



SAMPLING AND QUANTIZATION

EECE 2150 Circuits and Signals, Biomedical Engineering

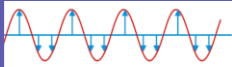
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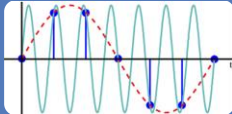
Learning Objectives



Learn the basics of analog to digital conversion



The Nyquist criteria and sampling a continuous time signal
Understanding the difference between continuous and discrete time frequency



Under-sampling and Aliasing



Quantization and mapping the discrete time signal into discrete levels

Analog to Digital Conversion, A/D



- ▶ To generate, store or process a continuous time signal on a computer, the signal must first be converted from analog to digital form.
- ▶ Samples are first taken from the continuous time signal at a given rate called the sampling rate or sampling frequency. The continuous time signal becomes a discrete time signal, but with continuous amplitude
- ▶ The next step is to convert the amplitude of the discrete time signal to a discrete set of levels, this process is referred to as quantization





Sampling



Sampling: Time Domain

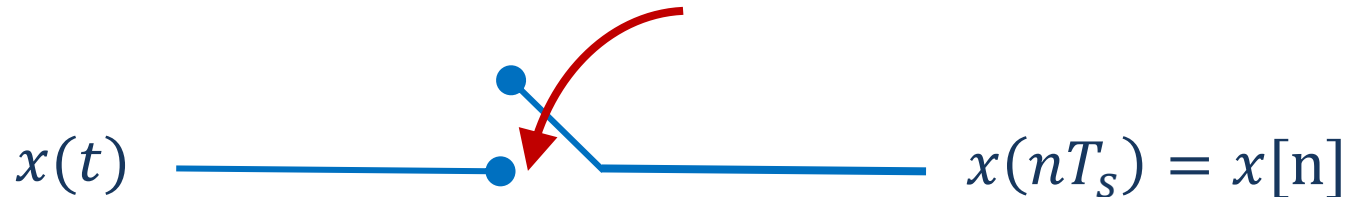
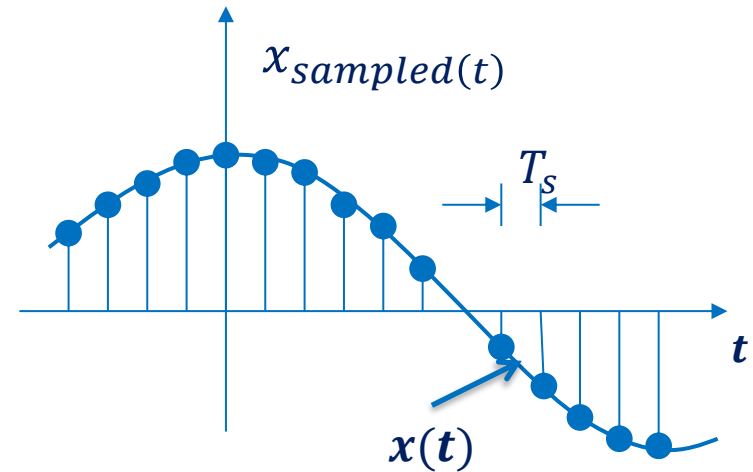
- ▷ A continuous-time signal is sampled at equally-spaced intervals in time,

$$x[n] = x(nT_s)$$

$$n \in \{..., -2, -1, 0, 1, 2, ...\}$$

T_s is the sampling interval

$$T_s = \frac{1}{f_s}$$



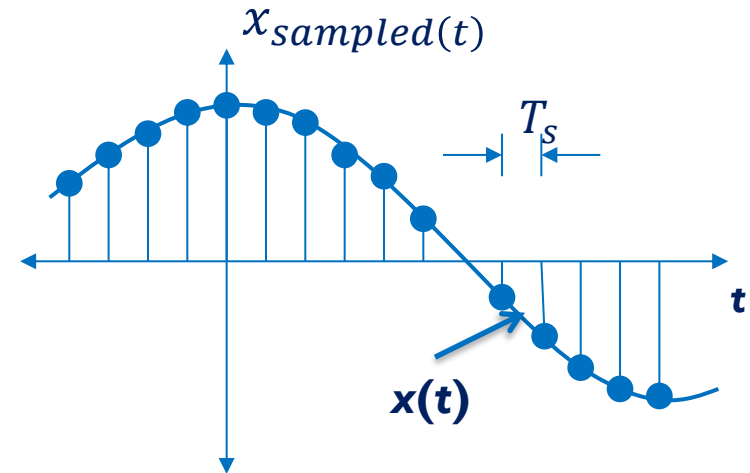
Switch closes momentarily every T_s to take a sample

How to choose the sampling rate



- For a given continuous-time signal, what is the minimum sampling rate? Or what is the largest sampling interval T_s ?

$$T_s = \frac{1}{f_s}$$



$$x(nT_s) = x[n]$$



Example: Sampling a Sine wave, choice of f_s

- ▶ $x(t) = A\sin(\omega_0 t + \phi) = A\sin(2\pi f_0 t + \phi)$
- ▶ The sampled version of the signal is given by =

$$x_{sampled}(nT_s) = x(nT_s) = A\sin(2\pi f_0 nT_s + \phi) = A\sin\left(2\pi \frac{f_0}{f_s} n + \phi\right)$$

The continuous time frequency f_0 is in $\frac{\text{cycles}}{\text{second}}$ and ω_0 is in $\frac{\text{radians}}{\text{second}}$

- ▶ The discrete signal can be expressed as

$$x[n] = x_{sampled}(nT_s) = A\sin(2\pi F n + \phi)$$

where $F = \frac{f_0}{f_s} = \frac{T_s}{T_0}$ = discrete time frequency, in cycles/sample

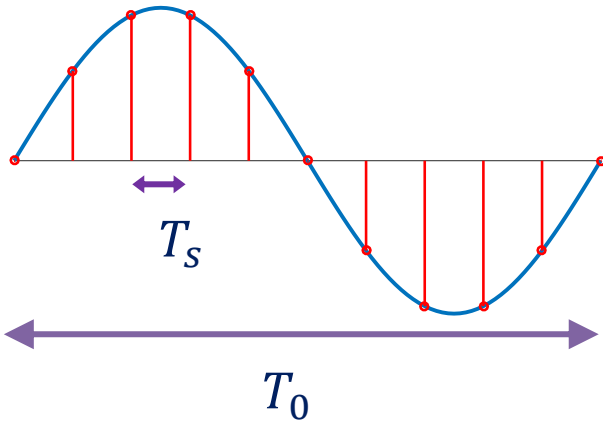
$\frac{T_0}{T_s}$ = number of samples per one cycle of the signal

$$x[n] = A\sin(\Omega n + \phi), \quad \Omega \text{ is in } \frac{\text{radians}}{\text{sample}} \quad \Omega = 2\pi F$$





Continuous time versus discrete time signals



Continuous time frequency

- $x(t) = A \sin(\omega_0 t + \phi)$
- $x(t) = A \sin(2\pi f_0 t + \phi)$
- f_0 is in $\frac{\text{cycles}}{\text{second}}$
- ω_0 is in $\frac{\text{radians}}{\text{second}}$

Discrete time frequency,

- $x[n] = A \sin(\Omega n + \phi)$
- $x[n] = A \sin(2\pi F n + \phi)$
- $F = \frac{f_0}{f_s} = \frac{T_s}{T_0} \frac{\text{cycles}}{\text{sample}}$
- $\Omega = 2\pi F \frac{\text{radians}}{\text{sample}}$

▷ $\frac{T_0}{T_s} = 10 \rightarrow \frac{f_s}{f_0} = 10 \rightarrow 10 \text{ samples per cycle}$, the sampling rate is 10 times the sinusoidal signal frequency

