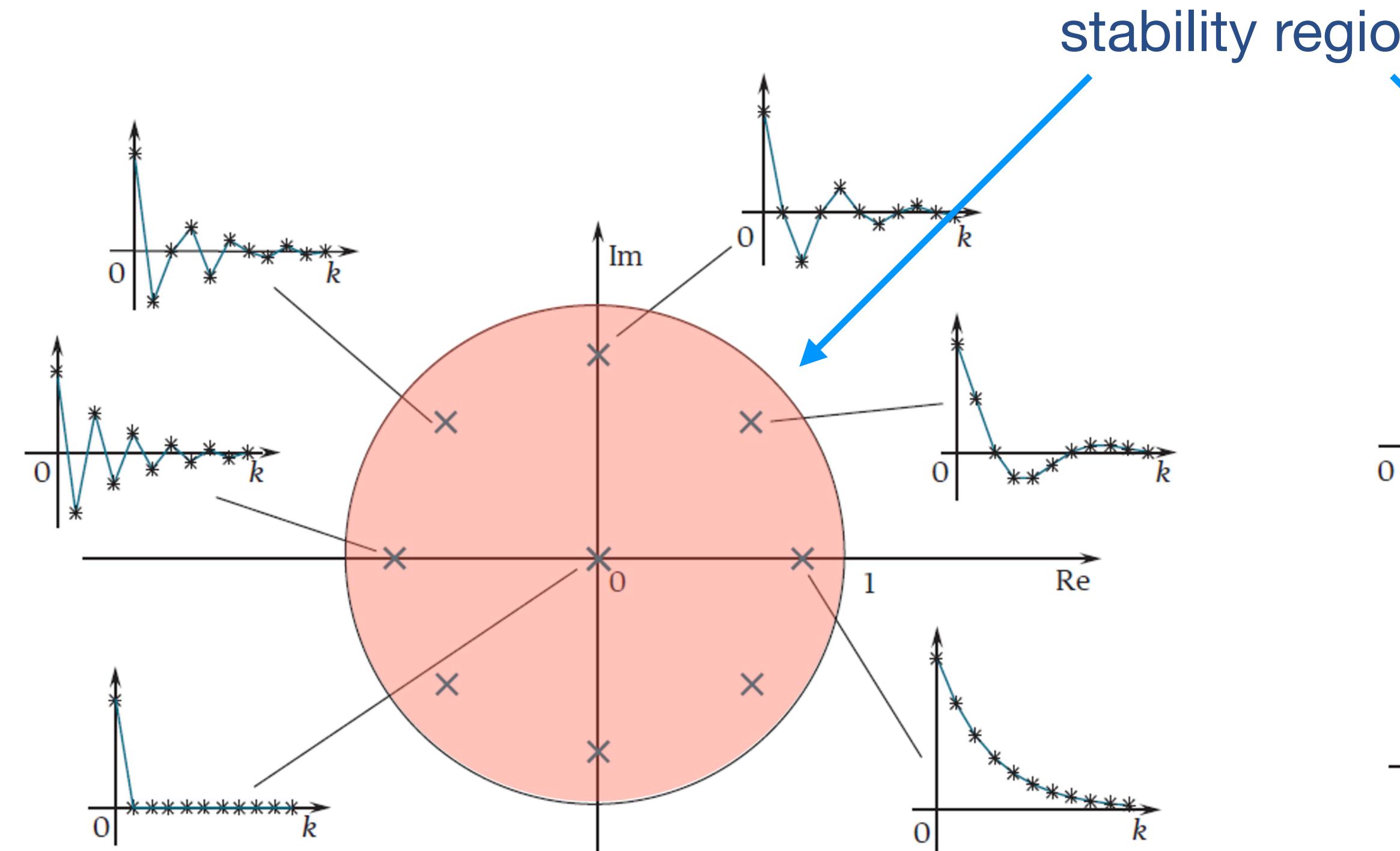
Linear Time-invariant

The stability property is referred to the system

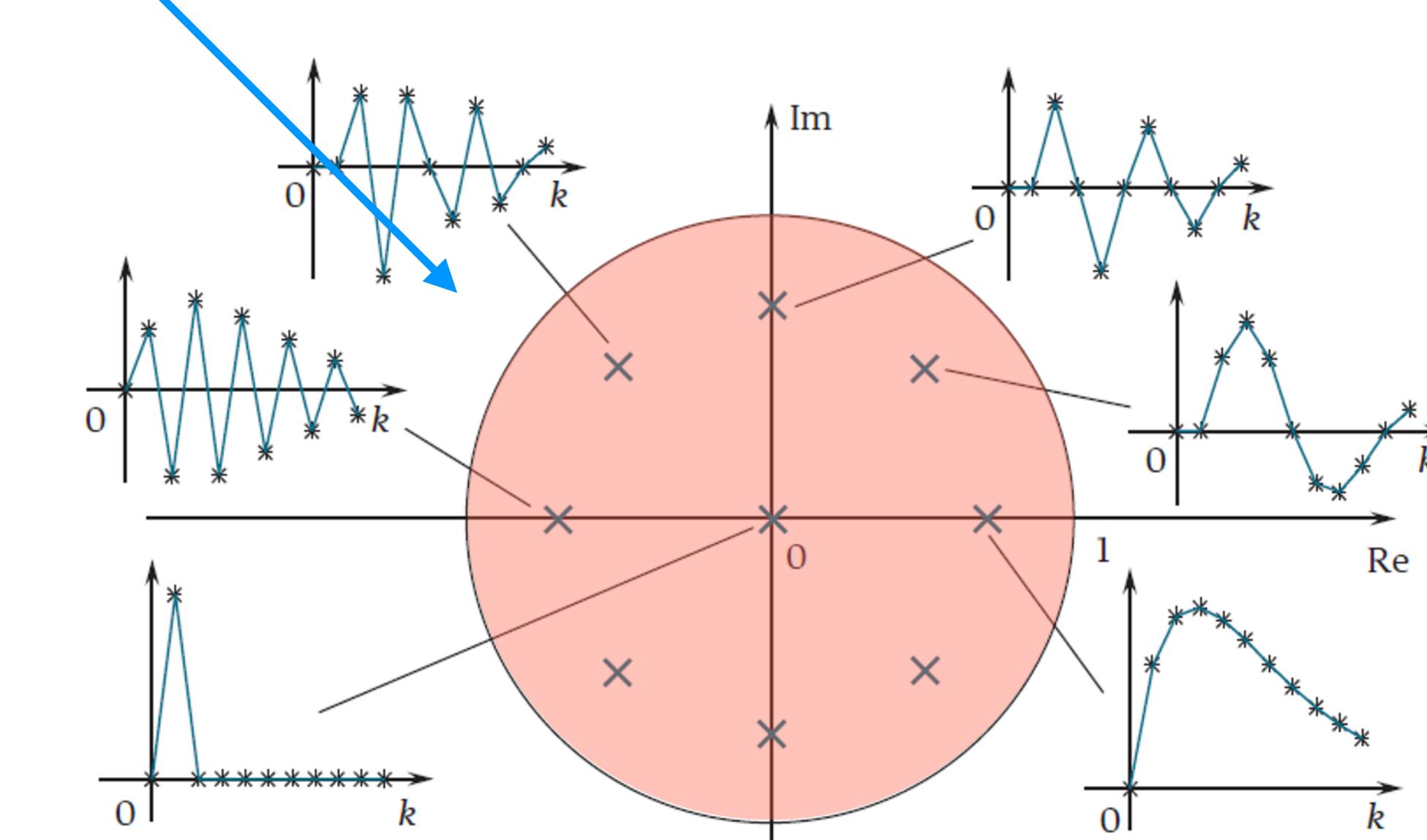
Stability only depends on the eigenvalues of matrix A



Discrete-time Systems: Stability of LTI Discrete-Time Systems



distinct eigenvalues with
magnitude < 1



double eigenvalues with
magnitude < 1



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \dots + \varphi_{n-1} z + \varphi_n \longrightarrow$$

To find the eigenvalues:

$$\varphi(z) = \det(zI - A) = 0$$



not straightforward if $n > 2$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \dots + \varphi_{n-1} z + \varphi_n \longrightarrow$$



Alternatives:

Necessary or sufficient conditions based on $\varphi(z)$

Stability test (Jury's Test) based on $\varphi(z)$

To find the eigenvalues:

$$\varphi(z) = \det(zI - A) = 0$$



not straightforward if $n > 2$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Necessary conditions based on $\varphi(z)$:

1. A necessary condition for a discrete-time dynamical system to be asymptotically stable is that

$$\left| \frac{\varphi_n}{\varphi_0} \right| < 1$$

2. A necessary condition for a discrete-time dynamical system to be asymptotically stable is that

$$\varphi_0 \varphi(1) > 0$$

3. A necessary condition for a discrete-time dynamical system to be asymptotically stable is that

$$(-1)^n \varphi_0 \varphi(-1) > 0$$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Sufficient conditions based on $\varphi(z)$:

1. A sufficient condition for a discrete-time dynamical system to be asymptotically stable is that

$$\varphi_0 > \varphi_1 > \varphi_2 > \dots > \varphi_{n-1} > \varphi_n > 0$$

2. A sufficient condition for a discrete-time dynamical system to be asymptotically stable is that

$$\sum_{i=1}^n |\varphi_i| < |\varphi_0|$$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Jury's Test:



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Jury's Test:

$$\varphi_0 \quad \varphi_1 \quad \varphi_2 \quad \dots \quad \varphi_{n-1} \quad \varphi_n$$

.....

$$h_0 \quad h_1 \quad \dots \quad h_{\nu-1} \quad h_\nu$$

$$l_0 \quad l_1 \quad \dots \quad l_{\nu-1}$$

.....



Build the Jury's Table



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Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Jury's Test:

$$\varphi_0 \quad \varphi_1 \quad \varphi_2 \quad \cdots \quad \varphi_{n-1} \quad \varphi_n$$

.....

$$h_0 \quad h_1 \quad \cdots \quad h_{\nu-1} \quad h_\nu$$

$$l_0 \quad l_1 \quad \cdots \quad l_{\nu-1}$$

.....

