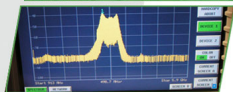




EURECOM

S o p h i a A n t i p o l i s



Radio Engineering

Lecture 1: Introduction

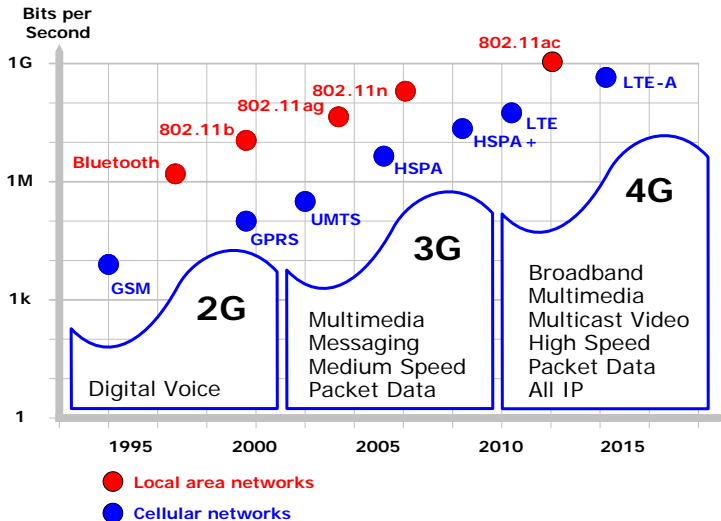
Florian Kaltenberger

- 1 Introduction
 - History of wireless communications
 - Types of services
 - Requirements for services
- 2 Technical challenges
 - Multipath propagation
 - Spectrum limitations
 - Limited Energy
 - User Mobility
- 3 Noise and Interference
 - Noise modeling
 - Link budget
 - Interference limited systems

- 1873: James Clark Maxwell develops theory of electromagnetic waves (Maxwell's Equations)
- 1886: Heinrich Hertz: fundamental experiments confirming Maxwell's theory
- 1890-1905: First experiments for wireless information transmission (Tesla, Bose, Marconi).
- 1920: First radio broadcasting services in England
- 1948: Claude Shannon: fundamental information theory for modern communication systems [1]

- 1950-1980: Development of cellular telephony, first analog systems
- GSM (Global System for Mobile Communications)
 - Second generation fully digital standard
 - First deployment 1990 in Europe
 - Today more than 5 billion users in 212 countries (mostly used standard)
- Other second generation systems
 - IS-95 (cdmaOne): used mainly in US and Korea
 - PDC (Pacific Digital Cellular)

Evolution of Wireless Standards



5G: The next big thing?

The 5G Infrastructure Public-Private Partnership



increasing wireless capacity
1,000 times



connecting
7 billion people



connecting
7 trillion "things"



saving 90% energy



perceiving zero downtime

Source: www.5g-ppp.eu

Main topic of this course:

- Local area networks (LAN)
 - includes body, personal, metropolitian, wide area networks
- Cellular networks
 - Macro, micro, picocells, femtocells

Also important

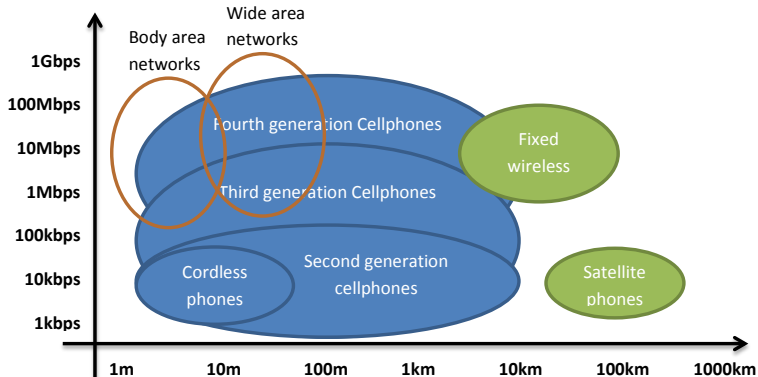
- Fixed wireless link
- Broadcast (TV, Radio)
- Ad-hoc and sensor networks
- Mesh networks
- Satellite Networks

- Data rate
- Range
- Mobility
- Number of users
- Spectrum Usage
- Power consumption

Trade-offs between requirements

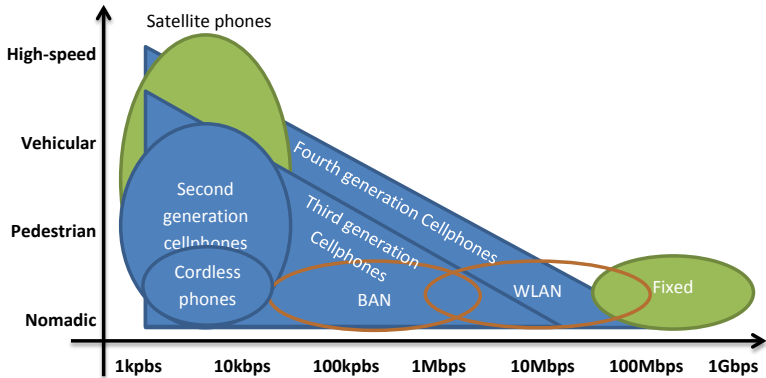
- Sensor networks: $< 1 \text{ kbit/s}$; central nodes need up to 10 Mbit/s
- Speech communications: $5\text{-}64 \text{ kbit/s}$, depending in speech coder (vocoder)
- Elementary data services (email, simple web pages): $10\text{-}100 \text{ kbit/s}$
- Video streaming: $1\text{-}40 \text{ Mbit/s}$
- Personal Area Networks: $> 100 \text{ Mbit/s}$

Data Rate vs Range



- Fixed devices: stay in one location; temporal variations due to moving objects in surroundings
- Nomadic devices: MS placed at certain location, stays there for a while (WLANs)
- Low mobility: pedestrian speeds (cordless phones)
- High speed: cellphones in cars
- Extremely high speed: high-speed trains, planes, ...

Data Rate vs Speed



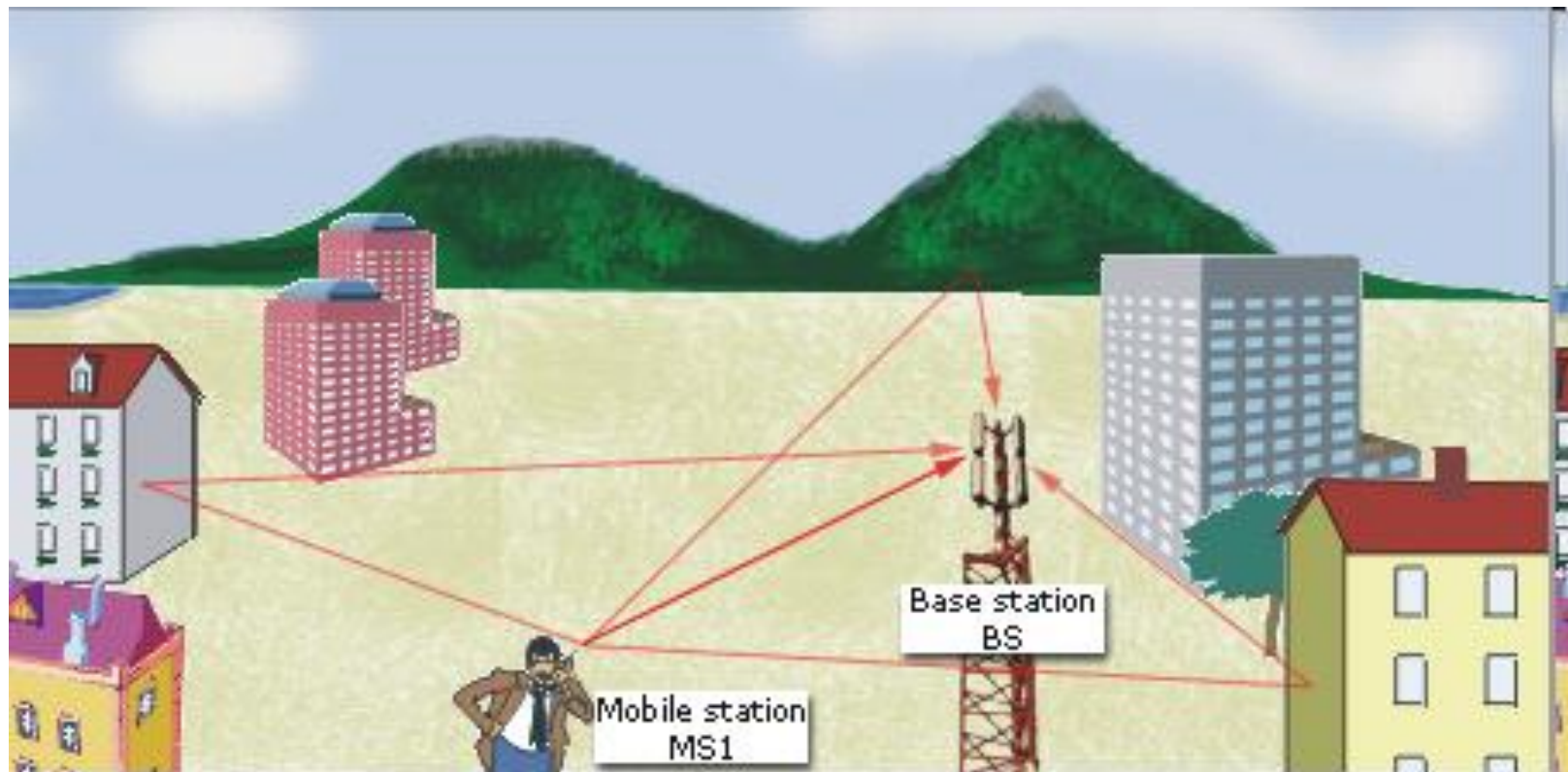
- Spectrum dedicated to specific service and operator
 - Cellular phones, TV, Military, etc
- Spectrum dedicated to specific service
- Free spectrum
 - Industrial, scientific and medical (ISM) bands: 915MHz, 2.4GHz, 5.8GHz
 - WiFi, Bluetooth, DECT, ZigBee, etc.
- Opportunistic Spectrum access (Cognitive radio)
 - Licensed Shared Access
 - Light Licensed Access (TVWS)
 - Spectral Overlay