▷ The discrete time frequency is defined as

▷ $F = \frac{f_0}{f_s} = \frac{1}{10} \text{ cycles per sample}, \Omega = 2\pi F = 0.2\pi$

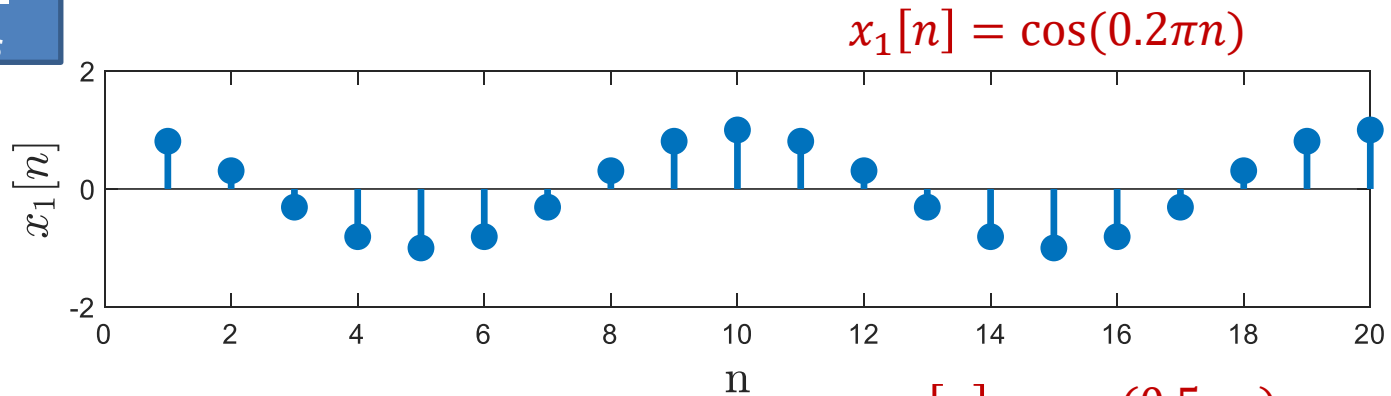


Discrete Time Signals at Different Frequencies

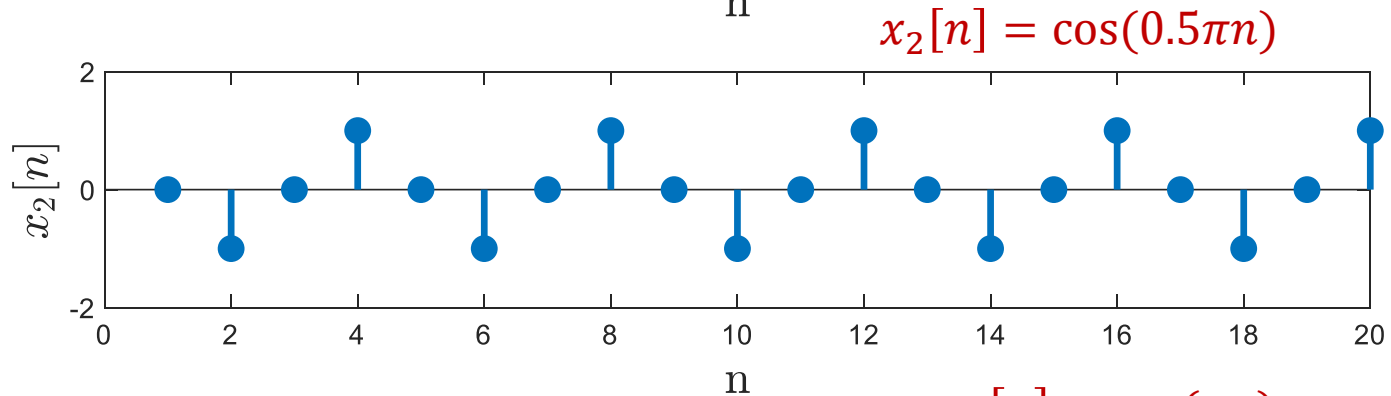


$$\Omega = 2\pi \frac{f_o}{f_s}$$

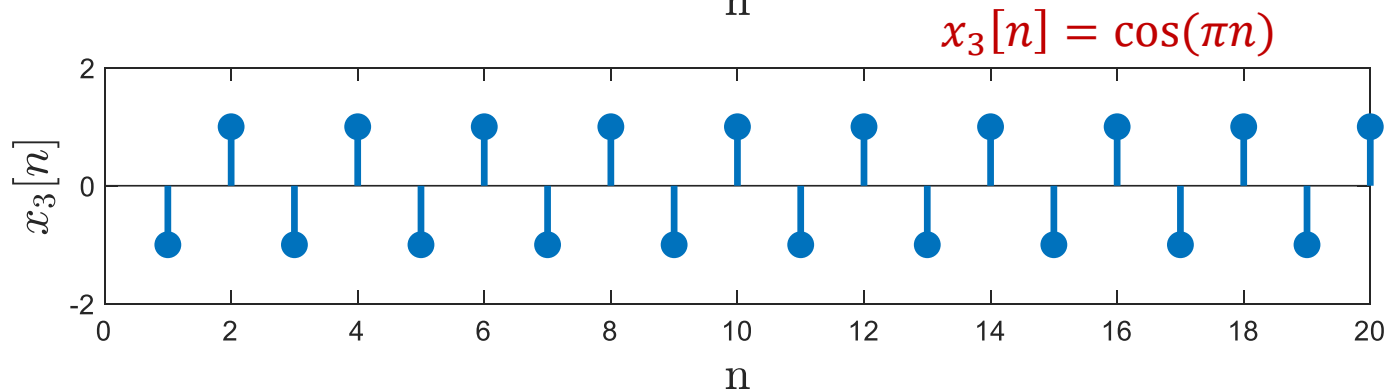
$$\Omega = 0.2\pi$$
$$\frac{f_o}{f_s} = \frac{1}{10}$$



