

Laplace Transform Examples

CPE 381 Foundations of Signals & Systems
for Computer Engineers
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One-sided Laplace Transform

Function of Time

Function of s , ROC

$$\delta(t)$$

$$1, \text{ whole } s\text{-plane}$$

$$u(t)$$

$$\frac{1}{s}, \operatorname{Re}[s] > 0$$

$$r(t)$$

$$\frac{1}{s^2}, \operatorname{Re}[s] > 0$$

$$e^{-at}u(t), a > 0$$

$$\frac{1}{s+a}, \operatorname{Re}[s] > -a$$

$$\cos(\Omega_0 t)u(t)$$

$$\frac{s}{s^2 + \Omega_0^2}, \operatorname{Re}[s] > 0$$

$$\sin(\Omega_0 t)u(t)$$

$$\frac{\Omega_0}{s^2 + \Omega_0^2}, \operatorname{Re}[s] > 0$$

$$e^{-at} \cos(\Omega_0 t)u(t), a > 0$$

$$\frac{s+a}{(s+a)^2 + \Omega_0^2}, \operatorname{Re}[s] > -a$$

$$e^{-at} \sin(\Omega_0 t)u(t), a > 0$$

$$\frac{\Omega_0}{(s+a)^2 + \Omega_0^2}, \operatorname{Re}[s] > -a$$

$$2A e^{-at} \cos(\Omega_0 t + \theta)u(t), a > 0$$

$$\frac{A \angle \theta}{s+a-j\Omega_0} + \frac{A \angle -\theta}{s+a+j\Omega_0}, \operatorname{Re}[s] > -a$$

One-sided Laplace Transform

Causal functions and constants: $\alpha f(t) \iff \alpha F(s)$

Linearity: $\alpha f(t) + \beta g(t) \iff \alpha F(s) + \beta G(s)$

Time shifting: $f(t - \alpha) \iff e^{-\alpha s} F(s)$

Frequency shifting: $e^{\alpha t} f(t) \iff F(s - \alpha)$

Multiplication by t: $tf(t) \iff -\frac{dF(s)}{ds}$

Derivative: $\frac{df(t)}{dt} \iff sF(s) - f(0-)$

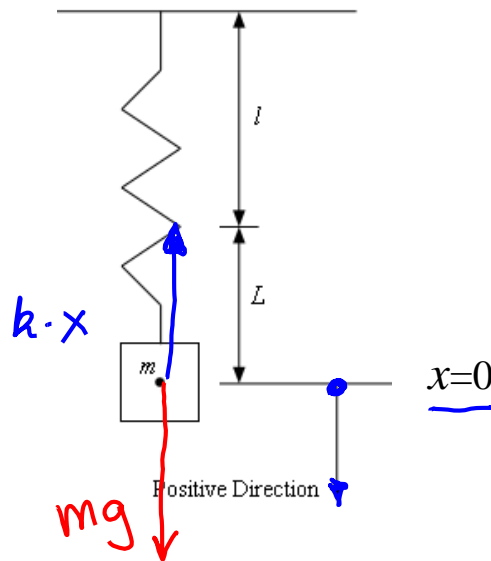
Second derivative: $\frac{d^2 f(t)}{dt^2} \iff s^2 F(s) - sf(0-) - f^{(1)}(0)$

Integral: $\int_{0-}^t f(t') dt \iff \frac{F(s)}{s}$

Expansion/Contraction: $f(\alpha t) \alpha \neq 0 \iff \frac{1}{|\alpha|} F\left(\frac{s}{\alpha}\right)$

Suspended Weight #1

- ◆ Write differential equation describing displacement x of suspended weight m on spring with elastic constant k .



