

Ex : 0.8 μm fabrication process
N MOS transistor

$$t_{ox} = 8 \text{ nm}$$

$$m_n = \frac{450 \text{ cm}^2}{V-s}$$

$$w = 8 \mu\text{m}$$

$$L = 0.8 \mu\text{m}$$

$$V_t = 0.7 \text{ V}$$

a) $C_{ox} = \frac{\epsilon_r \epsilon_0}{t_{ox}} = \frac{(3.9)(8.85 \times 10^{-12} \text{ F/m})}{8 \times 10^{-9} \text{ m}}$

$$C_{ox} = 4.31 \times 10^{-3} \frac{\text{F}}{\text{m}^2}$$

b) $k'_n = m_n C_{ox}$

$$= \frac{450 \text{ cm}^2}{V-s} \times \frac{4.31 \times 10^{-3} \text{ F}}{\text{m}^2} \times \frac{1 \text{ m}^2}{100^2 \text{ cm}^2}$$

$$= 194.15 \times 10^{-6} \frac{\text{F}}{\text{V-s}}$$

$$q = CV$$

$$C = \frac{q}{V} = \left[\frac{C}{V} \right]$$

↓

b) cont

$$R'_n = 194.15 \times 10^{-6} \quad \frac{C/V}{V-S} = \frac{C}{S} \cdot \frac{1}{V^2}$$

$$k'_n = 194.15 \times 10^{-6} \quad \frac{A}{V^2}$$

$$k'_n = 194.15 \frac{\mu A}{V^2}$$

c) NMOS $\Rightarrow i_D = 100 \mu A$

edge of saturation

$$V_{DS} = V_{GS} - V_t$$

$$i_D = \frac{1}{2} R'_n \frac{W}{L} (V_{GS} - V_t)^2 = 100 \mu A$$

$$i_D = \frac{1}{2} \left(194.15 \frac{\mu A}{V^2} \right) \left(\frac{8 \mu m}{.8 \mu m} \right) (V_{GS} - 0.7)^2$$

$$100 = \frac{1}{2} (194.15) (10) (V_{GS} - V_t)^2$$

$$V_{GS} = 1.02 V$$

$$V_{DS} = V_{GS} - V_t$$

$$= 1.02 - 0.7$$

$$V_{DS} = 0.321 \text{ V}$$

V_{OV}
 V_{OV}

} overdrive voltage

$V_{OV} = V_{GS} - V_t$

$$V_{OV} = V_{GS} - V_t$$

$\left. \begin{array}{l} V_{GS} \text{ or } i_D \Rightarrow \text{ac + dc components} \\ V_{GS} \text{ or } I_D \Rightarrow \text{dc only} \end{array} \right\}$
 $V_{GS} \text{ or } i_d \Rightarrow \text{ac only (small signal)}$

Ex 2

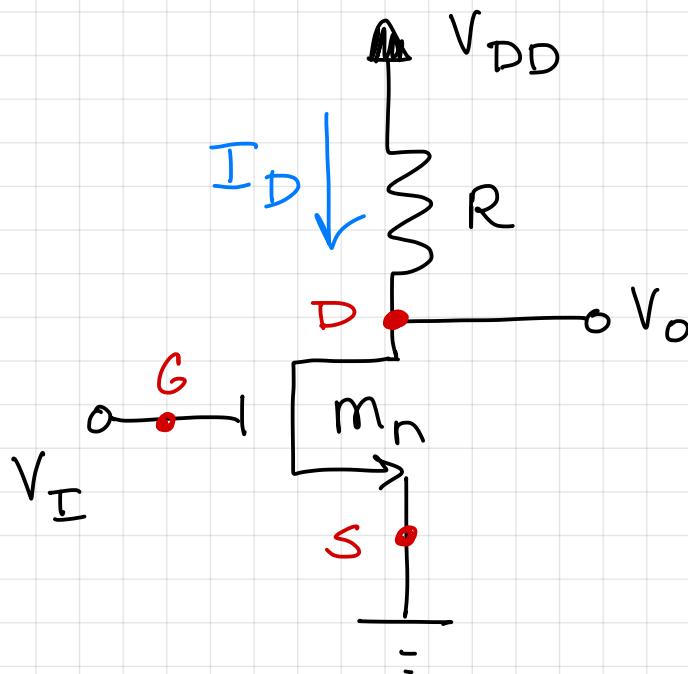
NMOS is in triode
 V_{DS} is very small
from Ex 1

NMOS device acts like 1000 Ω resistor. What is V_{GS} ?

$$r_{ds} = \frac{V_{DS}}{I_D} = \frac{1}{k'n\left(\frac{W}{L}\right)(V_{GS} - V_t)}$$

$$r_{ds} = \frac{1}{(194.15 \times 10^{-6})(10)(V_{GS} - 0.7)} = 1000$$

$$V_{GS} = 0.915 \text{ V}$$



$$V_t = 0.4V$$

$$k'_n = 0.5 \frac{mA}{V^2}$$

$$\text{if } V_{DD} = V_I = 1.3V$$

$$\text{and } V_O = 50mV$$

Find $\frac{W}{L}$ and R

$$V_{DS} = V_O - 0 = V_O = 0.05V$$

$$V_{GS} = V_I - 0 = V_I = 1.3V$$

$V_{DS} < V_{GS} - V_t$

\Rightarrow triode region

$0.05 < 1.3 - 0.4$
 $0.05 < 0.9$

we know $r_{DS} = 50\Omega$

$$r_{DS} = \frac{1}{k'_n \frac{W}{L} (V_{GS} - V_t)} = 50$$

$$\frac{1}{(0.5 \times 10^{-3}) \left(\frac{W}{L} \right) (1.3 - 0.4)} = 50$$

$$\frac{W}{L} = 44.4$$

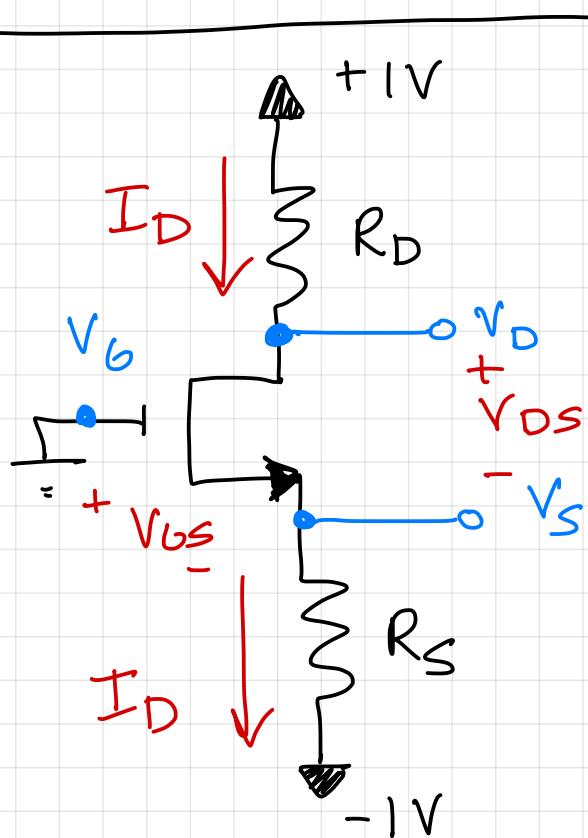
$$R = \frac{V_{DD} - V_0}{I_D}$$

$$I_D = kR'n \frac{W}{L} (V_{GS} - V_t)(V_{DS})$$

$$= 0.5 (44.4) (1.3 - 0.4) (.05)$$

$$I_D = 1 \text{ mA}$$

$$R = \frac{1.3 - 0.05}{1 \text{ m}} = 1.25 \text{ k}\Omega$$



Design R_D & R_S

$$I_D = 0.1 \text{ mA}$$

$$V_D = 0.2 \text{ V}$$

$$V_t = 0.5 \text{ V}$$

$$\mu n C_{ox} = 400 \frac{\mu \text{A}}{\text{V}^2}$$

$$W = 4 \mu \text{m}$$

$$L = 0.5 \mu \text{m}$$

$$\frac{W}{L} = \frac{4}{0.5} = 8$$

Know

$$\begin{aligned}V_{GS} &= 0 - V_S \\&= -V_S\end{aligned}$$

$$V_{DS} = 0.2 - V_S$$

$$\begin{aligned}V_{DS} &\bigcirc V_{GS} - V_t \\&\cdot 2 - V_S \bigcirc -V_S - 0.5 \\&\cdot 2 - V_S \geq -0.5 - V_S\end{aligned}$$