Technical challenges of wireless communications

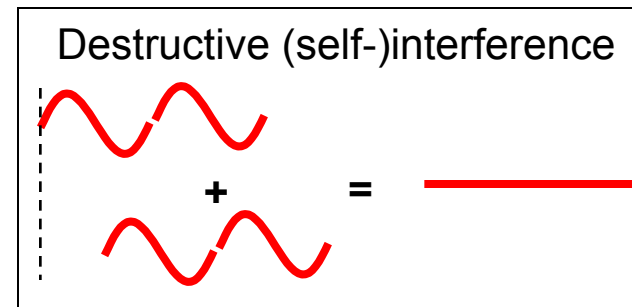
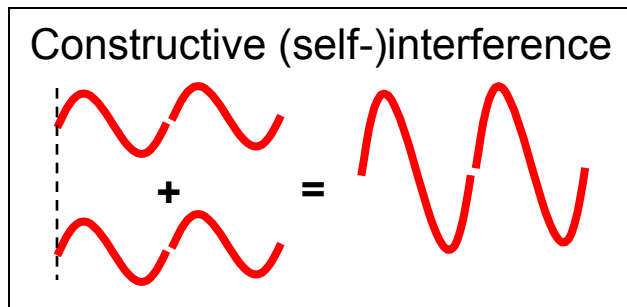
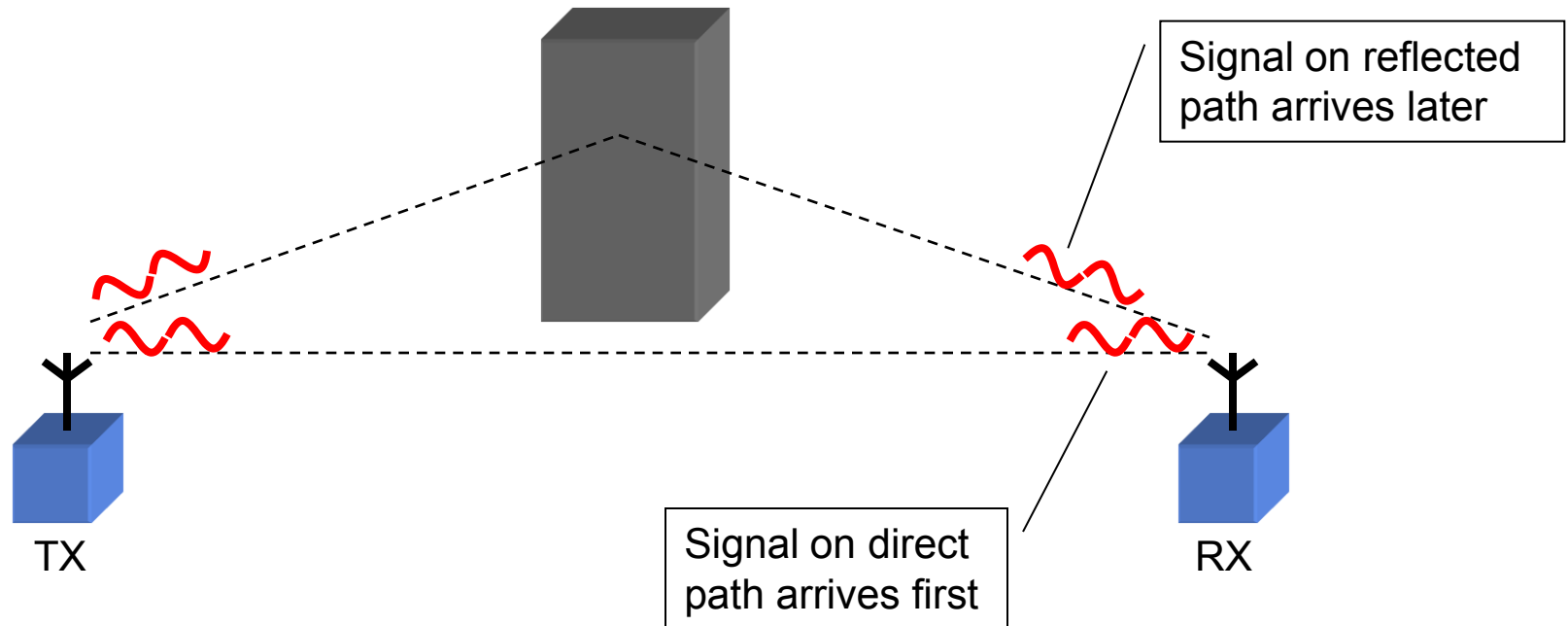
The major challenges

- Multipath propagation
- Spectrum limitations
- Limited energy
- User mobility

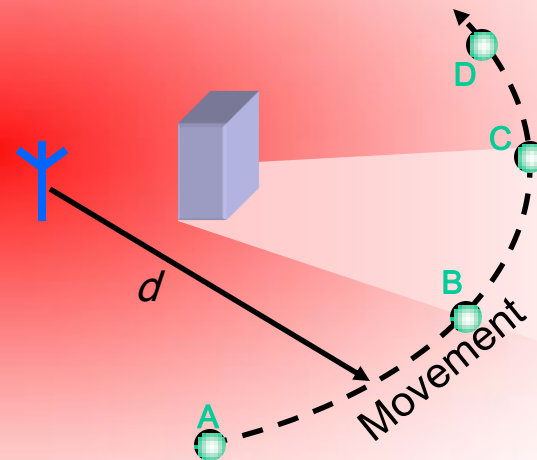
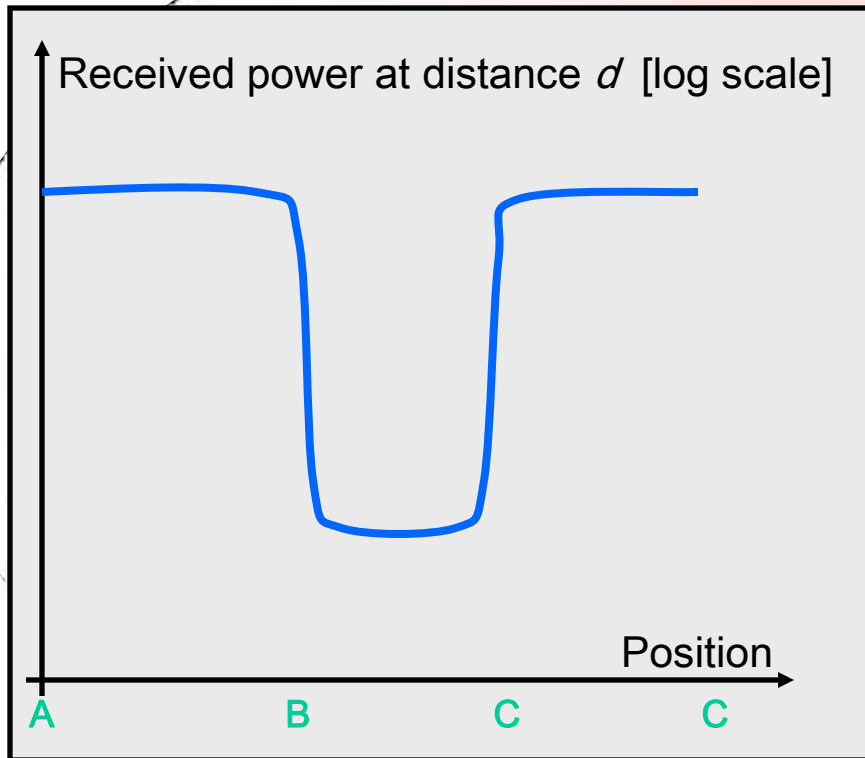
Multipath propagation



Small-scale fading



Large-scale fading

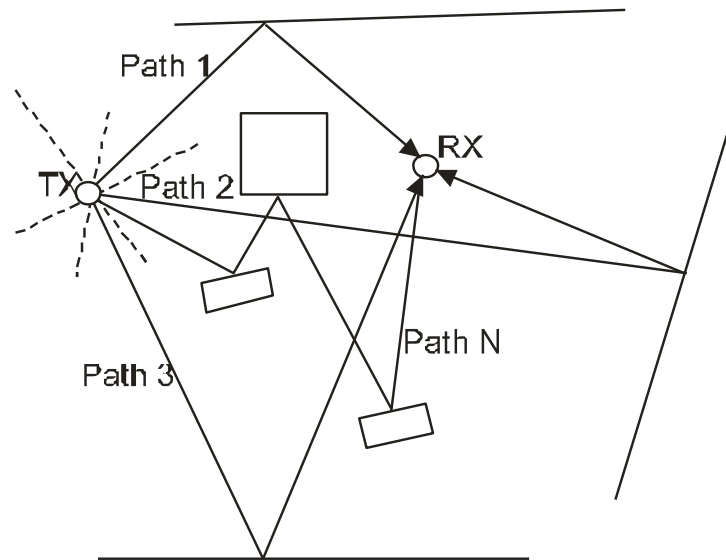


Consequences of fading

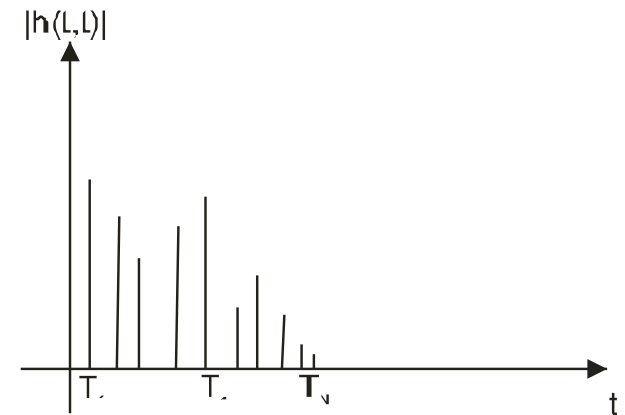
- Error probability is dominated by probability of being in a fading dip
- Error probability decreases only linearly with increasing SNR
- Fighting the effects of fading becomes essential for wireless transceiver design
- Deterministic modeling of channel at each point very difficult
- Statistical modeling of propagation and system behavior

Intersymbol interference (1)

- Channel impulse response is delay-dispersive

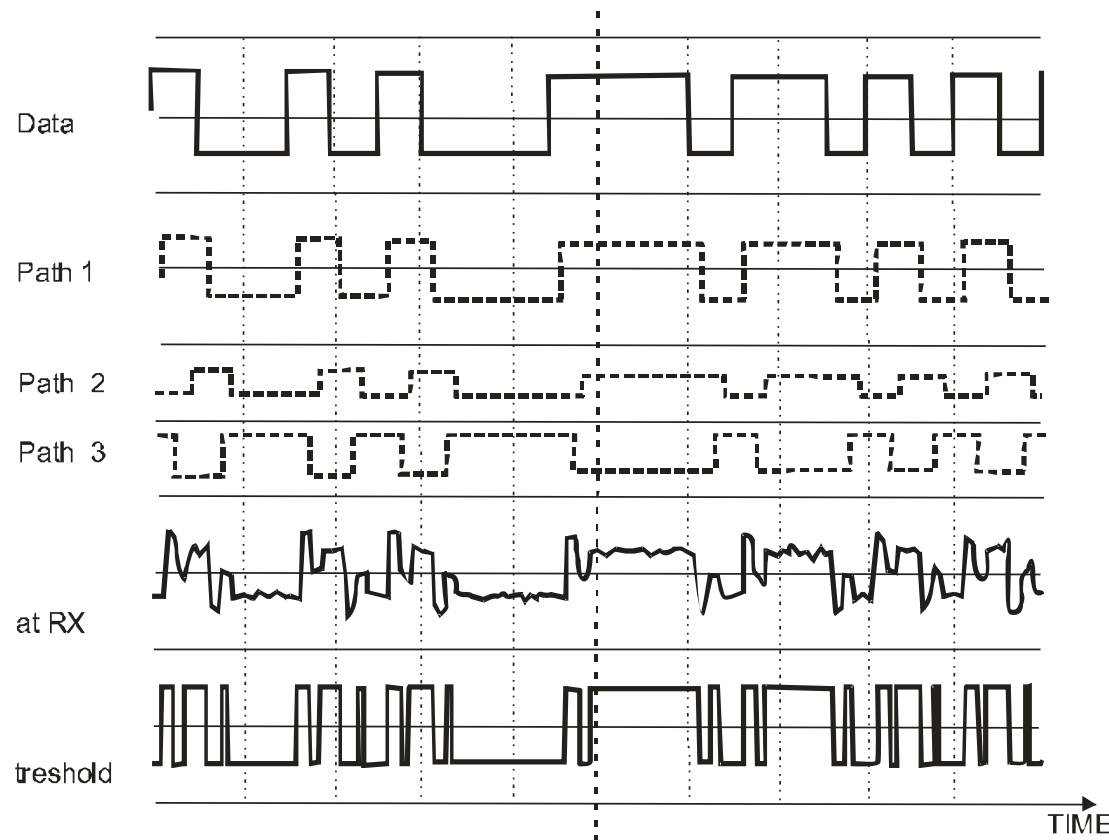


Multipath components with different runtimes



Channel impulse response

Intersymbol interference (2)



Spectrum assignment

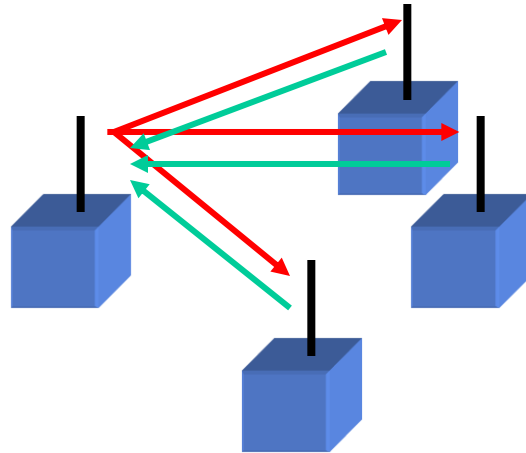
- **<100 MHz:** CB radio, pagers, and analogue cordless phones.
- **100-800 MHz:** broadcast (radio and TV)
- **400-500 MHz:** cellular and trunking radio systems
- **800-1000 MHz:** cellular systems (analogue and second-generation digital); emergency communications
- **1.8-2.0 GHz:** main frequency band for cellular and cordless
- **2.4-2.5 GHz:** cordless phones, wireless LANs and wireless PANs (personal area networks); other devices, e.g., microwave ovens.
- **3.3-3.8 GHz:** fixed wireless access systems
- **4.8-5.8 GHz:** wireless LANs
- **11-15 GHz:** satellite TV LTE frequency bands