At any time, sum of all forces is equal to zero

$$m\ddot{x} + c\dot{x} + kx = 0$$

With initial conditions

$$x(0)[m] \text{ and } \dot{x}(0)$$

By using Laplace transform

$$\mathcal{L}(m\ddot{x} + c\dot{x} + kx) = ms^2X(s) - msx(0) - m\dot{x}(0) + csX(s) - cx(0) + kX(s) = 0$$

$$(ms^2 + cs + k)X(s) = msx(0) + cx(0)$$

$$X(s) = \frac{msx(0) + cx(0)}{ms^2 + cs + k} = \frac{sx(0) + \frac{c}{m}x(0)}{s^2 + \frac{c}{m}s + \frac{k}{m}}$$

Suspended Weight #2

Example #1:

$$m = 1 \text{ [kg]}, k = 2 \left[\frac{\text{kg}}{\text{s}^2} \right], c = 2 \left[\frac{\text{kg}}{\text{s}} \right], x(0) = 1 \text{ [m]}, \text{ and } \dot{x}(0) = 0 \left[\frac{\text{m}}{\text{s}^2} \right]$$

and

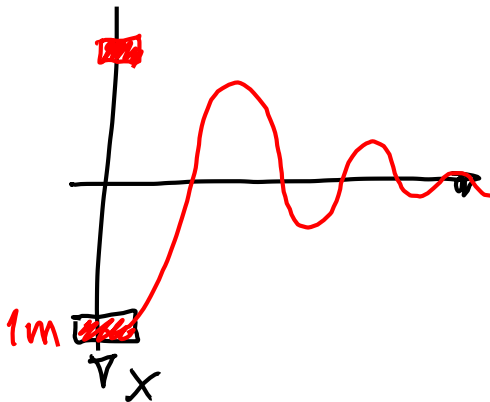
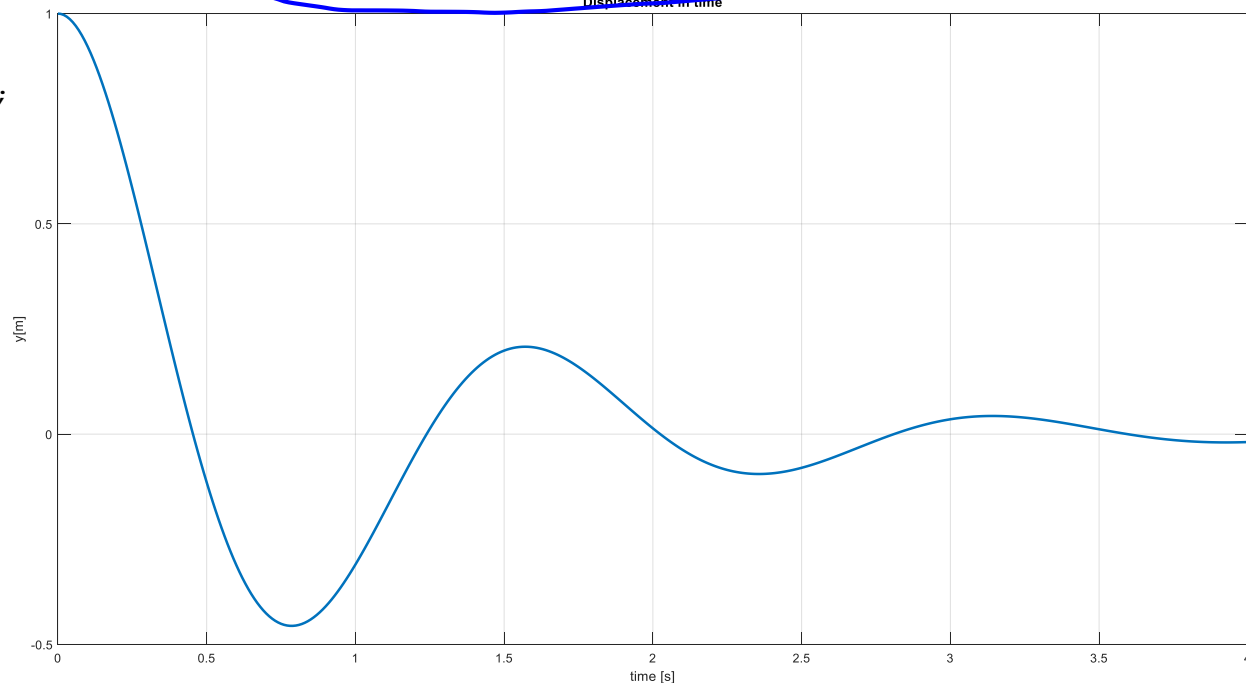
$$X(s) = \frac{s+2}{s^2+2s+17} = \frac{s+2}{(s+1)^2+16}$$