$$\Omega = 0.5\pi$$
$$\frac{f_o}{f_s} = \frac{1}{4}$$



$$\Omega = \pi$$
$$\frac{f_o}{f_s} = \frac{1}{2}$$

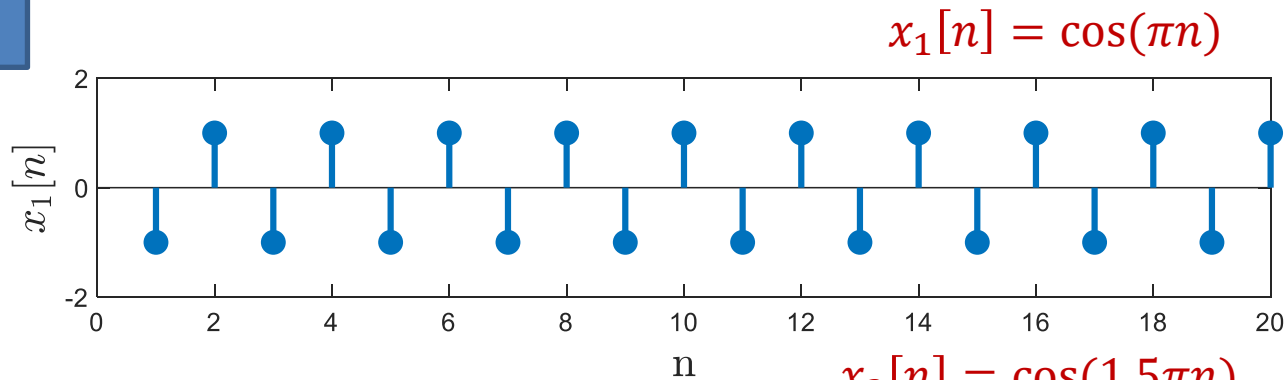


Discrete Time Signals at Different Frequencies

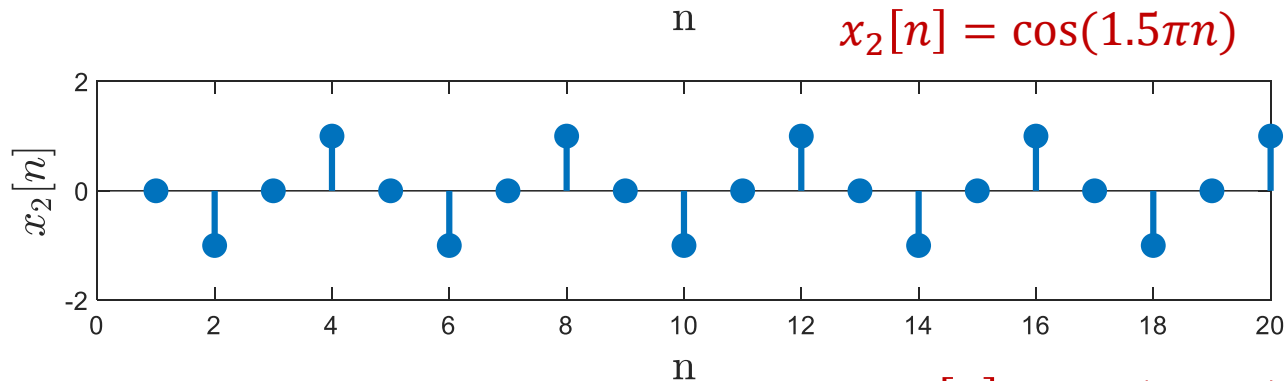


$$\Omega = 2\pi \frac{f_o}{f_s}$$

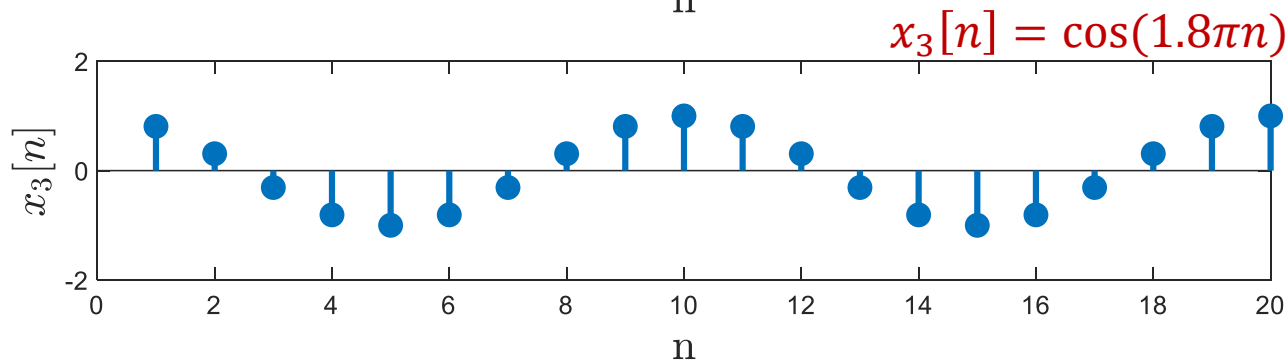
$$\Omega = \pi$$
$$\frac{f_o}{f_s} = \frac{1}{2}$$



$$\Omega = 1.5\pi$$
$$\frac{f_o}{f_s} = \frac{3}{4}$$



$$\Omega = 1.8\pi$$
$$\frac{f_o}{f_s} = \frac{9}{10}$$

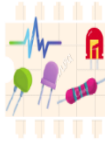




The discrete time Frequency Ω defined in the range $-\pi \leq \Omega \leq \pi$

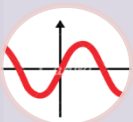
- ▷ $\cos(2.5\pi n) = \cos((2.5\pi - 2\pi)n) = \cos(0.5\pi n)$
- ▷ $\cos(1.5\pi n) = \cos((1.5\pi - 2\pi)n) = \cos(-0.5\pi n)$

Summary of the Difference between continuous time Frequency, ω , and discrete time frequency, Ω



Continuous time frequency, ω

- $x(t) = A\sin(\omega t + \phi)$
- The units of ω : rad/s,
- ω has a distinct value. As the magnitude of ω increases, the signal fluctuates at a higher rate



Discrete time frequency, Ω

- $x[n] = A\sin(\Omega n + \phi)$
- $\Omega = 2\pi F \frac{\text{radians}}{\text{sample}}$
- Ω has units of an angle(radians) and therefore it repeats after 2π .
- The discrete time frequency Ω increases in the range from $0 - \pi$ and then decreases from $\pi - 2\pi$ and then repeats.
- Discrete time frequencies near $0, 2\pi, 4\pi \dots 2n\pi$ are low frequencies, while frequencies in the vicinity of $\pi, 3\pi, \dots (2n + 1)\pi$ are high frequencies



The concept of Aliasing, definition of Alias (Oxford Living Dictionary)



- ▶ A false or assumed identity
- ▶ In Computing : an alternative name or label that refers to a file, command, address, or other item, and can be used to locate or access it.
- ▶ In Physics Telecommunications : Each of a set of signal frequencies which, when sampled at a given uniform rate, would give the same set of sampled values, and thus might be incorrectly substituted for one another when reconstructing the original signal.





Under-Sampling and Aliasing

- ▶ The higher the frequency content of the signal, the higher the sampling rate should be to preserve the full information in the signal.
- ▶ The Sampling Theorem states that a signal can be exactly reproduced if it is sampled at a frequency f_s (Hz) where f_s (Hz) is greater than twice the maximum frequency in the signal, which is known as the Nyquist rate..
- ▶ If the signal is sampled at a frequency that is lower than the Nyquist rate, when converted back into a continuous time signal, it will exhibit some form of distortion known as **aliasing**.
- ▶ Aliasing is the presence of unwanted frequency components in the reconstructed signal which were not present when the original signal was sampled.
- ▶ The process of aliasing describes the phenomenon in which components of the signal at high frequencies are mistaken for components at lower frequencies.



Frequency Limit

- ▶ The sampling theory says that we must capture at least two samples per cycle of a sinusoidal input waveform

$$f_s \geq 2f_0 \rightarrow T_s \leq \frac{T}{2} \quad \text{where } T = \frac{1}{f_0}$$

where f_s is the sampling frequency and f_0 is bandwidth or highest frequency component of the sampled signal





Sampling: Example

▷ Consider a signal $x(t) = 5 \cos(200\pi t)$

is sampled at a rate of $500 \frac{\text{samples}}{\text{sec}} \rightarrow f_s = 500 \text{ Hz} \quad T_s = \frac{1}{f_s}$

$$f_0 = \frac{200\pi}{2\pi} = 100 \text{ Hz}, \quad x[n] = 5 \cos(200\pi n T_s) =$$
$$5 \cos\left(\frac{200\pi n}{f_s}\right) =$$

$$x[n] = 5 \cos\left(\frac{200\pi n}{500}\right) = 5 \cos(0.4\pi n)$$

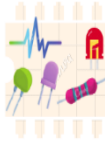
$$f_s = 500 \text{ Hz}, \quad F = \frac{f_0}{f_s} = \frac{100}{500} = 0.2$$

$$\text{Number of samples in one cycle } \frac{f_s}{f_0} = \frac{500}{100} = 5 = \frac{T_0}{T_s}$$

$$\text{Duration of 40 samples} = 40 T_s = \frac{40}{f_s} = \frac{40 \text{ sample}}{500 \frac{\text{samples}}{\text{s}}} = 0.08 \text{ sec}$$