Frequency reuse

- Available spectrum is limited
- -> the same frequency (range) has to be used at many different locations
- Regulated spectrum:
 - a single operator owns the spectrum, and can determine where to put TXs
 - cell planning so that interference adheres to certain limits
- Unregulated spectrum:
 - Often only one type of service allowed,
 - Nobody can control location of interferers
 - Power of interferers is limited by regulations
- Opportunistic Spectrum access:
 - IEEE 802.22 (TV white space, digital dividend)

Duplexing and multiple access

- Within each frequency band, multiple users need to communicate with one BS (multiple access)

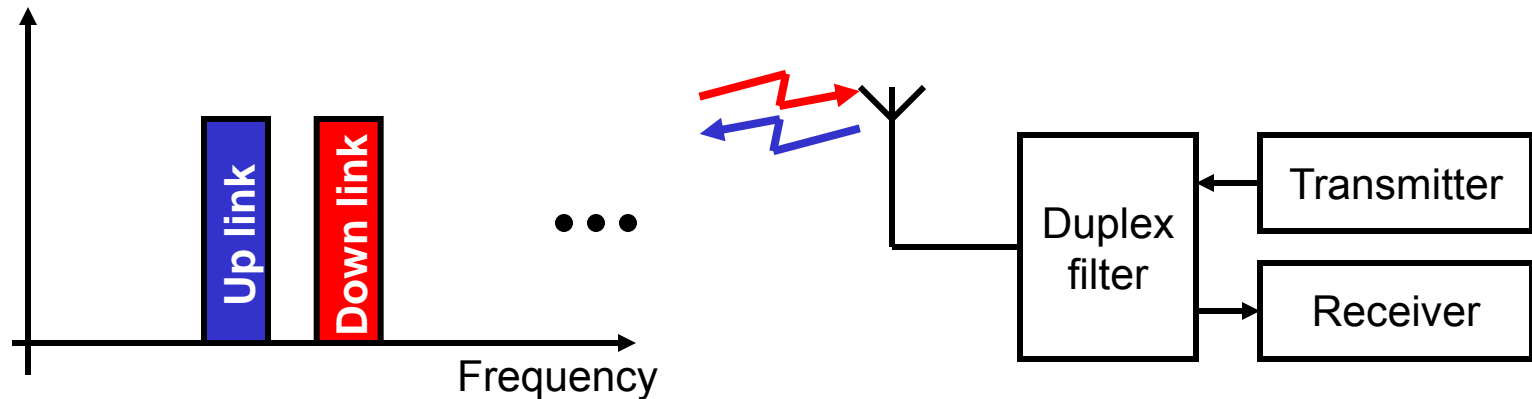


Mobile telephony, wireless LAN, ...

- Cellphones have to be able to transmit and receive voice communications (duplexing)

DUPLEX

Frequency-division Duplex (FDD)



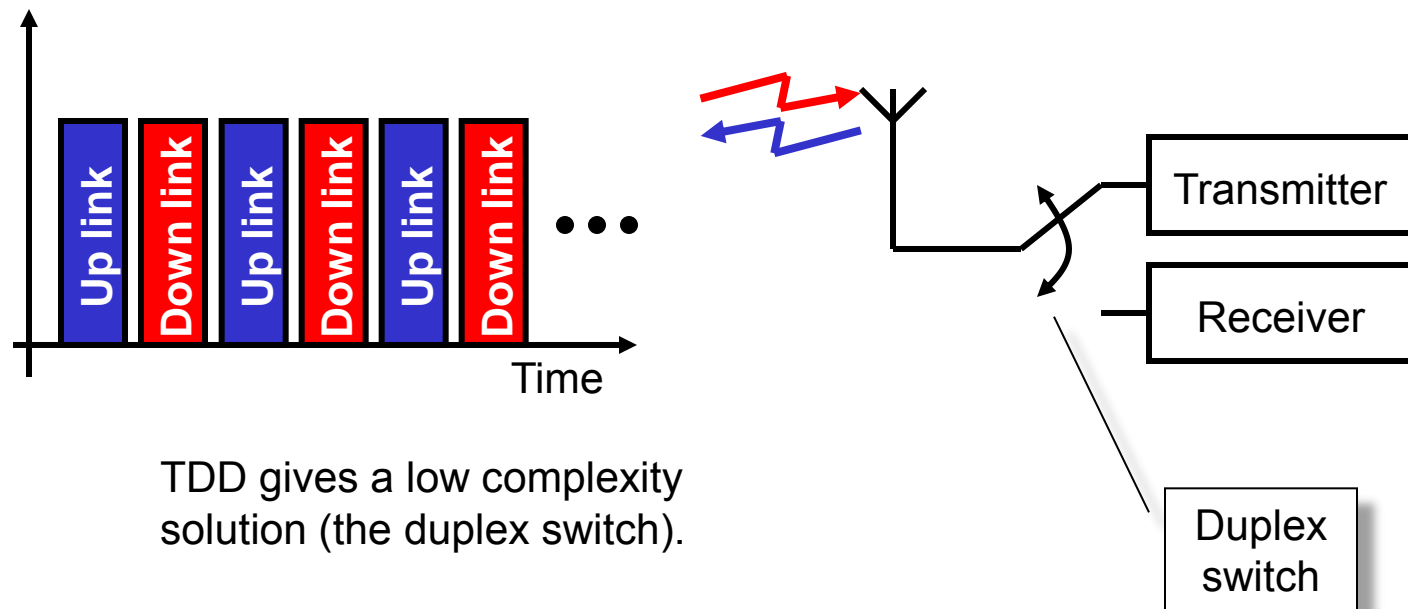
FDD gives a more complex solution (the duplex filter).

Can be used for continuous transmission.

Examples: Nodic Mobile Telephony (NMT), Global System for Mobile communications (GSM), Wideband CDMA (WCDMA)

DUPLEX

Time-division duplex (TDD)



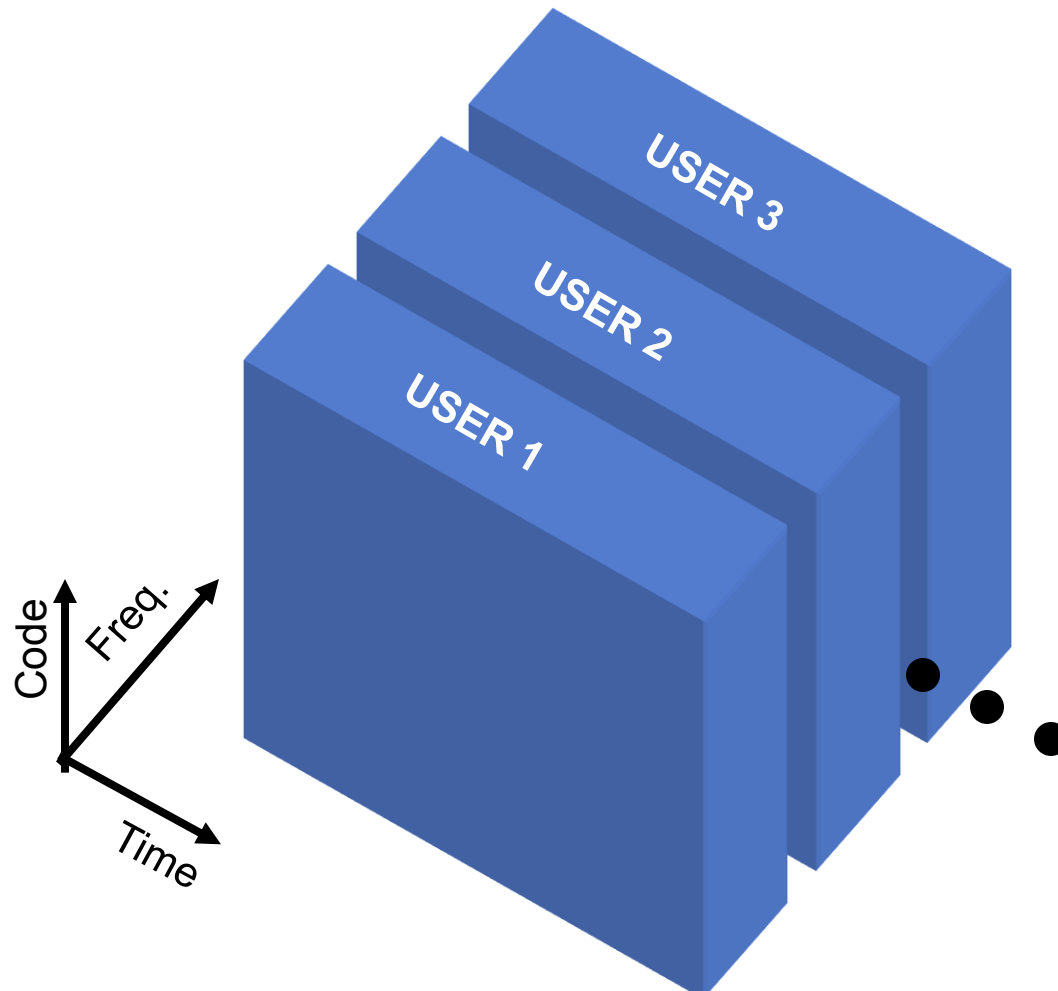
TDD gives a low complexity solution (the duplex switch).

Cannot be used for continuous transmission.

Examples: Global System for Mobile communications (GSM),
Wideband CDMA (WCDMA)

MULTIPLE ACCESS

Freq.-division multiple access (FDMA)

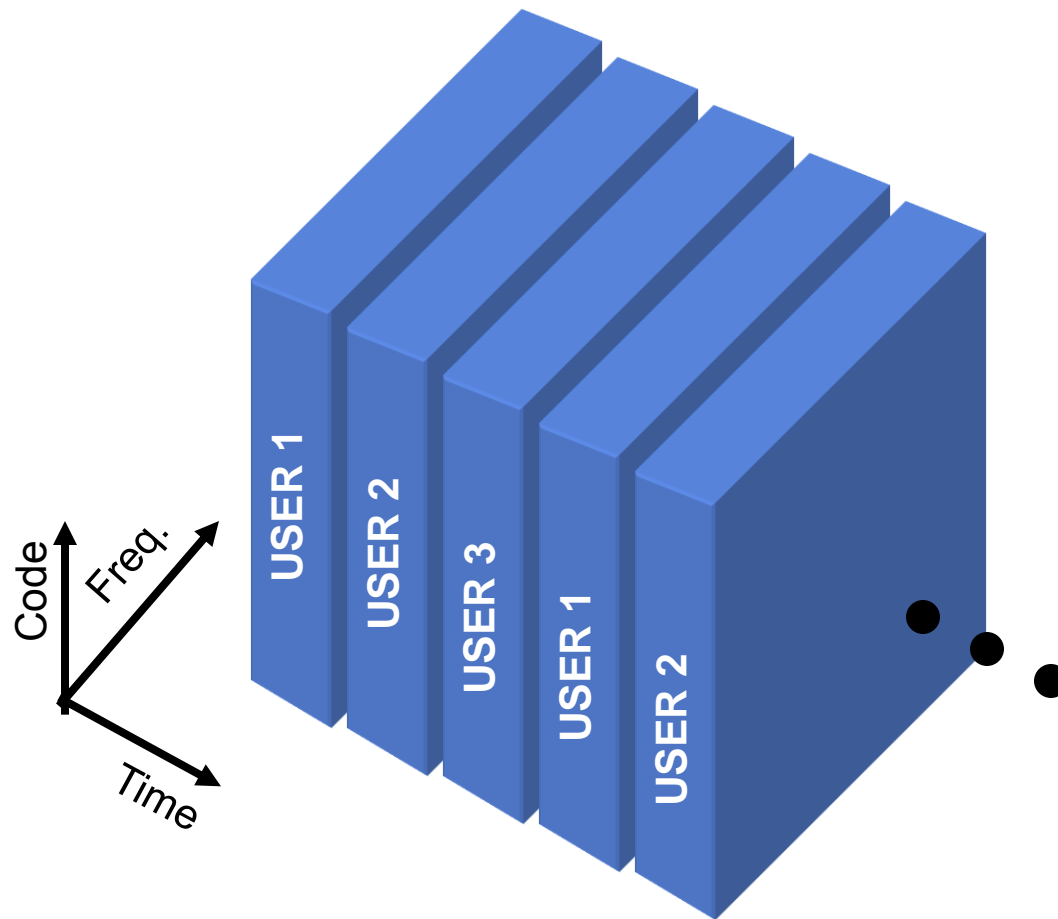


Users are separated in frequency bands.