$$x(t) = \mathcal{L}^{-1}(X(s)) = e^{-t} \left(\cos(4t) + \frac{1}{4} \sin(4t) \right)$$

$$= \frac{s+1}{(s+1)^2+16} + \frac{1}{s_0} \frac{1}{s_0^2} \frac{1}{(s+1)^2+16}$$

$\Omega_0 = 4$

```
% Example: differential equation
t=0:0.01:4;
x=exp(-t).*(cos(4*t)+0.25*sin(4*t));
plot(t,x),
title('Displacement in time'),
xlabel('time [s]'),ylabel('y[m]')
grid
```



Suspended Weight #3

Example #2: A 1 kg weight is hung on the end of a vertically suspended spring, thereby stretching the spring $L=10$ cm. The weight is raised 5 cm above its equilibrium position and released from rest at time $t=0$. Find the displacement x of the weight from its equilibrium position at time t . Use $g=10\text{m/s}^2$.

$$F = kL, \quad k = \frac{F}{L} = \frac{mg}{L} = \frac{1[\text{kg}] \ 10 \left[\frac{\text{m}}{\text{s}^2} \right]}{0.1[\text{m}]} = 100 \left[\frac{\text{kg}}{\text{s}^2} \right]$$

At any time, sum of all forces is equal to zero

$$m\ddot{x} + kx = 0$$

With initial conditions

$$x(0) = -0.05[\text{m}] \quad \dot{x}(0) = 0$$

By using Laplace transform

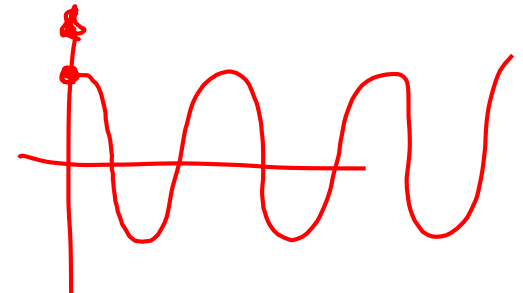
$$\mathcal{L}(m\ddot{x} + kx) = s^2X(s) - sx(0) - \dot{x}(0) + kX(s) = 0$$

$$(s^2 + 100)X(s) = -0.05s$$

$$X(s) = \frac{-0.05s}{s^2 + 100}$$

and

$$x(t) = \mathcal{L}^{-1}(X(s)) = -0.05 \cos(10t)$$



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Differential Equations

- ♦ A system with input $x(t)$ and output $y(t)$ is defined by the following differential equation:

$$\frac{d^2 y(t)}{dt^2} + 3 \frac{dy(t)}{dt} + 2y(t) = x(t)$$

Find the impulse response $h(t)$ and the unit-step response $s(t)$.

If $Y(s) = \mathcal{L}[y(t)]$ and $X(s) = \mathcal{L}[x(t)]$, then

$$Y(s) [s^2 + 3s + 2] = X(s)$$

To find impulse response, we let $x(t) = \delta(t)$, and $X(s) = 1$, then $Y(s)$ is equal to $H(s)$:

$$Y(s) = H(s) = \frac{1}{s^2 + 3s + 2} = \frac{1}{(s+1)(s+2)} = \frac{A}{s+1} + \frac{B}{s+2}$$