Note that in this example $f_s = 5f_0$ is larger than the Nyquist rate and the signal can be accurately recovered

Aliasing Example



- Consider a signal $x_1(t) = 5 \cos(200\pi t)$ is sampled at a rate of 500 *samples/sec*. Here $f_s = 5f_0$

$$f_0 = \frac{200\pi}{2\pi} = 100\text{Hz}, x_1[n] = 5 \cos\left(\frac{200\pi n}{f_s}\right) =$$

$$x_1[n] = 5 \cos\left(\frac{200\pi n}{500}\right) = 5\cos(0.4\pi n)$$

- Now another signal $x_2(t) = 5 \cos(800\pi t)$ signal sampled at the same rate

- Here $\frac{f_s}{f_0} = \frac{500}{400} = 1.25$ $x_2[n] = 5 \cos\left(\frac{800\pi n}{f_s}\right) =$

$$x_2[n] = 5 \cos\left(\frac{800\pi n}{500}\right) = 5 \cos(1.6\pi n) = 5\cos(1.6\pi n - 2\pi n)$$

$$x_2[n] = 5 \cos(-0.4\pi n) = 5\cos(0.4\pi n) = x_1[n] \Rightarrow \text{even function}$$

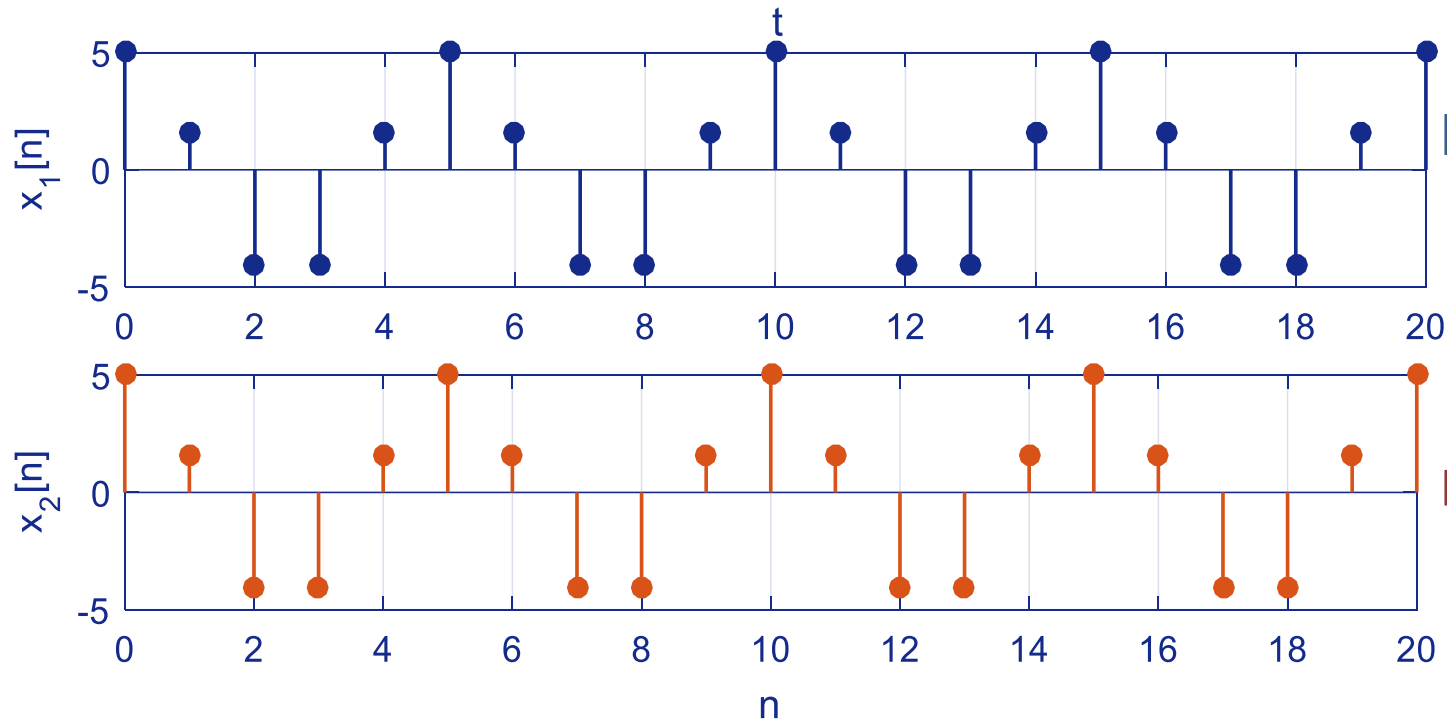
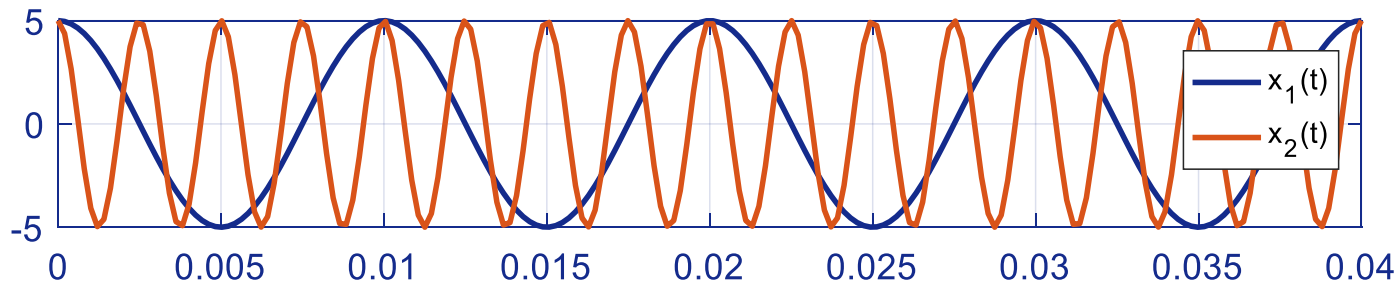
- The frequency $0.4\pi \frac{\text{rad}}{\text{sample}}$ is the alias frequency of the frequency $1.6\pi \text{ rad/sample}$
- The signal $x_2(t)$ which is under sampled will be incorrectly recovered as $x_1(t)$





Aliasing Example:

$$x_1(t) = 5 \cos(200\pi t) \quad x_2(t) = 5 \cos(800\pi t) \quad f_s = 500\text{Hz}$$



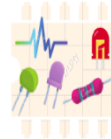
$$\Rightarrow f_s = 5f_{01}$$

$$\Rightarrow f_s = 1.25f_{02}$$

Two horizontal bars are positioned near the top of the slide. The left bar consists of a red segment followed by a blue segment. The right bar consists of a purple segment followed by a green segment.