Examples: Nordic Mobile Telephony (NMT), Advanced Mobile Phone System (AMPS)

MULTIPLE ACCESS

Time-division multiple access (TDMA)

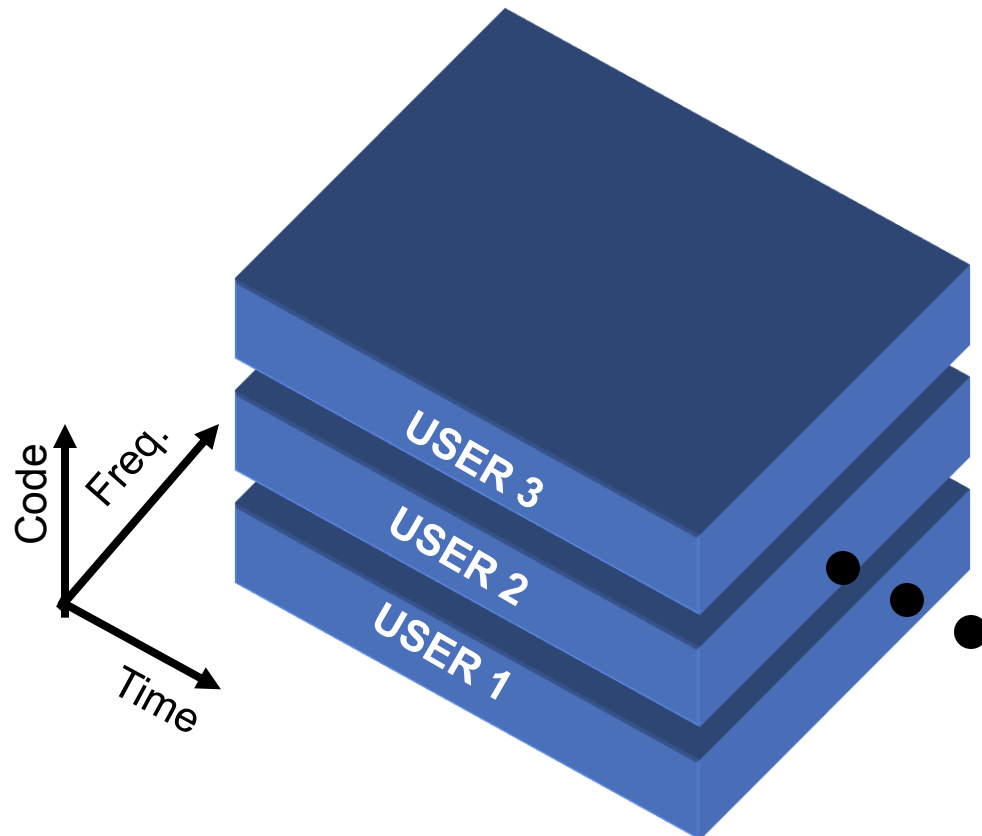


Users are separated
in time slots.

Example: Global System for Mobile communications (GSM)

MULTIPLE ACCESS

Code-division multiple access (CDMA)

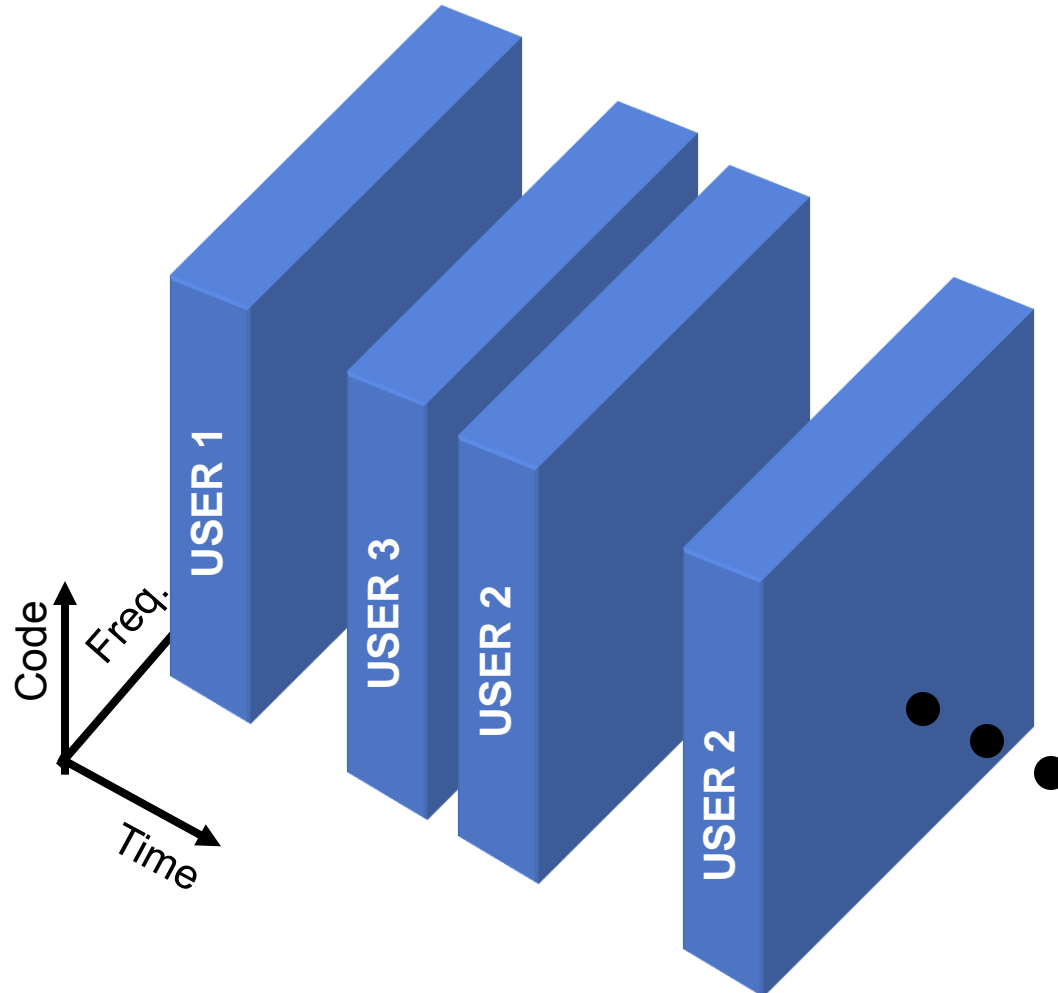


Users are separated by spreading codes.

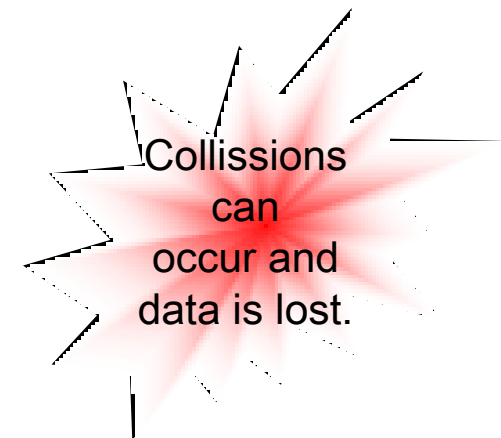
Examples: CdmaOne, Wideband CDMA (WCDMA), Cdma2000

MULTIPLE ACCESS

Carrier-sense multiple access (CSMA)



Users are separated in time but not in an organized way. The terminal listens to the channel, and transmits a packet if it's free.



Example: IEEE 802.11 (WLAN)

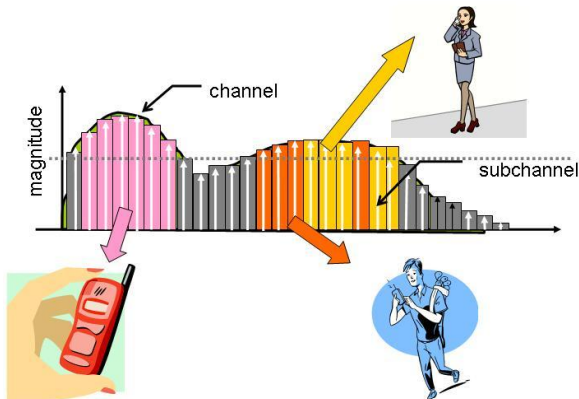
User mobility

- User can change position
- Mobility within one cell (i.e., maintaining a link to a certain BS): mostly effect on propagation channel (fading)
- Mobility from cell to cell: Handover, roaming, etc

MULTIPLE ACCESS

Orthogonal Frequency Division Multiplexing

- Based on Orthogonal frequency division duplexing (OFDM)
- Similar to FDMA, but more efficient
- Used in LTE



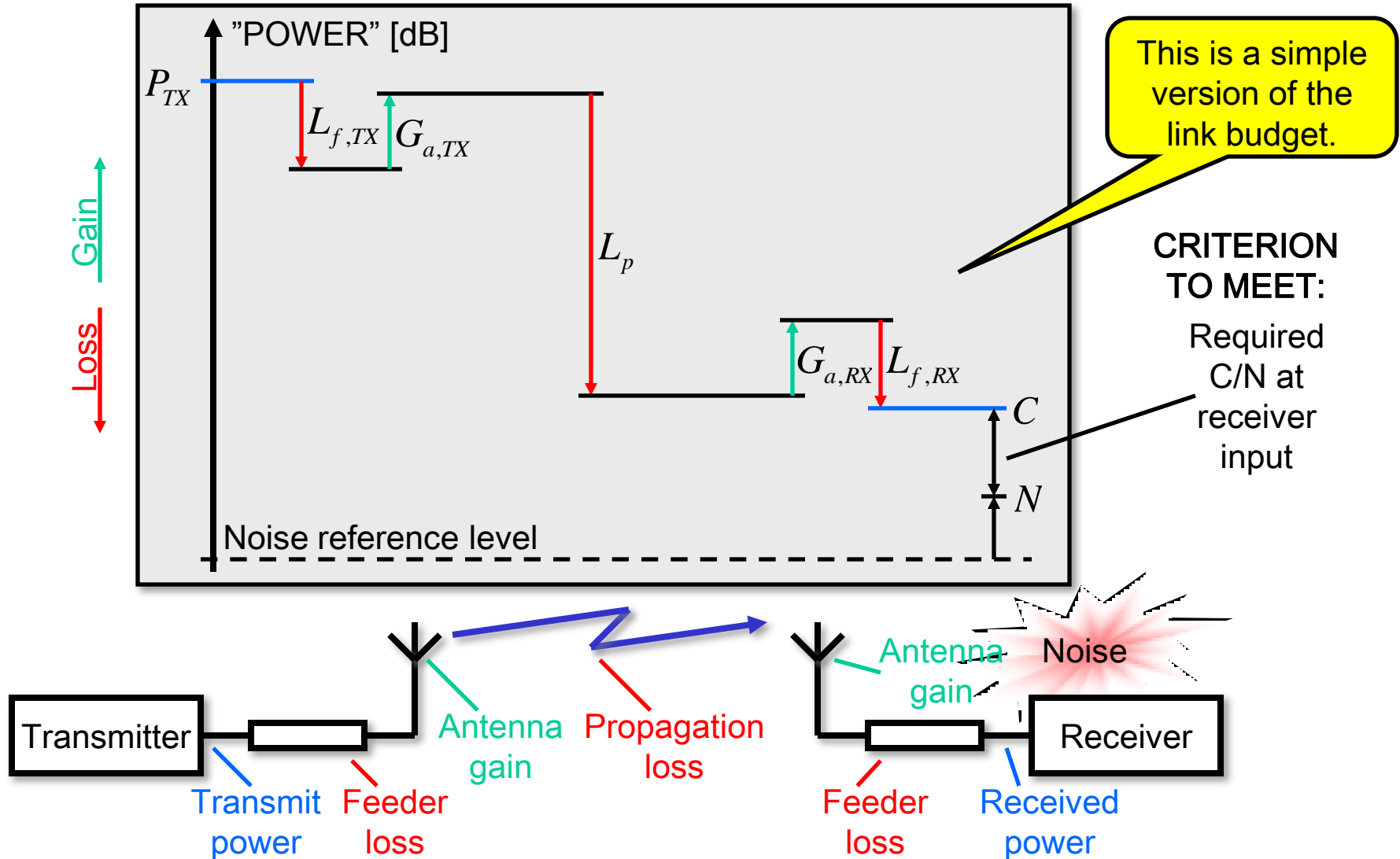
Noise- and interference limited systems

Basics of link budgets

- Link budgets show how different components and propagation processes influence the available SNR
- Link budgets can be used to compute, e.g., required transmit power, possible range of a system, or required receiver sensitivity
- Link budgets can be most easily set up using logarithmic power units (dB)

SINGLE LINK

The link budget – a central concept



dB in general

When we convert a measure X into decibel scale, we always divide by a reference value X_{ref} :

$$\frac{X \mid_{\text{non-dB}}}{X_{\text{ref}} \mid_{\text{non-dB}}}$$

Independent of the dimension of X (and X_{ref}), this value is always dimensionless.

