

## Fundamentals

**KCL:**  $\sum i_{in} = 0$ , **KVL:**  $\sum v_n = 0$

**Ohm's Law:**  $V = iR$

**Resistances:**  $\frac{1}{R_{eq}} = \sum \frac{1}{R_i}$  (parallel),  $R_{eq} = \sum R_i$  (series)

**Power:**  $P = iV = i^2R = \frac{V^2}{R}$

## Thevenin/Norton Equivalent

$V_{Th}$  = open circuit voltage

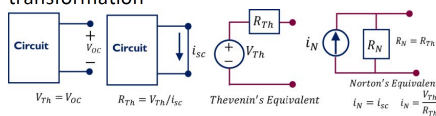
$R_{Th} = V_{th}/i_{sc}$

$R_{Th}$  is also the equivalent resistance seen

between the terminals when all independent sources are set to zero

Norton's Equivalent circuit can be obtained from

Thevenin's equivalent circuit using source transformation



## Voltage and Current Decision

### Voltage Division:

$$v_x = v_s \frac{R_x}{R_1 + R_2 + \dots + R_n}$$

Distributes voltage across resistors in series proportionally to their resistance.

### Current Division:

$$i_x = i_s \frac{R_{eq}}{R_x}$$

where  $R_{eq} = R_1 \parallel R_2 \parallel \dots \parallel R_n$ .

### Current through a branch:

$$i_x = i_s \frac{R_{total}}{R_x + R_{total}}$$

If resistances are combined in parallel and a current source is applied. These principles aid in determining how voltage or current is shared across circuit components.

## Capacitor and Inductor

### Capacitor

$$i = C \frac{dv}{dt} \quad v = \frac{1}{C} \int_{t=t_0}^t i(t) dt + v(t_0)$$

$$P = iv \quad w = \frac{1}{2} C v^2$$

Transient analysis for first order RC, general solution

$$v(t) = V_F + (V_0 - V_F) e^{-\frac{t}{\tau}} \quad t \geq 0$$

$$i(t) = I_0 e^{-\frac{t}{\tau}} \quad t \geq 0^+$$

$$\tau = RC$$

$$\Rightarrow v(0^-) = v(0) = v(0^+) = V_0$$

Sinusoidal Steady State Analysis

$$\tilde{V} = Z\tilde{I} \quad Z = \frac{1}{j\omega C}$$

$$v(t) = V_m \cos(\omega t), \quad i(t) = I_m \cos(\omega t + 90^\circ)$$

Current leads voltage by  $90^\circ$

Capacitors in parallel  $C_{eq} = \Sigma C_i$

Capacitor in series  $\frac{1}{C_{eq}} = \Sigma \frac{1}{C_i}$

### Inductor

$$v = L \frac{di}{dt} \quad i = \frac{1}{L} \int_{t=t_0}^t v(t) dt + i(t_0)$$

$$P = iv \quad w = \frac{1}{2} L i^2$$

Transient analysis for first order RL circuits, general solution

$$v(t) = V_0 e^{-\frac{t}{\tau}} \quad t \geq 0^+$$

$$i(t) = I_F + (I_0 - I_F) e^{-\frac{t}{\tau}} \quad t \geq 0$$

$$\tau = \frac{L}{R}$$

$$i(0^-) = i(0) = i(0^+) = I_0$$

Sinusoidal Steady State Analysis

$$\tilde{V} = Z\tilde{I} \quad Z = j\omega L$$

$$i(t) = I_m \cos(\omega t) \quad v(t) = V_m \cos(\omega t + 90^\circ)$$

Current lags voltage by  $90^\circ$

Inductors in parallel  $\frac{1}{L_{eq}} = \Sigma \frac{1}{L_i}$

Inductors in series  $L_{eq} = \Sigma L_i$

## Phasor Transform

$$v(t) = V_m \cos(\omega t + \theta_v) \rightarrow \tilde{V} = V_m \angle \theta_v = V_m e^{j\theta_v}$$

## Source Transformation



## Sampling and Quantization

Sampling rate  $f_s \geq 2f_{max}$ , minimum sampling rate is called the Nyquist rate

$N$  = number of bits,  $L$  = number of quantization Levels

$\Delta$  = quantization step,  $q_e$  = quantization error

$V_{FS}$  = full scale range of the analog to digital converter, A/D

For a digital system of  $N$  bits, the number of discrete levels  $L = 2^N$

Quantization step  $\Delta = \frac{V_{FS}}{L} = \frac{V_{FS}}{2^N}$

$V_{FS}$  is the full scale range of A/D,  $V_{FS} = V_{max} - V_{min}$

$V_{max}$  is the maximum allowed input level for A/D

$V_{min}$  is the minimum allowed input level for A/D

Quantization error  $q_e \leq \frac{\Delta}{2}$  if signal is within A/D range

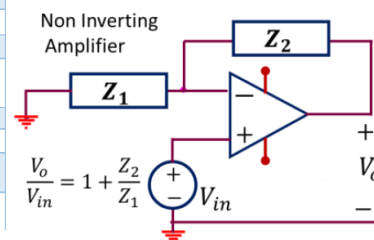
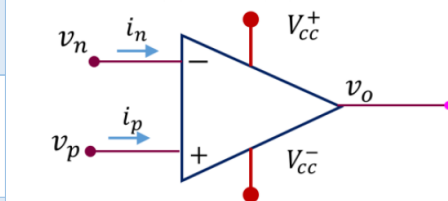
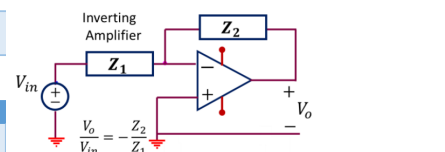
If the signal exceeds the A/D full scale range, the error will be given by

$$q_e = V_{signal} - \left(V_{max} - \frac{\Delta}{2}\right) \quad \text{for } V_{signal} > V_{max}$$

$$q_e = \left(V_{min} + \frac{\Delta}{2}\right) - V_{signal} \quad \text{for } V_{signal} < V_{min}$$

## Ideal Op-Amps

- $i_n = 0, i_p = 0 \quad v_n = v_p$
- Infinite resistance looking into the inverting or non-inverting terminal
- Infinite bandwidth
- For linear mode of operation  $V_{cc}^- \leq v_o \leq V_{cc}^+$



## Transfer Functions and the frequency response of linear systems

$$H(j\omega) = \frac{V_o}{V_{in}}$$

$$v_{in}(t) = A \cos(\omega_0 t) \xrightarrow{H(j\omega)} v_o(t) = A |H(j\omega_0)| \cos(\omega_0 t + \phi)$$

$$H(j\omega_0) = |H(j\omega_0)| \angle \phi(\omega_0)$$