Quantization

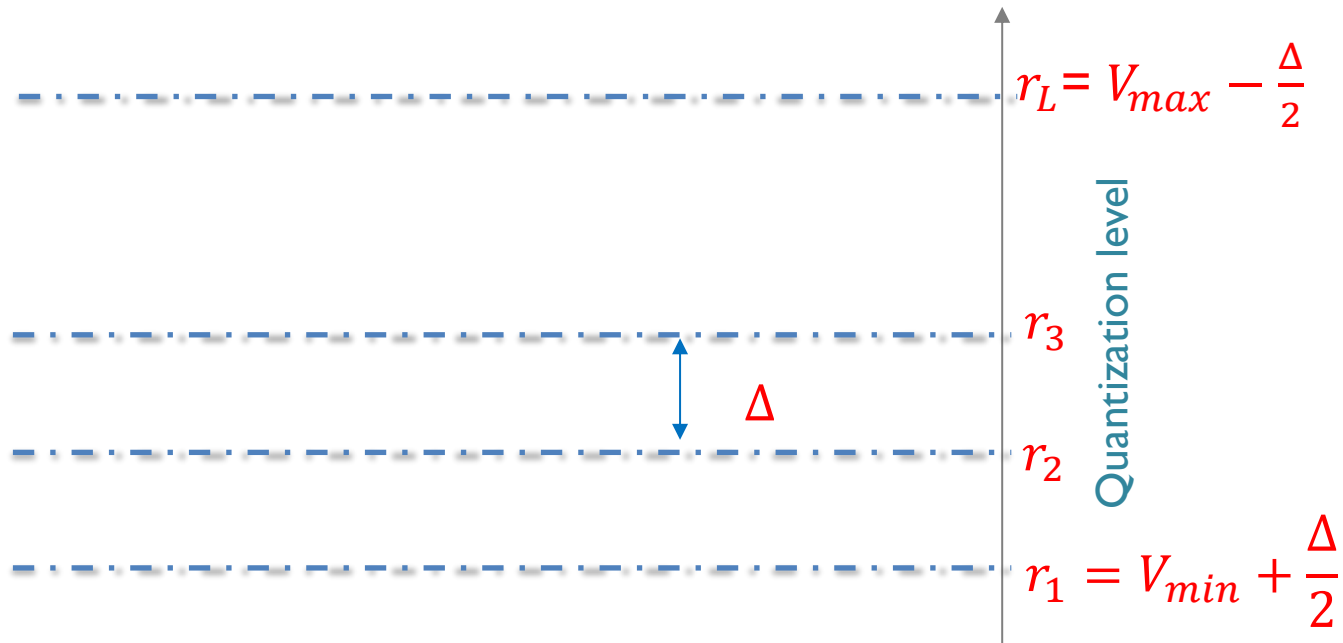
Quantization



- ▶ Quantization is the process of mapping continuous signal values to a discrete set of quantization levels.
- ▶ A quantized signal can only take on a discrete, usually finite, set of values.
- ▶ Unlike sampling, where exact reconstruction is possible under suitable conditions, quantization introduces irreversible errors, resulting in information loss.
- ▶ Quantization distortion is introduced into the quantized signal and cannot be completely eliminated.
- ▶ Increasing the number of discrete levels in an analog-to-digital converter (ADC) reduces quantization error but increases the required data storage.
- ▶ For an ADC system with N bits, the number of discrete levels L is given by $L = 2^N$



Uniform Quantization



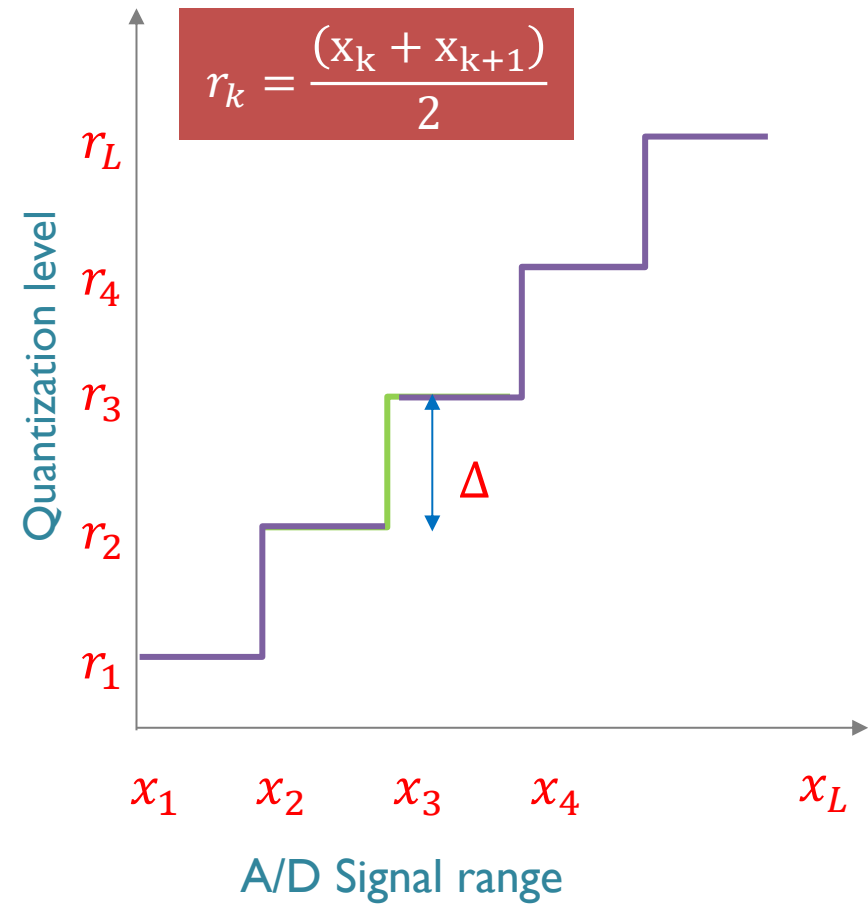
- ▶ **Uniform quantization** maps continuous signal values to a discrete set of quantization levels that are evenly spaced across the signal range
- ▶ V_{FS} is the full-scale voltage range of the A/D, $V_{FS} = V_{max} - V_{min}$
- ▶ V_{min} = minimum voltage level of A/D, V_{max} = maximum voltage level of A/D
- ▶ For a system of N bits, the number of quantization levels is given by $L = 2^N$
- ▶ The quantization step, also defined as the system resolution is given by $\Delta = \frac{V_{FS}}{2^N}$