assume saturation operation
and then check our assumption

$$R_D = \frac{1 - V_D}{I_D} = \frac{1 - 0.2}{0.1 \text{ m}} = 8 \text{ k}\Omega$$

$$I_D = \frac{1}{2} k' n \frac{W}{L} (V_{GS} - V_t)^2$$

[mA]

$$\frac{1}{2} (0.4)(8)(V_{GS} - 0.5)^2 = 0.1$$

[mA]

$$(V_{GS} - 0.5)^2 = 0.0625$$

$$V_{GS} = 0.75 \text{ V}$$

$$V_{GS} = -V_S$$

$$-V_S = 0.75 \text{ V}$$

$$\boxed{V_S = -0.75 \text{ V}}$$

$$0.2 - V_S \geq -0.5 - V_S$$

$$0.2 - (-0.75) \geq -0.5 - (-0.75)$$

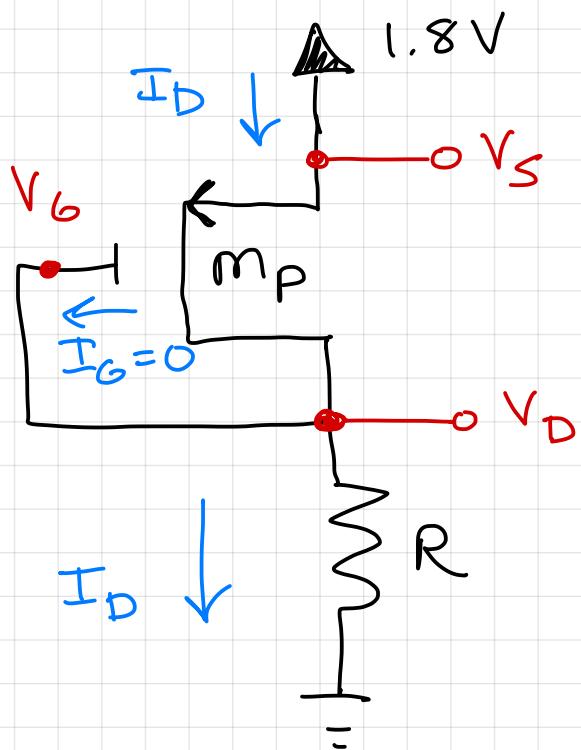
$$0.95 \geq 0.25$$

Assumption was correct

device is operating in saturation

$$R_S = \frac{V_S - (-1)}{I_D} = \frac{-0.75 + 1}{0.1 \text{ mA}} = 2.5 \text{ k}\Omega$$

$$R_S = 2.5 \text{ k}\Omega$$



$$k'p = \mu_p C_{ox} = 100 \frac{\mu\text{A}}{\text{V}^2}$$

$$V_{tp} = -0.5 \text{ V}$$

$$L = 0.18 \mu\text{m}$$

$$I_D = 160 \mu\text{A}$$

$$V_D = 0.8 \text{ V}$$

find w and R

$$V_S = 1.8 \text{ V}$$

$$\frac{V_{SG} = 1.8 - 0.8}{V_{SG} = 1 \text{ V}} \quad \boxed{V_{SG} = 1 \text{ V}}$$

$$V_G = V_D = 0.8 \text{ V}$$

when gate & drain are tied together ($V_G = V_D$), this "diode-connected" configuration.

$$V_{SD} \geq V_{SG} - |V_{tp}|$$

$$V_G = V_D$$

$$V_S - V_D \geq V_S - V_G - |V_{tp}|$$

$$V_S > V_S - |V_{tp}|$$

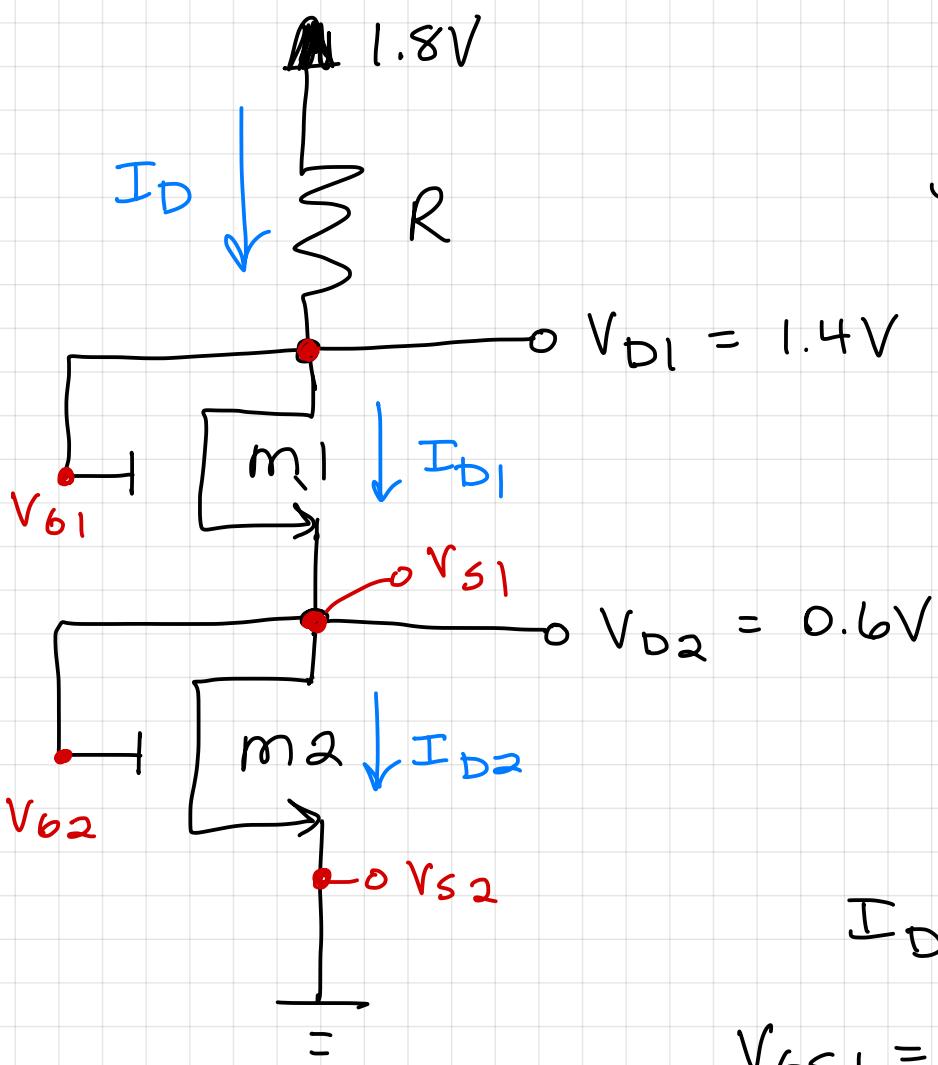
under diode connected conditions the MOSFET operates in saturation.

