

AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Burden

- today:
- course logistics: flipped lectures, (a)synchronous participation
 - Canvas: schedule, Syllabus, Announcements, Assignments
 - first assignment (HWO) due this Fri Apr 1
 - first lectures (week 1) available
 - questions / office hours

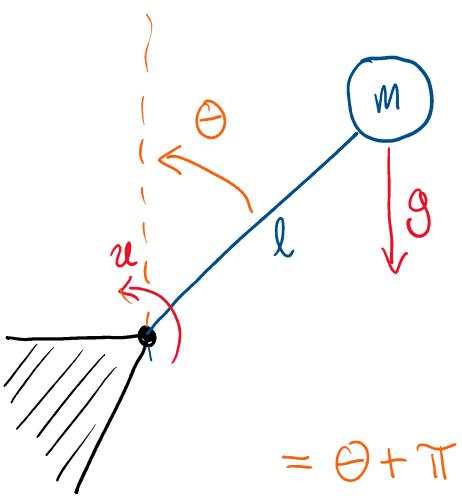
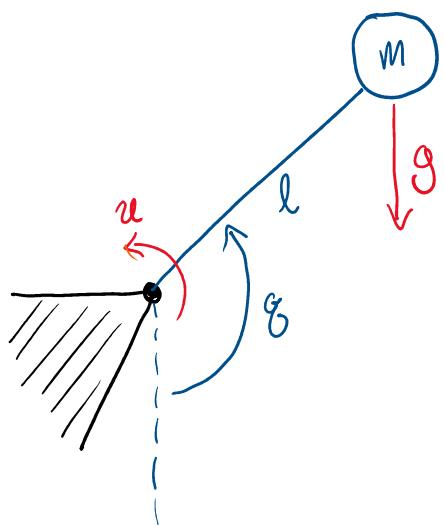
today: course logistics, Canvas, etc

Thu Mar 31 HwO

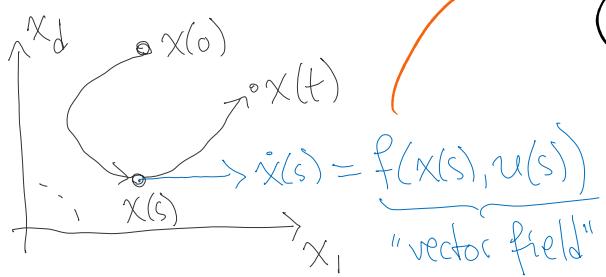
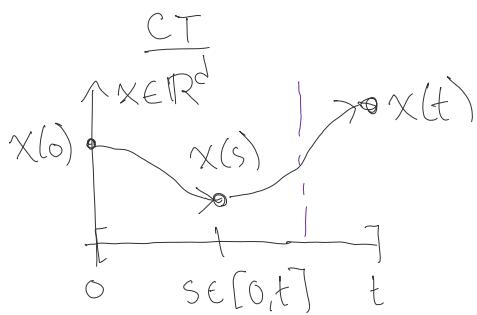
week 1 lectures

questions / office hours

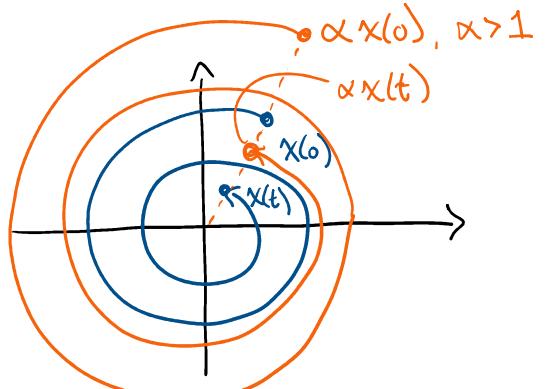
ex: inverted pendulum $ml^2 \ddot{\theta} = -mg \sin \theta + u - b\dot{\theta}$



$$\dot{x} = Ax \Leftrightarrow x(t) = e^{At} x(0)$$



what if f is linear?
(eg LTI?)



Linear systems:

$$z(0) = x(0) + y(0)$$

$$\downarrow$$

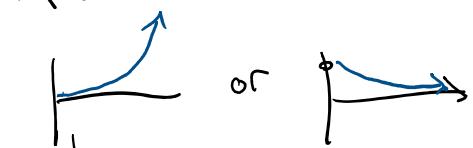
$$z(t) = x(t) + y(t)$$

Q: what can the flow of an LTI system do?

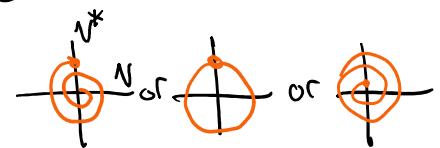
$$\dot{x} = Ax: \text{ if } Av = \lambda v \text{ then } x(t) = v e^{\lambda t}$$

$\alpha \in \mathbb{R}$

so there are 2 cases: 1: if $\lambda \in \mathbb{R}$ then $v \in \mathbb{R}^d$ and

$e^{\lambda t}$ looks like 

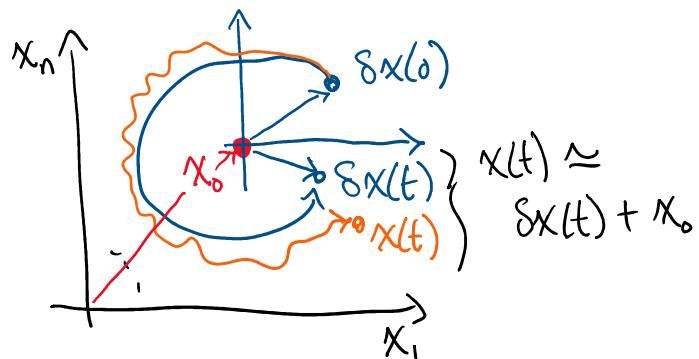
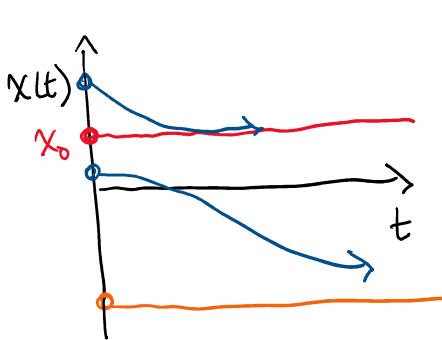
2: if $\lambda \in \mathbb{C}$ then $v \in \mathbb{C}^d$ and $Av^* = \lambda^* v^*$

so $e^{\lambda t}$ looks like 

$$(NL) \quad \dot{x} = f(x, u), \quad x \in \mathbb{R}^n, \quad u \in \mathbb{R}^m$$

def: $(x_0, u_0) \in \mathbb{R}^n \times \mathbb{R}^m$ is an equilibrium for (NL) if $f(x_0, u_0) = 0$

↳ if $x(0) = x_0, u(t) = u_0$, then $x(t) = x_0$



* letting $\delta \dot{x} = A \cdot \delta x + B \cdot \delta u$ where $A = \partial_x f(x_0, u_0)$, $B = \partial_u f(x_0, u_0)$.

then $x \approx x_0 + \delta x$ if $u = u_0 + \delta u$

this approx. gets better as $\|\delta x\|$ gets smaller

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TODO: post notes

consider $\dot{x} = Ax$, $A \in \mathbb{R}^{n \times n}$

fact: if A has n distinct eigenvalues $\lambda_1, \dots, \lambda_n \in \mathbb{C}$

with corresponding eigenvectors $v_1, \dots, v_n \in \mathbb{C}^n$

$$\text{then } D = VAV^{-1} = \begin{bmatrix} \lambda_1 & & \\ & \ddots & \\ & & \lambda_n \end{bmatrix} \Rightarrow A = V^{-1}DV$$

$$\text{where } V = [v_1, \dots, v_n]$$

$$\text{so } e^{At} = e^{V^{-1}DVt} = V^{-1}e^{Dt}V$$

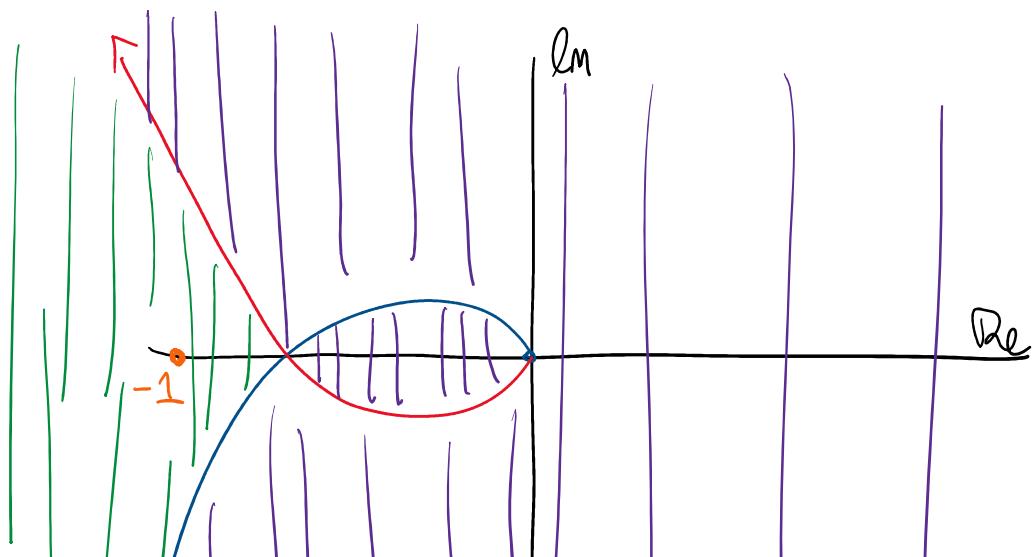
A is
diagonalizable

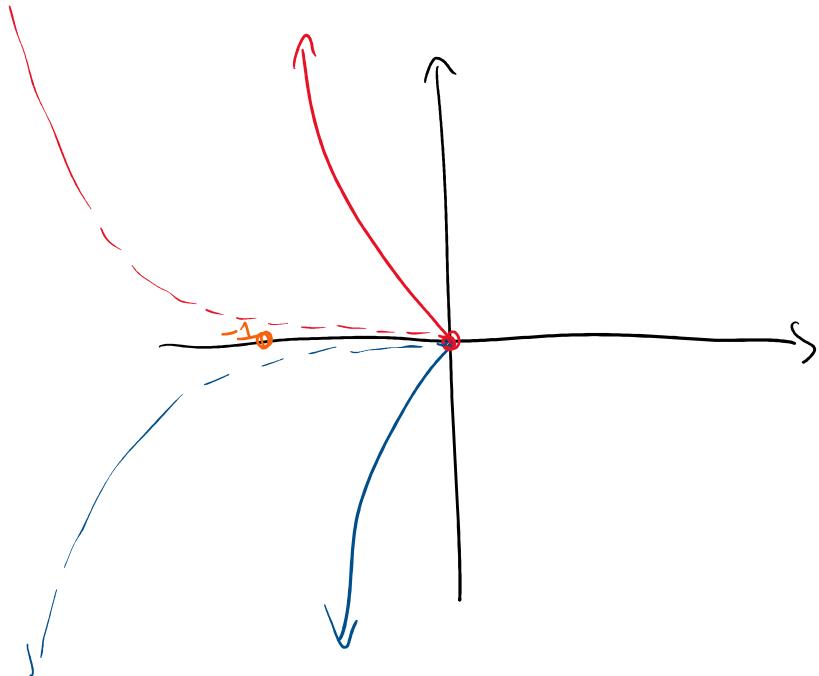
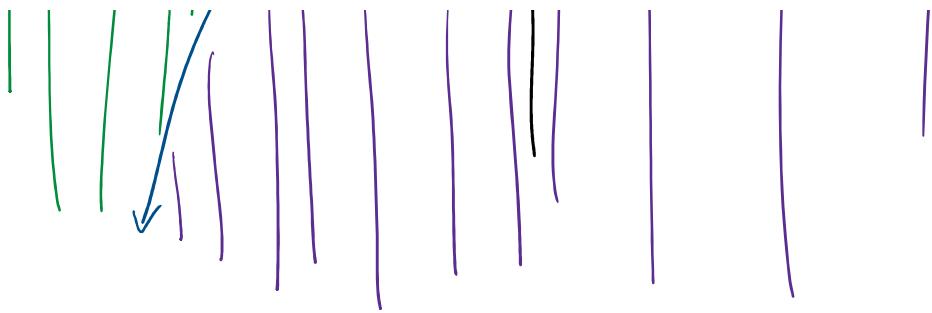
$$\text{so } e^{At} = e^{V^{-1}DVt} = V e^{Dt} V$$

$$\begin{bmatrix} e^{\lambda_1 t} & & \\ & \ddots & 0 \\ 0 & \ddots & e^{\lambda_n t} \end{bmatrix}$$

more generally, for any A there exists T^{-1} s.t.