Fill the generic line

$$l_i = \frac{1}{h_1} \det \begin{bmatrix} h_1 & h_{\nu-i+1} \\ h_\nu & h_i \end{bmatrix}$$



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Jury's Test:

	\leftarrow	Check the first column			
φ_0	φ_1	φ_2	\cdots	φ_{n-1}	φ_n
• • • • • • • • • • • •					
h_0	h_1	\cdots	$h_{\nu-1}$	h_ν	
l_0	l_1	\cdots	$l_{\nu-1}$		
• • • • • • •					

Discrete-time Systems: Stability of LTI Discrete-Time Systems

Characteristic Polynomial

$$\varphi(z) = \det(zI - A) = \varphi_0 z^n + \varphi_1 z^{n-1} + \cdots + \varphi_{n-1} z + \varphi_n$$

Jury's Test:

	← Check the first column				
φ_0	φ_1	φ_2	\cdots	φ_{n-1}	φ_n
• • • • • • • • • •					
h_0	h_1	\cdots	$h_{\nu-1}$	h_ν	
l_0	l_1	\cdots	$l_{\nu-1}$		
• • • • • •					

Jury's Criterion:

A LTI system is asymptotically stable if and only if the Jury Table associated with its characteristic polynomial is well defined and the elements appearing in the first column all have the same sign



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Example

Characteristic Polynomial $\varphi(z) = \det(zI - A) = z^3 + 2z^2 + z + 3$

Jury's Test:

$$\begin{matrix} 1 & 2 & 1 & 3 \\ -8 & & & \end{matrix}$$

Fill the second line

$$\begin{matrix} l_0 & l_1 & \dots & l_{\nu-1} \end{matrix}$$

Fill the generic line

$$l_1 = \frac{1}{1} \det \begin{bmatrix} 1 & 3 \\ 3 & 1 \end{bmatrix}$$

$$l_i = \frac{1}{h_1} \det \begin{bmatrix} h_1 & h_{\nu-i+1} \\ h_{\nu} & h_i \end{bmatrix}$$

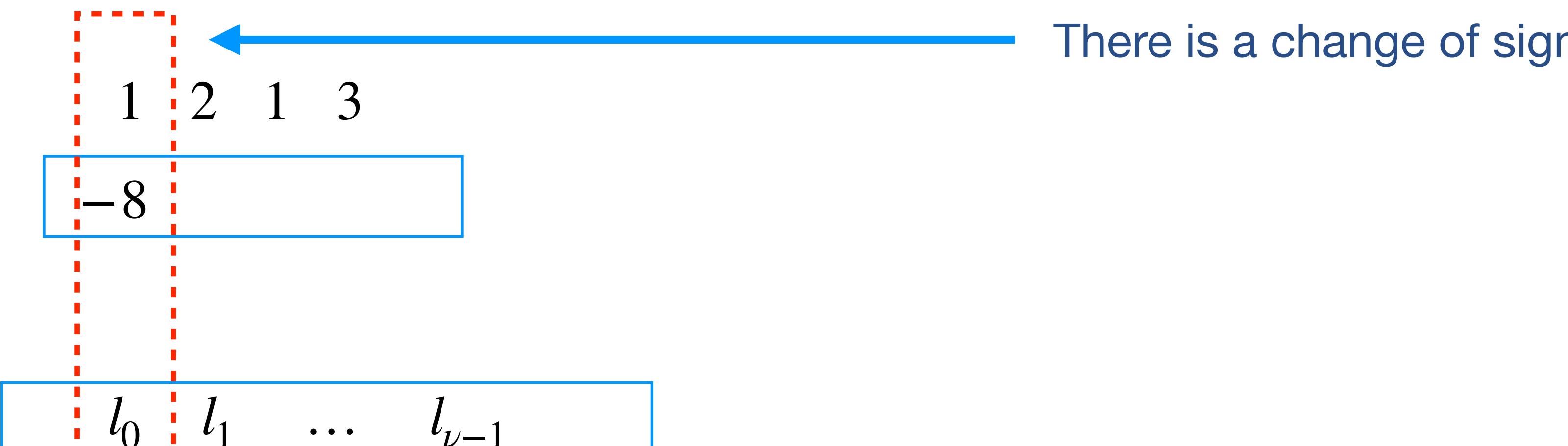


Discrete-time Systems: Stability of LTI Discrete-Time Systems

Example

Characteristic Polynomial $\varphi(z) = \det(zI - A) = z^3 + 2z^2 + z + 3$

Jury's Test:



Discrete-time Systems: Stability of LTI Discrete-Time Systems

Example

Characteristic Polynomial $\varphi(z) = \det(zI - A) = z^3 + 2z^2 + z + 3$

Jury's Test:

1	2	1	3
-8			
l_0	l_1	\dots	$l_{\nu-1}$

There is a change of sign

Jury's Criterion:

A LTI system is asymptotically stable if and only if the Jury Table associated with its characteristic polynomial is well defined and the elements appearing in the first column all have the same sign

The system with characteristic polynomial $\varphi(z)$ is NOT As. Stable

Digital Control Schemes



$$u(t) \in \mathbb{R}^m$$

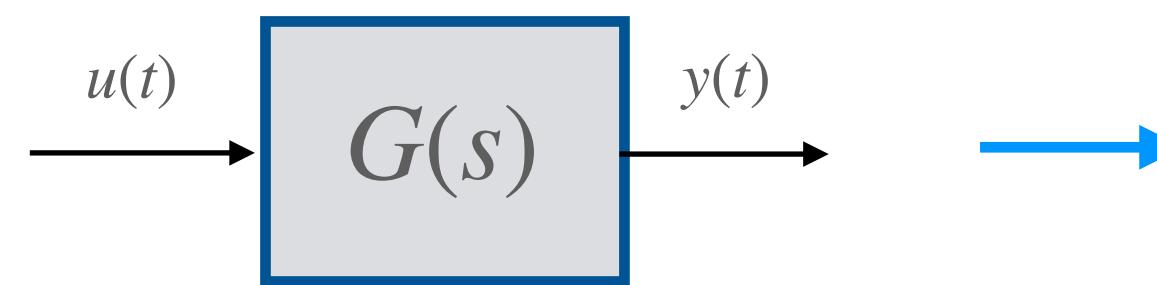
$$y(t) \in \mathbb{R}^p$$

$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System



Digital Control Schemes



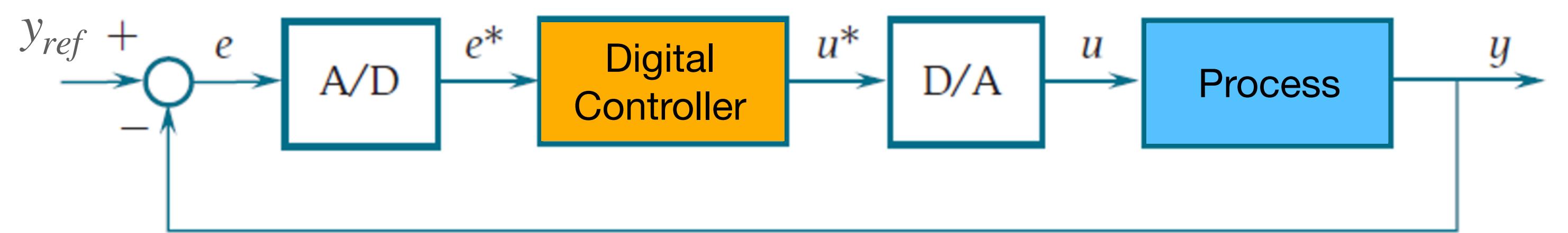
$$u(t) \in \mathbb{R}^m$$

$$y(t) \in \mathbb{R}^p$$

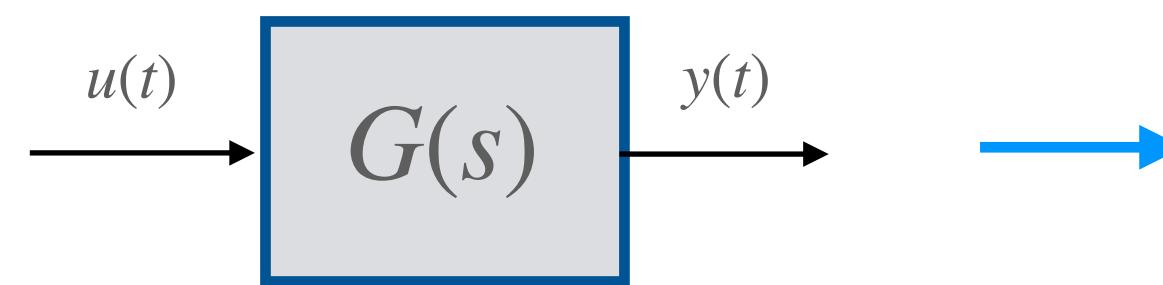
$$t \in \mathbb{R}_{\{0\}}^+$$

Continuous-time System

Digital control scheme with error sampling



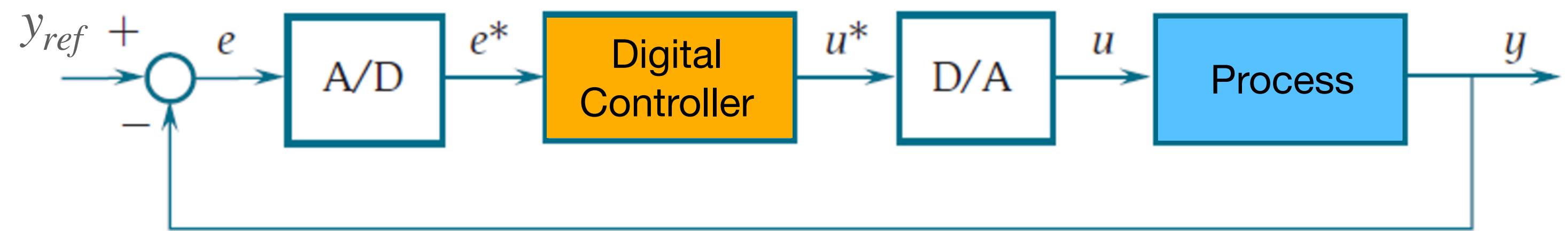
Digital Control Schemes



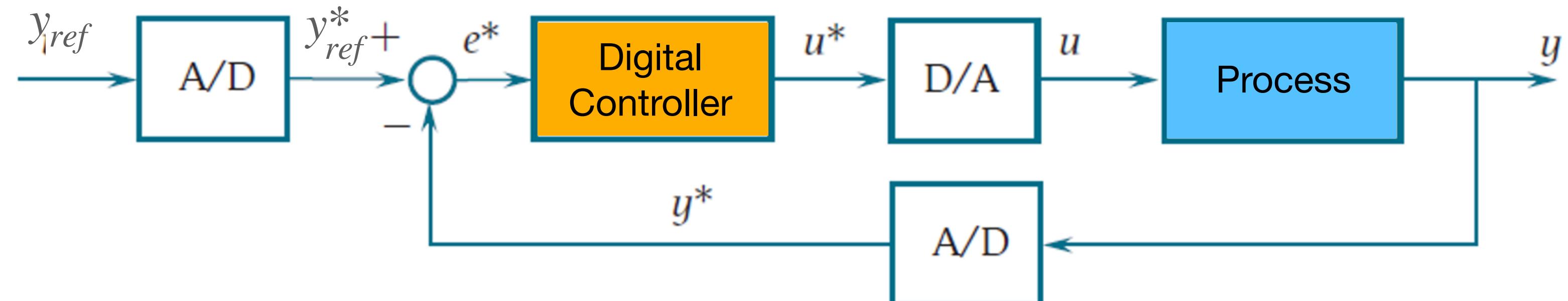
$$\begin{aligned} u(t) &\in \mathbb{R}^m \\ y(t) &\in \mathbb{R}^p \end{aligned}$$