Uniform Quantization

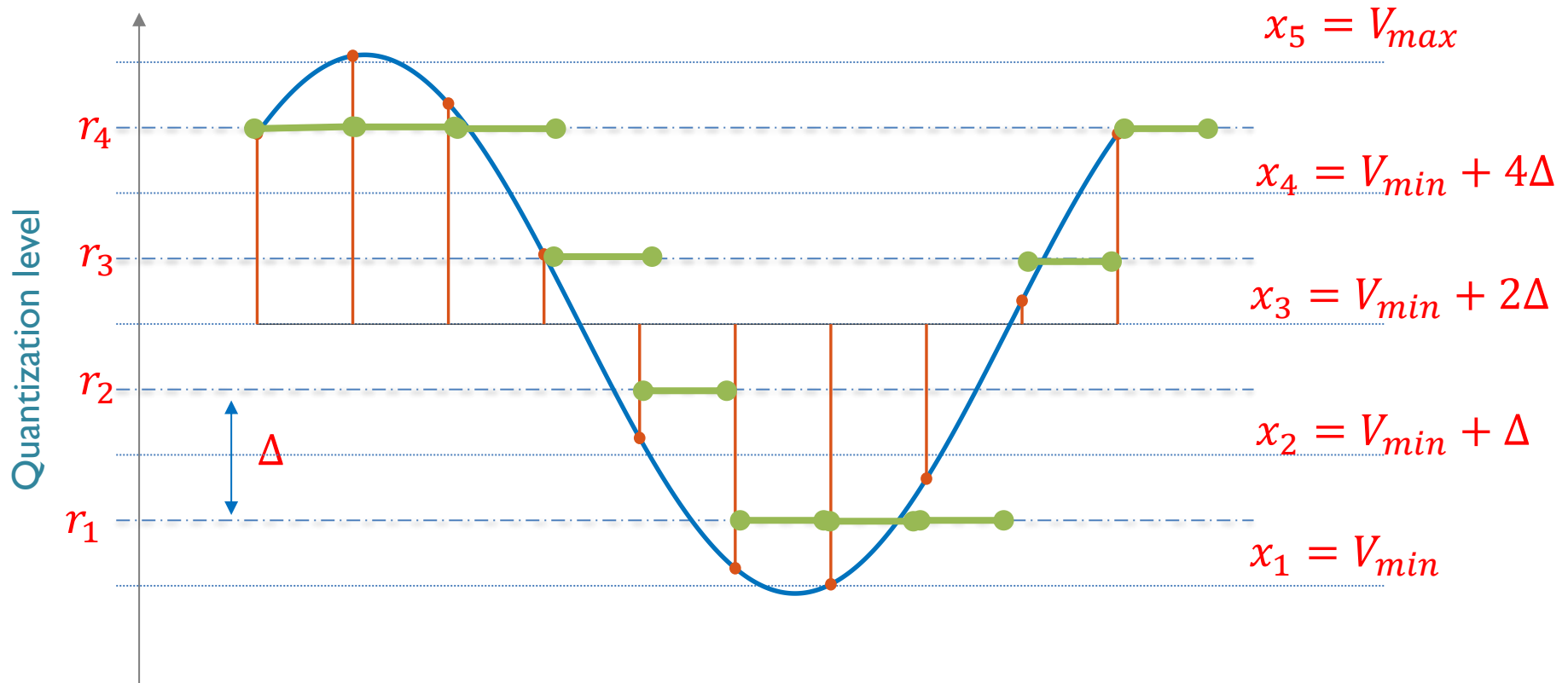


- ▷ $L = 2^N$
- ▷ $\Delta = \frac{V_{FS}}{2^N} = \text{system resolution}$
- ▷ $V_{FS} = V_{max} - V_{min}$
- ▷ V_{min} minimum voltage level of A/D
- ▷ V_{max} = maximum voltage level of A/D
- ▷ If the signal amplitude is between x_k and x_{k+1} , it's mapped into r_k
- ▷ $r_k = V_{min} + \frac{\Delta}{2} + (k - 1)\Delta, k = 1, 2, \dots, L$
- ▷ $x_k = V_{min} + (k - 1)\Delta. k = 1, 2, \dots, L + 1$





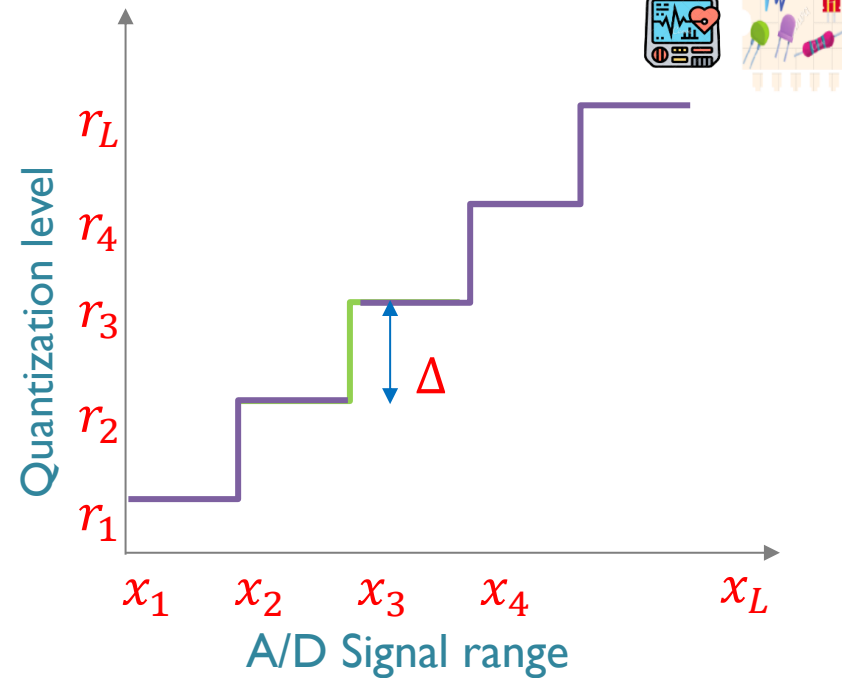
Quantization



$$r_k = \frac{(x_k + x_{k+1})}{2}$$

Quantization Error

- ▶ If the signal falls within the full-scale range of the A/D converter, the maximum absolute value of the quantization error is half the distance between two adjacent quantization levels, q_e is $\frac{\Delta}{2}$
- ▶ $|q_e| \leq \Delta/2$
- ▶ If the signal exceeds the full-scale range, it will be clipped to the nearest quantization level within that range, and the resulting quantization error will depend on the extent of the clipping.



A/D quantizer specifications

- $\Delta = \frac{V_{FS}}{2^N}$ = system resolution
- $L = 2^N$
- $V_{FS} = V_{max} - V_{min}$
- V_{min} minimum voltage level of A/D
- V_{max} = maximum voltage level of A/D

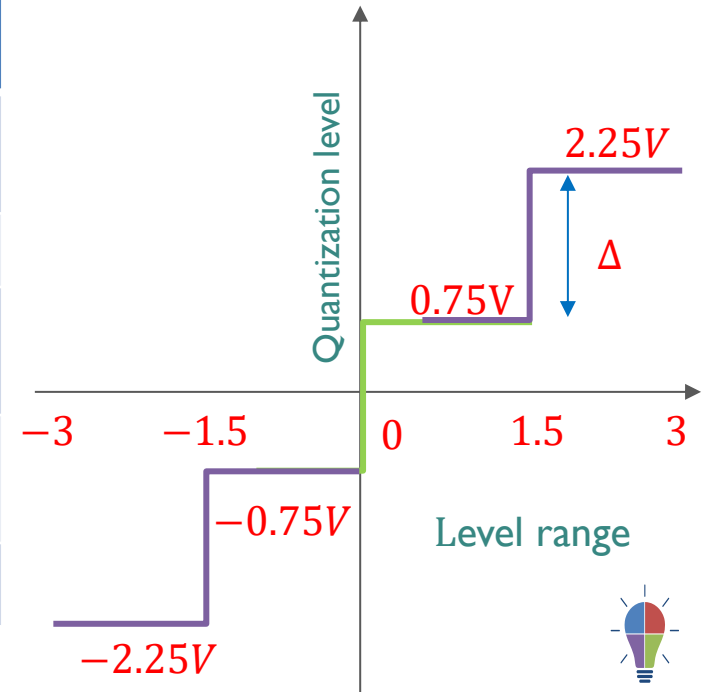




Simple Example

- ▶ Consider a 2-bit system, $N = 2$, $L = 2^N = 2^2 = 4$ Levels
- ▶ Consider the voltage range of the A/D is $\pm 3V$
- ▶ $V_{min} = -3V, V_{max} = 3V$ $V_{FS} = 3 - (-3) = 6V$ $\Delta = \frac{V_{FS}}{2^N} = \frac{6}{4} = 1.5V$
- ▶ $r_k = V_{min} + \frac{\Delta}{2} + (k - 1)\Delta, k = 1, 2, \dots, L$ $x_k = V_{min} + (k - 1)\Delta, k = 1, 2, \dots, L + 1$