$$J = T A T^{-1} = \begin{bmatrix} \Lambda_1 & 0 \\ 0 & \ddots & 0 \\ 0 & \ddots & \Lambda_n \end{bmatrix}, \quad \Lambda_i = \begin{bmatrix} \lambda_{i1} & & 0 \\ & \ddots & \\ 0 & \ddots & \lambda_{ii} \end{bmatrix}$$





$$D_2 \phi(s+1, \xi) = D_\xi [f(\phi(s, \xi))]$$

$$= Df(\phi(s, \xi)) \cdot D_2 \phi(s, \xi)$$

$$\xi = x_i$$

$$\phi = \phi_i$$

$$\begin{aligned} \phi: \mathbb{R} \times \mathbb{R}^d &\rightarrow \mathbb{R}^d & f: \mathbb{R}^d &\rightarrow \mathbb{R}^d \\ : (t, x_0) &\mapsto x_t & : x &\mapsto x^t \end{aligned}$$

$$\begin{aligned} &\phi(0, x) & \phi(1, x) & \phi(2, x) \\ &|| &|| &|| \\ &x \mapsto f(x) &\mapsto f(f(x)) &\mapsto \\ &&& \curvearrowright Df(x): \mathbb{R}^d \rightarrow \mathbb{R}^d \text{ linear} \\ &&& \in \mathbb{R}^{d \times d} \end{aligned}$$

$$\phi(s+1, x) = f(\phi(s, x))$$

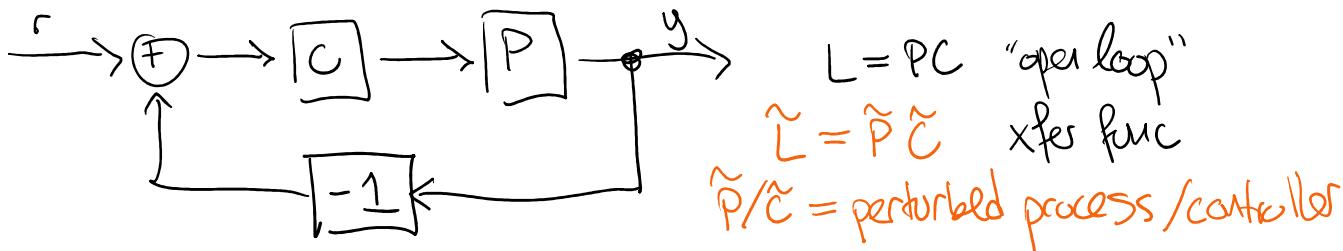
$$D_x [\phi(s+1, x)] = D_x [f(\phi(s, x))]$$

$$\begin{aligned} D_x[\phi(s+t, x)] &= D_x[t(\phi(s, x))] \\ &= Df \circ D_2 \phi \\ &= Df(\phi(s, x)) \cdot D_2 \phi(s, x) \end{aligned}$$

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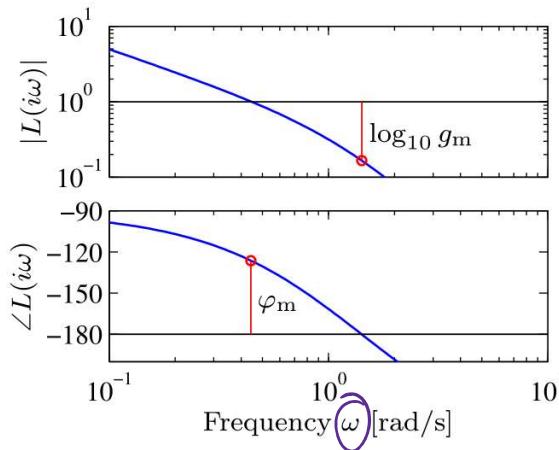
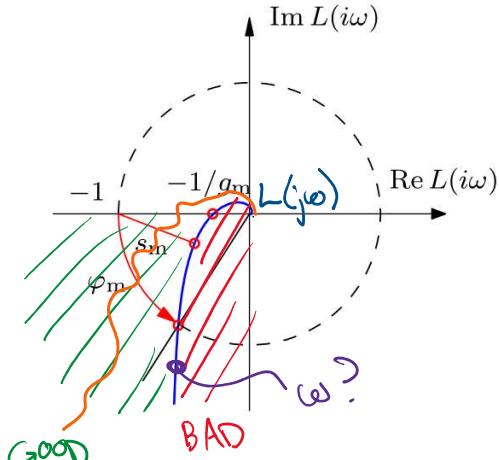
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 - questions / office hours

TODO: link notebooks / point to Python intro



* Nyquist tells me to look at $\Omega = \{L(j\omega) : \omega \in \mathbb{R}\} \subset \mathbb{C}$

↳ its relation to $-1 \in \mathbb{C}$



L strictly proper:

$$L(s) = \frac{b s^m + \dots}{a s^n + \dots}$$

$n > m$

$$\lim_{|\omega| \rightarrow \infty} |L(j\omega)| = 0$$

 // 

Frequency ω [rad/s]

Tue Apr 12

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- link notebooks / point to Python intro

AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Borden

- today:
- course logistics, Canvas, etc Thursday
 - HW 1 self-assessment - due next ~~Monday~~
 - HW 2 - due this Friday
 - week 3 lectures
 - questions / office hours

HW1 solution \rightsquigarrow Canvas 2π in p1(a)

TODO:

- link notebooks / point to Python intro
- ECE Colloq
- Robotics Colloquium
- Northwest Robotics Symp.

* how does continuous-time LQR relate to discrete-time?

$$\dot{x} = Ax + Bu \quad \rightsquigarrow \quad x^+ = e^{\Delta A} x = A_\Delta x, \quad \Delta > 0 \text{ timestep}$$

$$C = \int_0^{k \cdot \Delta} x^T Q x + u^T R u \quad \rightsquigarrow \quad \sum_{l=0}^{k-1} x_l^T Q_\Delta x_l + u_l^T R_\Delta u_l = C_\Delta$$

$\left. \begin{matrix} \\ \end{matrix} \right\} \text{LQR}$ $\left. \begin{matrix} \\ \end{matrix} \right\}$

$$u^* = -K_\Delta x \quad u^* = -K_\Delta x$$

then: $\lim_{\Delta \rightarrow 0} K_\Delta = K, \quad \lim_{\Delta \rightarrow 0} C_\Delta = C$

assuming $A_\Delta = A + \Delta I/\Delta$

$$\text{assuming } A_\Delta = A + O(\Delta)$$

$$Q_\Delta = Q + O(\Delta)$$

$$R_\Delta = R + O(\Delta)$$

* compare finite-horizon vs infinite-horizon

finite-horiz: $\dot{x}(t) \text{ or } x(t+1) = A_t x(t) + B_t u(t)$

cost is $\int_0^T \text{ or } \sum_{t=0}^T x_t^T Q_t x_t + u_t^T R_t u_t$

\rightsquigarrow opt ctrl is $u_t^* = -K_t x_t$

∞ -horiz: $\dot{x}/x^+ = Ax + Bu$

cost $\int_0^\infty \text{ or } \sum_{t=0}^\infty x_t^T Q x_t + u_t^T R u_t$

\rightsquigarrow opt ctrl is $u^* = -K x$

$$K_S = (B_S^T P_{S+1} B_S + R_S)^{-1} B_S^T P_{S+1} A_c$$

$$P_S = (A_S - B_S K_S)^T P_{S+1} (A_S - B_S K_S) + K_S^T R_S K_S + Q_S$$

$$u_S^* = -K_S x_S, \quad v_S^* = \frac{1}{2} x_S^T P_S x_S$$

$$K = (B^T P B + R)^{-1} B P A$$

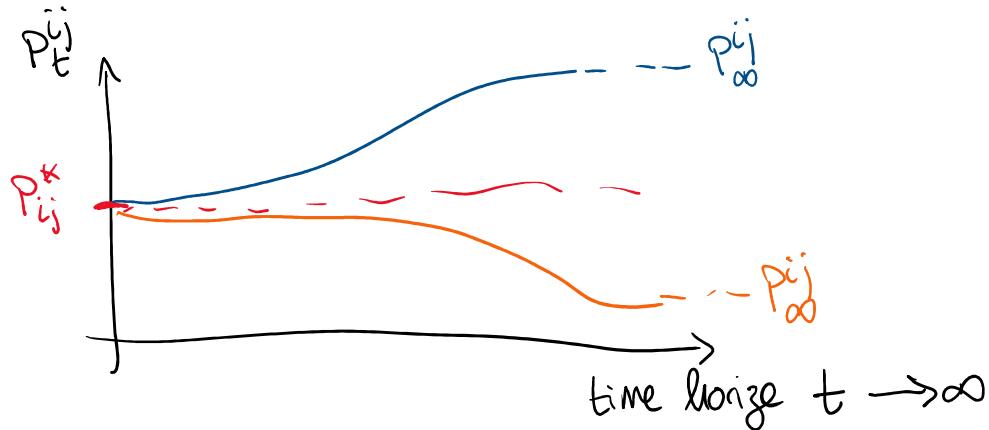
$$\rightsquigarrow (\text{LTI}) \quad P^- = (A - BK)^T P (A - BK) + K R K + Q$$

$$\hookrightarrow (\text{LTI}) \quad P^- = (A - BK)^T P (A - BK) + K R K + Q$$

* not linear, but: 1°. preserves symmetry & definiteness

2°. has unique GES equilibrium P^*

* P^* is solution of ∞ -horz LTI LQR!