$$R = \frac{V_D}{I_D} = \frac{0.8}{160 \times 10^{-6}} = 11.25 \text{ k}\Omega$$

$$I_D = \frac{1}{2} k'p \left(\frac{W}{L} \right) (V_{SG} - |V_{tp}|)^2 = 160 \mu\text{A}$$

$$\frac{1}{2} (100) \left(\frac{W}{.18} \right) (1 - |-.5|)^2 = 160$$

$$W = 2.3 \mu\text{m}$$



$$I_D = 0.1 \text{ mA}$$

$$V_t = 0.4 \text{ V} \text{ (both } M_1 \text{ & } M_2 \text{)}$$

$$MnCox = k'n = \frac{0.4 \text{ mA}}{\sqrt{2}}$$

$$L_1 = L_2 = 0.18 \mu\text{m}$$

Find R and w_1, w_2

M_1 & M_2 are
diode connected \rightarrow
saturation

$$I_D = I_{D1} = I_{D2} = 0.1 \text{ mA}$$

$$\begin{aligned} V_{GS1} &= V_{61} - V_{S1} \\ &= V_{D1} - V_{D2} \\ &= 1.4 - 0.6 \end{aligned}$$

$$V_{GS1} = 0.8 \text{ V}$$

$$\begin{aligned} R &= \frac{1.8 - V_{D1}}{I_D} \\ &= \frac{1.8 - 1.4}{0.1 \text{ m}} \end{aligned}$$

$$R = 4 \text{ k}\Omega$$

$$\begin{aligned} V_{GS2} &= V_{62} - V_{S2} \\ &= V_{D2} - 0 \end{aligned}$$

$$\begin{aligned} V_{GS2} &= 0.6 - 0 \\ V_{GS2} &= 0.6 \text{ V} \end{aligned}$$

for M1

$$I_{D1} = \frac{1}{2} k'n \left(\frac{w_1}{L_1} \right) (V_{GS1} - V_t)^2$$

$$\frac{1}{2} (0.4) \left(\frac{w_1}{.18} \right) (0.8 - 0.4)^2 = 0.1$$

$$\left(\frac{w_1}{.18} \right) = 3.125$$

$$w_1 = 0.56 \mu\text{m}$$

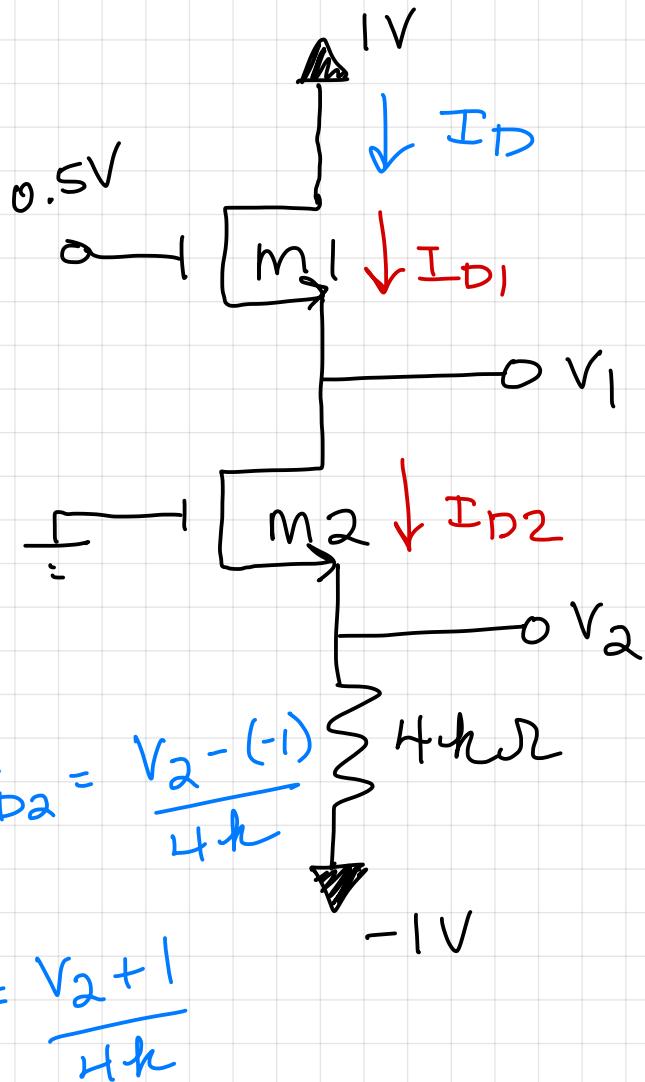
for M2

$$I_{D2} = \frac{1}{2} k'n \left(\frac{w_2}{L_2} \right) (V_{GS2} - V_t)^2$$

$$\frac{1}{2} (.4) \left(\frac{w_2}{.18} \right) (0.6 - 0.4)^2 = 0.1$$

$$\frac{w_2}{.18} = 12.5$$

$$w_2 = 2.25 \mu\text{m}$$



$$I_{D2} = \frac{V_2 - (-1)}{4k}$$

$$= \frac{V_2 + 1}{4k}$$

Find I_D , V_1 & V_2

for both M_1 & M_2

$$V_t = 0.4V$$

$$(4k \frac{W}{L}) = \frac{5mA}{V^2}$$

$$V_{GS1} = 0.5 - V_1$$

$$V_{GS2} = 0 - V_2$$

$$V_{DS1} = 1 - V_1$$

$$V_{DS2} = V_1 - V_2$$

for M_1

$$V_{DS1} \quad \bigcirc \quad V_{GS1} - V_t$$

assume

$$1 - V_1 \quad \bigcirc \quad 0.5 - V_1 - 0.4$$

M_1 is

$$1 - V_1 \quad \bigcirc \quad \text{---} \quad 0.1 - V_1$$

in saturation

for M_2

$$V_1 - V_2 \quad \bigcirc \quad -V_2 - 0.4$$

assume
 M_2 in
saturation

$I_D = I_{D1} = I_{D2}$ & devices are matched

$\left(k' n \frac{w}{L}, V_t \right)$
are same

$$I_{D1} = I_{D2}$$

$$\frac{1}{2} k' n \frac{w}{L} (V_{GS1} - V_t)^2 = \frac{1}{2} k' n \frac{w}{L} (V_{GS2} - V_t)^2$$

$$(V_{GS1} - V_t)^2 = (V_{GS2} - V_t)^2$$

$$V_{GS1} = V_{GS2}$$

$$0.5 - V_1 = 0 - V_2$$

$$V_1 - V_2 = 0.5V$$

$$I_{D2} = \frac{V_2 + 1}{4k}$$

$$\frac{1}{2} k' n \frac{w}{L} (V_{GS2} - V_t)^2 = \frac{V_2 + 1}{4k}$$

$$\frac{1}{2} (5m) (-V_2 - 0.4)^2 = \frac{V_2 + 1}{4k}$$

$$10(-V_2 - 0.4)^2 = V_2 + 1$$

$$10(0.16 + 0.8V_2 + V_2^2) = V_2 + 1$$

$$1.6 + 8V_2 + 10V_2^2 = V_2 + 1$$

$$10V_2^2 + 7V_2 + 0.6 = 0$$

$$V_2 = -0.1V$$

$$V_2 = -0.6V$$

$$V_1 = 0.4V$$

$$V_2 = -0.1V$$

$$V_{DS2} = 0.1V$$

$$V_{DS2} = 0.5V$$

$$V_{DS1} = 0.6V$$

$$V_{DS1} = 0.1V$$

$$V_{DS1} < V_t$$

$$0.1 < 0.4$$

$$V_{DS2} < V_t$$

$$0.1 < 0.4$$

↑

Both devices are off.