$t \in \mathbb{R}_{\{0\}}^+$
Continuous-time System

Digital control scheme with error sampling



Digital control scheme with output and reference sampling



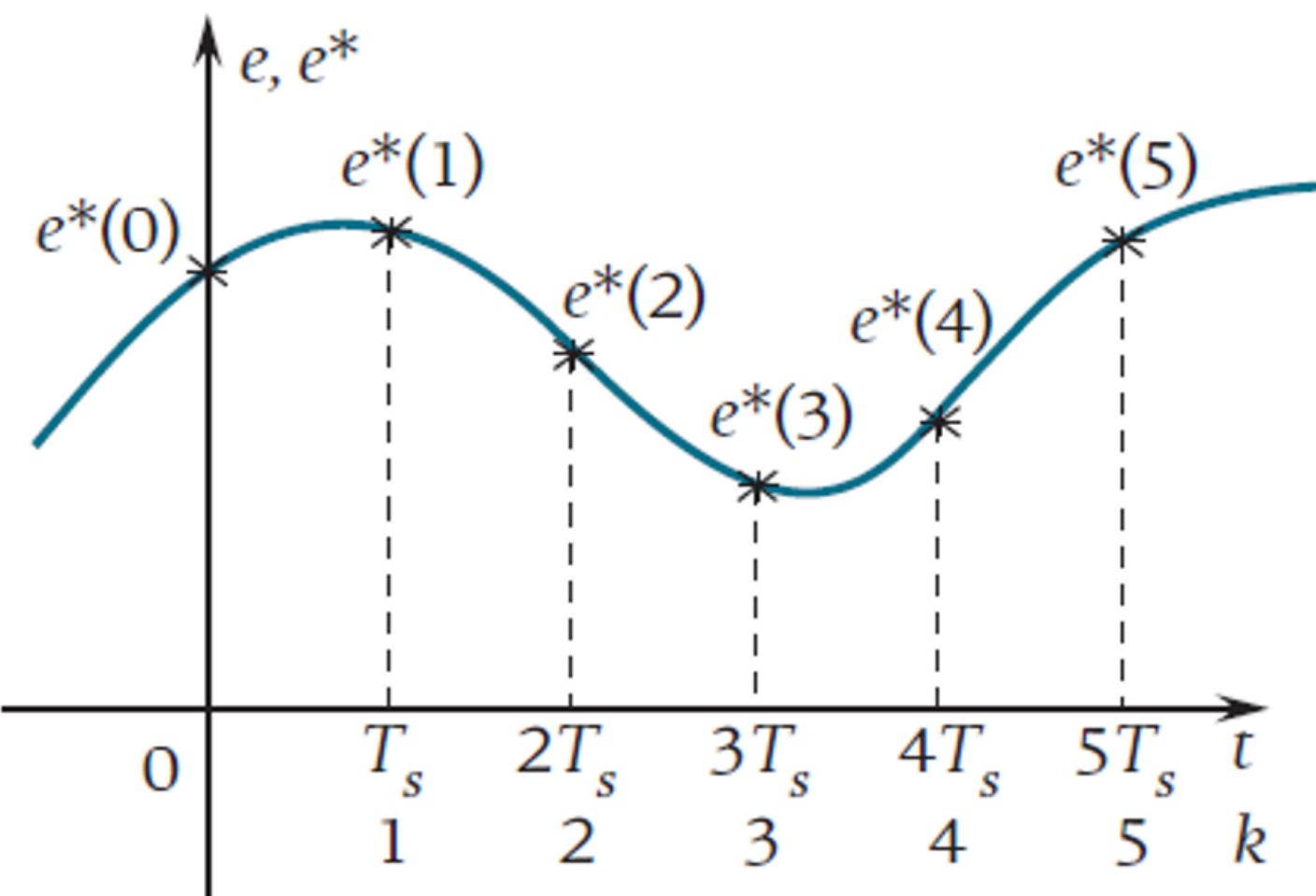
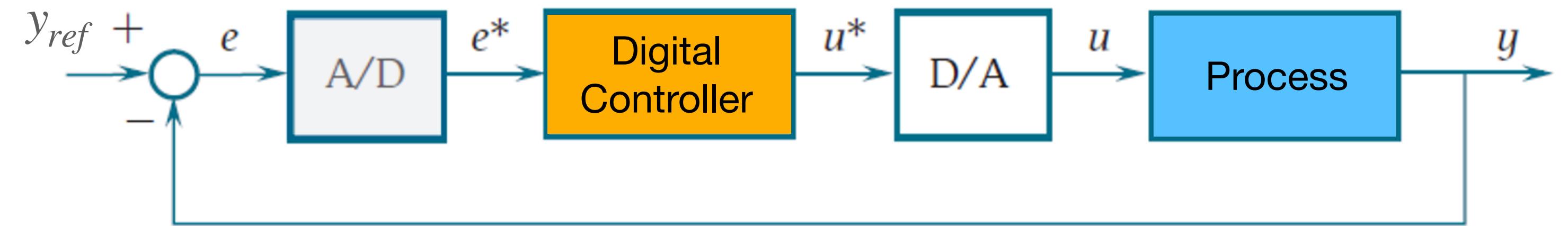
Digital Control Schemes