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today: course logistics, Canvas, etc

exam 1 next week

HW 2 self-assessment - due next Monday

HW 3 - due this Friday

week 4 lectures

questions / office hours

HW1 solution \rightsquigarrow Canvas 2π in p1(a)

TODO: link notebooks / point to Python intro

ECE Colloq Robotics Colloquium Northwest Robotics Symp.

$$w \in \mathbb{R}^2, z \in \mathbb{R}^2$$

$$\therefore T_{wz} = -\frac{1}{s+1} \begin{bmatrix} 1 & a \\ -a & 1 \end{bmatrix} = \begin{bmatrix} T_{w_1 z_1} & T_{w_1 z_2} \\ T_{w_2 z_1} & T_{w_2 z_2} \end{bmatrix}$$

This implies that $T_{w_1 z_1} = T_{w_2 z_2} = -\frac{1}{s+1}$.

$$w = T_{wz} \cdot z$$



$$\begin{bmatrix} w_1 \\ w_2 \end{bmatrix} = \begin{bmatrix} T_{w_1 z_1} \cdot z_1 + T_{w_1 z_2} \cdot z_2 \\ T_{w_2 z_1} \cdot z_1 + T_{w_2 z_2} \cdot z_2 \end{bmatrix}$$

multiple notions of stability:

$$\dot{x} = Ax + Bu \quad \xrightarrow{\text{output stability}} \quad y = Cx + Du$$

↓ state space stability

$$\text{spec } A \subset \mathbb{C}_0^-$$



asymptotically
exponentially stable
internally stable \Rightarrow BIBO stable

$$u \rightarrow \boxed{T} \quad y = T(u)$$

* $\max_{\substack{u \neq 0 \\ u \in L_2}} \frac{\|T(u)\|}{\|u\|} < \infty$

$$u: \mathbb{R} \rightarrow \mathbb{R}^m: \|u\|_2^2 = \int \|u(t)\|_2^2 dt$$

AA/ECE/ME 548 Linear Multivariable Control) Sp22 Prof Burden

- today:
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 - exam 1 next week
 - HW 2 self-assessment - due next Monday
 - HW 3 - due this Friday
 - week 4 lectures
 - questions / office hours probably I need to remove 2π from HW
 - HW1 solution \rightarrow Canvas 2π in p1(a)
- TODO:
- link notebooks / point to Python intro
 - ECE Colloq
 - Robotics Colloquium
 - Northwest Robotics Symp.
 - HW 2 1(e) explanation

Consider the following cost function of a scalar decision variable $u \in \mathbb{R}$:

$$J(u) = \frac{u^6}{6} - \frac{7u^5}{5} + \frac{17u^4}{4} - \frac{17u^3}{3} + 3u^2.$$

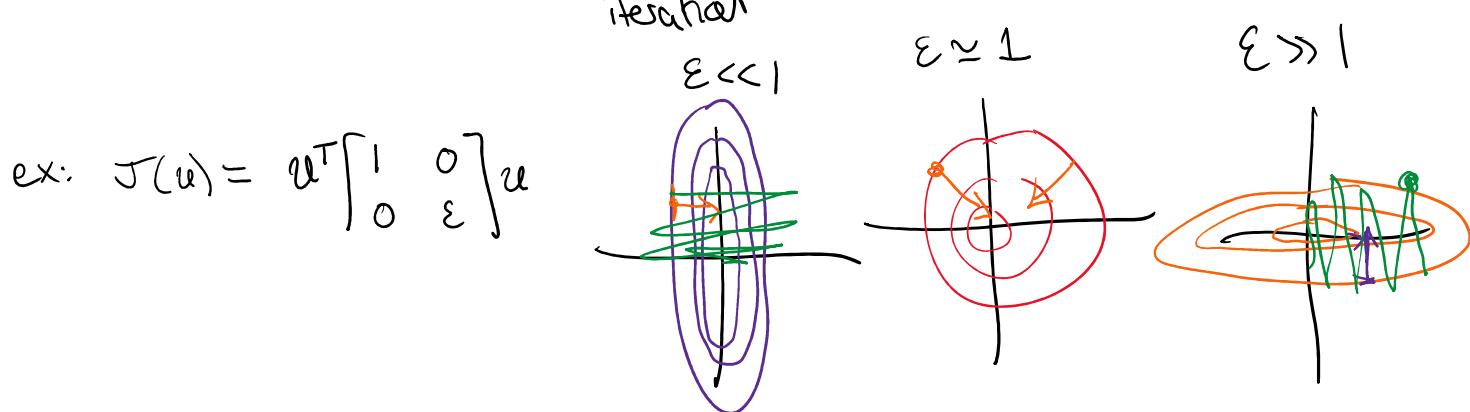
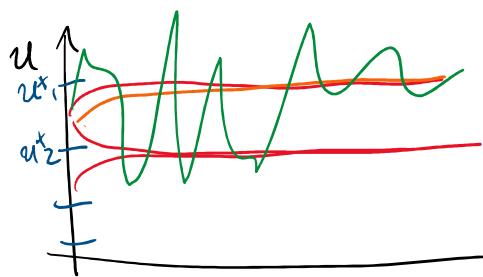
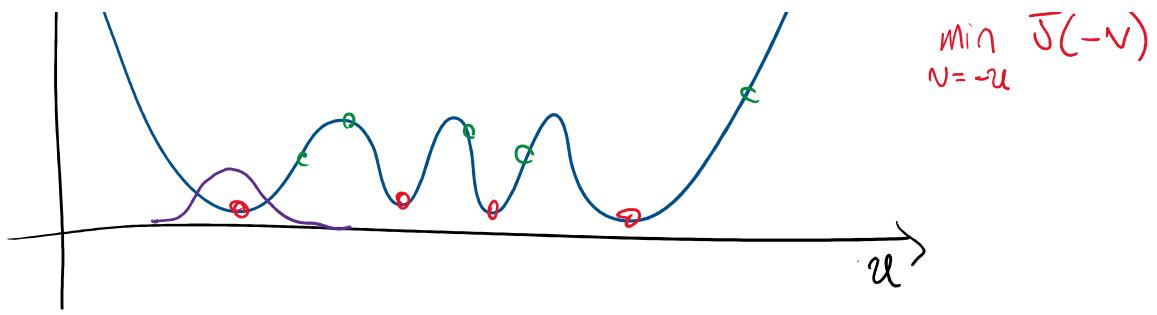
(a) Plot $J(u)$, $DJ(u)$, and $D^2J(u)$ versus u ; use the subplot(3,1,n) for $n = 1, 2, 3$ to align the u -axes of the three plots.

(b) Determine all local minima of J and the corresponding minimizing u .

(c) Run the gradient descent iteration $u^+ = u - \alpha DJ(u)$ starting from multiple initial u 's and with multiple values of the parameter $\alpha > 0$. Describe all of the outcomes (i.e. asymptotic behavior of the iteration) you observe and provide plots that illustrate these outcomes.



Q: do our curves change if we consider
 $\min_{v=-u} J(-v)$



Tue Apr 26

Tuesday, April 26, 2022 12:57 PM

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today: course logistics, Canvas, etc

exam 1

Hw 3 self-assessment - due next Monday

questions / office hours

TODO: link notebooks / point to Python intro

ECE Colloq Robotics Colloquium Northwest Robotics Symp.

Hw 2 1(e) explanation

Thu Apr 28

AA/ECE/ME 548 Linear Multivariable Control) Sp22 Prof Burden

- today:
- course logistics, Canvas, etc
 - exam 1
 - HW 3 self-assessment - due Monday after next
 - questions / office hours

AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Burden

- today:
- course logistics, Canvas, etc
 - exam 1 - solution date, working on grading
 - HW 3 self-assessment - due next Monday
 - HW 4 - due this Friday
 - week 6 lectures
 - questions / office hours

* connection between stability \nexists optimal control
 i.e. Lyapunov i.e. LQR

consider $\min \int_0^\infty x^T Q x + u^T R u \quad \text{s.t. } \dot{x}/x^+ = Ax + Bu$

* know that optimal $u^* = -Kx$ where $K = R^{-1}B^T P$ (\dot{x})
 or $K = (B^T P B + R)^{-1} B^T P A$ (x^+)

where $P = P^T > 0$ solves a Riccati equation

fact: closed-loop dynamics $\dot{x}/x^+ = Ax + Bu = (A - BK)x$
 are (exponentially) stable and $v(x) = \underbrace{x^T P x}_{\sim \sim \dots \sim \sim}$ is a Lyapunov function

are (exponentially) stable and $v(x) = x^T P x$ is a Lyapunov function
 * also the (optimal) cost-to-go

$$\text{ex: } \dot{y} = u \leftrightarrow x = \begin{bmatrix} y \\ \dot{y} \end{bmatrix} \Rightarrow \dot{x} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} y \\ \dot{y} \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \\ = Ax + Bu$$

$$\text{so } u = -Kx = -[k_p \ k_D] \begin{bmatrix} y \\ \dot{y} \end{bmatrix}$$

* if we choose K to $\min \int_0^\infty x^T Q x + u^T R u$

then we've synthesized the "optimal" PD controller

$$\text{ex: } \dot{y} = u \leftrightarrow x = \begin{bmatrix} y \\ \dot{y} \\ \ddot{y} \end{bmatrix} \Rightarrow \dot{x} = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} y \\ \dot{y} \\ \ddot{y} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u$$

$$\text{so } u = -Kx = -[k_p \ k_I \ k_D] \begin{bmatrix} y \\ \dot{y} \\ \ddot{y} \end{bmatrix}$$

* if we choose K to $\min \int_0^\infty x^T Q x + u^T R u$

then we've synthesized the "optimal" PID controller
 (alternatively of Zeigler-Nichols)

given nonzero time-varying reference r to track,

might want to choose K to min $\int_0^\infty g \|\dot{r} - \dot{y}\|^2 + r \|u\|^2$

but this problem doesn't have a solution!

$$\text{let } e = r - y$$

$$= (r - y)^T \cdot g \cdot (r - y)$$
$$= \begin{bmatrix} r - y \\ \ddot{y} \end{bmatrix}^T \begin{bmatrix} g & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} r - y \\ \ddot{y} \end{bmatrix}$$

$$\ddot{e} = \ddot{r} - u$$

$$\dot{e} = e$$

$$\leftrightarrow x = \begin{bmatrix} e \\ \dot{e} \end{bmatrix} \Rightarrow \dot{x} = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} e \\ \dot{e} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ -1 \end{bmatrix} u$$
$$+ \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} \ddot{r}$$

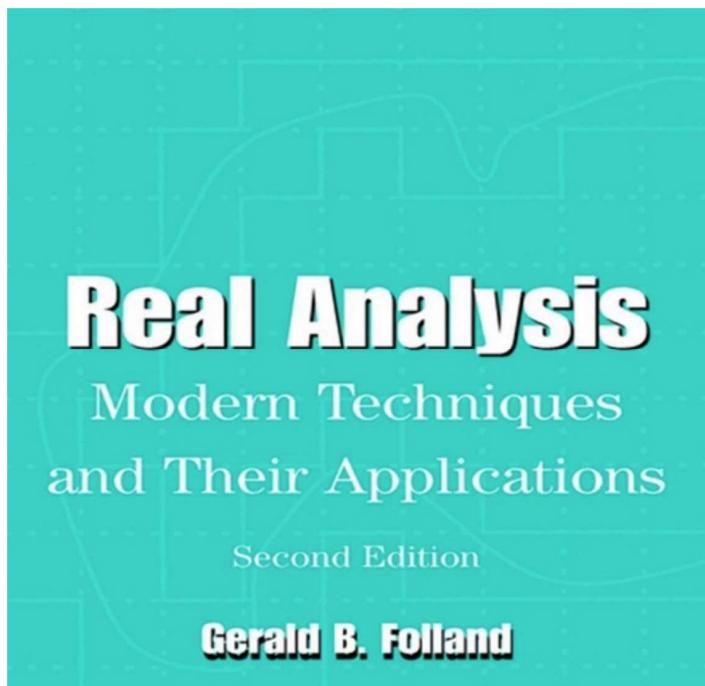
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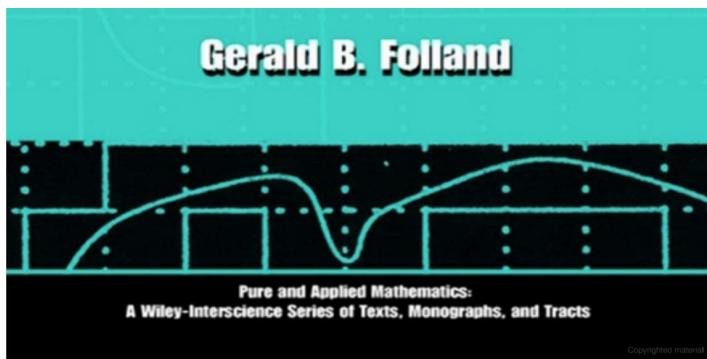
AA/ECE/ME 548 Linear Multivariable Control) sp22 Prof Burden