The corresponding dB value is calculated as:

$$X \mid_{\text{dB}} = 10 \log \left(\frac{X \mid_{\text{non-dB}}}{X_{\text{ref}} \mid_{\text{non-dB}}} \right)$$

Power

We usually measure power in Watt (W) and milliWatt [mW]

The corresponding dB notations are dB and dBm

	Non-dB	dB
Watt:	$P _W$	$P _{dB} = 10 \log \left(\frac{P _W}{1 _W} \right) = 10 \log(P _W)$
milliWatt:	$P _{mW}$	$P _{dBm} = 10 \log \left(\frac{P _{mW}}{1 _{mW}} \right) = 10 \log(P _{mW})$
RELATION:	$P _{dBm} = 10 \log \left(\frac{P _W}{0.001 _W} \right) = 10 \log(P _W) + 30 _{dB} = P _{dB} + 30 _{dB}$	

Example: Power

Sensitivity level of GSM RX: $6.3 \times 10^{-14} \text{ W} = -132 \text{ dB}$ or -102 dBm

Bluetooth TX: $10 \text{ mW} = -20 \text{ dB}$ or 10 dBm

GSM mobile TX: $1 \text{ W} = 0 \text{ dB}$ or 30 dBm

GSM base station TX: $40 \text{ W} = 16 \text{ dB}$ or 46 dBm

Vacuum cleaner: $1600 \text{ W} = 32 \text{ dB}$ or 62 dBm

Car engine: $100 \text{ kW} = 50 \text{ dB}$ or 80 dBm

TV transmitter (Hörby, SVT2): $1000 \text{ kW ERP} = 60 \text{ dB}$ or 90 dBm ERP

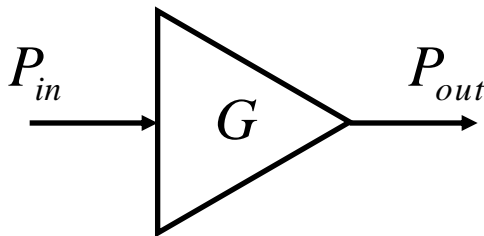
Nuclear powerplant (Barsebäck): $1200 \text{ MW} = 91 \text{ dB}$ or 121 dBm



ERP – Effective
Radiated Power

Amplification and attenuation

(Power) Amplification:

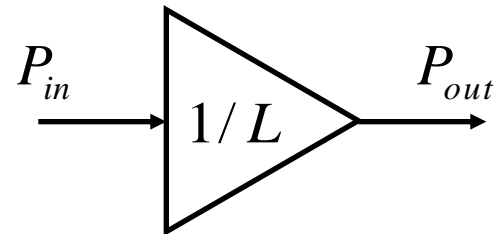


$$P_{out} = GP_{in} \Rightarrow G = \frac{P_{out}}{P_{in}}$$

The amplification is already dimension-less and can be converted directly to dB:

$$G|_{dB} = 10 \log_{10} G$$

(Power) Attenuation:



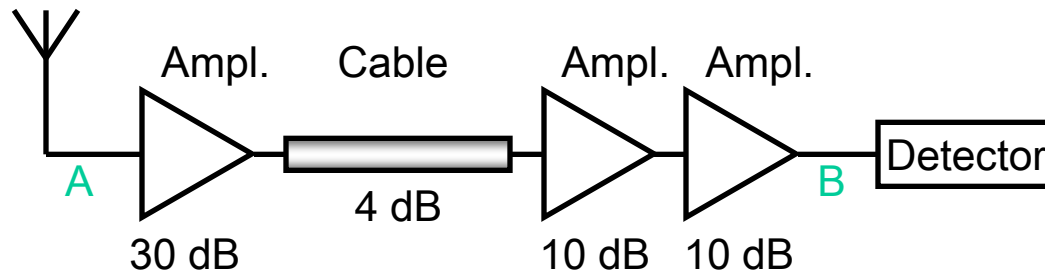
$$P_{out} = \frac{P_{in}}{L} \Rightarrow L = \frac{P_{in}}{P_{out}}$$

The attenuation is already dimension-less and can be converted directly to dB:

$$L|_{dB} = 10 \log_{10} L$$

Note: It doesn't matter if the power is in mW or W. Same result!

Example: Amplification and attenuation

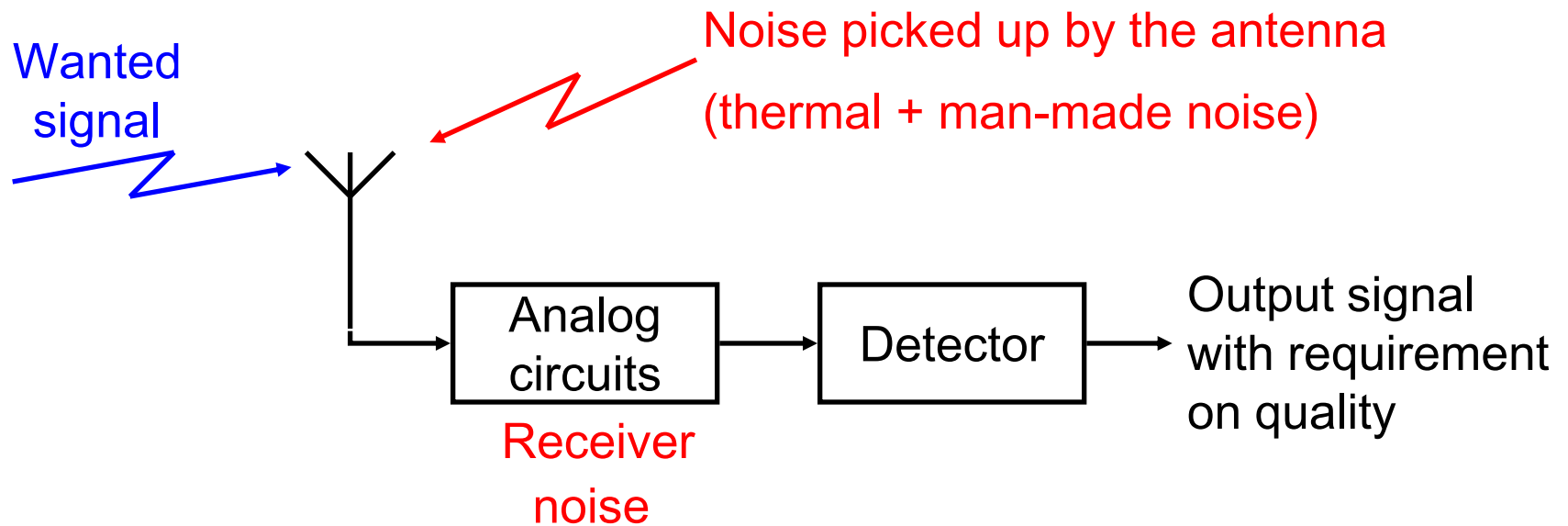


The total amplification of the (simplified) receiver chain (between A and B) is

$$G_{A,B} \mid_{dB} = 30 - 4 + 10 + 10 = 46$$

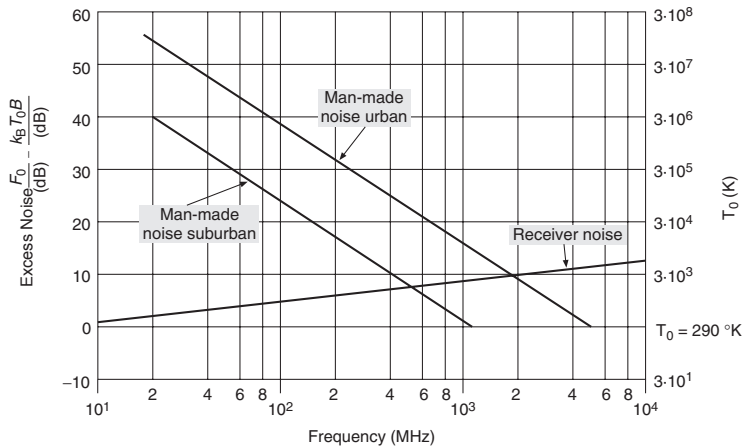
Noise sources

The noise situation in a receiver depends on several noise sources



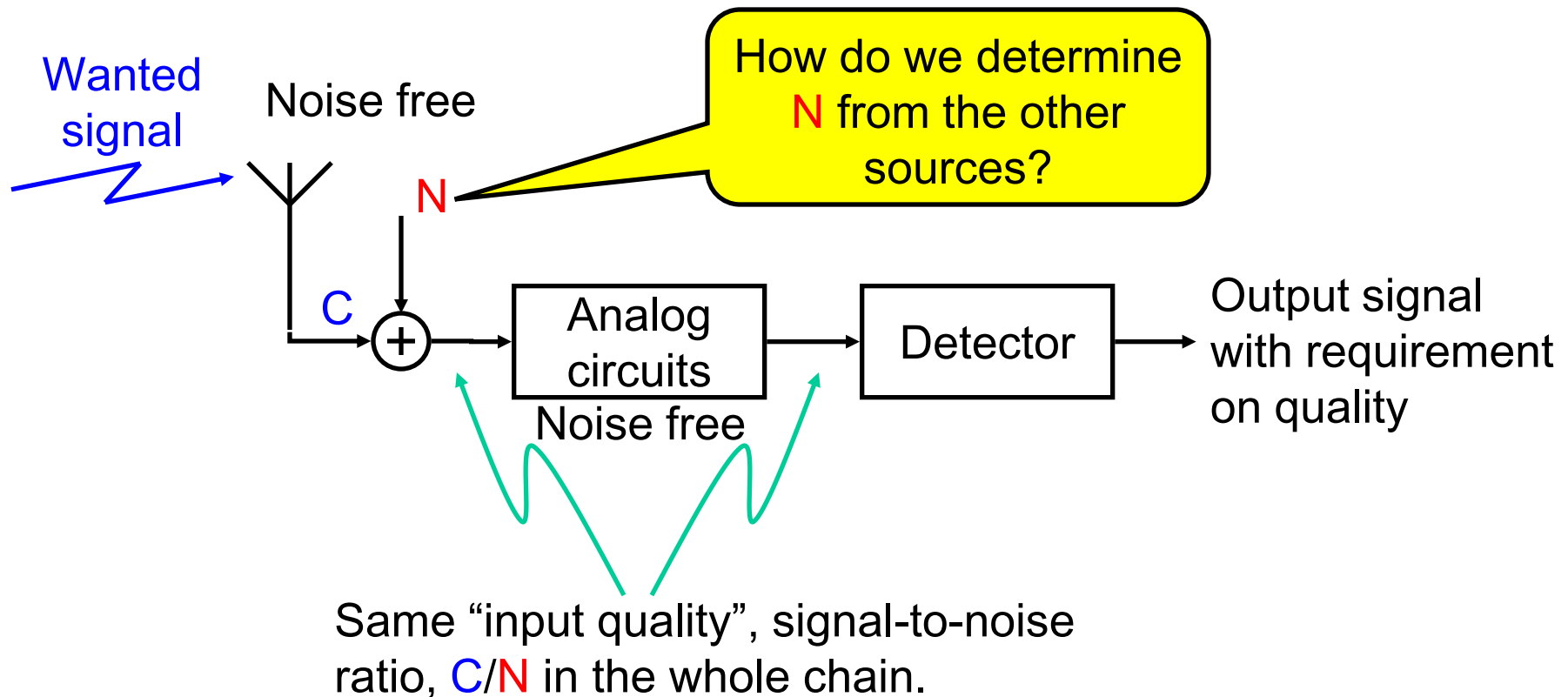
$$N_0 = k_B T_e$$

- N_0 thermal noise in Watt/Hz,
- $k_B = 1.38 \times 10^{-23}$ Joule/Kelvin Boltzman constant (Joule = Watt \times second),
- T_e environment temperature in Kelvin
- Average temperature on surface of the earth $T_e \approx 300\text{K}$
 $\Rightarrow N_0 \approx -174\text{dBm/Hz}$
- For a receiver with a total bandwidth of B Hz, the total received noise is $P_n = N_0 B$
- Example: $B = 200\text{kHz} \Rightarrow P_n = -174 + 10 \log_{10}(B) \approx -121 \text{ dBm}$



Receiver noise: Equivalent noise source

To simplify the situation, we replace all noise sources with a single equivalent noise source.