We find

$$A = H(s)(s+1) \Big|_{s=-1} = \frac{1}{-1+2} = 1$$

and

$$B = H(s)(s+2) \Big|_{s=-2} = \frac{1}{-2+1} = -1$$

therefore:

$$h(t) = [e^{-t} - e^{-2t}] \cdot u(t)$$

Similarly, unit step response is:

$$S(s) = \frac{H(s)}{s} = \frac{1}{s \cdot (s+1)(s+2)} = \frac{A}{s} + \frac{B}{s+1} + \frac{C}{s+2}$$

and $A=0.5$, $B=-1$, $C=0.5$, therefore:

$$s(t) = 0.5 \cdot u(t) - e^{-t} \cdot u(t) + 0.5e^{-2t} \cdot u(t)$$

$$x(t) = \delta(t) \quad X(s) = 1$$

$$y(t) = h(t)$$

$$Y(s) = H(s) \cdot X(s)$$

$$Y(s) = \frac{1}{s+1} - \frac{1}{s+2} = H(s)$$

$$h(t) = \mathcal{L}^{-1}[\quad]$$

$$x(t) = u(t) \Rightarrow X(s) = \frac{1}{s}$$

$$Y(s) = H(s) \cdot \frac{1}{s}$$

$$y(t) = \mathcal{L}^{-1}\{Y(s)\} = \mathcal{L}^{-1}\left\{H(s) \cdot \frac{1}{s}\right\}$$

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and $A=0.5$, $B= -1$, $C=0.5$, therefore:

$$s(t) = 0.5 \cdot u(t) - e^{-t} \cdot u(t) + 0.5e^{-2t} \cdot u(t)$$

Differential Equations

- ◆ The Laplace transform of the response is:

$$S(s) = H(s)X(s) = \frac{s}{s(s^2 + s + 1)} = \frac{1}{(s + 1/2)^2 + 3/4}$$

$\Omega_0 = \sqrt{\frac{3}{4}} = \frac{\sqrt{3}}{2}$

since (take a look at page 199)

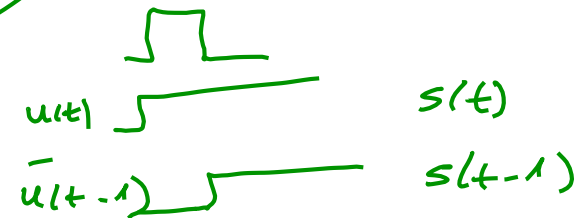
$$\mathcal{L}[Ae^{-\alpha t} \sin(\Omega_0 t \cdot u(t))] = \frac{A\Omega_0}{(s + \alpha)^2 + \Omega_0^2}$$

Therefore, the Inverse Laplace transform of the response is:

$$s(t) = \frac{2}{\sqrt{3}} e^{-0.5t} \sin(\sqrt{3}t/2) u(t)$$

a) $y_1(t) = s(t) - s(t-1)$ ←

b) $y_2(t) = \frac{d}{dt}(s(t) - s(t-1))$



Feedback Stabilization

An unstable system can be stabilized by using negative feedback with a gain K in the feedback loop. For instance, consider an unstable system with transfer function

$$H(s) = \frac{2}{s-1} = 2 \cdot e^{1 \cdot t}$$

which has a pole in the right-hand s -plane, making the impulse response of the system $h(t)$ grow as t increases. Use negative feedback with a gain $K > 0$ in the feedback loop, and put $H(s)$ in the forward loop. Draw a block diagram of the system. Obtain the transfer function $G(s)$ of the feedback system and determine the value of K that makes the overall system BIBO stable (i.e., its poles in the open left-hand s -plane).

$$\mathcal{L}\left[\frac{1}{s+a}\right] = e^{-at}$$

General solution:

$$Y(s) = (X(s) - G(s)Y(s))F(s) = \frac{F(s)}{1 + F(s)G(s)}X(s)$$

$$H(s) = \frac{F(s)}{1 + F(s) \cdot G(s)}$$

In this particular case system output is:

$$\begin{aligned} Y(s) &= (X(s) - K Y(s)) H(s) \\ &= X(s) H(s) - K H(s) Y(s) \end{aligned}$$