$$V_{GS1} = 0.5 - V_1$$

$$V_{GS2} = 0 - V_2$$

$$V_{DS1} = 1 - V_1$$

$$V_{DS2} = V_1 - V_2$$

$$V_1 - V_2 = 0.5V$$

$$V_{GS1} = 0.5 - V_1$$

$$V_2 = -0.6 \text{ V}$$

$$V_1 = -0.1 \text{ V}$$

$$V_{GS2} = 0 - V_2$$

$$V_{GS1} = 0.6 \text{ V}$$

$$V_{GS2} = 0.6 \text{ V}$$

$$V_{DS1} = 1 - V_1$$

$$V_{DS2} = 0.5 \text{ V}$$

$$V_1 - V_2 = 0.5 \text{ V}$$

greater
than V_t

$$V_{DS1} \bigcirc V_{GS1} - V_t$$

$$V_{DS2} \bigcirc V_{GS2} - V_t$$

$$1.1 \bigcirc 0.6 - 0.4$$

$$0.5 \bigcirc 0.6 - 0.4$$

confirmed saturation
for both!

$$I_D = \frac{V_2 + 1}{4k} = \frac{-0.6 + 1}{4k}$$

$$I_D = 0.1 \text{ mA}$$

Ex

npn transistor

$$V_T = 25mV$$

$$V_{BE1}, i_{C1} \Rightarrow 0.7V, 1mA$$

$$V_{BE2}, i_{C2} \Rightarrow \underline{\quad}, 0.1mA$$

$$V_{BE3}, i_{C3} \Rightarrow \underline{\quad}, 10mA$$

$$\textcircled{a} \quad V_{BE2} - V_{BE1} = V_T \ln \left(\frac{i_{C2}}{i_{C1}} \right)$$

$$V_{BE2} - 0.7 = 0.025 \cdot \ln \left(\frac{0.1}{1} \right)$$

$$V_{BE2} = 0.64V$$

$$\textcircled{b} \quad V_{BE3} - V_{BE1} = V_T \ln \left(\frac{i_{C3}}{i_{C1}} \right)$$

$$V_{BE3} - 0.7 = 0.025 \ln \left(\frac{10}{1} \right)$$

$$V_{BE3} = 0.76V$$

EX

npn transistor

β range 50 to 150

α range ?

$$\alpha = \frac{\beta}{\beta + 1}$$

$\alpha = 0.98$ to 0.993

EX

npn transistor

$$i_B = 14.46 \mu A$$

$$i_E = 1.46 mA$$

$$V_{BE} = 0.7 V$$

$$i_C + i_B = i_E$$

$$\begin{aligned} i_C &= i_E - i_B \\ &= 1.446 mA \\ &= 1.45 mA \end{aligned}$$

$$\begin{aligned} I_S &= \frac{i_C}{\exp(V_{BE}/V_T)} \\ &= \frac{1.446 \times 10^{-3}}{\exp(-0.7/0.025)} \end{aligned}$$

Find i_C, I_S, β, α

$$i_B = \frac{i_C}{\beta}$$

$$\beta = \frac{i_C}{i_B} = \frac{1.446 \times 10^{-3}}{14.46 \times 10^{-6}}$$

$$\beta = 99.97 \approx 100$$

$$\alpha = \frac{\beta}{\beta + 1} = 0.99$$

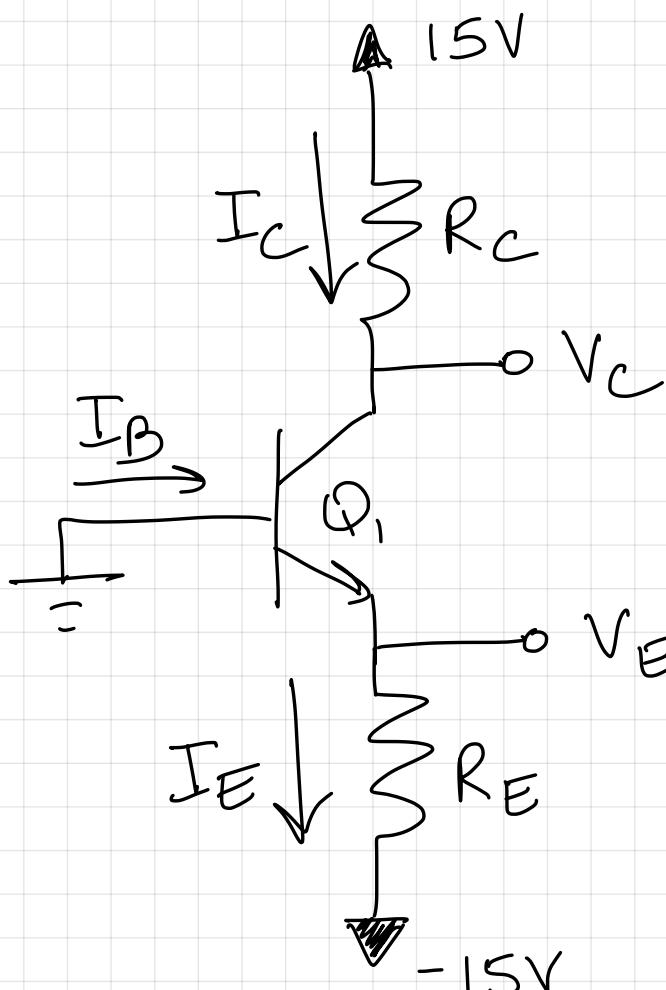
$$I_S = 9.995 \times 10^{-16} A$$

$$= 10 \times 10^{-16} A$$

$$= 10^{15} A$$

$$n-p-n \Rightarrow Q_1, Q_2$$

$$V_C = V_c$$



$$B = 100$$

$$\boxed{V_{BE} = 0.7V}$$

$$I_C = 1mA$$

what are R_C and R_E for

$$I_C = 2mA$$

$$V_C = 5V$$

$$V_{CB} > 0 \quad \left. \right\} 5 - 0 = 5V$$

$$R_C = \frac{15 - V_C}{I_C} = \frac{15 - 5}{2mA} = 5k\Omega$$

$$V_{BE2} - V_{BE1} = V_T \ln \left(\frac{I_{C2}}{I_{C1}} \right)$$

$$V_{BE2} - 0.7 = (0.025) \ln \left(\frac{2mA}{1mA} \right)$$

$$\boxed{V_{BE2} = 0.717V}$$

$$V_{BE2} = V_B^0 - V_E = 0.717V$$

$$\boxed{V_E = -0.717V}$$

$$R_E = \frac{V_E - (-15)}{I_E}$$

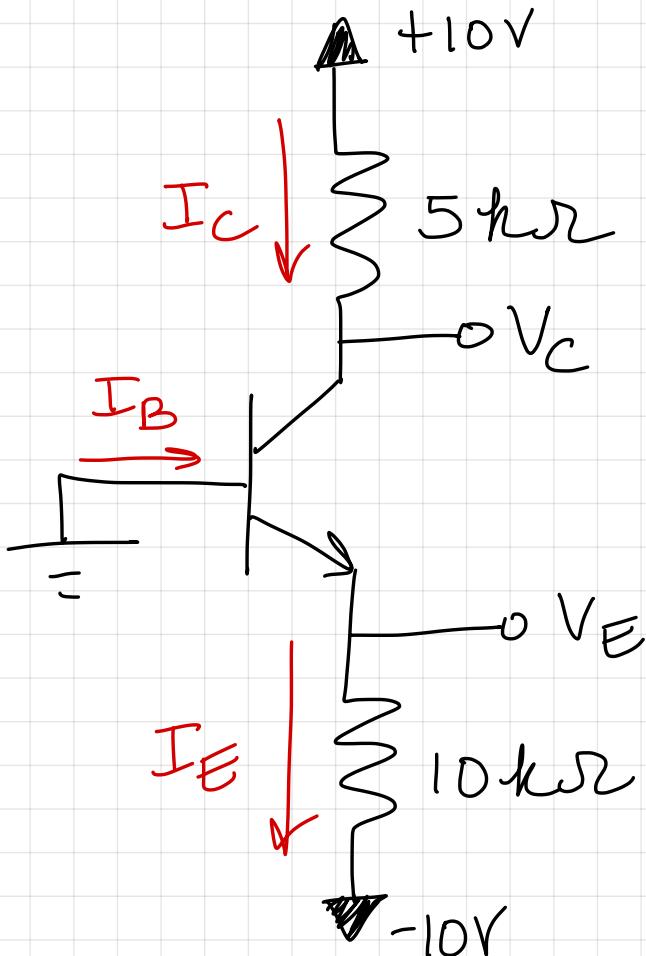
$$I_C = \alpha I_E \quad \alpha = \frac{\beta}{\beta+1} = \frac{100}{101}$$