Receiver noise: Noise sources (1)

The power spectral density of a noise source is usually given in one of the following three ways:

1) Directly [W/Hz]:

N_s

2) Noise **temperature** [Kelvin]:

T_s

3) Noise **factor** [1]:

F_s

This one is sometimes given in dB and called **noise figure**.

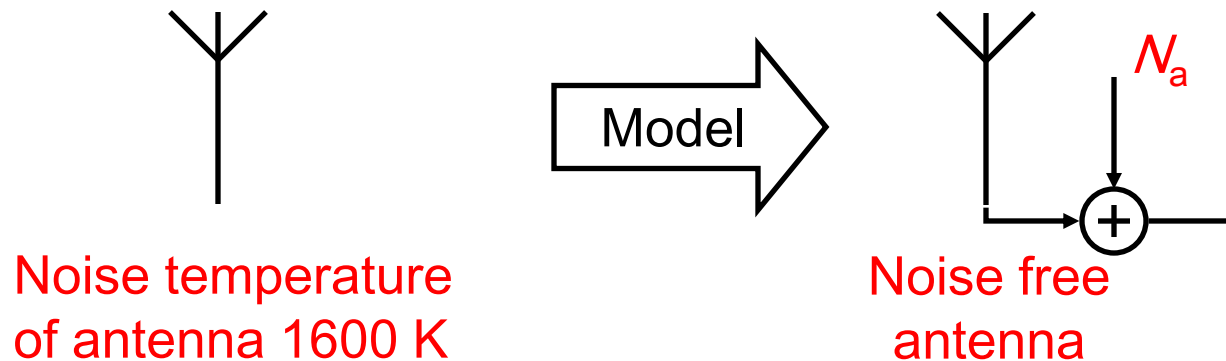
The relation between the three is

$$N_s = kT_s = kF_sT_0$$

where k is **Boltzmann's constant** (1.38×10^{-23} J/?) and T_0 is the, so called, **room temperature** of 290 K (17° C).

Receiver noise: Noise sources (2)

Antenna example



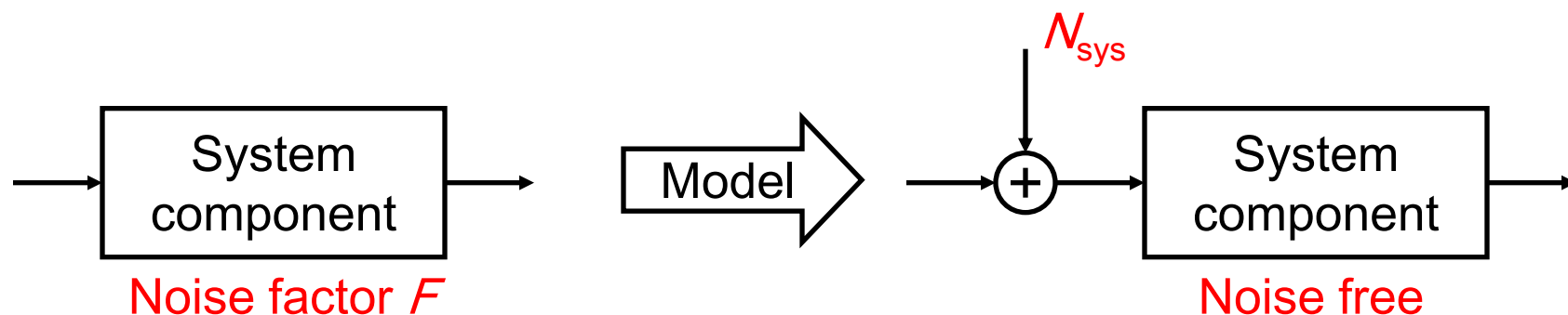
Power spectral density of antenna noise is

$$N_a = 1.38 \times 10^{-23} \times 1600 = 2.21 \times 10^{-20} \text{ W/Hz} = -196.6 \text{ dB[W/Hz]}$$

and its noise factor/noise figure is

$$F_a = 1600 / 290 = 5.52 = 7.42 \text{ dB}$$

Receiver noise: System noise



Due to a definition of noise factor (in this case) as the ratio of noise powers on the output versus on the input, when a resistor in room temperature ($T_0=290$ K) generates the input noise, the PSD of the equivalent noise source (placed **at the input**) becomes

$$N_{sys} = k \underbrace{(F - 1) T_0}_{\text{Equivalent noise temperature}} \text{ W/Hz}$$

Don't use dB value!

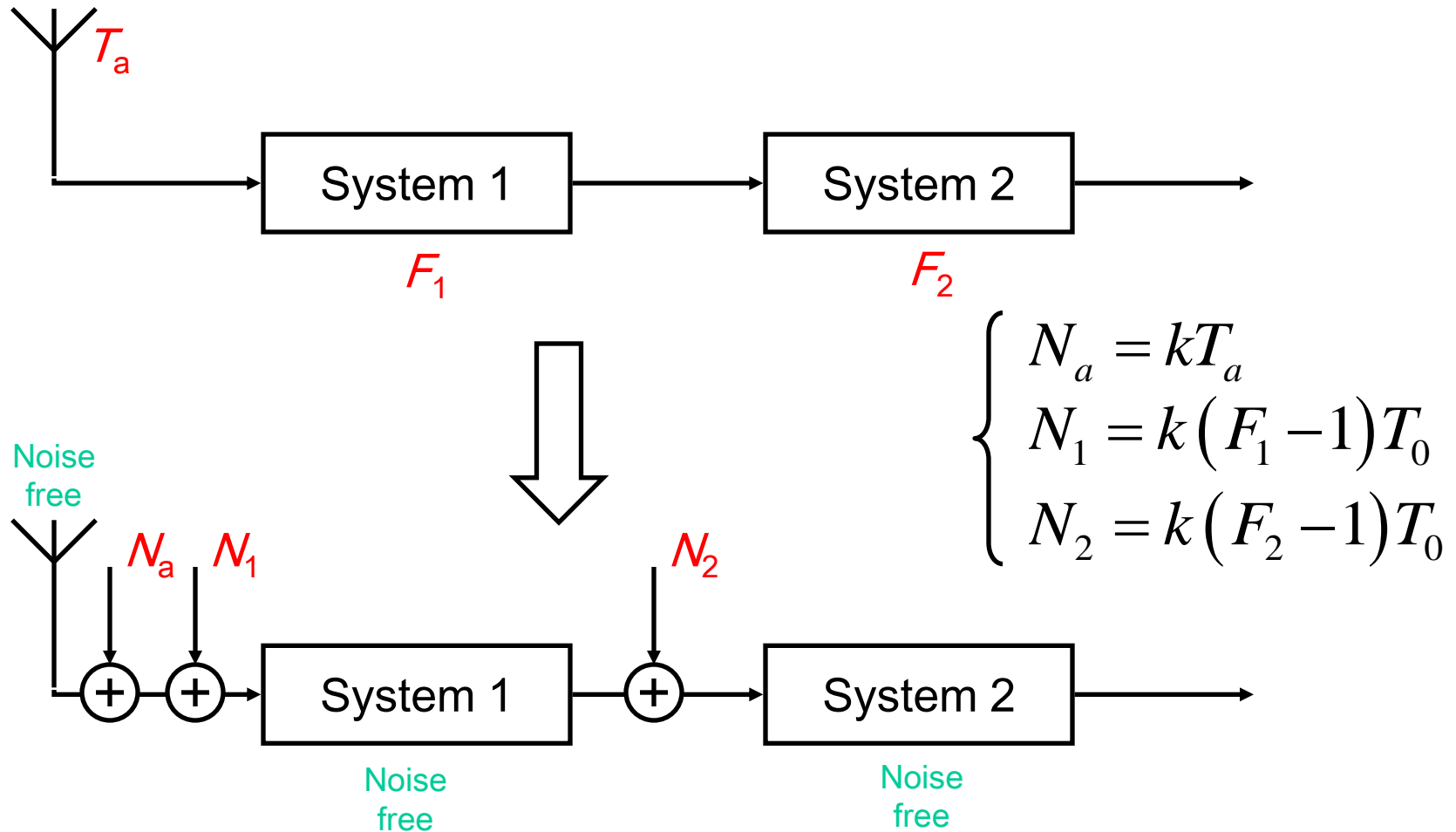
Equivalent noise temperature

A GSM receiver has a bandwidth of 200kHz and requires that its input SNR is at least 10dB when the input signal is -104dBm.

- 1 What is the thermal noise level at the receiver?
- 2 What is the maximum permitted value of the receiver noise figure?
- 3 What is the equivalent input noise temperature of such a receiver?

Receiver noise: Sev. noise sources (1)

A simple example

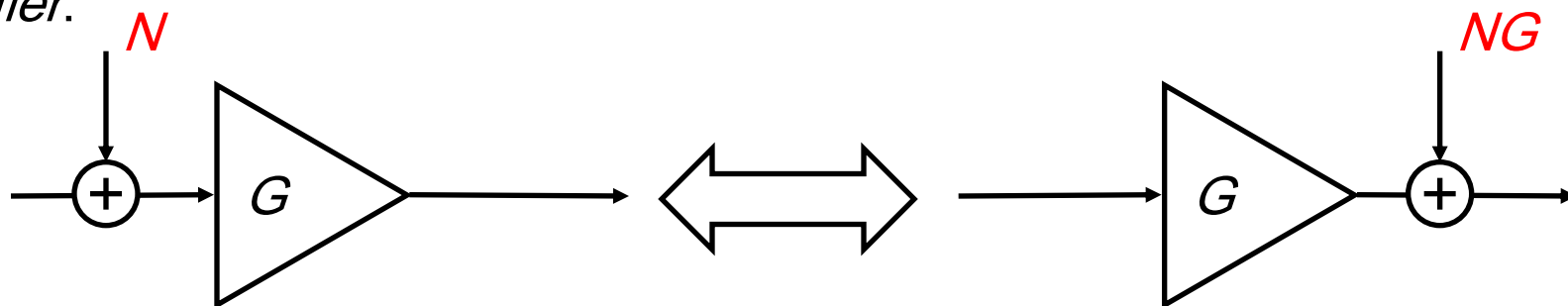


Receiver noise: Sev. noise sources (2)

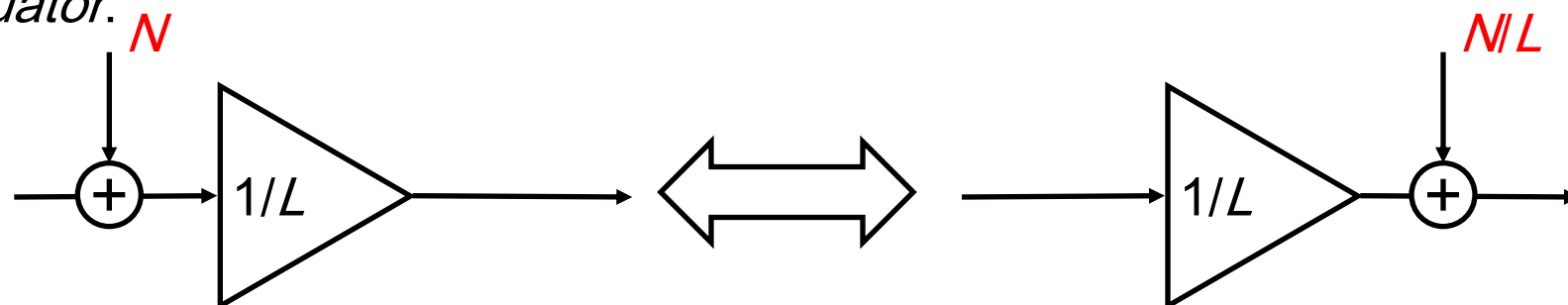
After extraction of the noise sources from each component, we need to move them to one point.

When doing this, we must compensate for amplification and attenuation!

Amplifier.