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TODO: class notes (Tue & today)

* real analysis textbook - if you know measure theory &
want to learn probability theory





AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Burden

today: course logistics, Canvas, etc

HW 4 self-assessment - due next Monday

HW 6 - due this Friday

week 7 lectures

questions / office hours

exam 1 solution @ 2pm today

given rv $x: \Omega \rightarrow \mathbb{R}^n$

define expectation $E[x] = \begin{bmatrix} E[x_1] \\ \vdots \\ E[x_n] \end{bmatrix} \in \mathbb{R}^n$

covariance $\text{Cov}[x] = E[(\underbrace{x - E[x]}_{\in \mathbb{R}^{n \times 1}})(\underbrace{x - E[x]}_{\in \mathbb{R}^{1 \times n}})^T] \in \mathbb{R}^{n \times n}$

$$= \left[\text{Cov}(x_i, x_j) \right]_{i,j=1}^n$$

$\text{diag}(\text{Cov}[x]) = (\text{Var}(x_1), \text{Var}(x_2), \dots, \text{Var}(x_n))$

$$\text{Cov}(a, b) = E[(a - E[a]) \cdot (b - E[b])]$$

$$x_1 = Ax_0 + Bu_0 + \varepsilon_0, \quad x_0 \notin u_0 \text{ deterministic}$$
$$\varepsilon_0 \sim \mathcal{N}(d, \Delta)$$

$$\text{then } x_1 \sim \mathcal{N}(Ax_0 + Bu_0 + d, \Delta)$$

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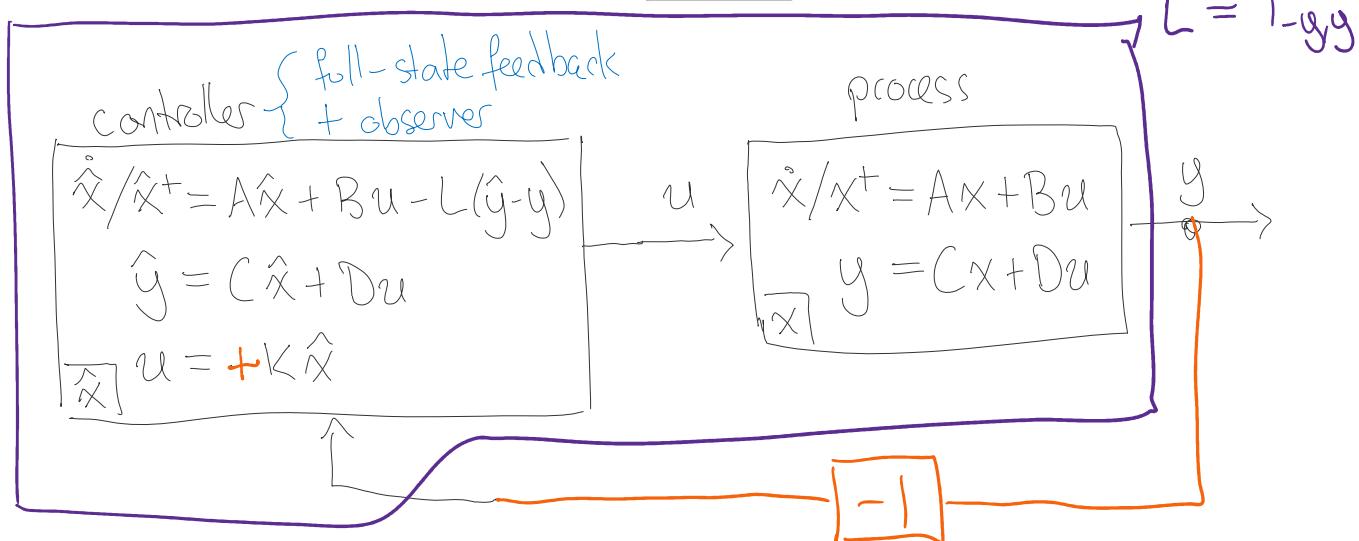
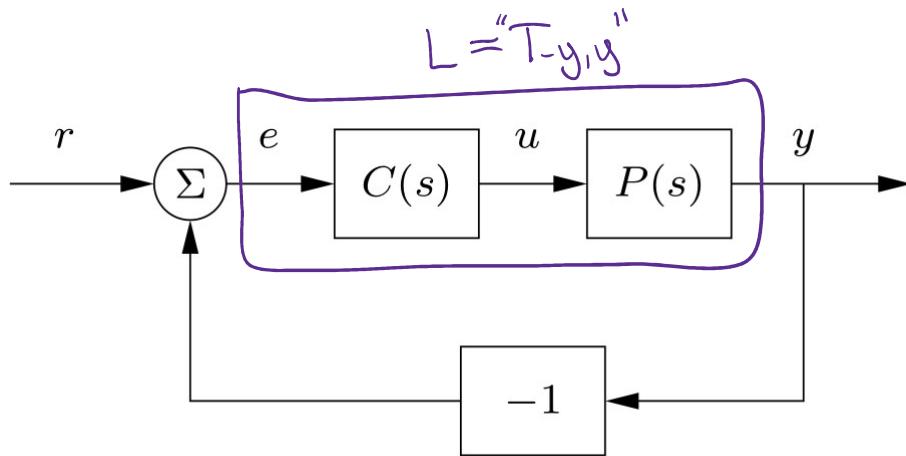
exam 1 solution

HW 4 self-assessment - due next Monday

HW 6 - due this Friday

week 7 lectures

questions / office hours





$$\begin{bmatrix} \hat{x} \\ e \end{bmatrix}^+ = \begin{bmatrix} -BK\hat{x} & = -Ce \\ A\hat{x} + Bu - L(\hat{y} - y) & \\ (A - LC)e & \end{bmatrix} = \begin{bmatrix} (A - BK) & | & LC \\ -\bar{O} & | & -\bar{O} \\ 0 & | & (A - LC) \end{bmatrix} \begin{bmatrix} \hat{x} \\ e \end{bmatrix}$$

$$e = x - \hat{x}$$

$$y = Cx = C(e - \hat{x}) = [-c \ c] \begin{bmatrix} \hat{x} \\ e \end{bmatrix} = \bar{A} \begin{bmatrix} \hat{x} \\ e \end{bmatrix}$$

$L = "T_{-y,y}"$

$$\begin{bmatrix} \hat{x} \\ e \end{bmatrix}^+ = \begin{bmatrix} \bar{A} & \\ (A - BK) & | & LC \\ -\bar{O} & | & -\bar{O} \\ 0 & | & (A - LC) \end{bmatrix} \begin{bmatrix} \hat{x} \\ e \end{bmatrix} + \begin{bmatrix} -L \\ 0 \end{bmatrix} (-y)$$

$y = \underbrace{[-c \ c]}_{\bar{C}} \begin{bmatrix} \hat{x} \\ e \end{bmatrix} \quad \bar{D} = 0$

? need to check...

$$x_{st1} = A_s x_s + B_s u_s + \delta_s$$

$$y_s = C_s x_s + D_s u_s + \gamma_s$$

$$\hat{x}_{st1} = TBD$$

$$\hat{y}_s = C_s \hat{x}_s + D_s u_s$$

$$\min \sum_{s=1}^t (\hat{x}_s - x_s)^T Q_s^{-1} (\hat{x}_s - x_s) + (\hat{y}_s - y_s)^T R_s^{-1} (\hat{y}_s - y_s)$$

$$\min \sum_{s=0}^k (\hat{x}_s - x_s)^\top Q_s^{-1} (\hat{x}_s - x_s) + (\hat{y} - y)^\top R_s^{-1} (\hat{y} - y)$$

consistency w/ process model consistency w/ observations

* \hat{x}^*

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today: course logistics, Canvas, etc

survey about class experience - anonymous / ungraded

HW 6 self-assessment - due next Monday

HW 7 - due this Friday

week 8 lectures

questions / office hours

* won't meet on Thu this week (only) - instead 9:30-11a Fri

$$\text{Cov}[x] = E\left[\underbrace{(x - E[x])(x - E[x])^T}_{=: X}\right] \in \mathbb{R}^{n \times n}$$

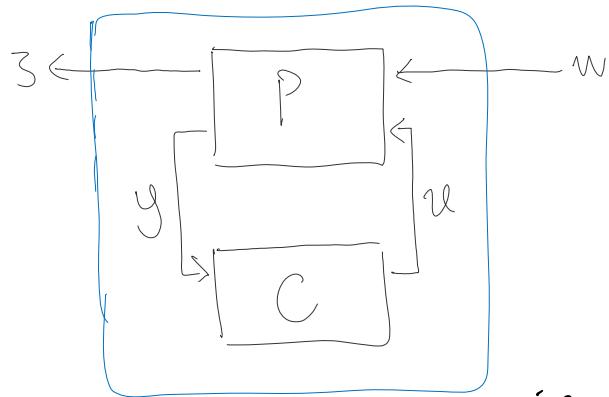
$=: X$, so $X: \Omega \rightarrow \mathbb{R}^{n \times n}$ is a rv
 \uparrow
 \mathbb{R}^{n^2}

$$x_{ij} = (x_i - E[x_i])(x_j - E[x_j])$$

$$E[x] = [E[x_{ij}]]_{i,j=1}^n$$

AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Borden

- today:
- course logistics, Canvas, etc
 - HW 6 self-assessment - due next Monday
 - HW 7 - due this Friday
 - week 8 lectures
 - questions / office hours



Q: what is T_{zw} ?

$$u = Cy$$

$$C = T_{uy} \text{ i.e. } \cancel{y = C \cdot u}$$

$$P = \begin{bmatrix} P_{11} & P_{12} \\ P_{21} & P_{22} \end{bmatrix} = \begin{bmatrix} P_{3w} & P_{3u} \\ P_{yw} & P_{yu} \end{bmatrix}$$

$$\text{i.e. } z = P_{11} \cdot w + P_{12} \cdot u \quad \Leftrightarrow \quad \begin{bmatrix} z \\ y \end{bmatrix} = P \cdot \begin{bmatrix} w \\ u \end{bmatrix}$$

$$y = P_{21} \cdot w + P_{22} \cdot u = P_{21} w + P_{22} Cy \Leftrightarrow (I - P_{22} C) y = P_{21} w$$

$$\Leftrightarrow y = (I - P_{22} C)^{-1} P_{21} w$$

$$z = P_{11} w + P_{12} u = P_{11} w + P_{12} Cy = P_{11} w + P_{12} C(I - P_{22} C)^{-1} P_{21} w$$

$$\begin{aligned}z &= P_{11}w + P_{12}u = P_{11}w + P_{12}Cg = P_{11}w + P_{12}C(I - P_{22}C)^{-1}P_{21}w \\&\iff T_{gw} = P_{3w} \cdot w + P_{3u}C(I - P_{gu}C)^{-1}P_{yu}\end{aligned}$$

AA/ECE/ME 548 Linear Multivariable Control Sp22 Prof Borden

today: course logistics, Canvas, etc

results from "check-in" survey Sat Jun 4 (for +2 bonus)

exam 2 next week - due Fri Jun 3 (?)