$\{e^*(k)\}, \{u^*(k)\}$ sequences

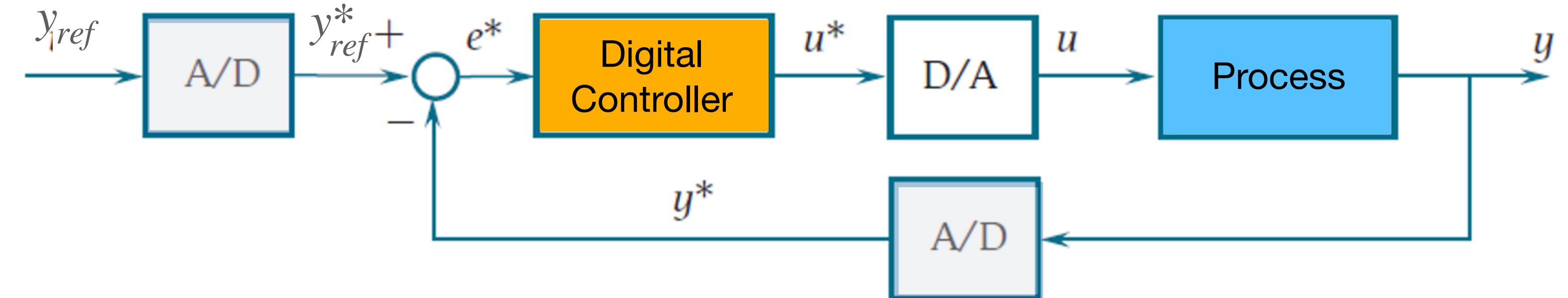
A/D converter: it is a sampler



Digital control scheme with error sampling



Digital control scheme with output and reference sampling

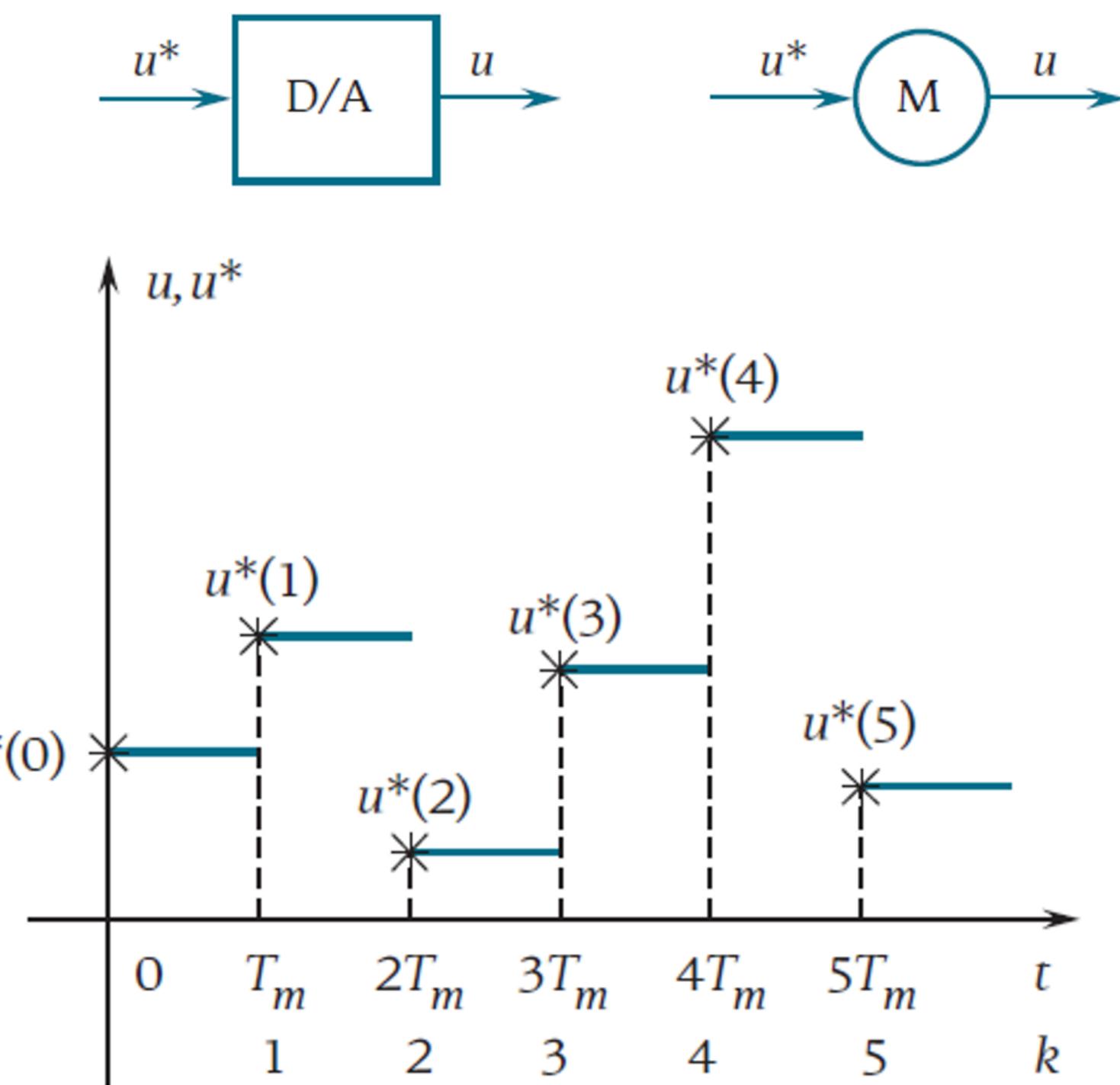


$\{y^*_{ref}(k)\}, \{y^*(k)\}$ sequences

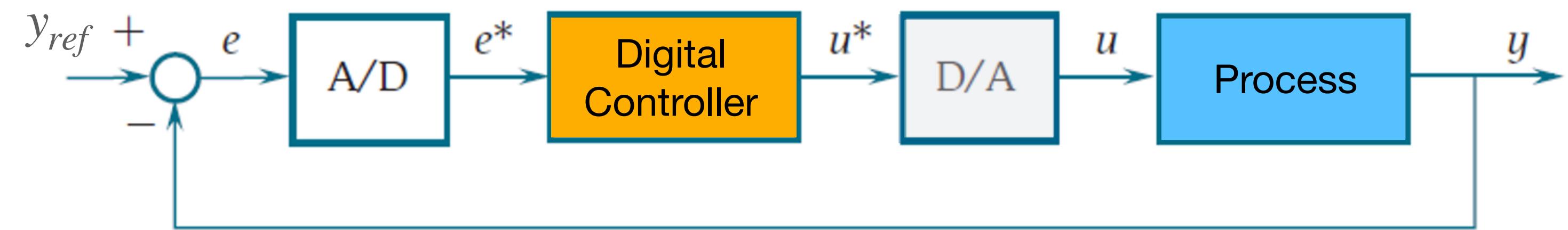


Digital Control Schemes

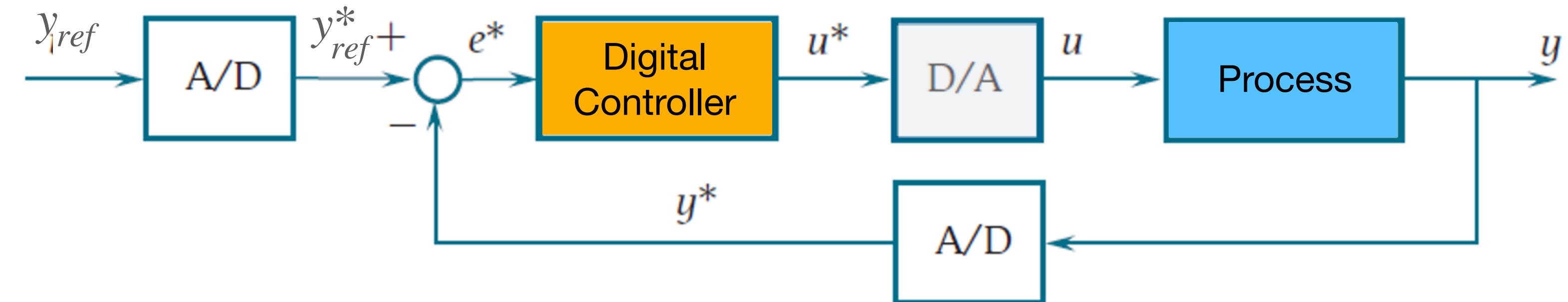
D/A converter: it is a hold



Digital control scheme with error sampling

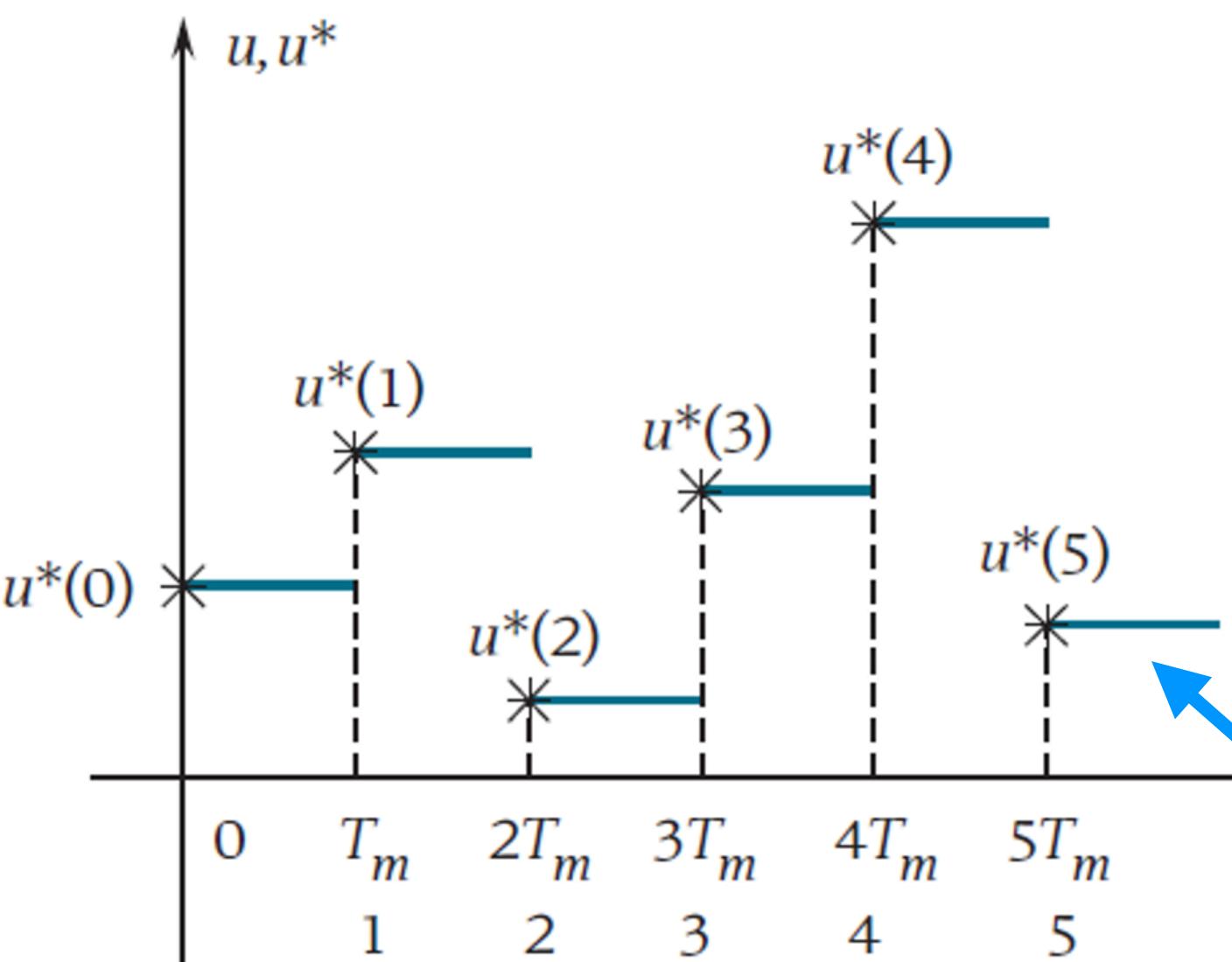


Digital control scheme with output and reference sampling

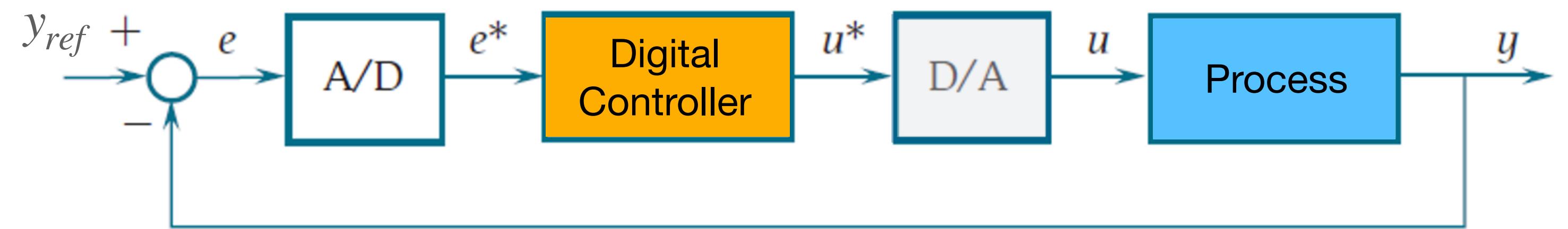


Digital Control Schemes

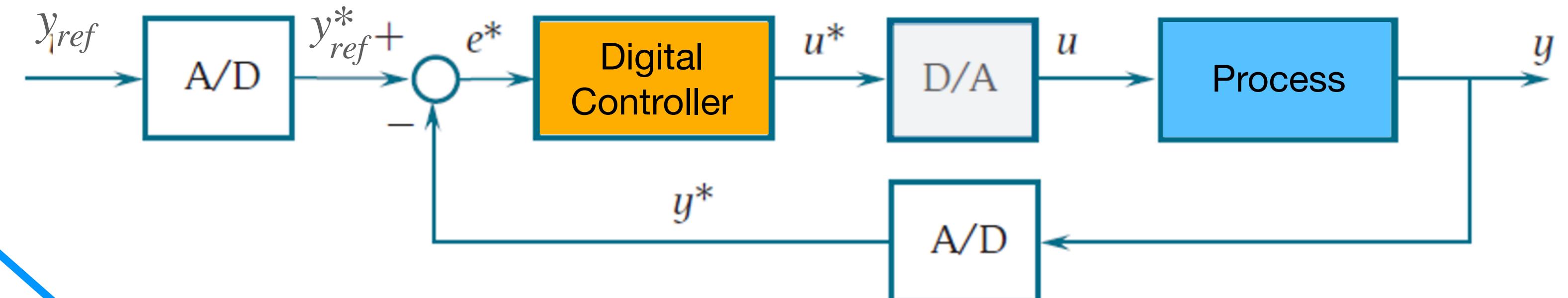
D/A converter: it is a hold



Digital control scheme with error sampling



Digital control scheme with output and reference sampling

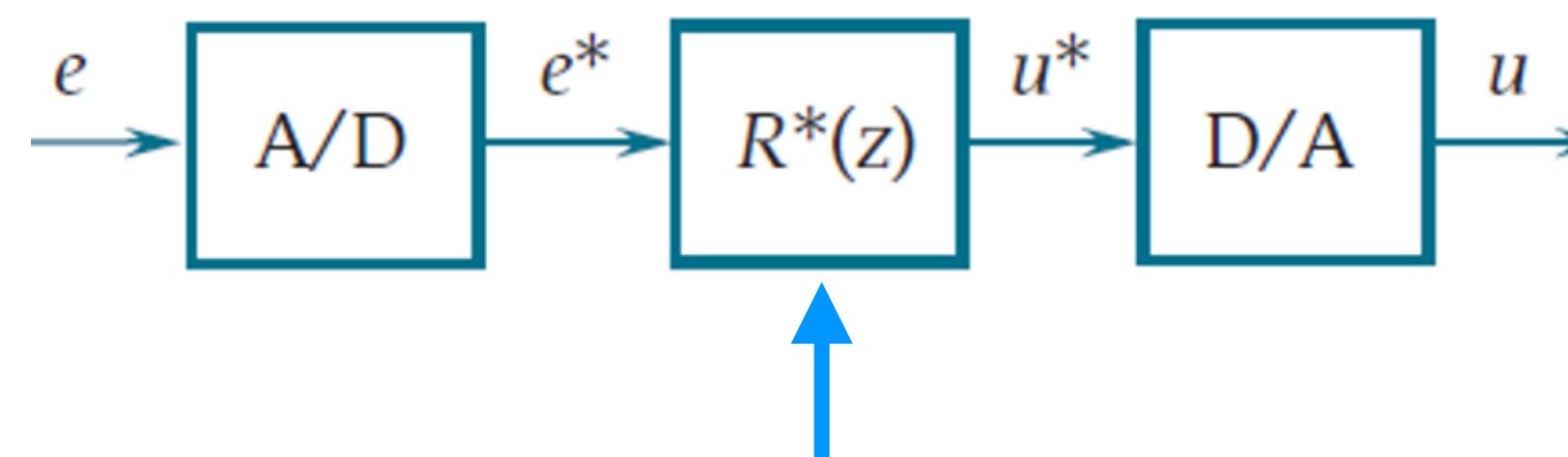


ZOH: zero-order hold



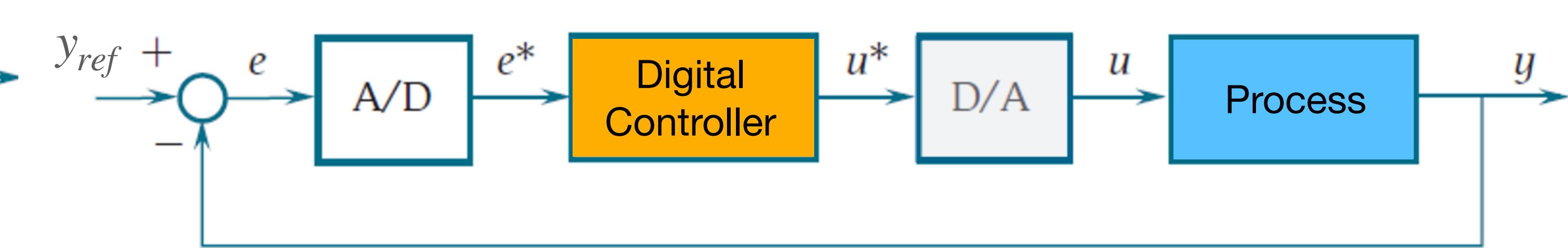
Digital Control Schemes

The control element:

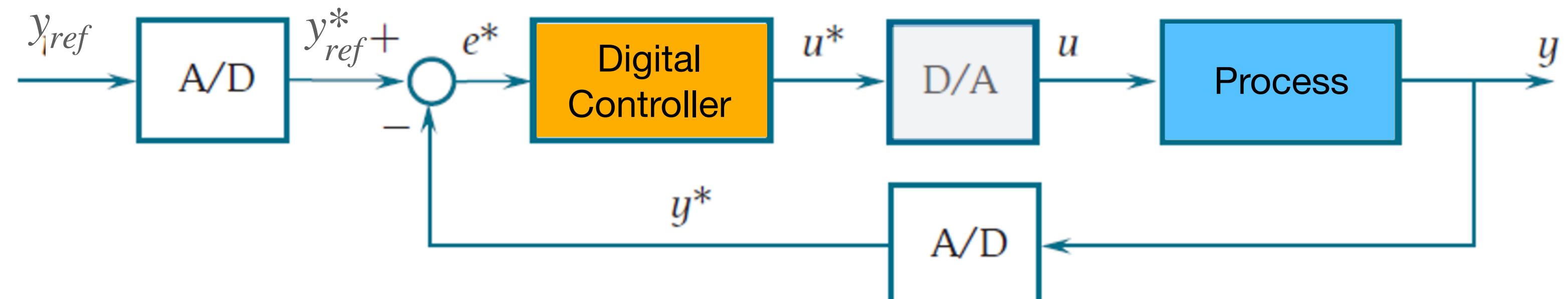


Digital Controller

Digital control scheme with error sampling

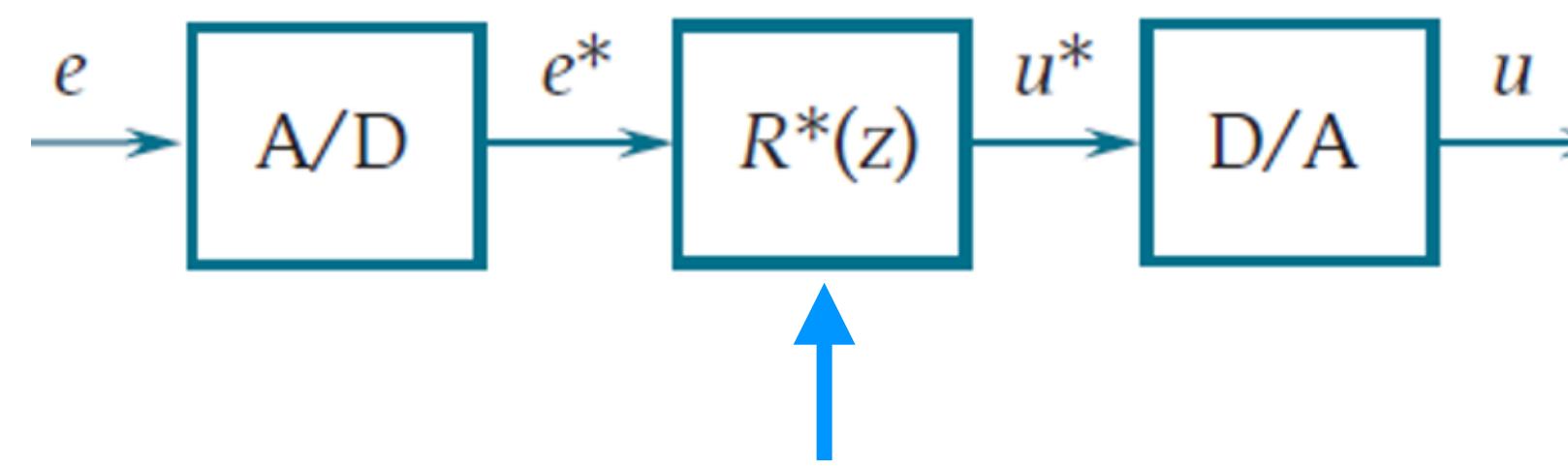


Digital control scheme with output and reference sampling

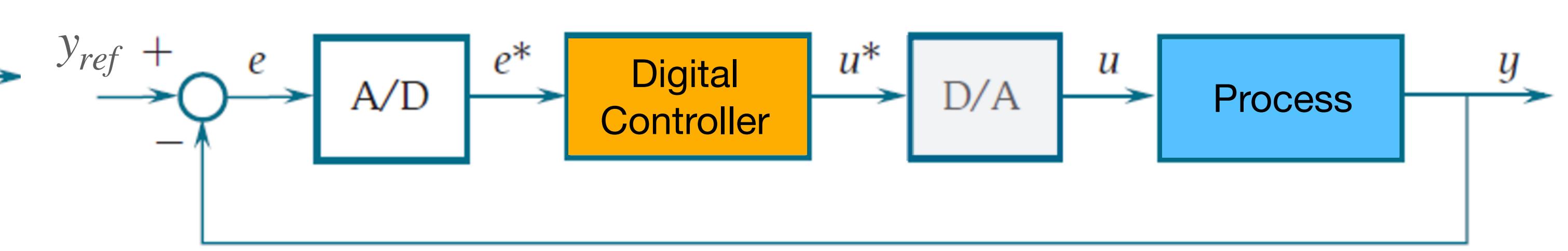


Digital Control Schemes

The control element:



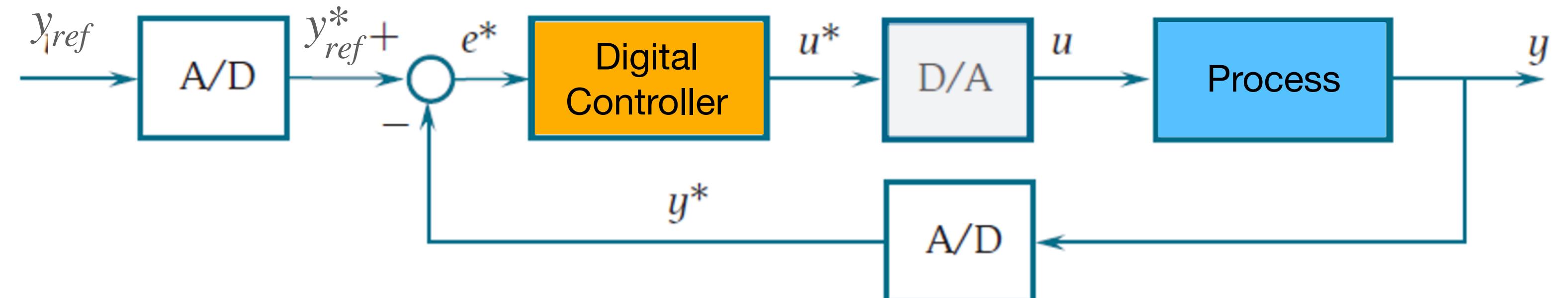
Digital control scheme with error sampling



Digital Controller

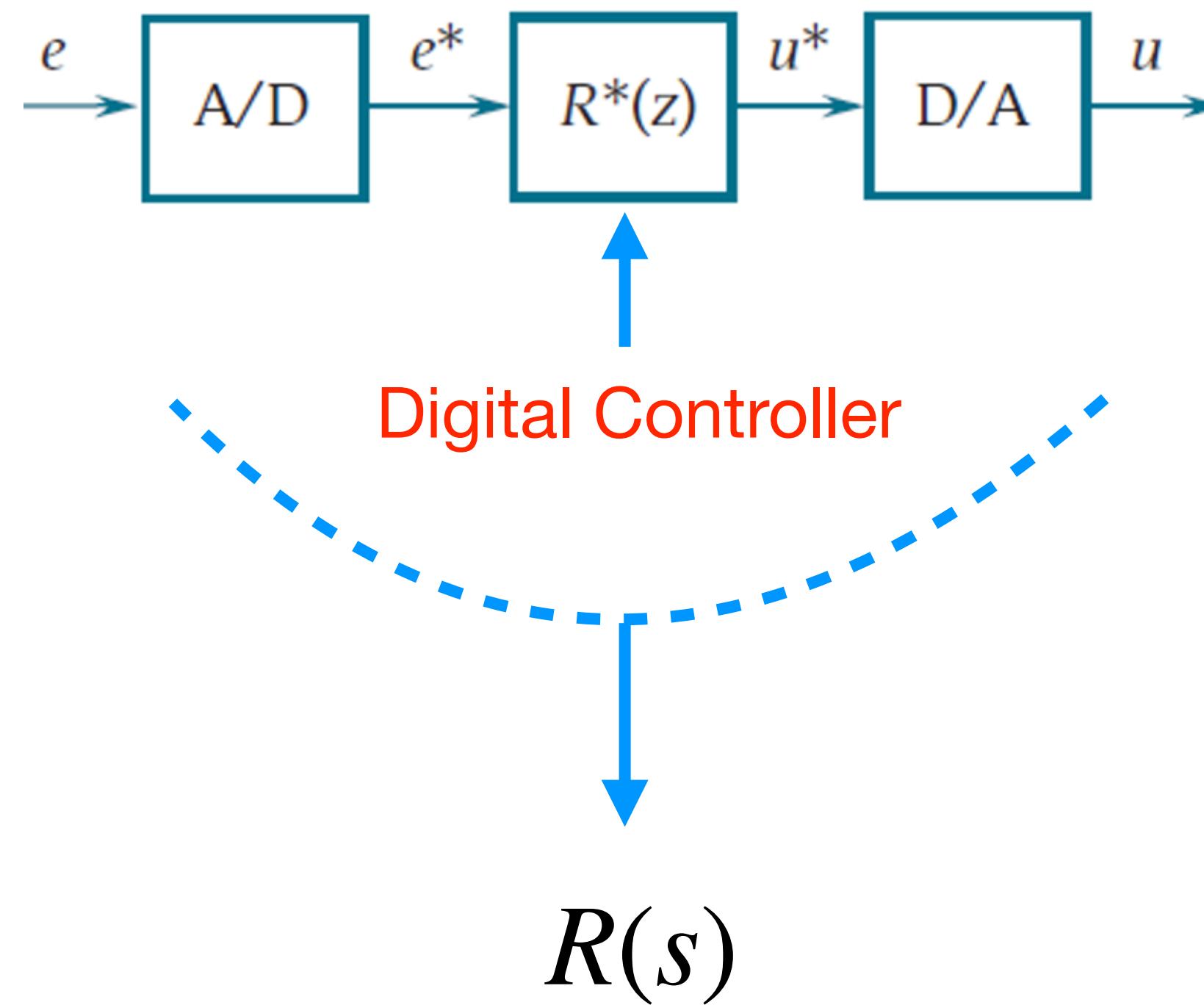
$R(s)$

Digital control scheme with output and reference sampling

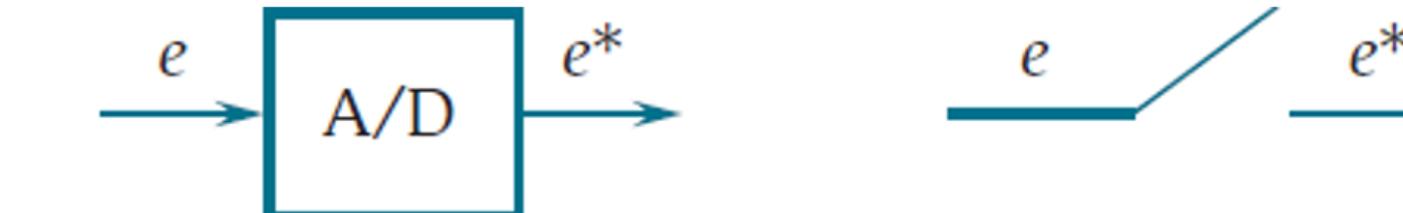


Digital Control Schemes: Problems to solve

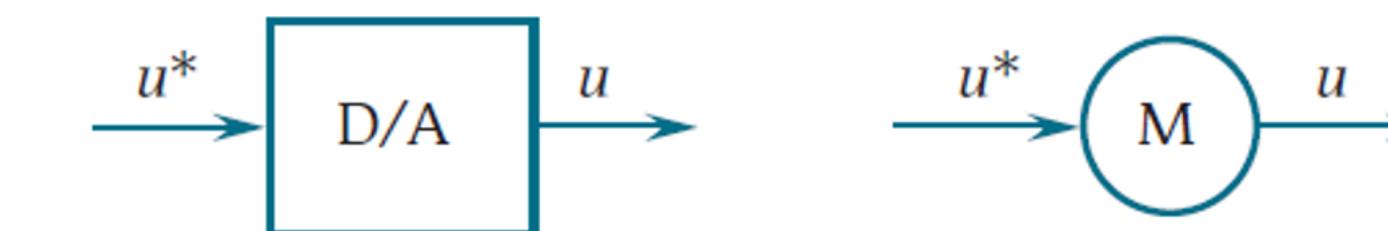
The control element:



A/D converter: it is a sampler



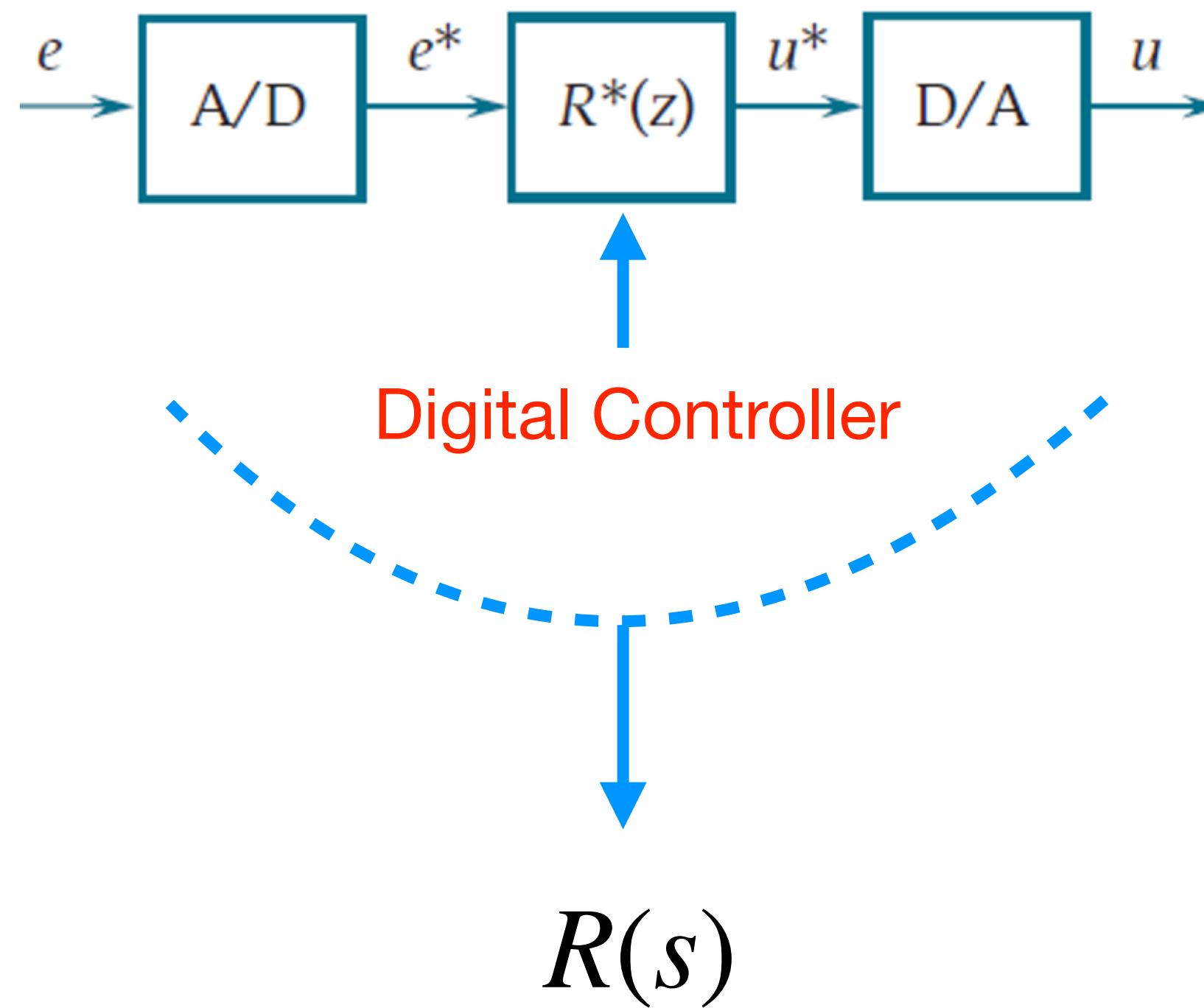
D/A converter: it is a hold



Digital Control Schemes: Problems to solve

1. How to determine the transfer function of a discrete-time LTI system

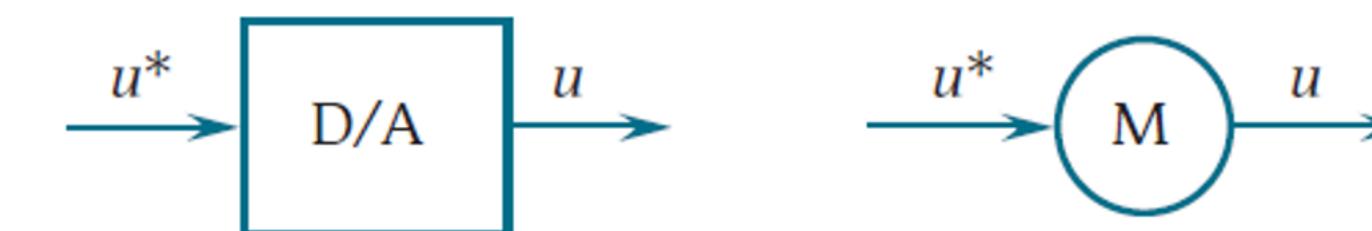
The control element:



A/D converter: it is a sampler



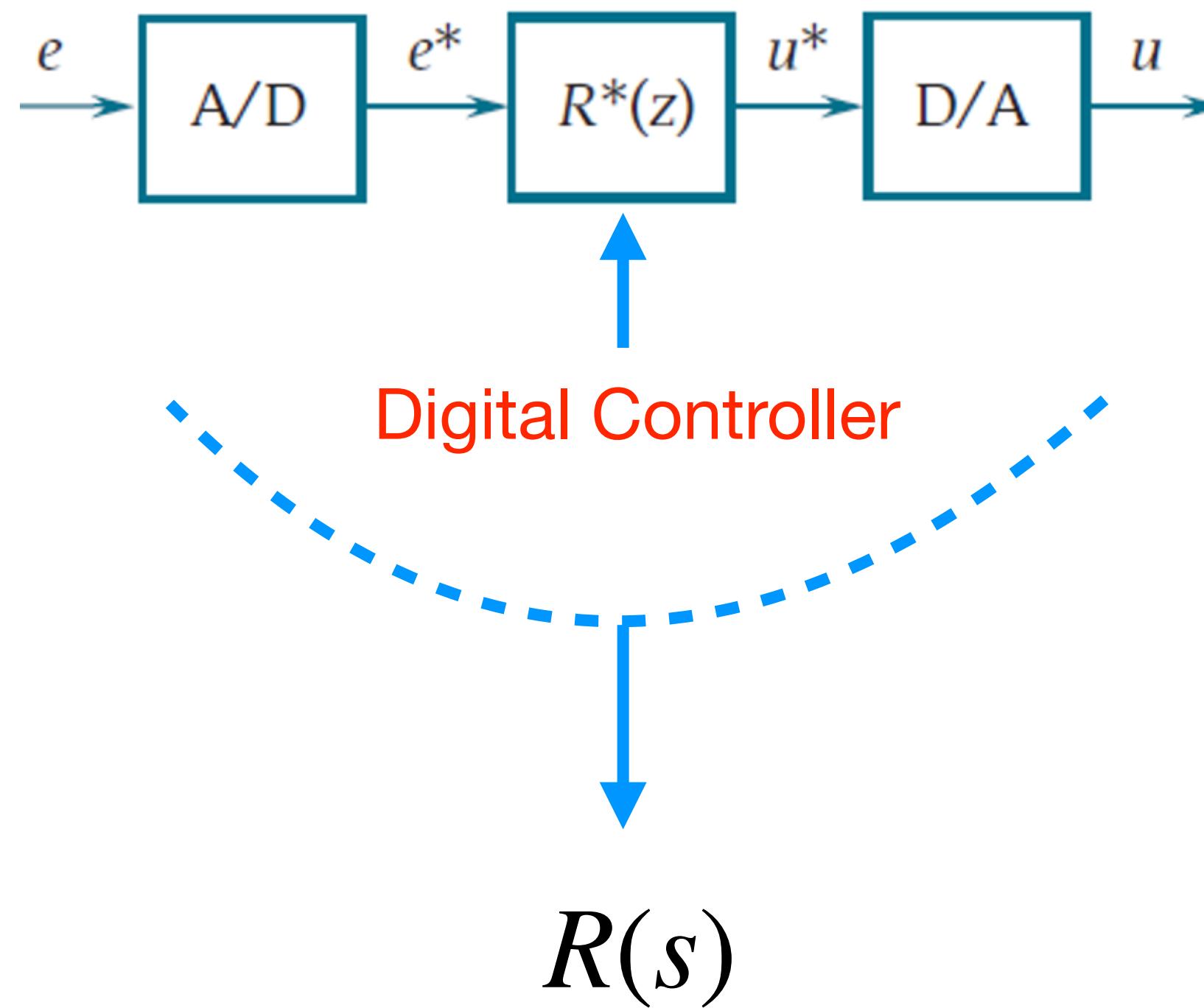
D/A converter: it is a hold



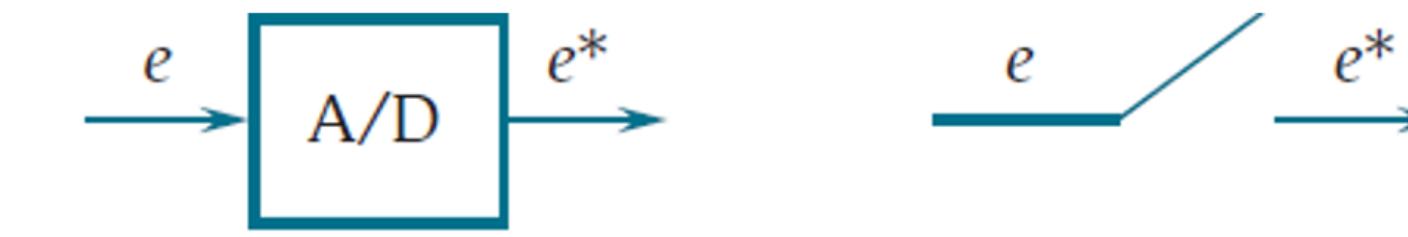
Digital Control Schemes: Problems to solve

1. How to determine the transfer function of a discrete-time LTI system

The control element:

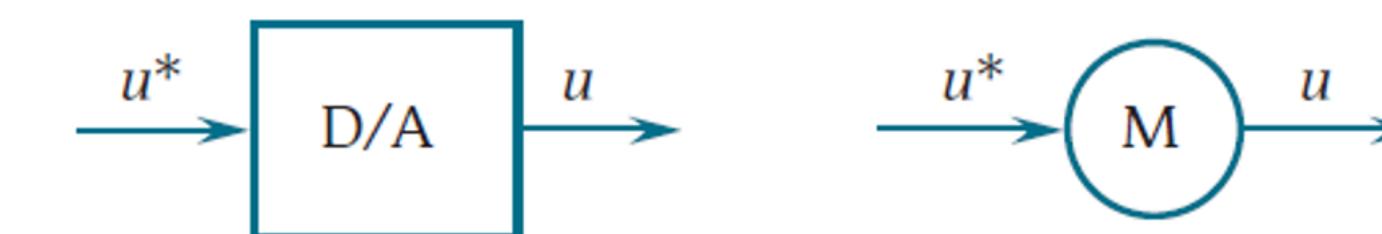


A/D converter: it is a sampler



2. How to determine the sampling time T_s

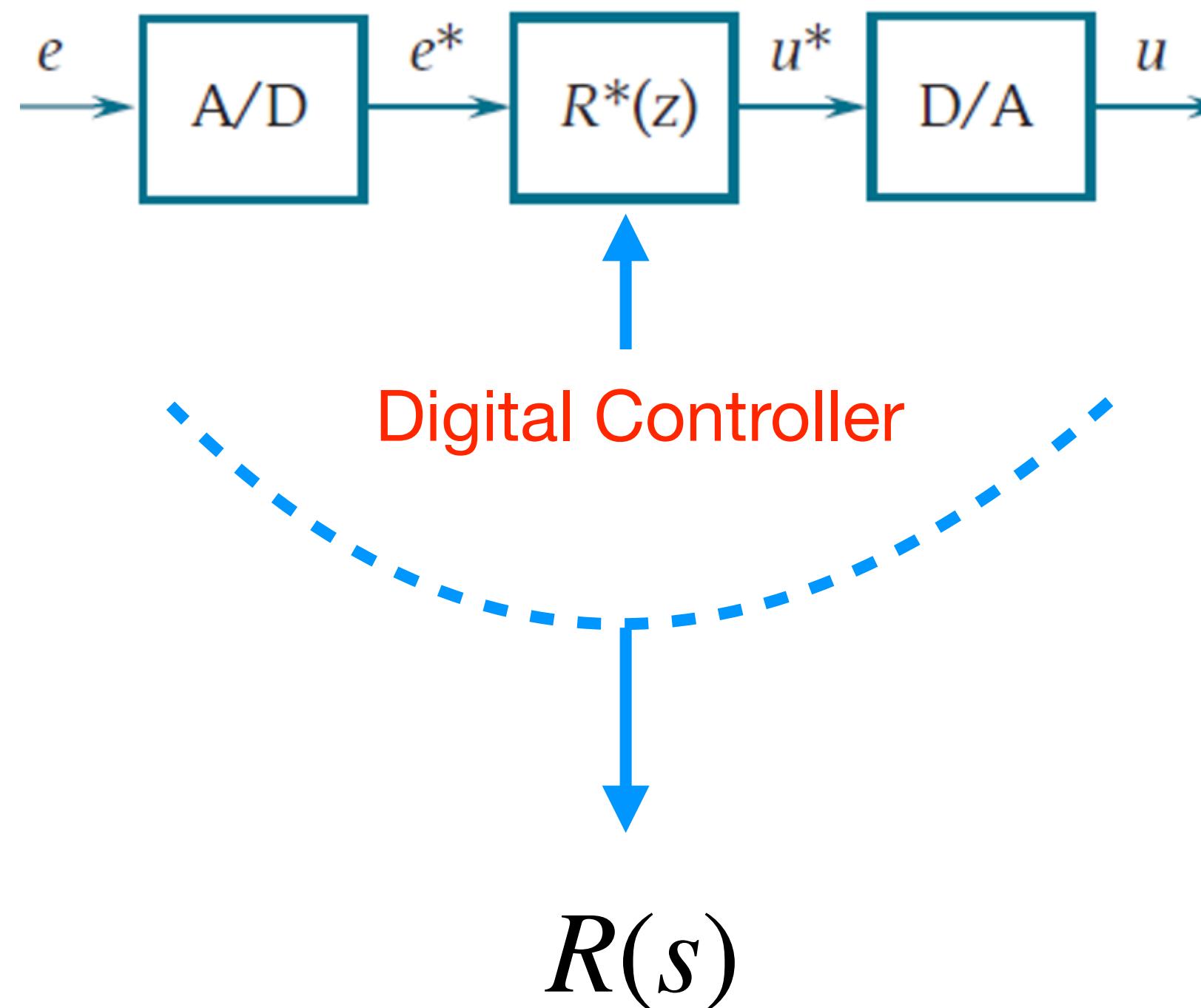
D/A converter: it is a hold



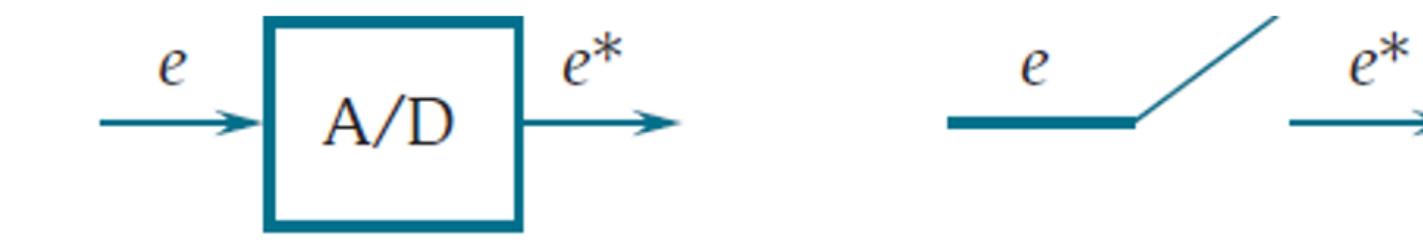
Digital Control Schemes: Problems to solve

1. How to determine the transfer function of a discrete-time LTI system

The control element:

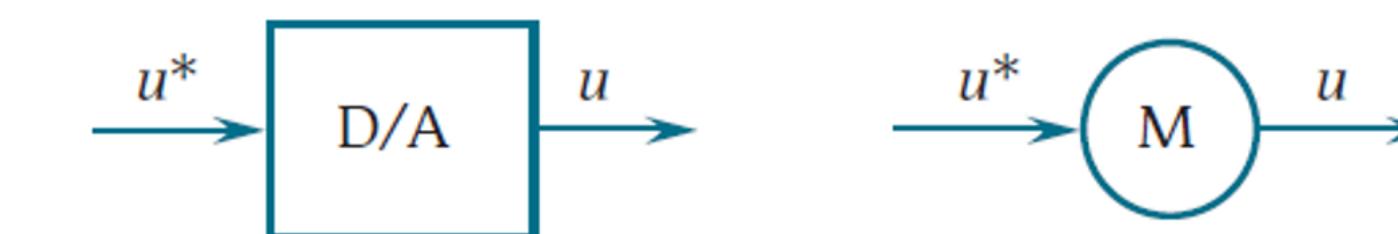


A/D converter: it is a sampler



2. How to determine the sampling time T_s

D/A converter: it is a hold



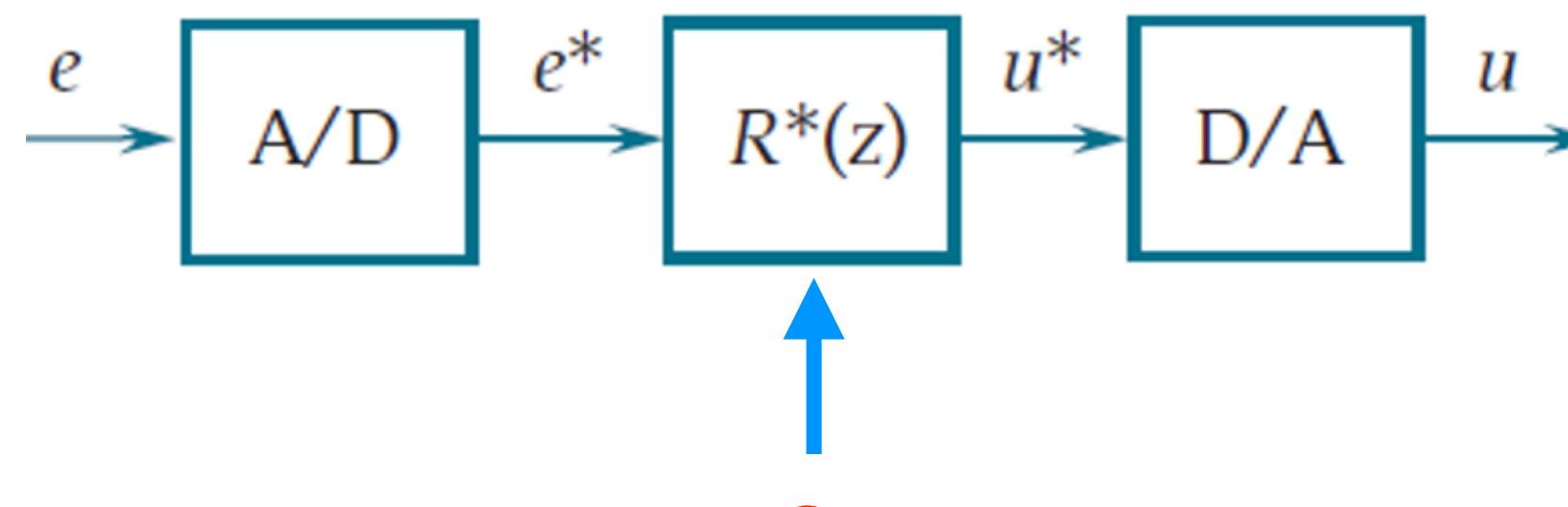
3. How to model the ZOH effect



Digital Control Schemes: Problems to solve

1. How to determine the transfer function of a discrete-time LTI system

The control element:

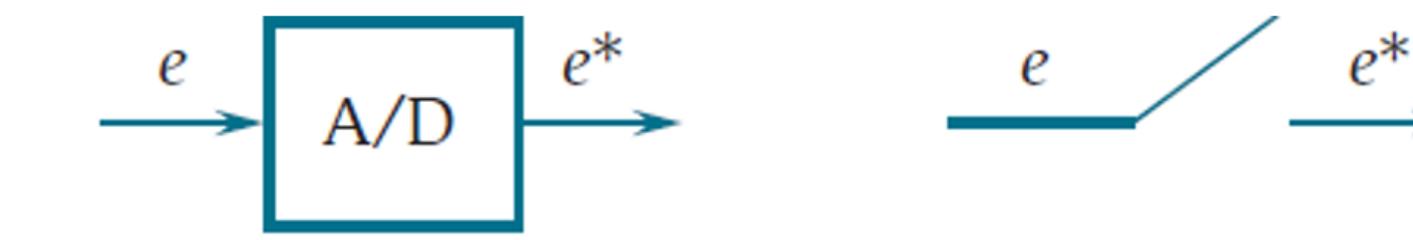


Digital Controller

$R(s)$

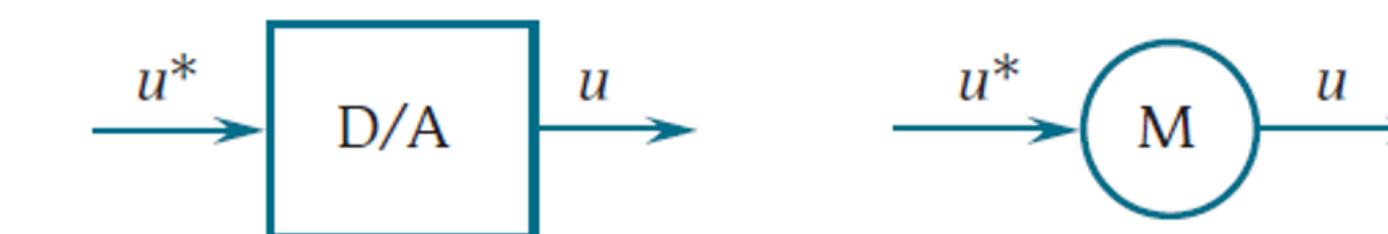
4. How to discretize
 $R(s)$ to obtain $R^*(z)$