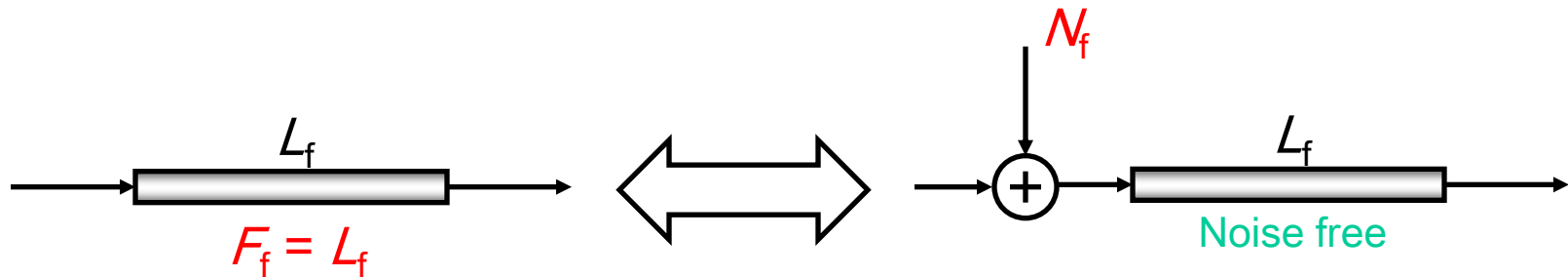


Attenuator.



Pierce's rule

A passive attenuator, in this case a feeder, has a noise figure equal to its attenuation.



$$N_f = k(F_f - 1)T_0 = k(L_f - 1)T_0$$

Remember to
convert *from* dB!

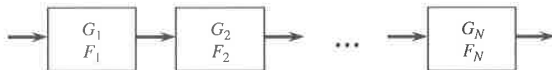


Figure 5.3: A cascade of two-port elements

- Total Gain: $G = G_1 \cdot G_2 \cdots G_N$
- Total Noise Figure: $F = F_1 + \frac{F_2 - 1}{G_1} + \frac{F_3 - 1}{G_1 G_2} + \cdots + \frac{F_N - 1}{G_1 \cdots G_{N-1}}$

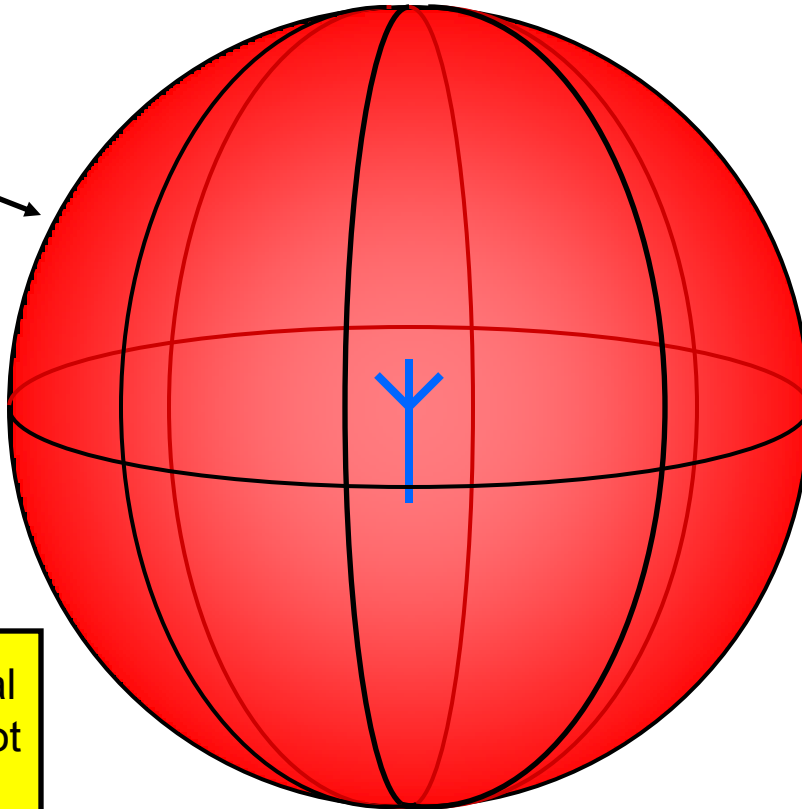
A receiver is made up of three main elements: a preamplifier, a mixer, and an IF amplifier with noise figures of 3, 6, and 10 dB.

- If the overall gain of the receiver is 30 dB, and the IF amplifier gain is 10 dB, what is the minimum gain of the preamplifier to achieve an overall noise figure of no more than 5 dB?
- If its gain is set to this minimum, what would the system noise figure become if the noise figure of the amplifier is increased to 20 dB?

The isotropic antenna

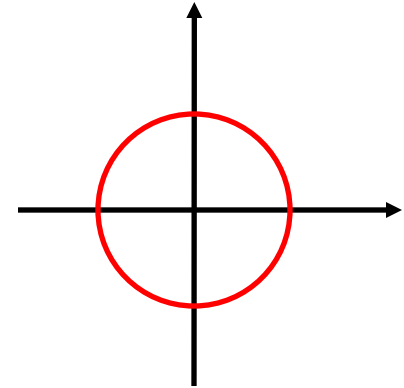
The isotropic antenna radiates equally in all directions

Radiation pattern is spherical

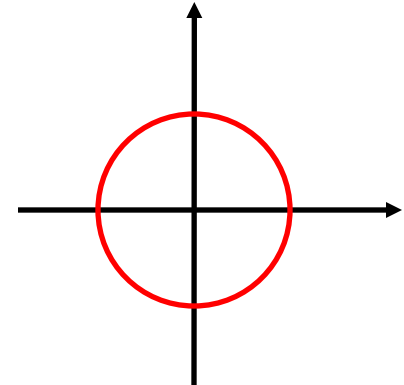


This is a theoretical antenna that cannot be built.

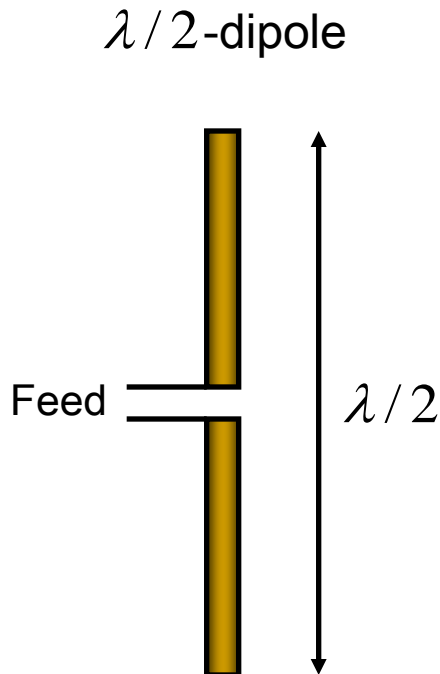
Elevation pattern



Azimuth pattern



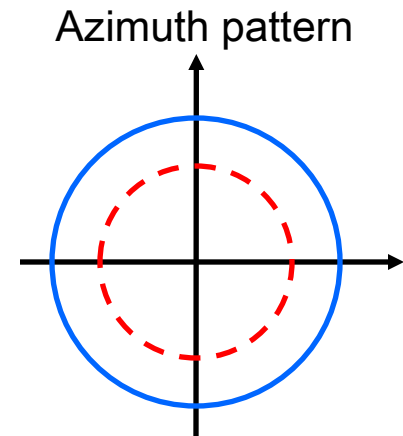
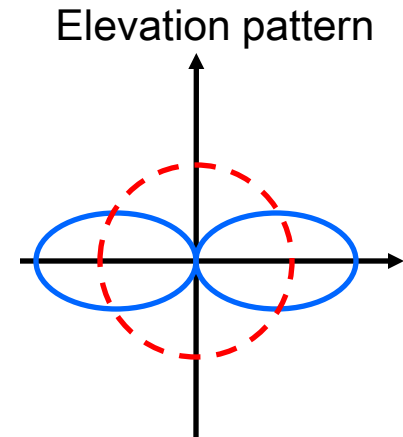
The dipole antenna



This antenna does not radiate straight up or down. Therefore, more energy is available in other directions.

THIS IS THE PRINCIPLE
BEHIND WHAT IS CALLED
ANTENNA GAIN.

A dipole can be of any length, but the antenna patterns shown are only for the $\lambda/2$ -dipole.

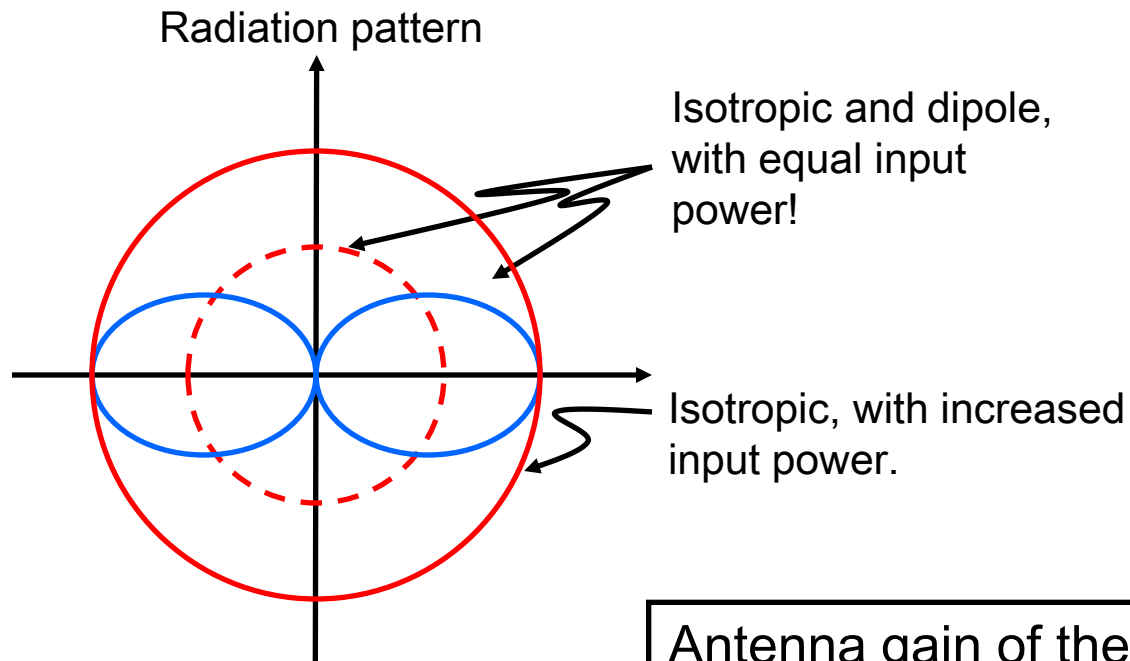


-- Antenna pattern of isotropic antenna.

Antenna gain (principle)

Antenna gain is a relative measure.

We will use the isotropic antenna as the reference.



The amount of increase in input power to the isotropic antenna, to obtain the same maximum radiation is called the **antenna gain!**

Antenna gain of the $\lambda/2$ dipole is **2.15 dB.**

A note on antenna gain

Sometimes the notation **dB*i*** is used for antenna gain (instead of dB).

The "i" indicates that it is the gain relative to the isotropic antenna (**which we will use in this course**).

Another measure of antenna gain frequently encountered is **dB*d***, which is relative to the $\lambda/2$ dipole.

$$G|_{dB_i} = G|_{dB_d} + 2.15$$

Be careful! Sometimes it is not clear if the antenna gain is given in dBi or dBd.

EIRP: Effective Isotropic Radiated Power

EIRP = Transmit power (fed to the antenna) + antenna gain

$$EIRP|_{dB} = P_{TX|dB} + G_{TX|dB}$$

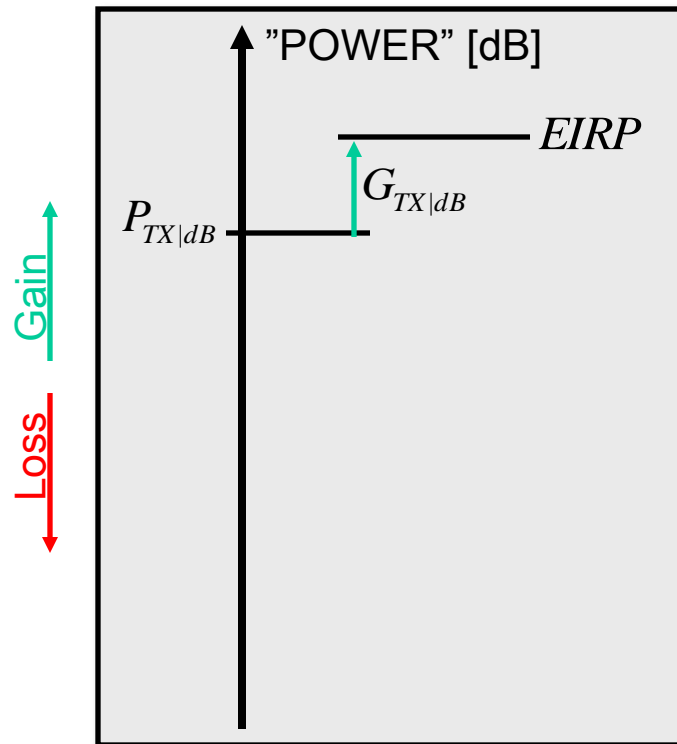
Answers the questions:

How much transmit power would we need to feed an isotropic antenna to obtain the same maximum on the radiated power?