HW 7 self-assessment - due ~~next Monday~~ Mon Jun 6

HW 8 - due this Friday

week 9 lectures

questions / office hours

$$\text{ex: } L_p(-\infty, \infty) = \left\{ u : \underline{(-\infty, \infty)} \rightarrow \mathbb{C}^n \mid \|u\|_p < \infty \right\}$$

$$= \mathbb{R} \text{ or } \mathbb{Z} = \{\dots, -2, -1, 0, 1, 2, \dots\}$$

given $u, v \in L_p(-\infty, \infty)$, $\alpha \in \mathbb{C}$

define $w = u + \alpha v$ where $\forall t \in (-\infty, \infty) : w(t) = u(t) + \alpha \cdot v(t) \in \mathbb{C}^n$

ex: $W = \{f : S \rightarrow V\}$ where V is vec space, S is set
 $\text{eg } V = \mathbb{C}^n \quad \text{eg } S = (-\infty, \infty)$

$u, v \in W, \alpha \in \mathbb{C} : w = u + \alpha \cdot v$ defined $\forall a \in S : w(a) = u(a) + \alpha \cdot v(a)$

$$\|u\|_p = \left(\int_{-\infty}^{\infty} \|u(t)\|_p^p dt \right)^{1/p} \quad \|u\|_{\infty} = \text{ess sup}_{t \in \mathbb{R}} \|u(t)\|_{\infty} = \max_{t \in \mathbb{R}} \|u(t)\|_{\infty}$$

ex: $V = \left\{ u: \underbrace{\{0, 1, 2, \dots, t-1\}}_{=: S} \rightarrow \mathbb{R}^n \right\}$

given $u \in V$ define $\bar{u} \in \mathbb{R}^{t \cdot n}$ by $\bar{u} = \begin{bmatrix} u(0) \\ u(1) \\ \vdots \\ u(t-1) \end{bmatrix}$

$$\text{then } \|\bar{u}\|_2^2 = \bar{u}^T \cdot \bar{u} = \sum_{s=0}^{t-1} u(s)^T \cdot u(s)$$

$$\text{so I might propose } \|w\|_2^2 := \int_{-\infty}^{\infty} w^T(s) \cdot \underbrace{w(s)}_{\in \mathbb{R}^n} ds = \int_{-\infty}^{\infty} \|w(s)\|_2^2 ds$$

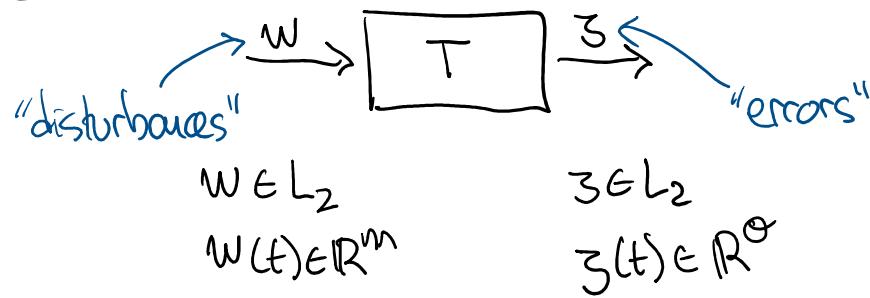
for any $w: (-\infty, \infty) \rightarrow \mathbb{R}^n$

$$\text{more generally, } \|w\|_p = \left(\int_{-\infty}^{\infty} \|w(s)\|_p^p ds \right)^{1/p}$$

$p \in [1, \infty)$

$\|T\|_{L_2 \rightarrow L_2}$ is the induced norm of LTI system T

$\| \cdot \|_{L_2 \rightarrow L_2}$ is the induced norm of LTI system T



→ quantifies the maximum amount of "power" (quantified as $\|w\|_2$) that T transfers from w to z (quantified as $\|z\|_2$)

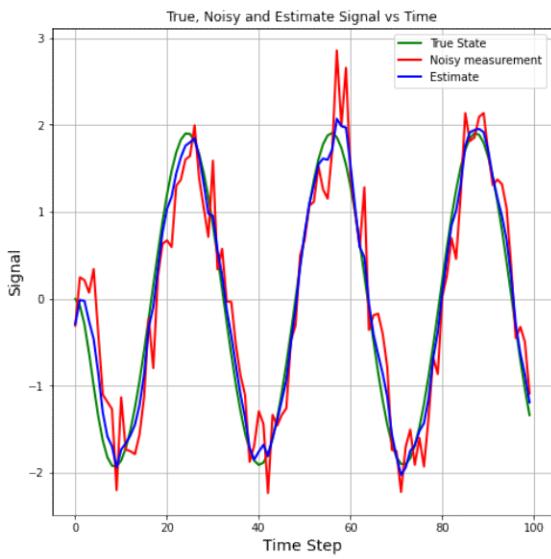
In contrast, $\|\hat{T}\|_2$ quantifies power in output ($\|z\|_2$) produced by one specific input ($\|w^*\|_2$)

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 - exam 2 next week — due Sat Jun 4
 - HW 7 self-assessment — due ~~next Monday~~ Mon Jun 6
 - HW 8 — due this Friday — self-assessment due
 - week 9 lectures
 - questions / office hours

- TODO:
- mixed H₂/H_∞ control paper
 - post today's notes & recording

example results from applying Kalman Filter (HW7):



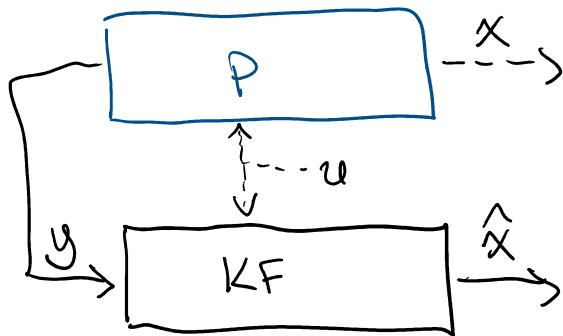
* looks like a low-pass filter
 → is it? if so, why? if not, why not?

• our Kalman filter is defined by:

$$\hat{x}_s = \tilde{x}_s + L_s (y_s - C_s \tilde{x}_s)$$

$$\tilde{x}_s = A_{s-1} \hat{x}_{s-1} + B_{s-1} u_{s-1}$$

$$\text{i.e. } \hat{x}^+ = \hat{A} \hat{x} + \hat{B} u + L u$$



$$\text{i.e. } \hat{x}^+ = \hat{A}\hat{x} + \hat{B}u + Lg$$

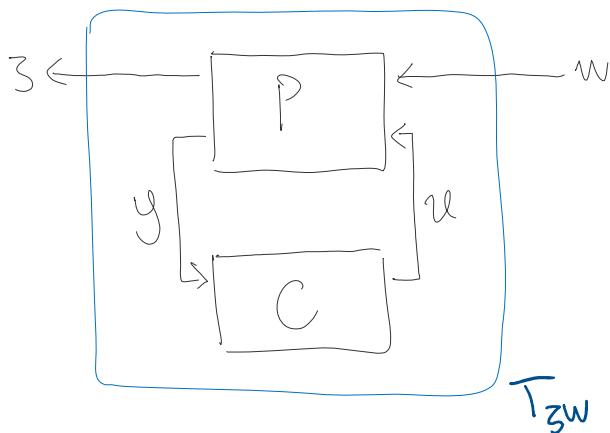
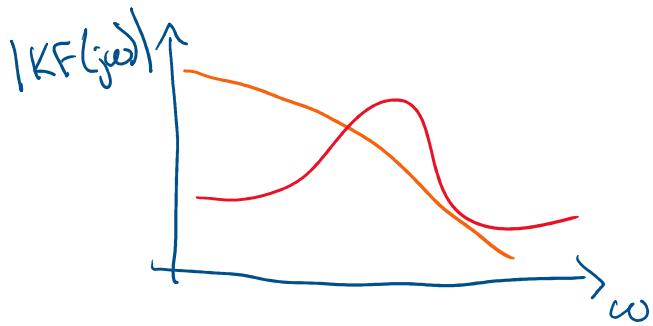
$$\Updownarrow$$

$$\hat{x} = \underbrace{[(sI - \hat{A})^{-1}L]}_{KF} \cdot y$$

$$x^+ = Ax + Bu + g$$

$$y = Cx + Du + v$$

$$x = \underbrace{[C(sI - A)^{-1}B + D]}_P y$$



LQG problem:

$$w = \begin{bmatrix} s & u \end{bmatrix}$$

$$P: z \left| \begin{array}{c|cc} A & [E \ 0] & B \\ \hline Q^{1/2} & 0 & 0 \\ 0 & C & [0 \ F] \end{array} \right| \begin{array}{c} O \\ O \\ R^{1/2} \end{array} \right.$$

ex: LQG $w = (s, u)$, $s, u \sim \mathcal{N}(0, I)$, $z = (Q^{1/2}x, R^{1/2}u)$

$$x^+ = Ax + Bu + Es \quad \Rightarrow \text{Cov}[Es] = E^T E$$

$$y = Cx + F\eta$$

WANT: $\sum_{s=0}^{\infty} x_s^T Q_s x_s + u_s^T R_s u_s \text{ small} \Leftrightarrow \|T_{zw}\|_2 \text{ small}$

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today:

- course logistics, Canvas, etc

- exam 2

- questions / office hours

$$\min_{x_0, u_0} \|f(x_0, u_0)\|_2^2$$

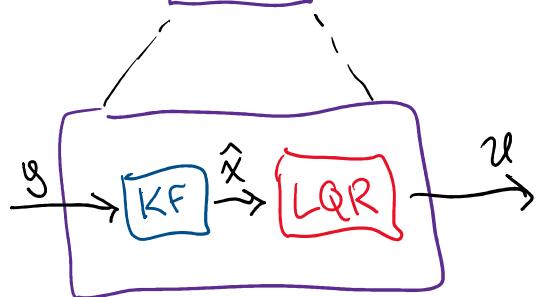
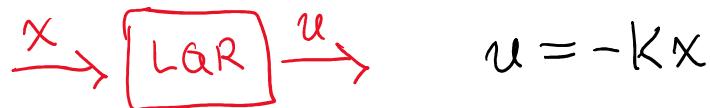
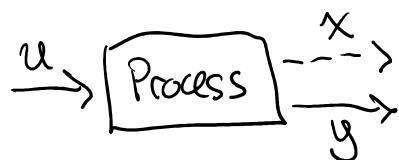
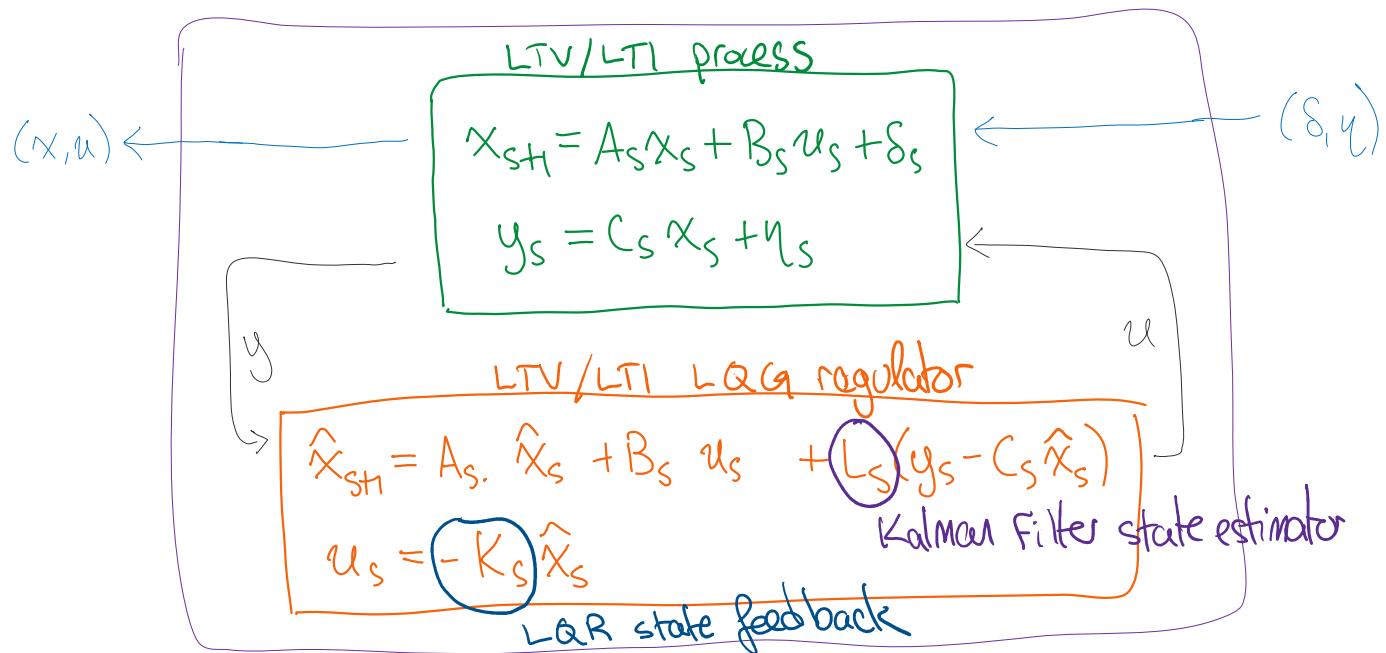
$$\dot{x}/x^+ = f(x, u) \quad y = h(x, u) \quad \text{suppose } \begin{matrix} f(x_0, u_0) = 0 \\ h(x_0, u_0) = x_0 \end{matrix} \quad (\dot{x})$$