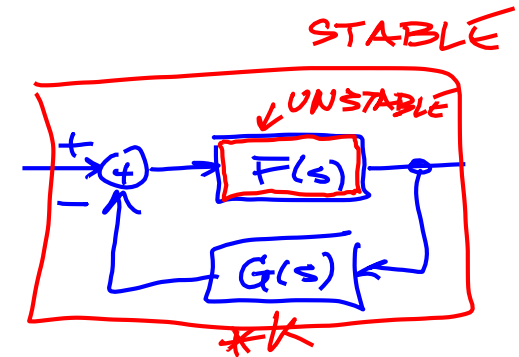
and

$$\frac{Y(s)}{X(s)} = \frac{H(s)}{1 + K \cdot H(s)} = \frac{2}{s + \underline{2K - 1}}$$

In order to have the pole in the left-hand s -plane we need $2K - 1 > 0 \rightarrow K > 0.5$

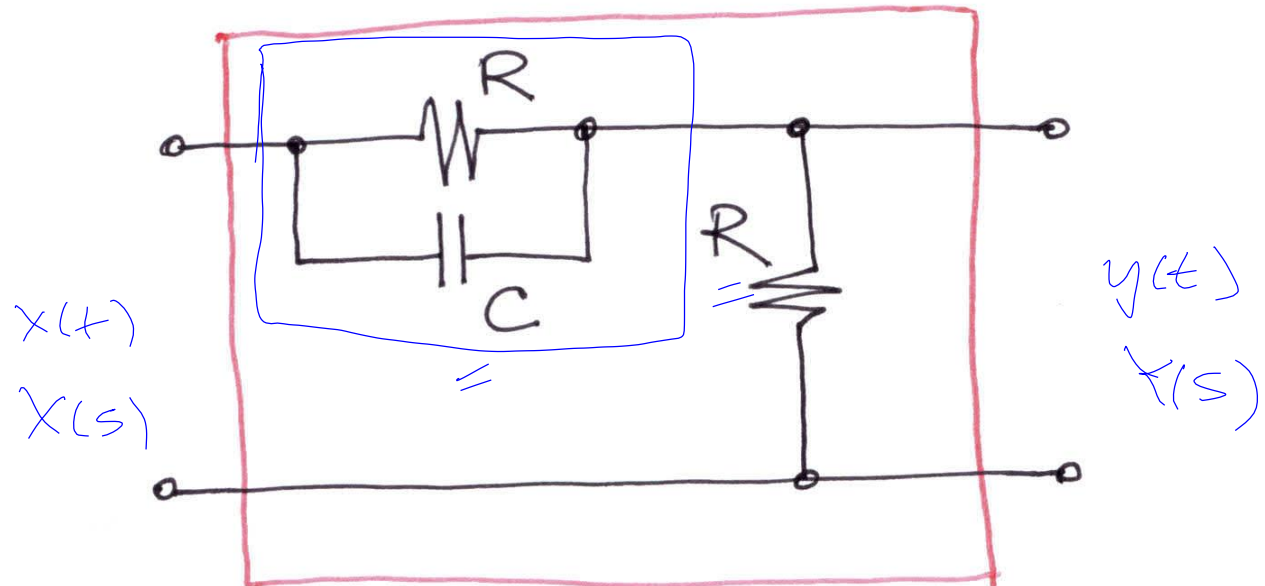
For example, $K = 1 \rightarrow$ pole at $s = -1$ and impulse response

$$g(t) = 2e^{-t}u(t)$$



Transfer Functions

- ◆ What is the transfer function of the following circuit:



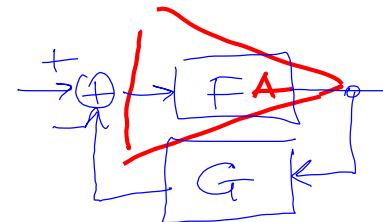
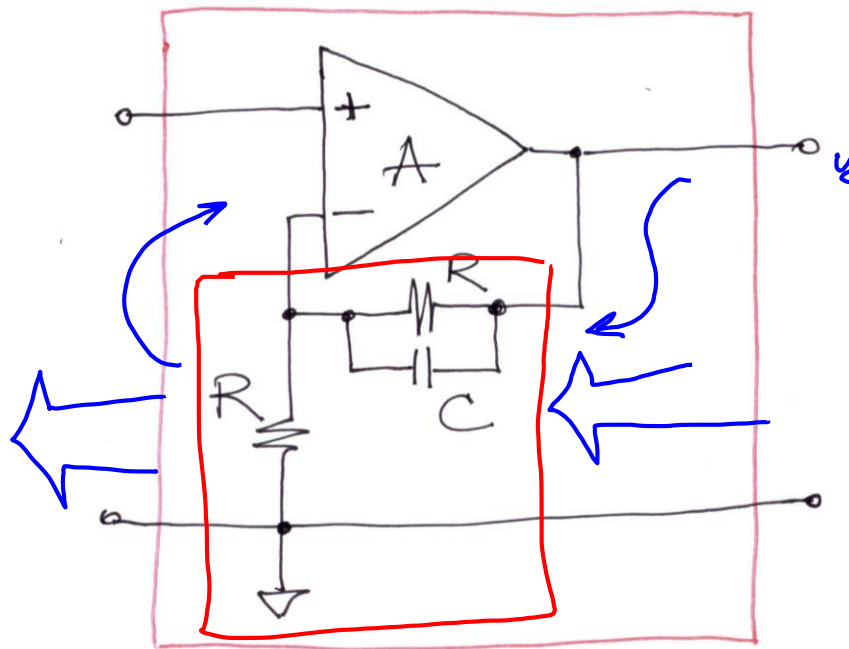
$$H(s) = \frac{R}{R + R \parallel \frac{1}{Cs}} = \frac{R}{R + \frac{R}{RCs + 1}} = \frac{RCs + 1}{RCs + 2} = \frac{s + \frac{1}{RC}}{s + \frac{2}{RC}}$$

Transfer Functions

b) What is the transfer function of the following negative feedback circuit

Hints:

- to simplify the result you can assume that $A \rightarrow \infty$



Since

$$H(s) = \frac{F(s)}{1 + F(s) \cdot G(s)}$$

$$F(s) = A \quad \text{and} \quad G(s) = \frac{s + \frac{1}{RC}}{s + \frac{2}{RC}}$$

$$H(s) = \frac{A}{1 + A \left(\frac{s + \frac{1}{RC}}{s + \frac{2}{RC}} \right)} \quad \text{for } A \rightarrow \infty \quad H(s) = \frac{s + \frac{2}{RC}}{s + \frac{1}{RC}}$$

$$\frac{A}{1 + A \left(\frac{s + \frac{1}{RC}}{s + \frac{2}{RC}} \right)} = \frac{1}{\frac{1}{A} + \left(\frac{s + \frac{1}{RC}}{s + \frac{2}{RC}} \right)} = \frac{1}{G(s)} = \frac{1}{G(s)}$$

$A \rightarrow \infty$

Transfer Functions

c) Find and plot the unit-step response $s(t)$ of the system?

$$S(s) = \frac{1}{s} \cdot \frac{s + \frac{2}{RC}}{s + \frac{1}{RC}} = \frac{A}{s} + \frac{B}{s + \frac{1}{RC}} = \frac{2}{s} - \frac{1}{s + \frac{1}{RC}} =$$

$$s(t) = (2 - e^{-\frac{t}{RC}}) \cdot u(t)$$

Block diagram showing input $X(s) = \frac{1}{s}$ entering a block $H(s)$, resulting in output $Y(s) = \frac{1}{s} \cdot H(s)$.

$$s(t) = \mathcal{L}^{-1}[Y(s)]$$

$$t=0 \quad (2 - e^0) = 1$$

$$t \rightarrow \infty \quad 2 - \frac{1}{e^\infty} = 2$$