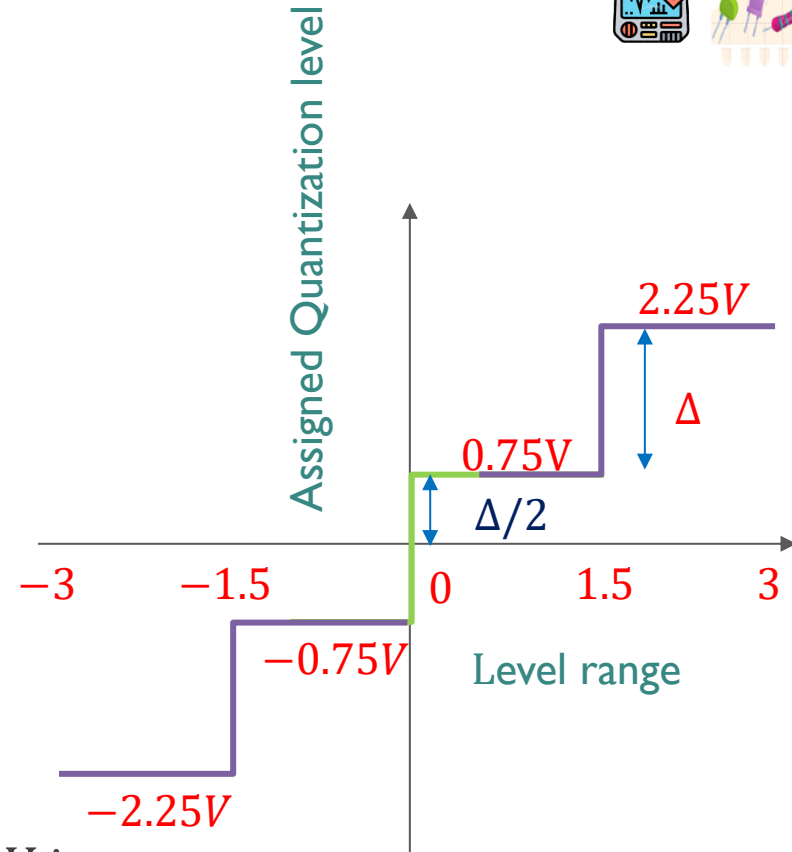
$r_k = r_{k-1} + \Delta$	$x_k = x_{k-1} + \Delta$
$r_1 = V_{min} + \frac{\Delta}{2} = -2.25V$	$x_1 = V_{min} = -3V$
$r_2 = r_1 + 1.5 = -0.75V$	$x_2 = x_1 + 1.5 = -1.5V$
$r_3 = r_2 + 1.5 = 0.75V$	$x_3 = x_2 + 1.5 = 0V$
$r_4 = r_3 + 1.5 = 2.25V$	$x_4 = x_3 + 1.5 = 1.5V$
	$x_5 = x_4 + 1.5 = 3V$





Simple Example, continued

- ▶ 2-bit A/D system, $N = 2, L = 2^N = 2^2 = 4$ Levels
- ▶ The voltage range of the A/D is $\pm 3V$
- ▶ $V_{FS} = 3 - (-3) = 6V$
- ▶ $\Delta = \frac{V_{FS}}{2^N} = \frac{6}{4} = 1.5V$
- ▶ If the signal range is within ± 3 , $q_e = \frac{\Delta}{2} = 0.75V$ is the maximum quantization error.
- ▶ If the signal has a range $\pm 5V$, then the maximum quantization error will be $5V - 2.25V = 2.75V$





Quantization error

- ▶ If the signal range is within the A/D full scale range, $q_e = \frac{\Delta}{2}$ is the maximum quantization error.
- ▶ If the signal exceeds the A/D full scale range, the error will be given by
- ▶

Smallest quantization level

Largest quantization level

$$q_e = V_{\text{signal}} - \left(V_{\text{max}} - \frac{\Delta}{2} \right) \quad \text{for } V_{\text{signal}} > V_{\text{max}}$$
$$q_e = \left(V_{\text{min}} + \frac{\Delta}{2} \right) - V_{\text{signal}} \quad \text{for } V_{\text{signal}} < V_{\text{min}}$$

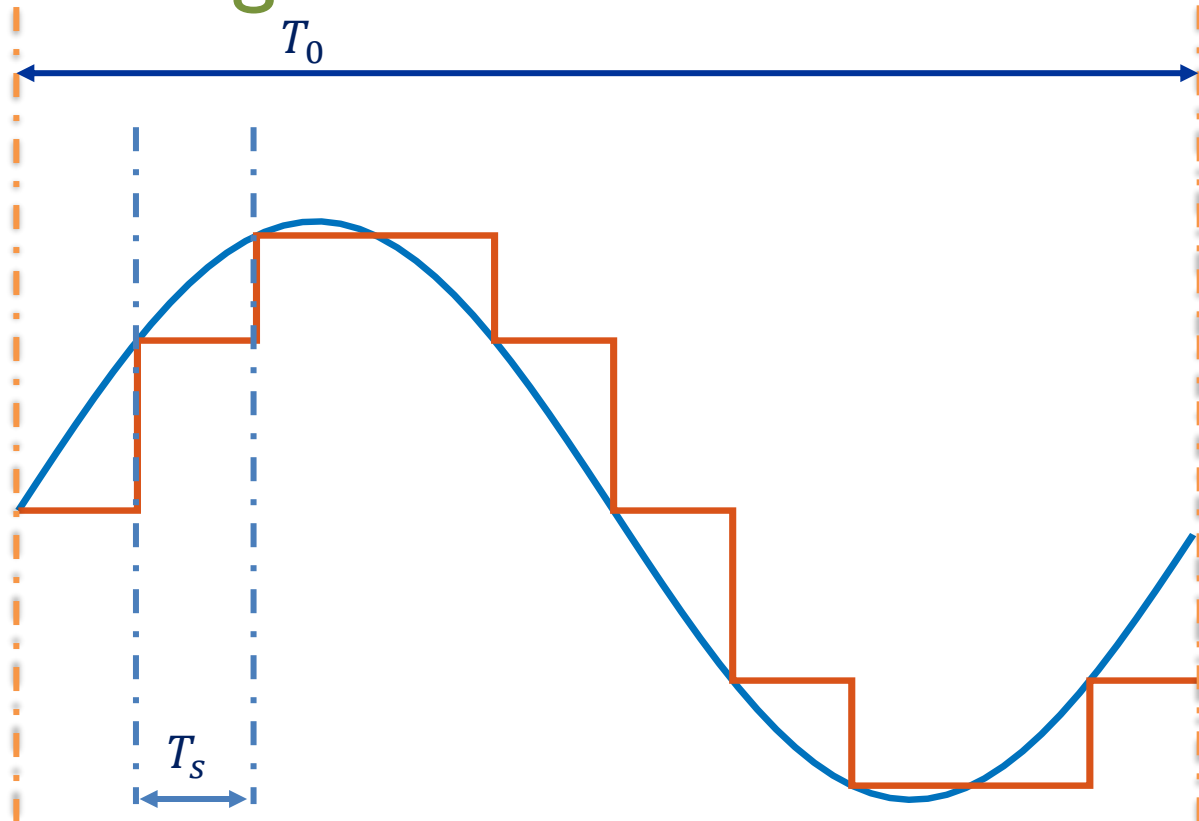
V_{max} is the maximum allowed input level for A/D