How "strong" is our radiation in the maximal direction of the antenna?

This is the more important one, since a limit on EIRP is a limit on the radiation in the maximal direction.

EIRP and the link budget



$$EIRP|_{dB} = P_{TX|dB} + G_{TX|dB}$$

- Path loss

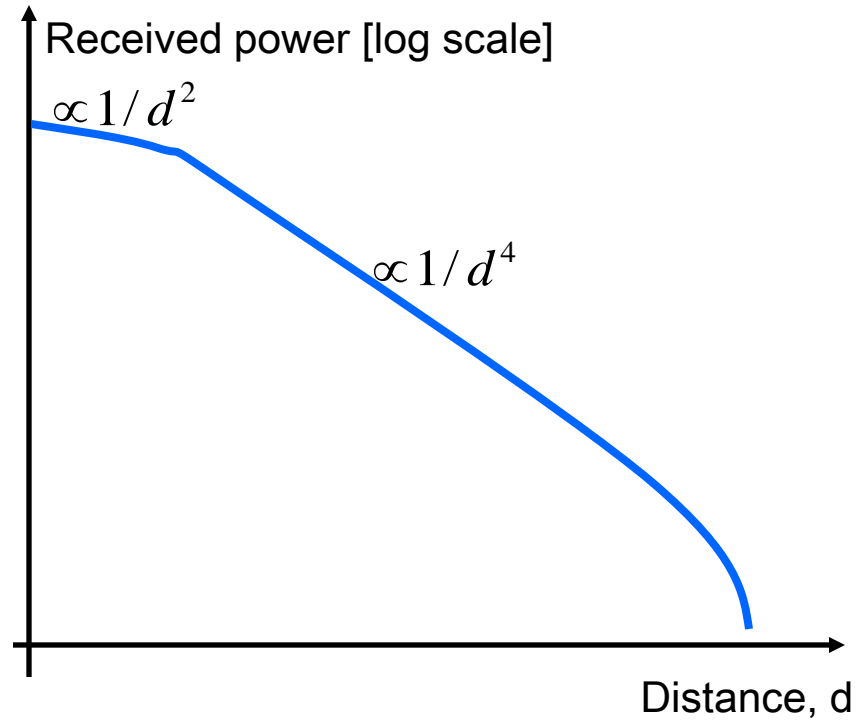
$$P_{\text{RX}}(d) = \frac{P_{\text{TX}}}{PL(d)} \quad \text{or} \quad P_{\text{RX}}(d)|_{\text{dB}} = P_{\text{TX}}|_{\text{dB}} - PL(d)|_{\text{dB}}$$

- Simple model

$$PL(d) = \left(\frac{4\pi d}{\lambda} \right)^2 \quad 0 \leq d \leq d_{\text{break}}$$

$$PL(d) = PL(d_{\text{break}}) \left(\frac{d}{d_{\text{break}}} \right)^n \quad d > d_{\text{break}}$$

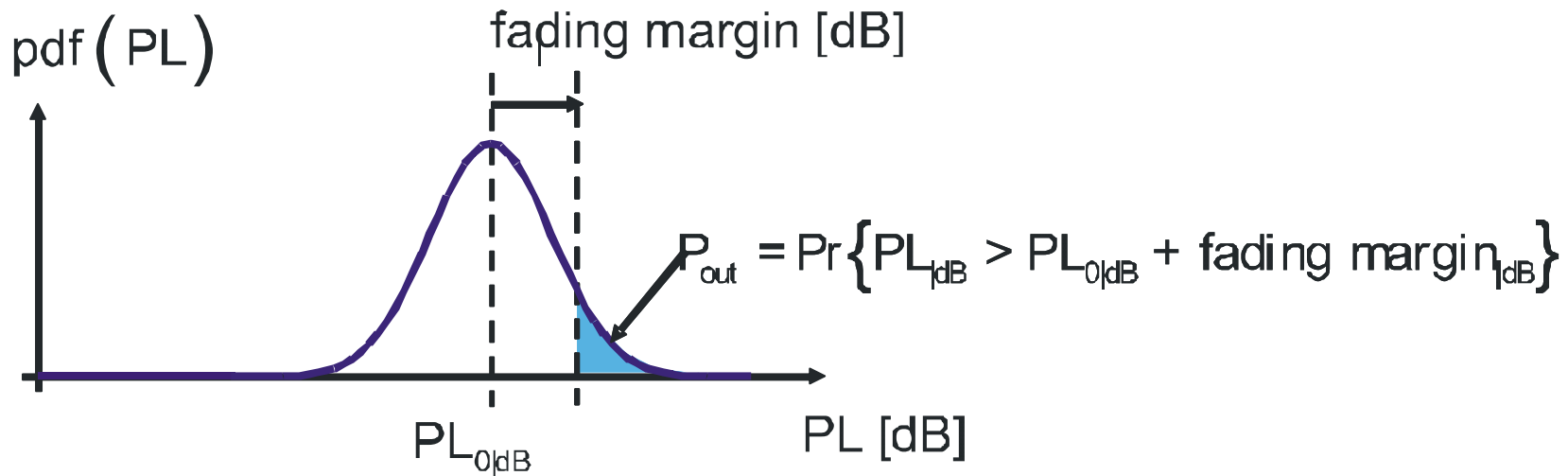
Path loss



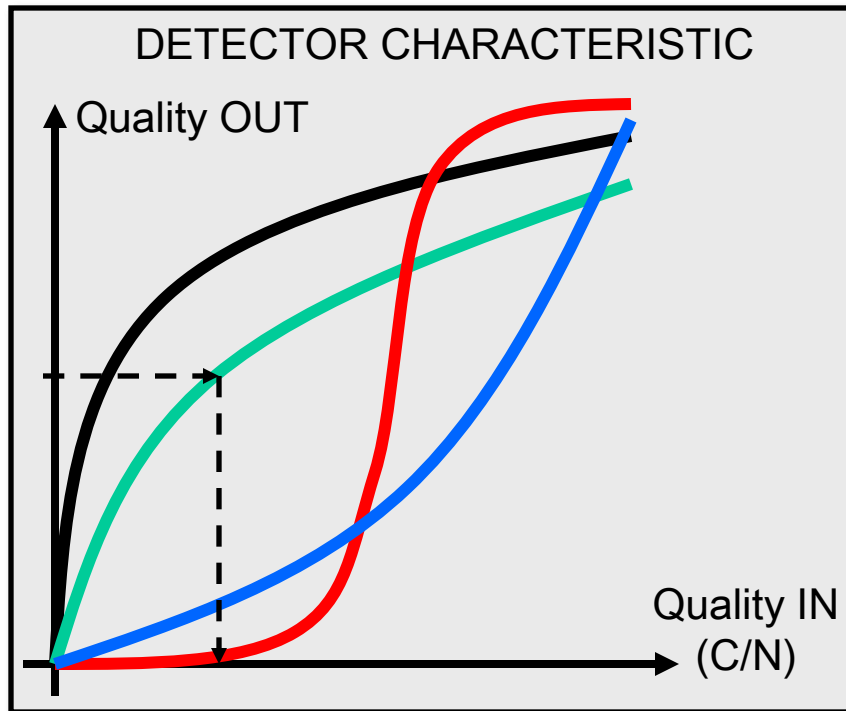
Fading margin

Received signal strength is not deterministic, more like a random process

-> include Fading margin to account for this uncertainty



Required C/N – another central concept

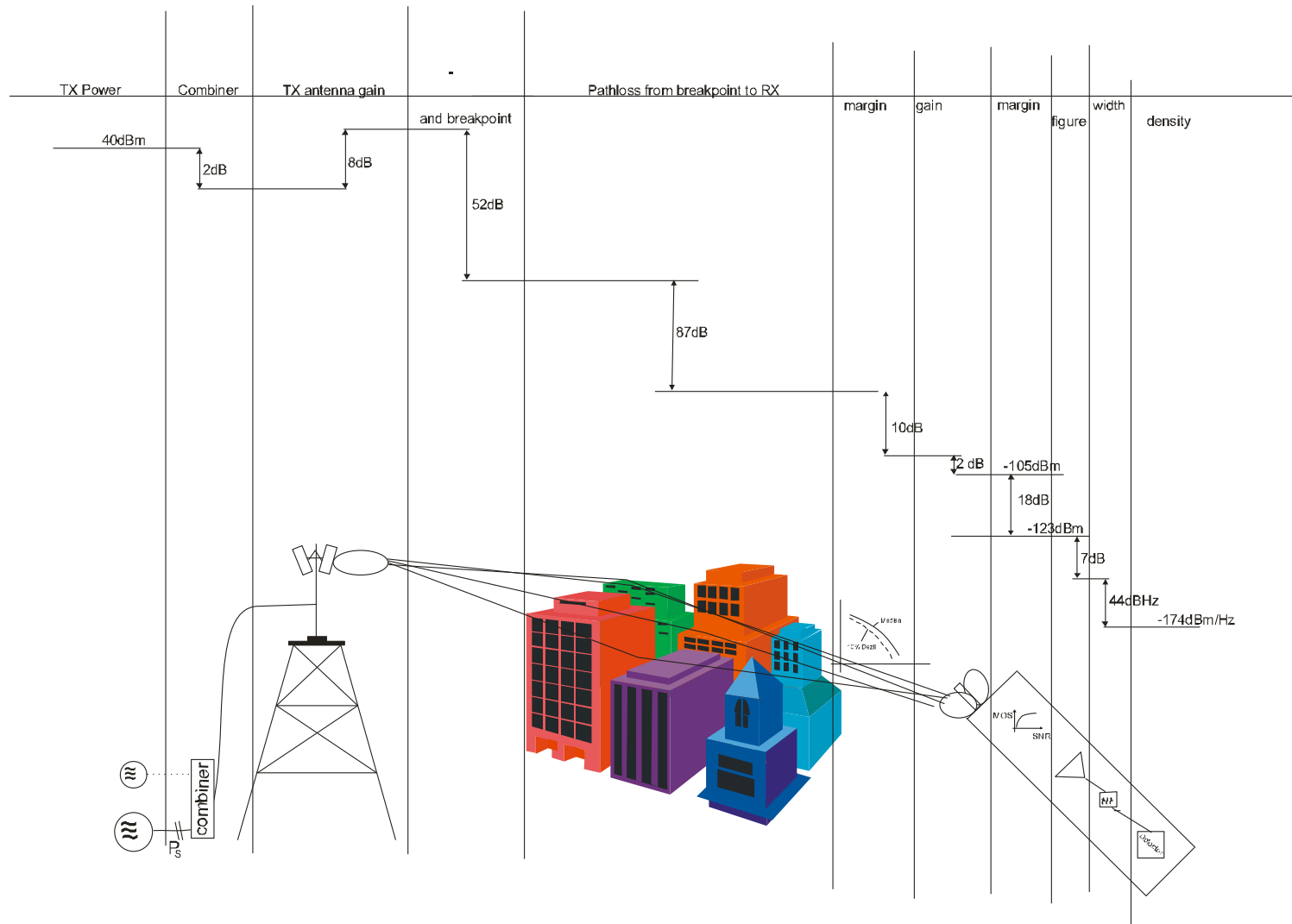


The detector characteristic is different for different system design choices.

REQUIRED QUALITY OUT:

Audio SNR
Perceptive audio quality
Bit-error rate
Packet-error rate
etc.

Example for link budget



Consider a GSM system with the following characteristics:

- Carrier frequency $f_c = 900\text{MHz}$,
- Bandwidth $B = 200\text{kHz}$,
- Operating temperature $T = 300\text{ K}$,
- Antenna gains $G_{\text{TX}} = 8\text{ dB}$ and $G_{\text{RX}} = -2\text{ dB}$,
- Cable losses at TX $L_{\text{TX}} = 2\text{ dB}$,
- Receiver noise figure $F = 7\text{ dB}$.

The propagation characteristics are

- The path loss exponent is $n = 3.8$,
- the breakpoint distance is 10 m ,
- the fading margin is 10 dB .

The required operating SNR is 8 dB , the desired range of coverage 2 km . **What is the minimum TX power?**



Claude Elwood Shannon,
“A mathematical theory of communication,”
The Bell System Technical Journal, vol. 27, pp. 379–432, July
1948.