$$I_E = \frac{I_C}{\alpha} \quad \alpha = 0.99$$

$$I_E = \frac{2 \times 10^{-3}}{0.99} = 2.02 \text{ mA}$$

$$R_E = \frac{-0.717 + 15}{2.02 \times 10^{-3}}$$

$$\boxed{R_E = 7.07 \text{ k}\Omega}$$



$$\beta = 50$$

$$V_E = -0.7 \text{ V}$$

Find I_C , I_B , I_E & V_C .

$$I_E = \frac{V_E - (-10)}{10 \times 10^3}$$

$$= \frac{-0.7 + 10}{10 \times 10^3}$$

$$I_E = 0.93 \text{ mA}$$

$$I_C = \alpha I_E$$

$$= \left(\frac{\beta}{\beta+1} \right) I_E$$

$$= \left(\frac{50}{51} \right) (0.93 \text{ mA})$$

$$I_C = 0.912 \text{ mA}$$

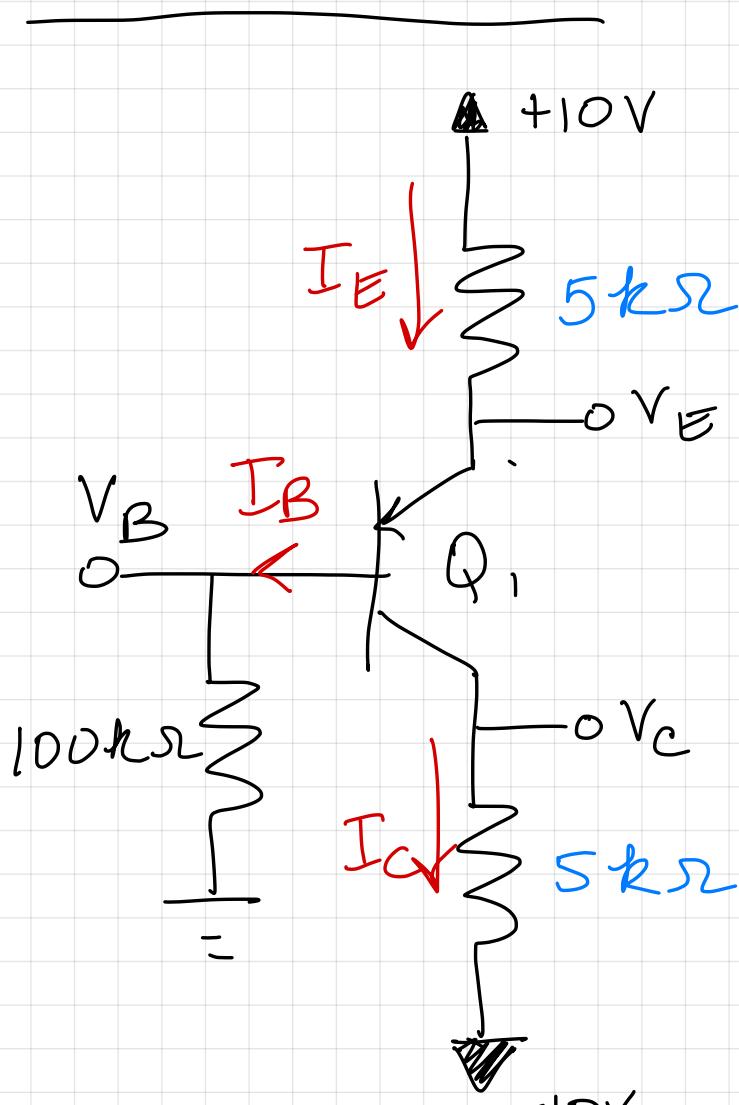
$$I_B = \frac{I_C}{\beta} = \frac{0.912 \text{ mA}}{50} = 0.018 \text{ mA}$$

$$I_B = 0.018 \text{ mA}$$

$$V_C = 10 - 5 I_C^{\text{mA}}$$

$$= 10 - 5(0.912)$$

$$V_C = 5.44 \text{ V}$$



$$V_B = 1 \text{ V}$$

$$V_E = 1.7 \text{ V}$$

$$\beta, \alpha \rightarrow V_C$$

$$V_{EB} = 0.7 \text{ V}$$

$$I_B = \frac{V_B}{100 \times 10^3} = \frac{1}{100 \times 10^3}$$

$$I_B = 0.01 \text{ mA}$$

$$I_E = \frac{10 - V_E}{5 \times 10^3}$$

$$I_E = \frac{10 - 1.7}{5 \times 10^3} = 1.66 \text{ mA}$$

$$I_C = I_E - I_B = 1.65 \text{ mA}$$

$$I_C = \alpha I_E$$

$$\alpha = \frac{I_C}{I_E} = 0.994$$

$$\beta = \frac{I_C}{I_B} = \frac{1.65}{.01} = 165$$

$$I_C = \frac{V_C + 10}{5 \times 10^3}$$

$$V_{BC} = 1 - (1.75)$$

$$V_C = 5 I_C - 10$$

$$V_{BC} = 2.75 \text{ V}$$

$$V_C = -1.75 \text{ V}$$

↑
active.

A common source (CS) amplifier
w/ NMOS transistor:

$$k'_n = 0.2 \frac{mA}{V^2} \quad \frac{W}{L} = 40 \quad V_t = 0.5V$$

$$V_A = 50V$$

$$R_D = 20k\Omega$$

$$R_L = 20k\Omega$$

$$R_{sig} = 100k\Omega$$

a) If $I_{DQ} = 0.25mA$

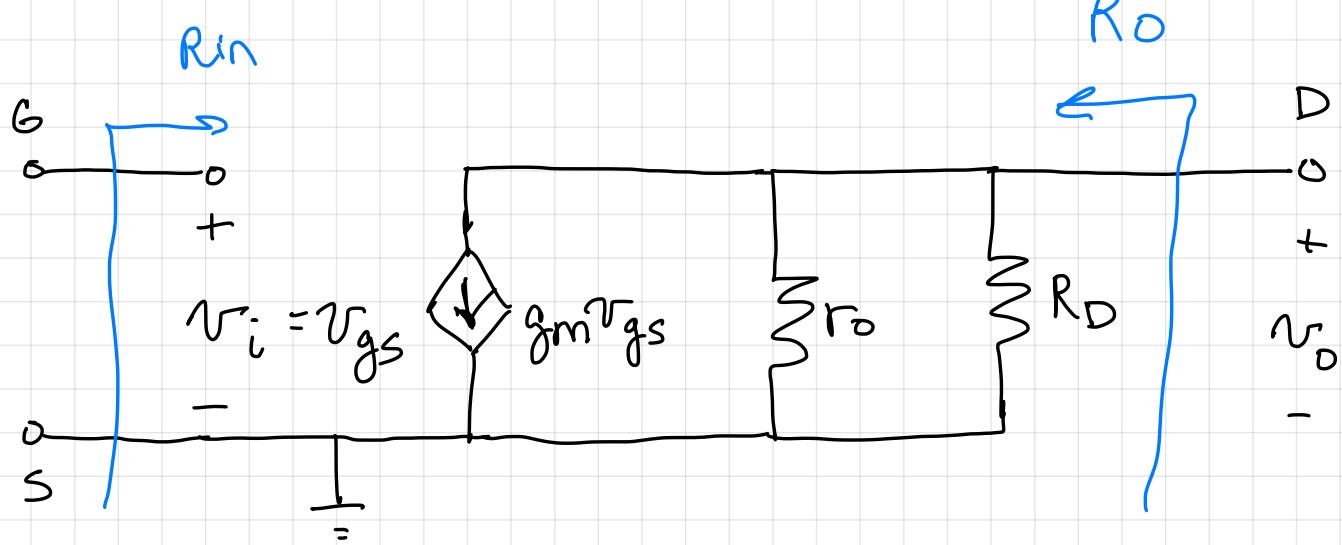
$$V_{GSQ} = ?$$

$$I_{DQ} = \frac{1}{2} k'_n \frac{W}{L} (V_{GSQ} - V_t)^2$$

$$0.25mA = \frac{1}{2} (0.2)(40)(V_{GSQ} - 0.5)^2$$

$$\boxed{V_{GSQ} = 0.75V}$$

b) R_{in} , g_m , R_o , $\underline{A_{ro}}$, r_o



$$R_{in} = \infty$$

$$g_m = k'n \frac{W}{L} (V_{GSQ} - V_t) = 2 \text{ mA/V}^2$$

$$= 0.2(40)(.25) = 2 \text{ mA/V}^2$$

$$r_o = \frac{V_A}{I_{DQ}} = \frac{50}{.25} = 200 \text{ k}\Omega$$

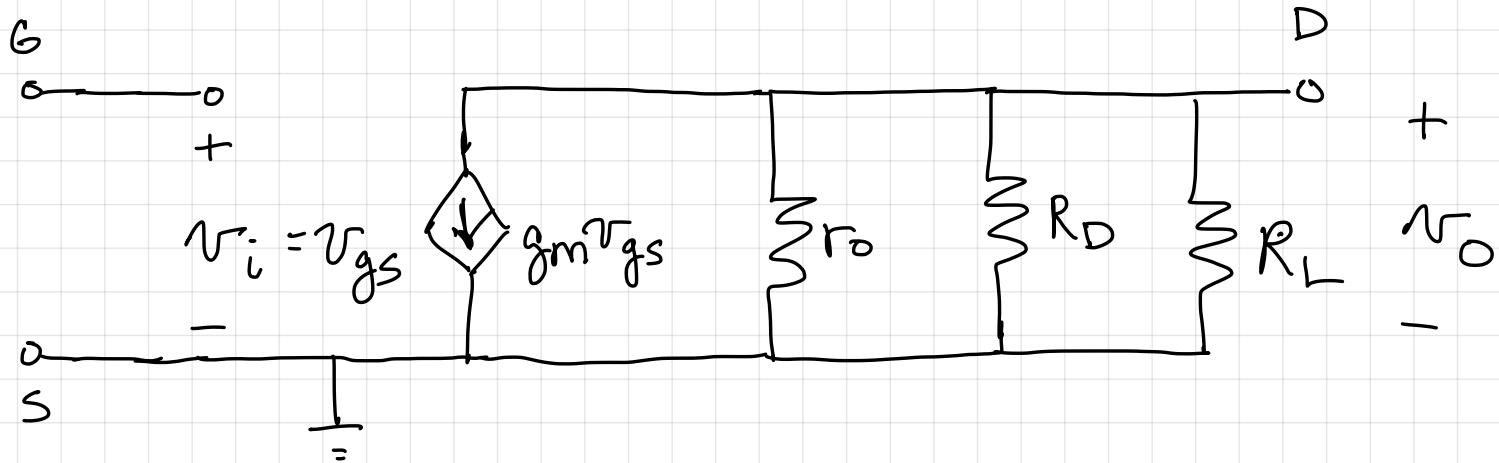
$$R_o = r_o \parallel R_D = (200 \parallel 20) \text{ k}\Omega = 18.2 \text{ k}\Omega$$

$$A_{vo} = \left. \frac{V_o}{V_i} \right|_{R_L \rightarrow \infty} = -g_m(r_o \parallel R_D)$$