V_{min} is the minimum allowed input level for A/D





Stair-case approximation of a sampled signal

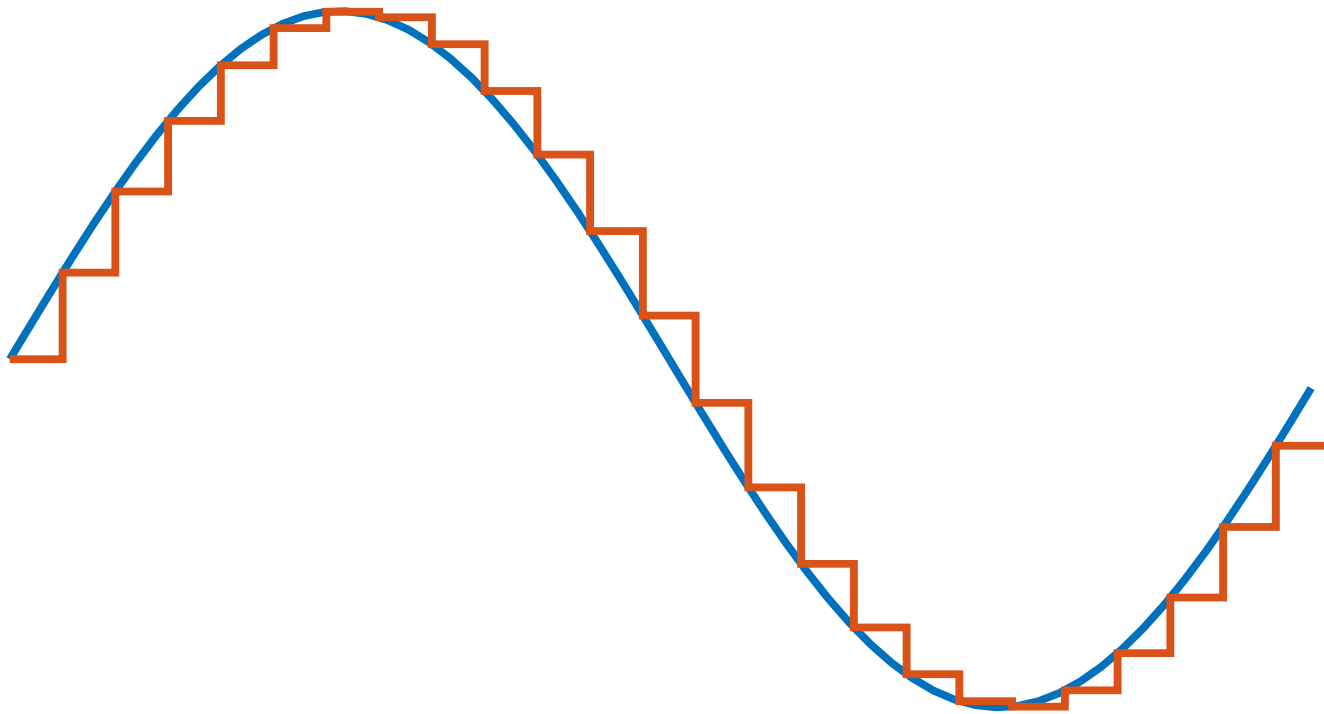


T_s = sampling interval T_0 = period of signal

In the figure shown $f_s = 10f_0$

Note that the difference between adjacent steps in this figure is not representative of the quantization step size; instead, it is a function of the sampling frequency

Stair-case Approximation as the Sampling Rate Increases



In the figure shown $f_s = 25f_0$

Notice that the difference between adjacent steps is becoming smaller as the sampling frequency increases. The quantization step is characteristic of the A/D

$$\Delta = \frac{V_{FS}}{2^N}$$



NI USB-6001

Data Acquisition, DAQ, USB device

- ▶ Analog-to-digital converter (ADC) with 14-bit Resolution
- ▶ Maximum Sample 20 kS/s
- ▶ The NI DAQ device has an input signal range of ± 10 V.
- ▶ Provides eight single-ended analog input (AI) channels, which may also be configured as four differential channels. It also includes two analog output (AO) channels, 13 digital input/output (DIO) channels, and a 32-bit counter.

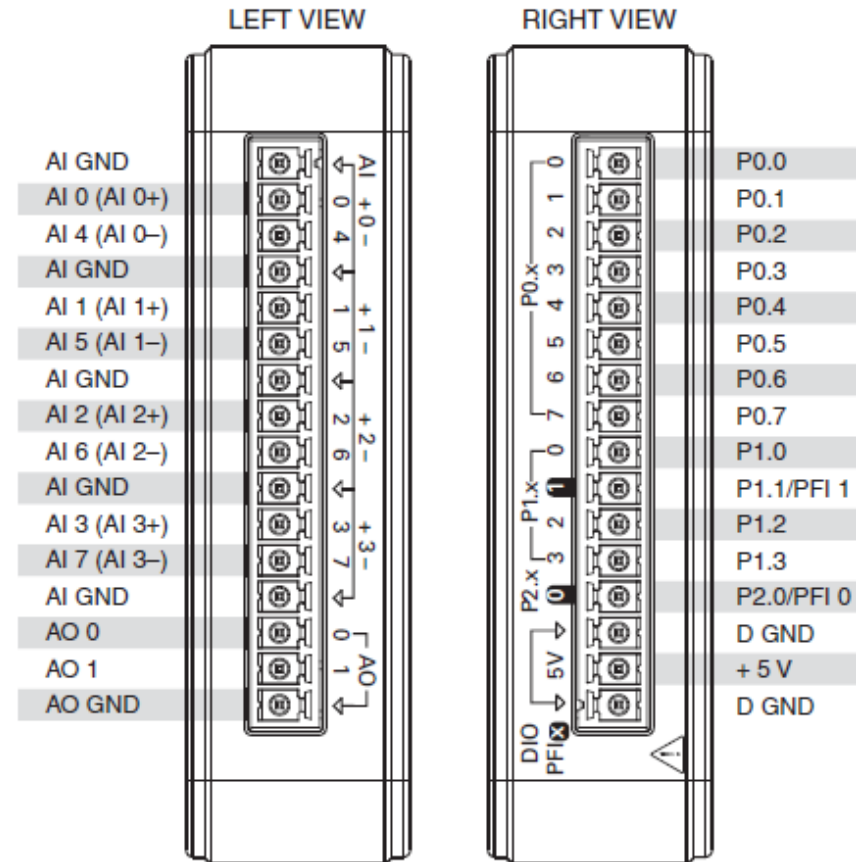


Pinout and signal description



- ▶ **AI GND : Analog Input Ground**—The reference point for single-ended analog input measurements.
- ▶ **AI <0..7> AI GND Input Analog Input Channels 0 to 7**—For single-ended measurements, each signal corresponds to one analog input voltage channel.
- ▶ For differential measurements, AI 0 and AI 4 are the positive and negative inputs of differential analog input channel 0.
- ▶ The following signal pairs also form differential input channels: AI <1,5>, AI <2, 6>, and AI <3, 7>.

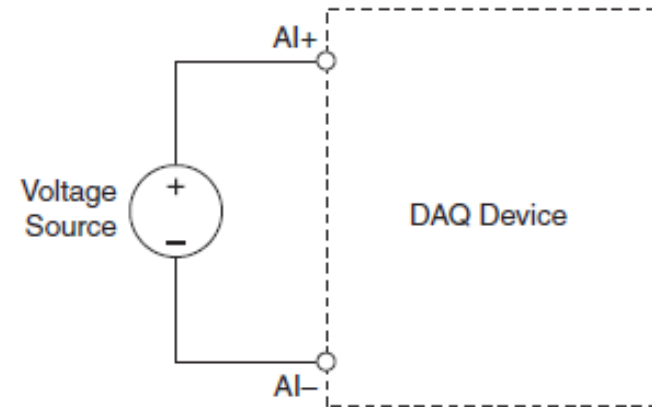
Figure 4. NI USB-6001/6002/6003 Pinout





Taking differential measurement

Figure 6. Connecting a Differential Voltage Signal



- ▶ For differential signals, connect the positive lead of the voltage signal or source to the AI+ terminal, and the negative lead to the AI- terminal.

Figure 7. Example of a Differential ± 10 V Measurement

