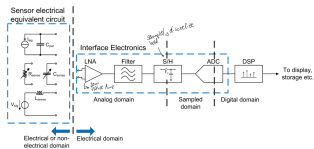


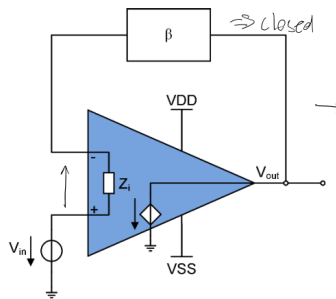
Sensor Principles

A generic sensor interface



OPAMP Basics

Opamps in feedback



Gain : $T(\omega) = \beta \cdot A_0(\omega)$

Phase Margin:

$$PM = 180^\circ + \angle(T(\omega_1))|_{|T(\omega_1)|=1}$$

Gain margin: $GM =$

$$20 \cdot \log_{10}\left(\frac{1}{|T(\omega_2)|}\right)|_{\angle(T(\omega_2))=-180^\circ}$$

Generic Transfer function

$$V_x = \beta \cdot V_{out} =$$

$$\beta \cdot A(\omega) \cdot (V_{in} - V_x) =$$

$$\beta \cdot A(\omega) \cdot (V_{in} - \beta \cdot V_{out})$$

$$\frac{V_{out}}{V_{in}} = \frac{A(\omega)}{1 + \beta \cdot A(\omega)} \approx \frac{1}{\beta}$$

Negative Feedback and linear operation

$A(\omega) \gg 1 \Rightarrow$ virtual short at the

input $\Delta V \approx 0$

$Z_i \rightarrow \infty \Rightarrow i_{oa} \approx 0$

Voltage Drive \rightarrow negative feedback

Current drive \rightarrow positive feedback

Errors and Noise

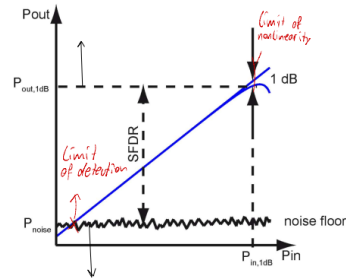
Limit of detection(LOD):

minimum measurable input

amplitude ($SNR \approx 0$)

Dynamic range(DR): ratio of

max and min amplitude within inaccuracy levels.



Error types:

Deterministic: source loading, offset, gain error

Random: thermal noise, 1/f noise

Quantification:

Absolute : $\Delta x = |\hat{x} - x_0|$

Relative: $\left| \frac{\Delta x}{x_0} \right| = \left| \frac{\hat{x} - x_0}{\hat{x}} \right|$

Max inaccuracy:

$$\Delta x_{max} |x \in [\hat{x} - \Delta x_{max}, \hat{x} + \Delta x_{max}]$$

Error Propagation

$$y = f(x_1, x_2, \dots, x_N)$$

Deterministic fluctuations of $x_i \rightarrow$

total error:

$$\Delta y \approx \sum_{i=1}^N \frac{\partial f}{\partial x_i} \cdot \Delta x_i$$

Partial derivative $\frac{\partial f}{\partial x_i}$ is called

sensitivity

Additive errors are best specified absolute and multiplicative errors are best specified relative

Interference:

Unwanted coupling of external signal

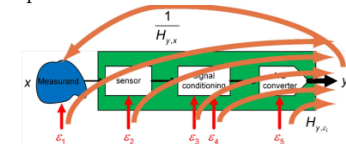
Noise: random fluctuations from setup \rightarrow can be modeled as error sources

Combining Error sources

Output referred noise

Effect of an error-source on the output

Input referred noise Equivalent effect of the error-source on the input



Find TF from ES to output

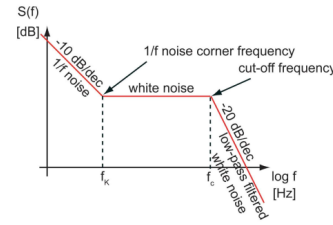
$$y : H_{y,ES} \Rightarrow y_{out, \epsilon_2} = H_{y, \epsilon_2} \cdot \epsilon_2$$

Refer result back to input with

$$H_{y,x} : x_{\epsilon_2} = \frac{y_{out, \epsilon_2}}{H_{y,x}} = H_{y, \epsilon_2} \cdot \frac{\epsilon_2}{H_{y,x}}$$

Lin. System noise:

$$PSD = S_y(f) = |H(f)|^2 \cdot S_x(f)$$



Noise Types

Thermal noise: excitation of charge carriers(white)

Shot noise: carriers randomly crossing the barrier, dependent on DC bias and white

Flicker Noise: due to traps in semiconductor. 1/f spectral density.

MOS trans at low freq.

Thermalnoise Theorem: Every

closed system at temp. T has

average. Energy of $kT/2 \rightarrow$

$$S_u(f) = 2kTR \text{ (double-sided) } \text{ math}$$

$$S_u(f) = 4kTR \text{ (single-sided) } \text{ physics}$$

Langevin Approach kt/C noise

PSD noise voltage of V_n

$$= S_{v_n}(f) = 4kT \cdot \Re\{Z(j2\pi f)\}$$

$$\rightarrow \overline{V_n^2} = kT \left[\frac{1}{C_\infty} - \frac{1}{C_0} \right] = \frac{kT}{C}$$

MOS IRN

$$S_{\Delta V_{nG-tot}^2} = 4kT \cdot$$

$$R_{nG-tot}, R_{nG-tot} = \frac{\rho}{W \cdot L \cdot f} + \frac{\gamma_{nD}}{G_m}$$

with GateExcess Noise factor

$$\gamma_{nD} = (n = 1.3) \cdot [0.5WI; 2/3SI]$$

Bipolar Trans IRN:

$$S_{\Delta V_{nR}^2} = 4kT \cdot R_B$$

Noise Analysis

small-signal-equivalent is valid.

Total ORN:

$$S_{n, out}(f) = \sum_{k=1}^N |H_k(f)|^2 \cdot S_{nk}(f)$$

N uncorrelated NS.

$H_k(f)$ TF from NS to output. IRN:

$$S_{V_{neq}}(f) = \frac{S_{V_{nout}}(f)}{|A(f)|^2} \text{ with } A(f) \text{ is TF}$$