$$= -2(200 \parallel 20)$$

$$A_{vo} = -36.4 \text{ V/V}$$

$$c) A_v = \frac{V_o}{V_i}$$

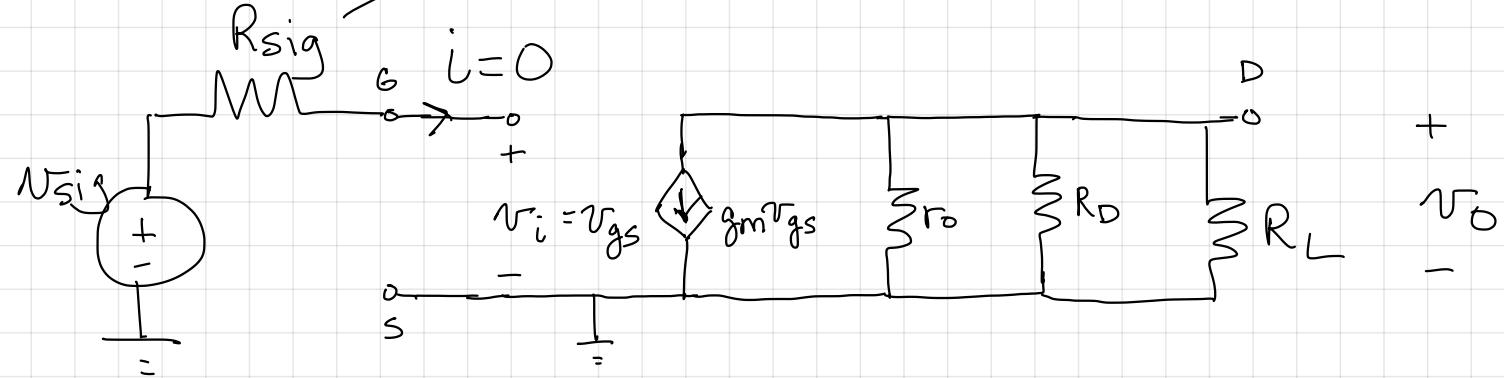


$$A_v = -g_m (r_o \parallel R_D \parallel R_L)$$

$$= -2 (200 \parallel 20 \parallel 20)$$

$$= -19.05 \text{ V/V}$$

d) G_V does not affect G_V



$$G_V = A_{Vr} = -g_m(r_o) | R_D | | R_L |$$
$$= -19.05 \text{ V/V}$$

BJT Amp Configuration Examples

CE amplifier

$$\beta = 100$$

$$I_{CQ} = 1 \text{ mA}$$

$$R_L = 5 \text{ k}\Omega$$

$$V_A = 100$$

$$R_C = 5 \text{ k}\Omega$$

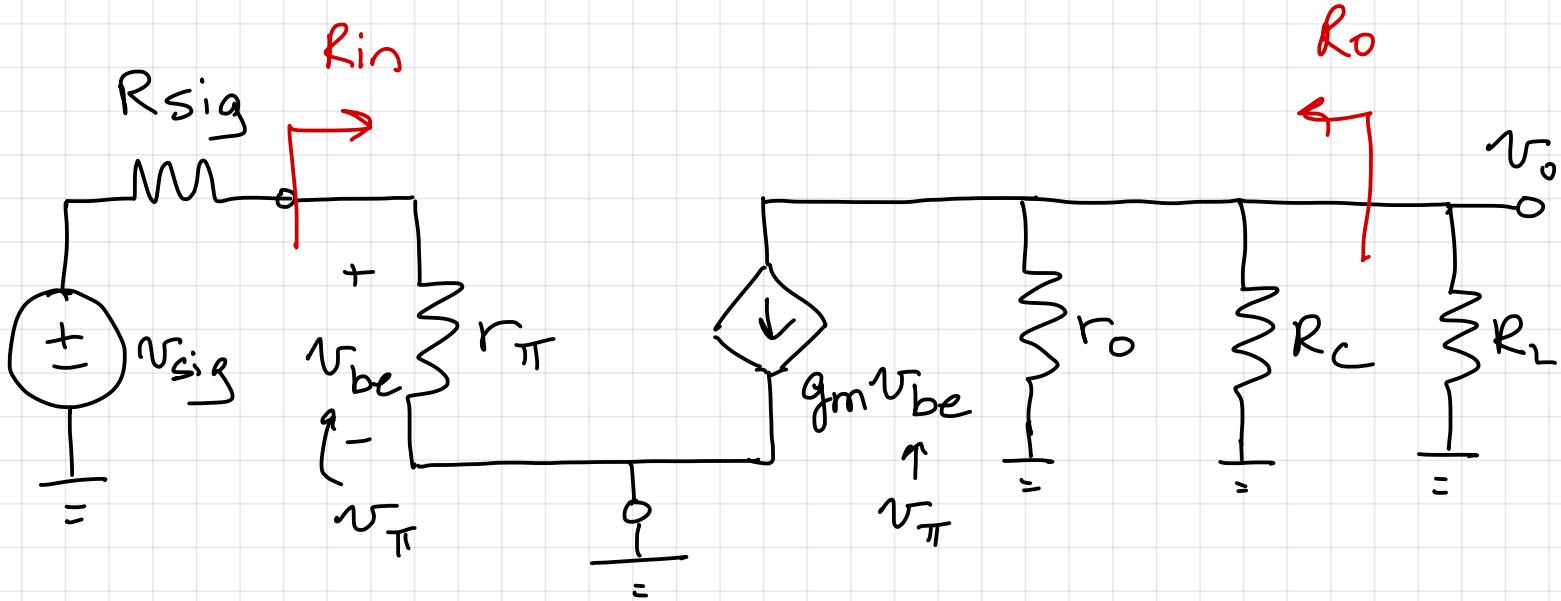
$$R_{sig} = 10 \text{ k}\Omega$$

Finding g_m , r_π , r_o

$$g_m = \frac{I_{CQ}}{V_T} = \frac{1 \text{ mA}}{0.025 \text{ V}} = 40 \frac{\text{mA}}{\text{V}}$$

$$r_\pi = \frac{\beta}{g_m} = \frac{100}{40 \times 10^{-3}} = \frac{2.5 \times 10^3}{2.5 \text{ k}\Omega}$$

$$r_o = \frac{V_A}{I_{CQ}} = \frac{100}{1} = 100 \text{ k}\Omega$$



$$R_{in} = r_{\pi} = 2.5 k\Omega$$

$$R_o = r_{ol} || R_c = 100 || 5 = 4.76 k\Omega$$

$$G_V = \frac{V_o}{V_{sig}} = -g_m(r_{ol} || R_c || R_L) \left(\frac{r_{\pi}}{r_{\pi} + R_{sig}} \right)$$

$$= -40(100 || 5 || 5) \left(\frac{2.5}{12.5} \right)$$

$$G_V = -19.52 \text{ V/V}$$

Let's do a redesign :

$I_{CQ} = 0.5 \text{ mA} \Rightarrow$ reduction to increase R_{in}

$$g_m = \frac{I_{CQ}}{V_T} = 20 \text{ mA}$$

$$r_{\pi} = 5 k\Omega$$

$$r_o = \frac{V_A}{I_{CQ}} = 200 k\Omega$$

$$G_V = -g_m(r_{ol} || R_c || R_L) \left(\frac{r_{\pi}}{r_{\pi} + R_{sig}} \right)$$

$$= -20(200 || R_c || 5) \left(\frac{5}{15} \right)$$

$$= -20(R_c || 4.88) (1/3)$$

$$= -6.67(R_c || 4.88)$$

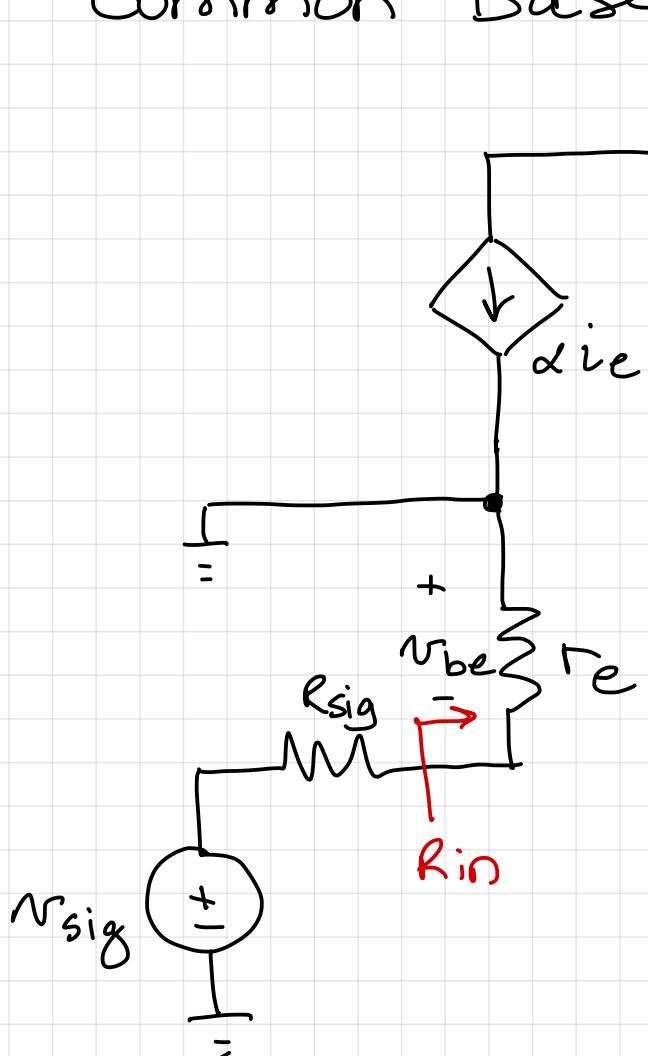
$$-6.607 \left(\frac{R_C (4.88)}{R_C + 4.88} \right) = -19.52$$