A/D converter: it is a sampler



2. How to determine the sampling time T_s

D/A converter: it is a hold



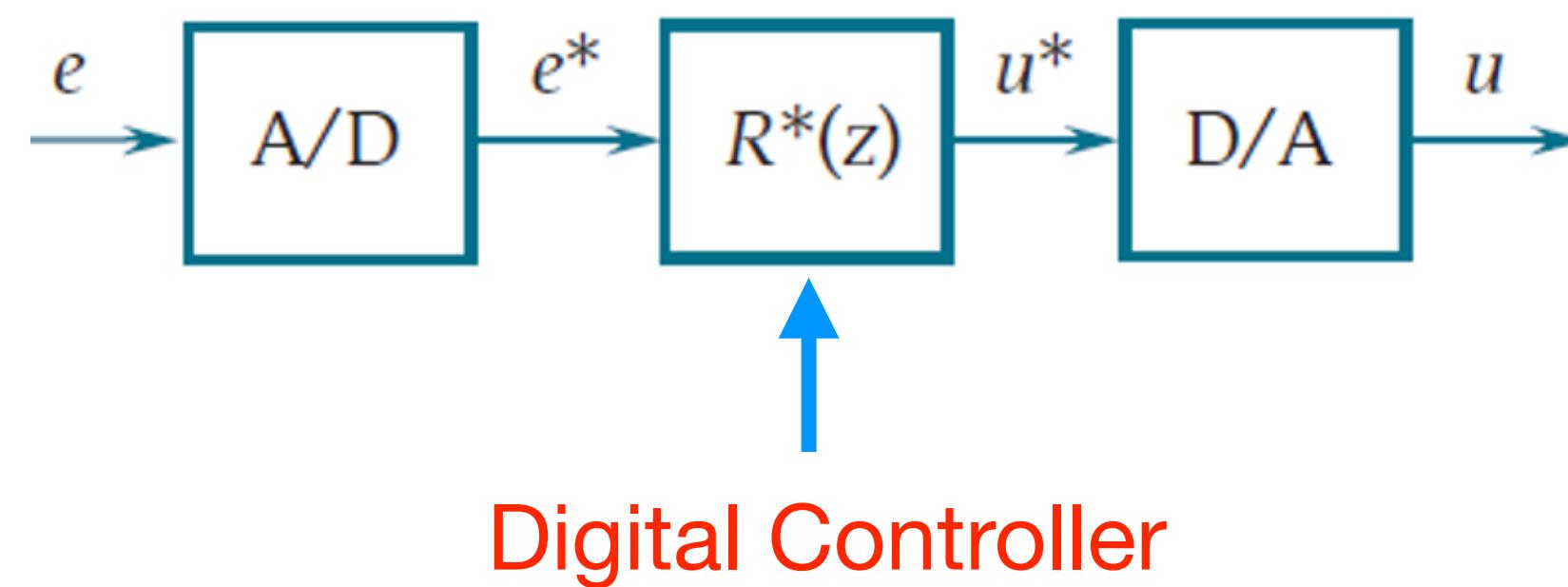
3. How to model the ZOH effect



Digital Control Schemes: Z-transform

1. How to determine the transfer function of a discrete-time LTI system

The control element:



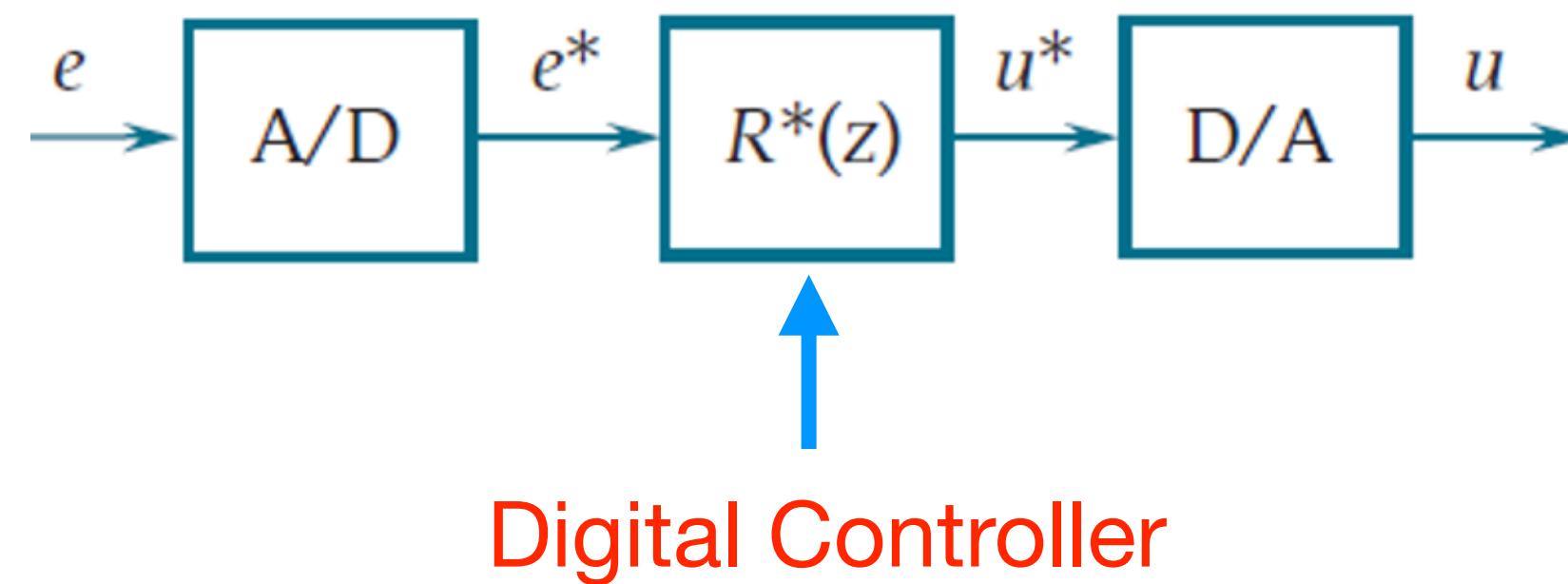
Given the discrete sequence $\{x(k)\}$, $k \geq 0$, its Z-transform is defined as:

$$X(z) = \mathcal{Z} [\{x(k)\}] = \sum_{k=0}^{\infty} x(k) z^{-k}, \text{ with } z \in \mathbb{C}$$

Digital Control Schemes: Z-transform

1. How to determine the transfer function of a discrete-time LTI system

The control element:



Given the discrete sequence $\{x(k)\}$, $k \geq 0$, its Z-transform is defined as:

$$X(z) = \mathcal{Z} [\{x(k)\}] = \sum_{k=0}^{\infty} x(k) z^{-k}, \text{ with } z \in \mathbb{C}$$

Properties:

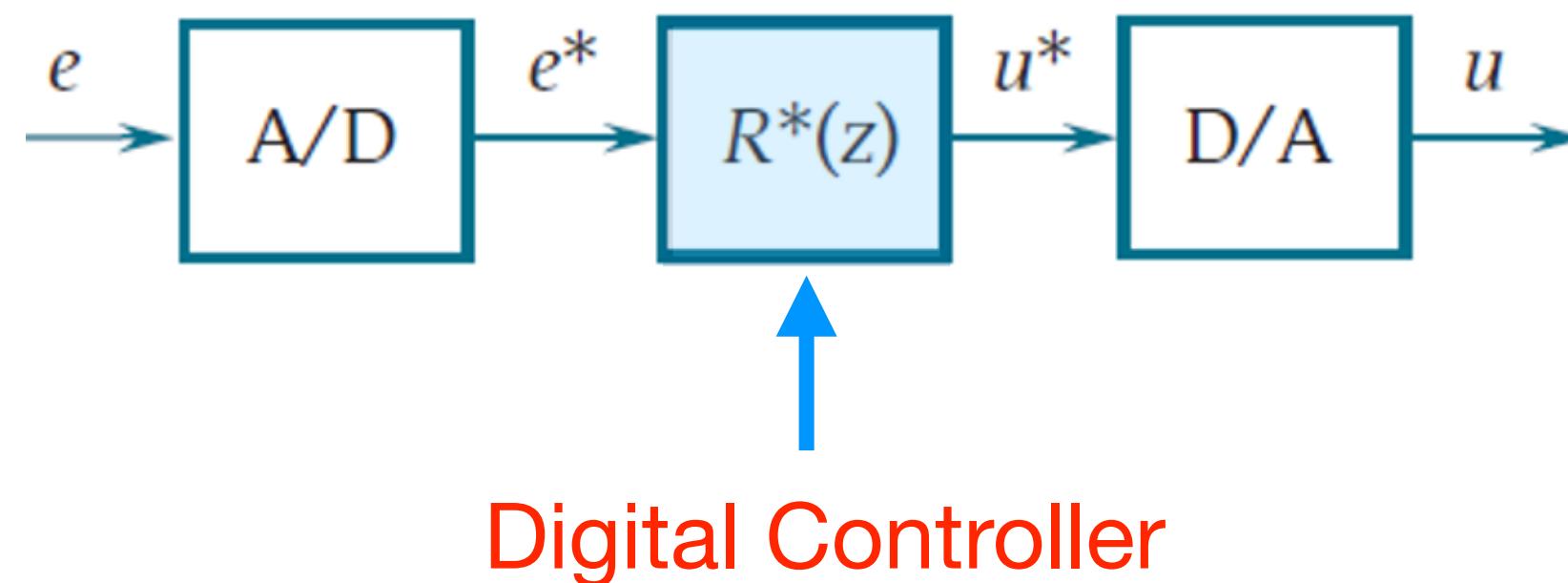
- 1) $\mathcal{Z} [a \{x(k)\} + b \{y(k)\}] = a X(z) + b Y(z)$
- 2) $\mathcal{Z} [\{x(k - n)\}] = z^{-n} X(z)$
- 3) $\mathcal{Z} [\{x(k + 1)\}] = z X(z)$



Digital Control Schemes: Z-transform

1. How to determine the transfer function of a discrete-time LTI system

The control element:



Given the discrete sequence $\{x(k)\}$, $k \geq 0$, its Z-transform is defined as:

$$X(z) = \mathcal{Z} [\{x(k)\}] = \sum_{k=0}^{\infty} x(k) z^{-k}, \text{ with } z \in \mathbb{C}$$

Properties:

Transfer Function:

$$R^*(z) = \frac{U^*(z)}{E^*(z)}$$



- 1) $\mathcal{Z} [a \{x(k)\} + b \{y(k)\}] = a X(z) + b Y(z)$
- 2) $\mathcal{Z} [\{x(k - n)\}] = z^{-n} X(z)$
- 3) $\mathcal{Z} [\{x(k + 1)\}] = z X(z)$