$$f(x_0, u_0) = x_0 \quad (x^+)$$

$$\begin{aligned} \delta \dot{x}/\delta x^+ &= A \delta x + B \delta u & A = D_x f(x_0, u_0) & B = D_u f(x_0, u_0) \\ \delta y &= C \delta y + D \delta u & C = D_x h(x_0, u_0) & D = D_u h(x_0, u_0) \end{aligned}$$

$$\Rightarrow C = \left[\begin{array}{cccc} \partial_{x_1} h_1 & \partial_{x_2} h_1 & \cdots & \partial_{x_n} h_1 \\ \vdots & & & \\ \partial_{x_1} h_O & \partial_{x_2} h_O & \cdots & \partial_{x_n} h_O \end{array} \right] \Bigg|_{\begin{array}{l} x=x_0 \\ u=u_0 \end{array}}$$

$$\begin{vmatrix} \partial_{x_1} h_0 & \partial_{x_2} h_0 & \cdots & \partial_{x_n} h_0 \end{vmatrix} \Big|_{\substack{x=x_0 \\ u=u_0}}$$



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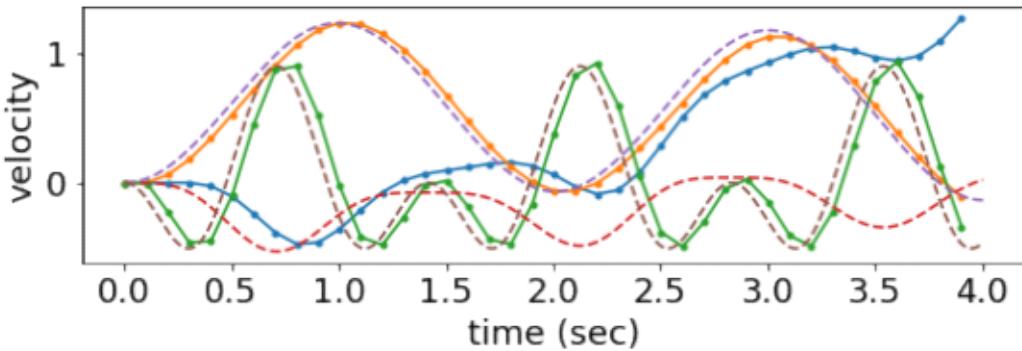
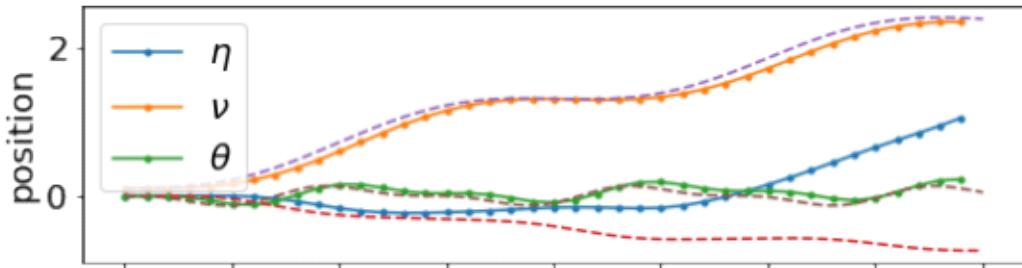
- today:
- course logistics, Canvas, etc
 - exam 2 - let me know if you want more time
 - course evaluation
 - questions / office hours

subproblem (a)

Discretize the nonlinear system: choose a step size $\Delta > 0$ and time horizon $T \gg \Delta$ so that the forward Euler approximation $x(t + \Delta) \approx x(t) + \Delta f(x(t), u(t))$ yields reasonably good approximation of trajectories on the time interval $t \in [0, T]$. Provide plots that show the approximation is reasonable by simulating the nonlinear system twice -- once with your chosen step size Δ , and once with step size $\nabla \ll \Delta$; the second simulation will be used as the "ground truth". Use non-constant input signals F and τ in your simulation.

Note: there is no one right choice for Δ , but it will be helpful in what follows if you choose a large step size.

ex:



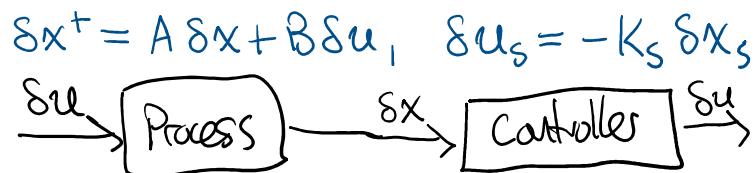
subproblem (b)

Find an equilibrium (x_0, u_0) for the discretized nonlinear system and linearize the dynamics around this equilibrium to obtain a discrete-time linear time-invariant (DT-LTI) control system model: determine A, B, C, D such that, with $\delta x = x - x_0$, $\delta u = u - u_0$, and $\delta y = y - y_0$ where $y_0 = h(x_0, u_0)$, we have $x^+ \approx x_0 + \delta x^+$ and $y \approx y_0 + \delta y$ where

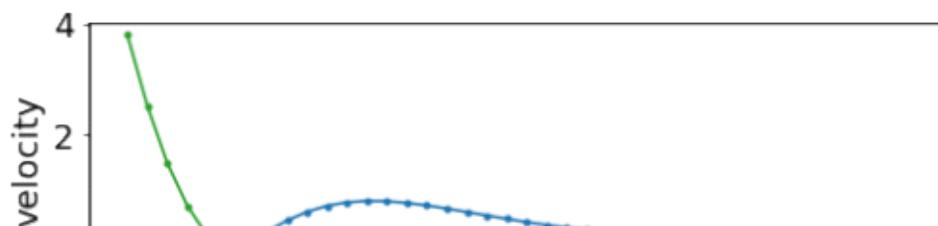
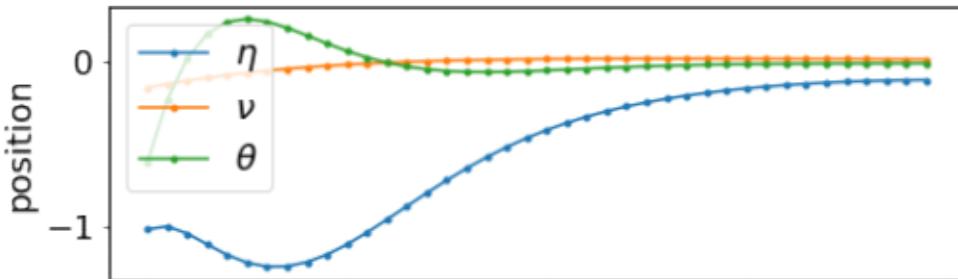
$$(\text{DT-LTI}) \quad \delta x^+ = A \delta x + B \delta u, \quad \delta y = C \delta x + D \delta u.$$

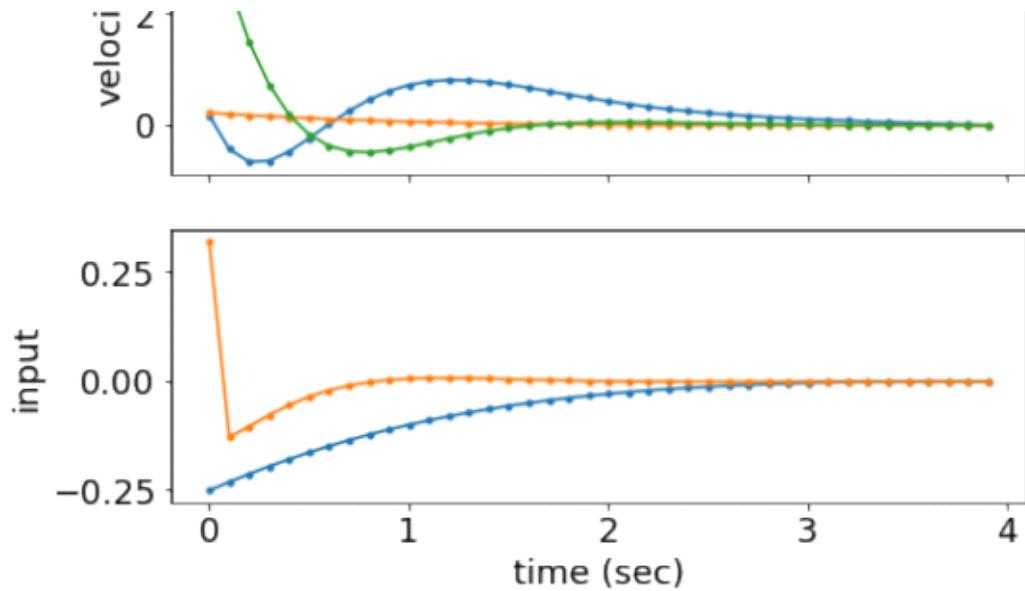
subproblem (c)

Choose Q and R matrices and solve the finite-horizon LQR problem to synthesize an LTV state-feedback controller for the LTI control system on the time horizon T you chose above. Run a simulation of the controlled system and provide plots of the state δx , control δu , and output δy time series.



ex:





subproblem (d)

Choose square nonsingular matrices V_1 and V_2 that multiply standard normal random vectors $w_1 \in \mathbb{R}^3$ and $w_2 \in \mathbb{R}^2$ to model process and measurement noise,

$$\delta x^+ = A \delta x + B \delta u + \begin{bmatrix} 0 \\ V_1 \end{bmatrix} w_1, \quad \delta y = C \delta x + D \delta u + V_2 w_2.$$

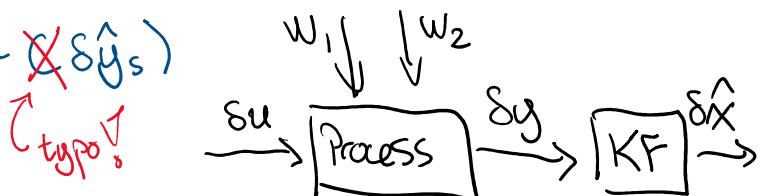
Solve the finite-horizon Kalman Filtering problem to synthesize an LTV observer for the LTI control system. Run a simulation of the LTI system including samples from the process and measurement noise distributions you chose and non-constant input signals, use the resulting data to run a simulation of your LTV Kalman Filter, and provide plots of: the state δx and your estimate $\hat{\delta x}$; the noisy output δy , noise-free output, and observer output $\hat{\delta y}$.

$$\delta x^+ = A \delta x + B \delta u + \begin{bmatrix} 0 \\ V_1 \end{bmatrix} w_1 \quad w_1 \sim \mathcal{N}(0, I)$$

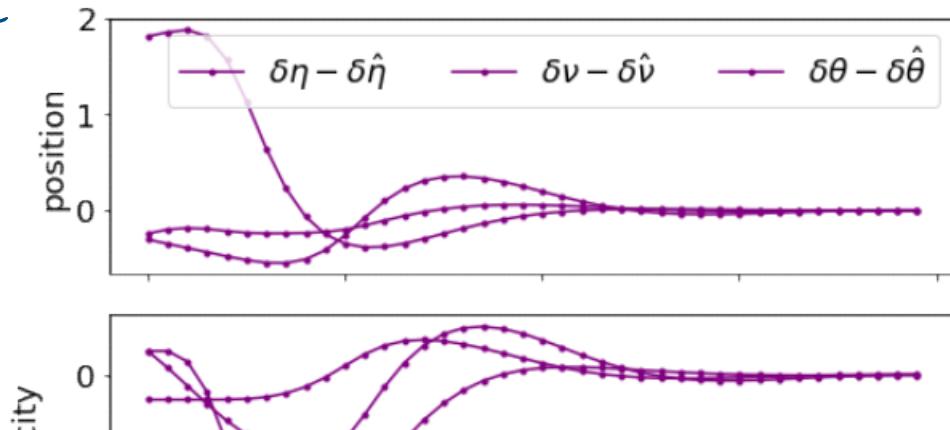
$$\delta y = C \delta x + D \delta u + V_2 w_2 \quad w_2 \sim \mathcal{N}(0, I)$$

$$\delta \hat{x}_{st+1} = A \delta \hat{x}_s + B \delta u_s + L_s (\delta y_s - \cancel{\delta \hat{y}_s})$$

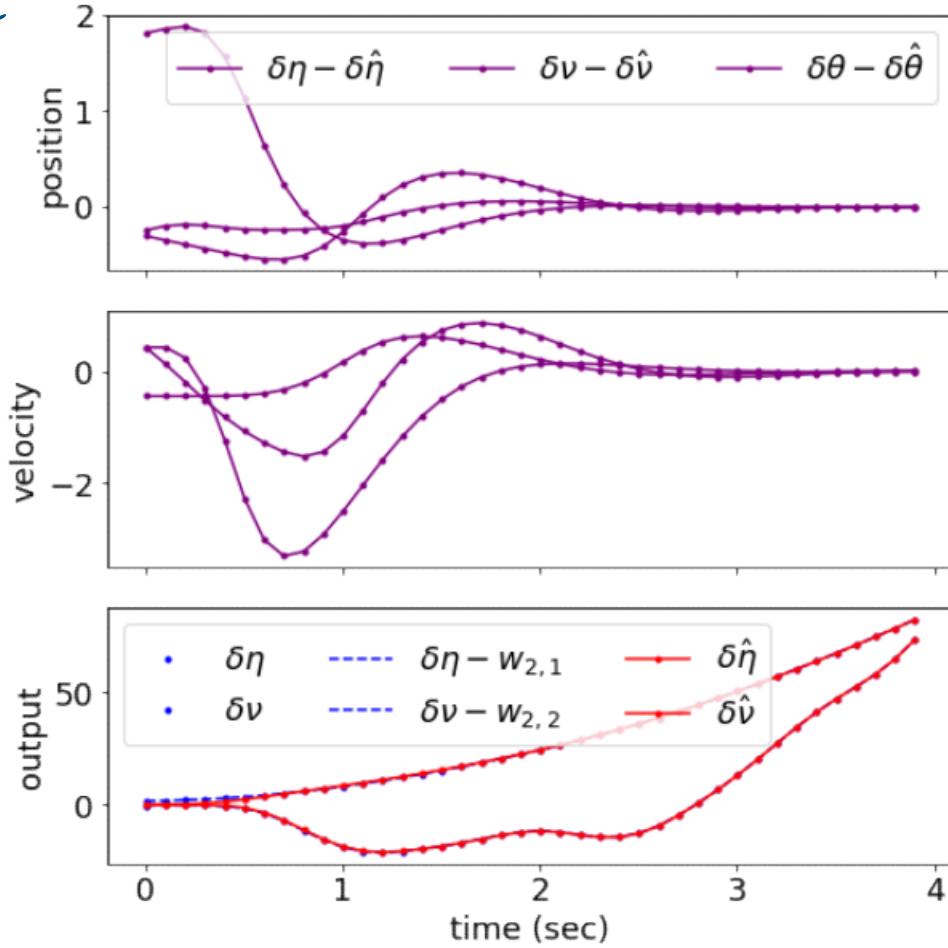
$$\delta \hat{y} = C \delta \hat{x} + D \delta u$$



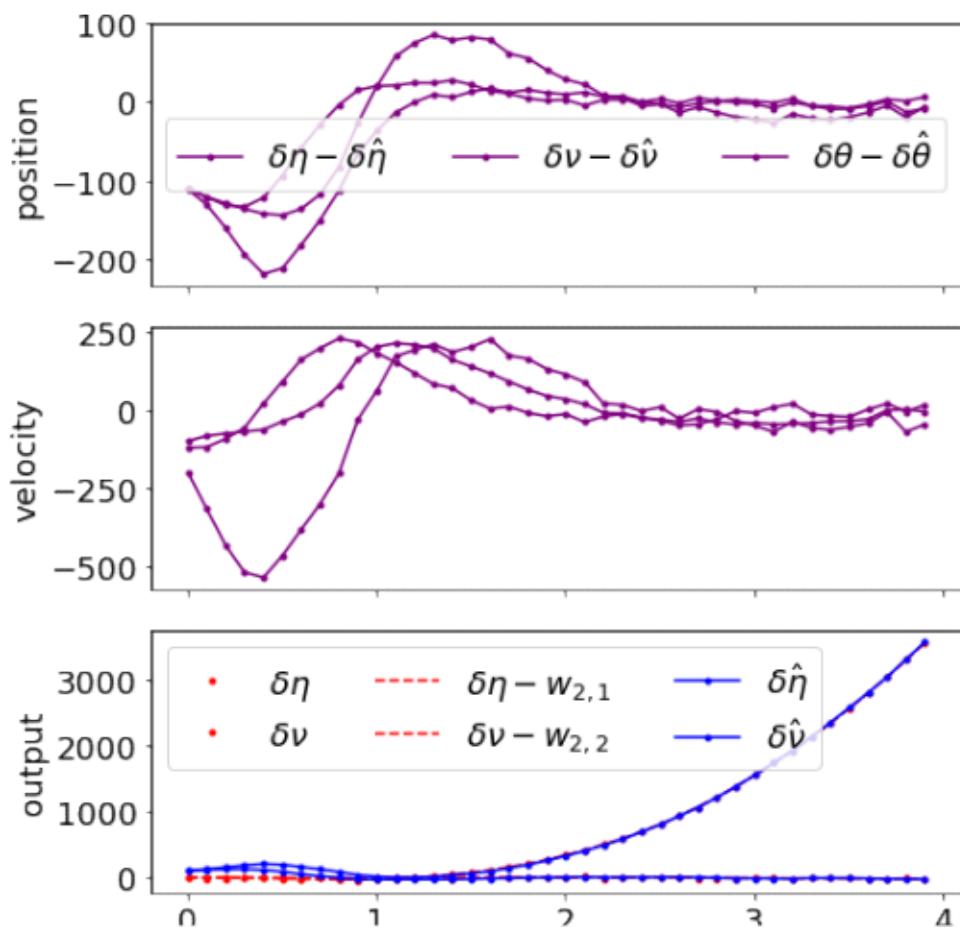
ex: low noise

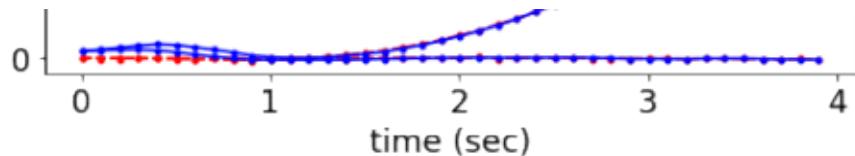


ex: low noise



ex: high noise





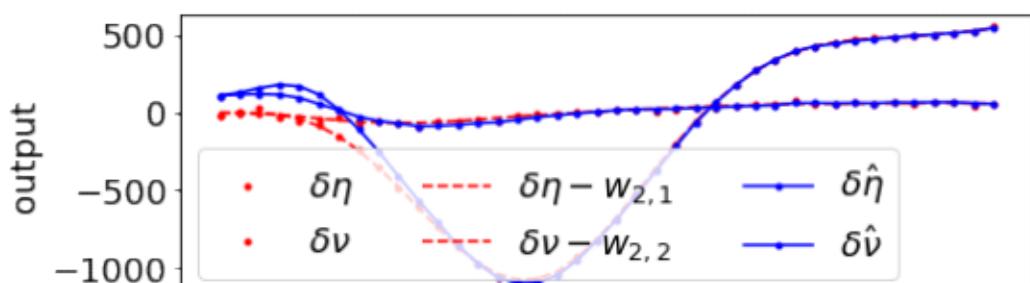
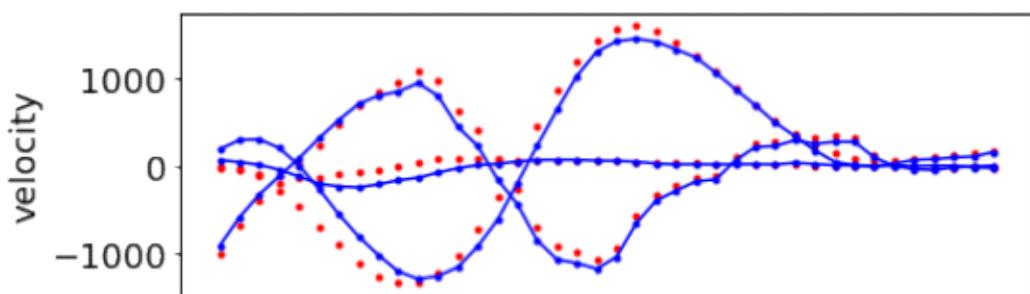
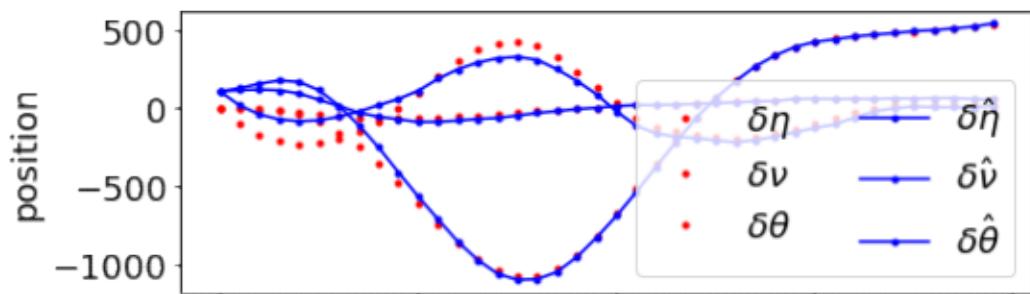
subproblem (e)

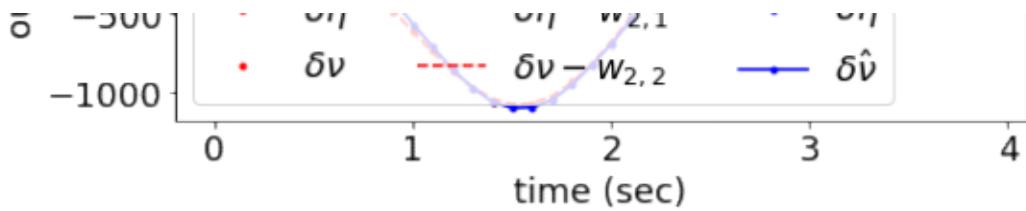
Combine your LQR state-feedback law and Kalman Filter observer to obtain a time-varying linear-quadratic Gaussian (LQG) regulator. Run a simulation of the closed-loop system obtained from the feedback interconnection of the LTI system and your LQG regulator. Provide plots of: the state δx and your estimate $\hat{\delta x}$; the noisy output δy , noise-free output, and observer output $\hat{\delta y}$.

$$\left\{ \begin{array}{l} \delta x^+ = A \delta x + B \delta u + \begin{bmatrix} 0 \\ V_1 \end{bmatrix} w_1 \quad w_1 \sim \mathcal{N}(0, I) \\ \delta y = C \delta x + D \delta u + V_2 w_2 \quad w_2 \sim \mathcal{N}(0, I) \\ \hat{\delta x}_{st1} = A \delta \hat{x}_s + B \delta u_s + L_s (\delta y_s - \cancel{C \hat{\delta y}_s}) \\ \hat{\delta y} = C \delta \hat{x} + D \delta u \\ \delta u_s = -K_s \delta \hat{x}_s \end{array} \right.$$

typo!

ex.:



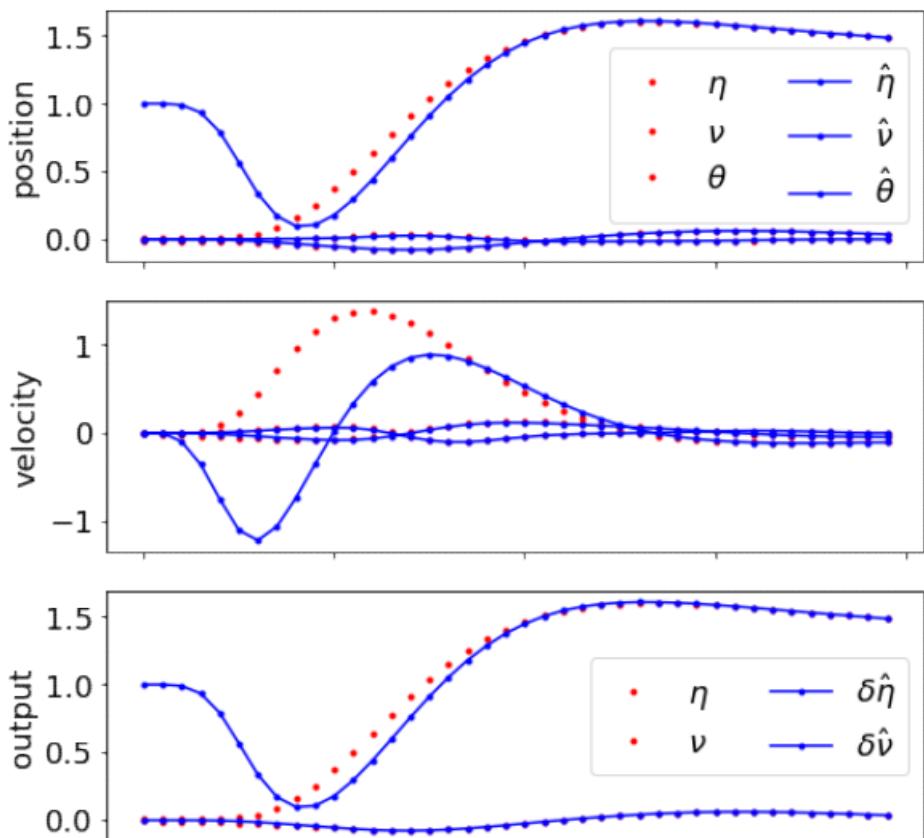


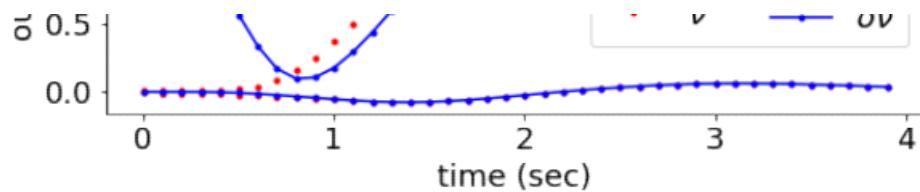
subproblem (f)

Now apply your LQG regulator to the discretized nonlinear system. Run a simulation of the closed-loop system obtained from the feedback interconnection of the DT-NL system and your LQG regulator. Provide plots comparing: the state x and your estimate \hat{x} ; the noisy output y , noise-free output, and observer output \hat{y} .

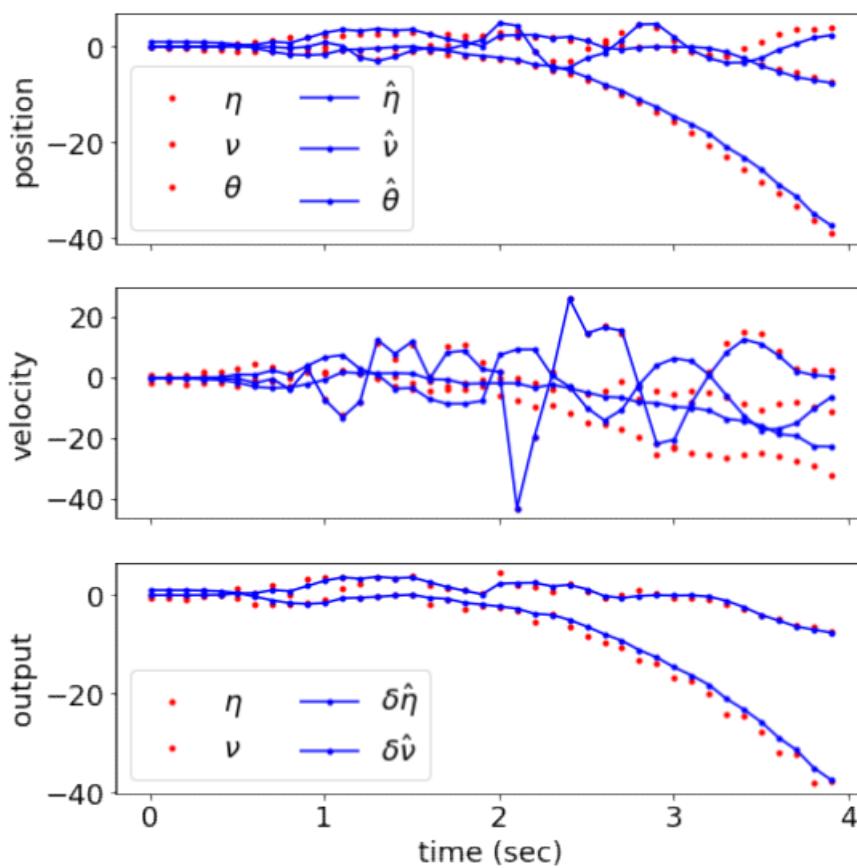
$$\begin{aligned} \text{NL} & \left\{ \begin{array}{l} \dot{x} = f(x, u) + \begin{bmatrix} 0 \\ v_1 \end{bmatrix} w_1 \quad w_1 \sim \mathcal{N}(0, I) \\ y = h(x, u) + V_2 w_2 \quad w_2 \sim \mathcal{N}(0, I) \end{array} \right. \\ \text{L} & \left\{ \begin{array}{l} \delta \hat{x}_{st1} = A \delta \hat{x}_s + B \delta u_s + L_s (sy_s - \cancel{\delta \hat{y}_s}) \\ \delta \hat{y} = C \delta \hat{x} + D \delta u \\ \delta u_s = -K_s \delta \hat{x}_s \end{array} \right. \end{aligned}$$

ex: low noise





ex: high(er) noise



interlude

Reinterpreting the DT-LTI system above as a continuous-time process P in the form expected for robust controller synthesis with inputs $w, \delta u$ and outputs $z, \delta y$ yields

$$\begin{aligned}\delta \dot{x} &= \bar{A} \delta x + B_1 w + B_2 \delta u, \\ z &= C_1 \delta x + D_{11} w + D_{12} \delta u, \\ \delta y &= C_2 \delta x + D_{21} w + D_{22} \delta u\end{aligned}$$

TODO: missing $\frac{1}{\Delta}$?

where $\bar{A} = \frac{1}{\Delta}(A - I)$ and

$Q \in \mathbb{R}^{6 \times 6}$ from (c)

$R \in \mathbb{R}^{2 \times 2}$ from (c)

$\sqrt{\lambda} \in \mathbb{R}^{3 \times 3}$ from (d)

$\sqrt{\lambda} \in \mathbb{R}^{2 \times 2}$ from (d)

$s_N \in \mathbb{R}^6 \dots \mathbb{R}^3 \dots \mathbb{R}^2 \dots \mathbb{R}^5$

$$\begin{aligned}E \in \mathbb{R}^{6 \times 6} &\quad B_1 = \begin{bmatrix} \frac{3}{\Delta} & \frac{2}{\Delta} \\ 0 & 0 \\ \frac{1}{\Delta} V_1 & 0 \end{bmatrix}, B_2 = B, \quad E \in \mathbb{R}^{6 \times 2} \\ E \in \mathbb{R}^{2 \times 6} &\quad C_1 = \begin{bmatrix} \sqrt{\frac{1}{\Delta} Q} \\ 0 \end{bmatrix}, D_{11} = 0, D_{12} = \begin{bmatrix} 0 \\ \sqrt{\frac{1}{\Delta} R} \end{bmatrix}, \quad E \in \mathbb{R}^{6 \times 2} \\ E \in \mathbb{R}^{2 \times 6} &\quad C_2 = [C \ X], D_{21} = \begin{bmatrix} 3 & 2 \\ 0 & \frac{1}{\Delta} V_2 \end{bmatrix}, D_{22} = D, \quad E \in \mathbb{R}^{2 \times 2} \\ s_N \in \mathbb{R}^6 \dots \mathbb{R}^3 \dots \mathbb{R}^2 \dots \mathbb{R}^5 &\quad \dots [w_i] \in \mathbb{R}^5 \quad e_1 \in \mathbb{R}^2 \quad e_0 \in \mathbb{R}^2 \end{aligned}$$

$v_2 \in \mathbb{R}^+$ from (a)

$$\delta x \in \mathbb{R}^6, \quad w_1 \in \mathbb{R}^3, \quad w_2 \in \mathbb{R}^2, \quad w = \begin{bmatrix} w_1 \\ w_2 \end{bmatrix} \in \mathbb{R}^5, \quad \delta u \in \mathbb{R}^2, \quad \delta y \in \mathbb{R}^2$$

$$z = \begin{bmatrix} \sqrt{\frac{1}{\Delta}} Q \delta x \\ \sqrt{\frac{1}{\Delta}} R \delta u \end{bmatrix} \in \mathbb{R}^8 \Rightarrow \|z\|_2^2 = \delta x^\top \frac{Q}{\Delta} \delta x + \delta u^\top \frac{R}{\Delta} \delta u$$

C_{H_2} = control. b12 sign (P , $n_{\text{meas}} = 2$, $n_{\text{cont}} = 2$)

$\hookrightarrow A_{H_2}, B_{H_2}, C_{H_2}, D_{H_2}$ define

$$\boxed{\begin{array}{l} \dot{\delta \hat{x}} = A_{H_2} \delta \hat{x} + B_{H_2} \delta y \\ \delta u = C_{H_2} \delta \hat{x} + D_{H_2} \delta y \end{array}}$$

discretize

$$\delta \hat{x}^+ = (I + \Delta A_{H_2}) \delta \hat{x} + \Delta B_{H_2} \delta y$$

$$\delta u = C_{H_2} \delta \hat{x} + D_{H_2} \delta y$$