$$R_C = 7.33 \text{ k}\Omega$$

$$R_O = 200 \parallel 7.33 = 7.07 \text{ k}\Omega$$

$$R_{in} = 5 \text{ k}\Omega$$

Common Base



$$I_{CQ} = 1 \text{ mA}$$

$$R_C = 5 \text{ k}\Omega$$

$$R_L = 5 \text{ k}\Omega$$

$$R_{sig} = 5 \text{ k}\Omega$$

$$g_m = \frac{I_{CQ}}{V_T}$$

$$= \frac{1}{0.025} = \frac{40 \text{ mA}}{V}$$

$$r_e = \frac{\alpha}{g_m} \approx \frac{1}{g_m}$$

$$= 25 \Omega$$

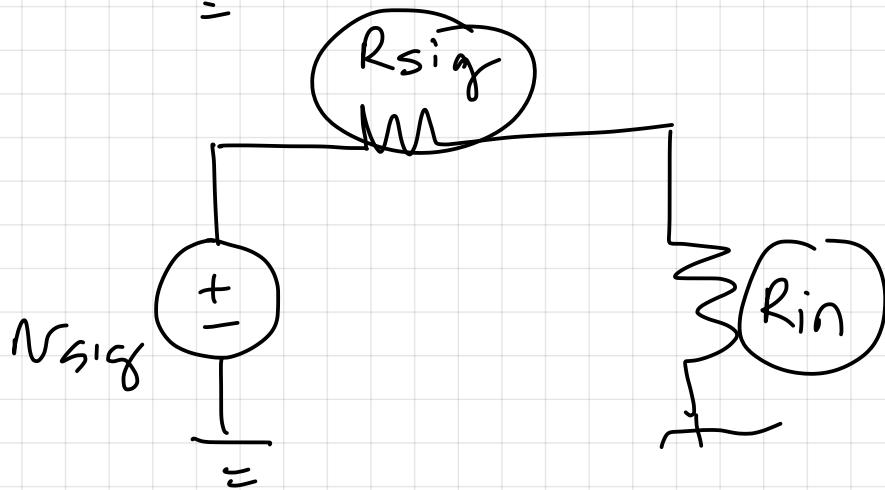
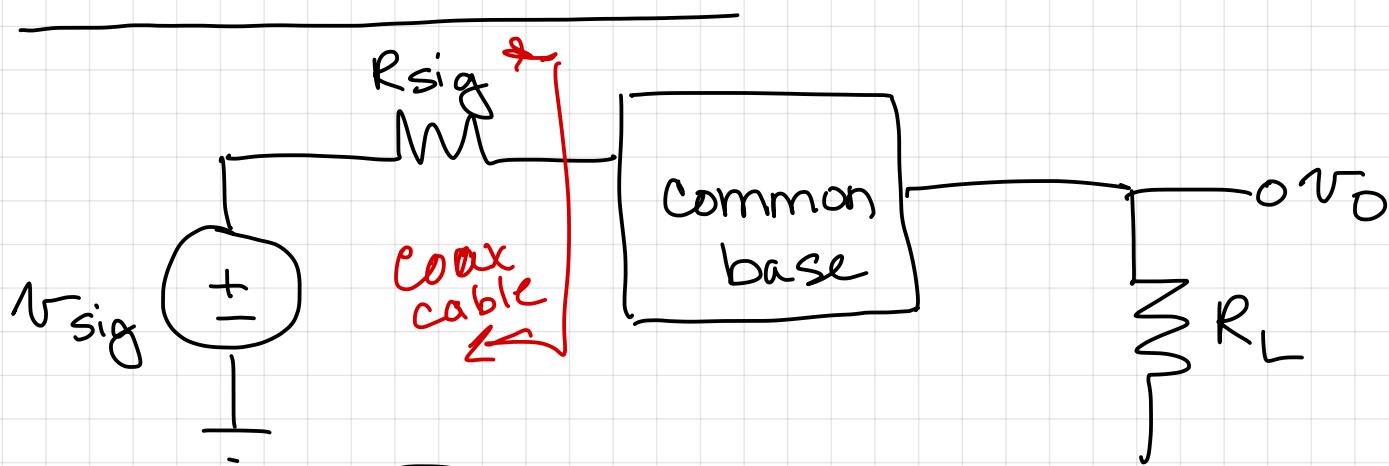
$$R_{in} = r_e = 25 \Omega$$

$$R_o = R_c = 5k\Omega$$

$$\frac{V_o}{V_{sig}} = \left(\frac{r_e}{r_e + R_{sig}} \right) \left(g_m (R_c || R_L) \right)$$

$$= \left(\frac{25}{25 + 5000} \right) \left(40 (5 || 5) \right)$$

$$\frac{V_o}{V_{sig}} = 0.5 \text{ V/V}$$



max power

$$R_{TH} = R_L$$

$$\text{Design} \\ R_{in} = 50\Omega$$

$$r_e = 50\Omega$$

$$R_{in} = r_e$$

$$r_e \approx \frac{1}{g_m} = 50$$

$$g_m = \frac{20\text{mA}}{\sqrt{V}}$$

Design

$$G_V = 40 \text{ V/V}$$

$$40 = \frac{20}{g_m (R_C || R_L)} \left(\frac{r_e}{50 + 50} \right) \frac{1/2}{R_L}$$

$$\frac{40}{20} = (R_C || R_L) \left(\frac{1}{2} \right)$$

$$g_m = \frac{I_{CQ}}{V_T}$$

$$\frac{80}{20} = R_C || R_L$$

$$I_{CQ} = g_m V_T$$

$$= 20(.025)$$

$$I_{CQ} = 0.5\text{mA}$$

$$R_C || R_L = 4$$

Common Collector

CC amp

$$\beta = 100$$

$$R_{sig} = 10\text{k}\Omega$$

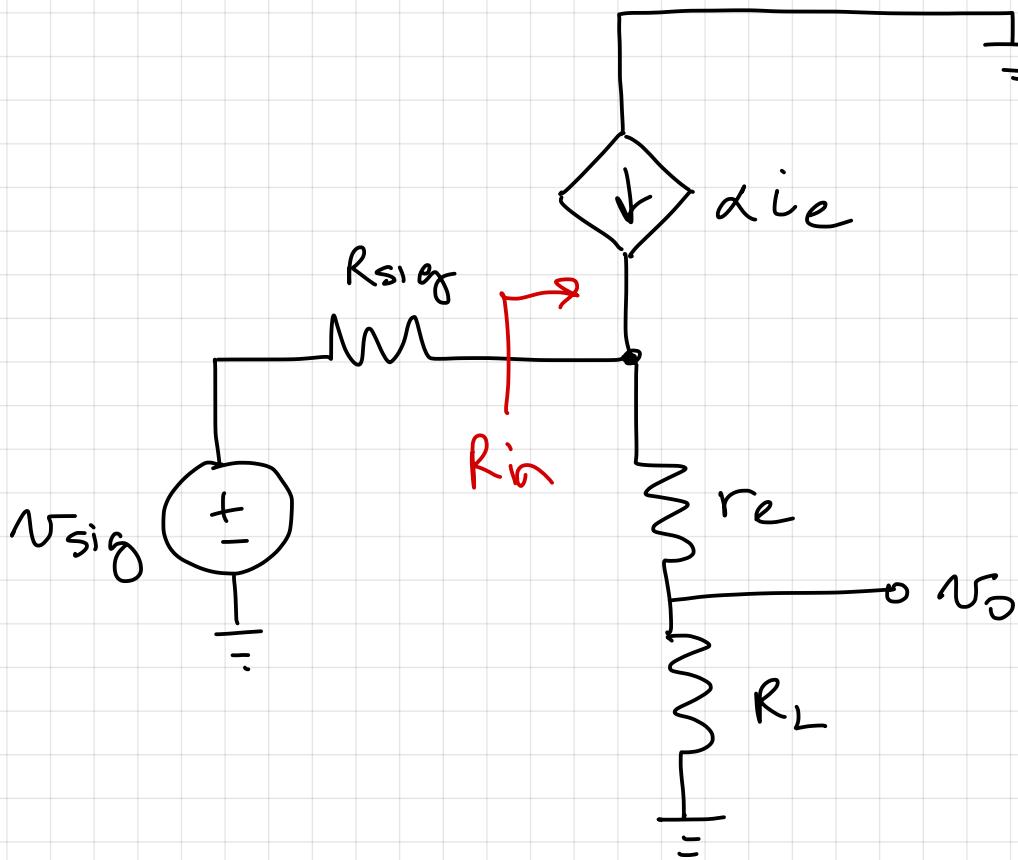
$$I_{CQ} = 5\text{mA}$$

$$R_L = 1\text{k}\Omega$$

$$g_m = \frac{I_{CQ}}{V_T}$$

$$= \frac{5}{.025}$$

$$= \frac{200\text{mA}}{\sqrt{V}}$$



$$r_e \equiv \frac{1}{g_m}$$

$$= 5\Omega$$

$$R_{in} = (1 + \beta)(r_e + R_L)$$

$$= (1 + 100)(5 + 1000)$$

$$= (101)(1005)$$

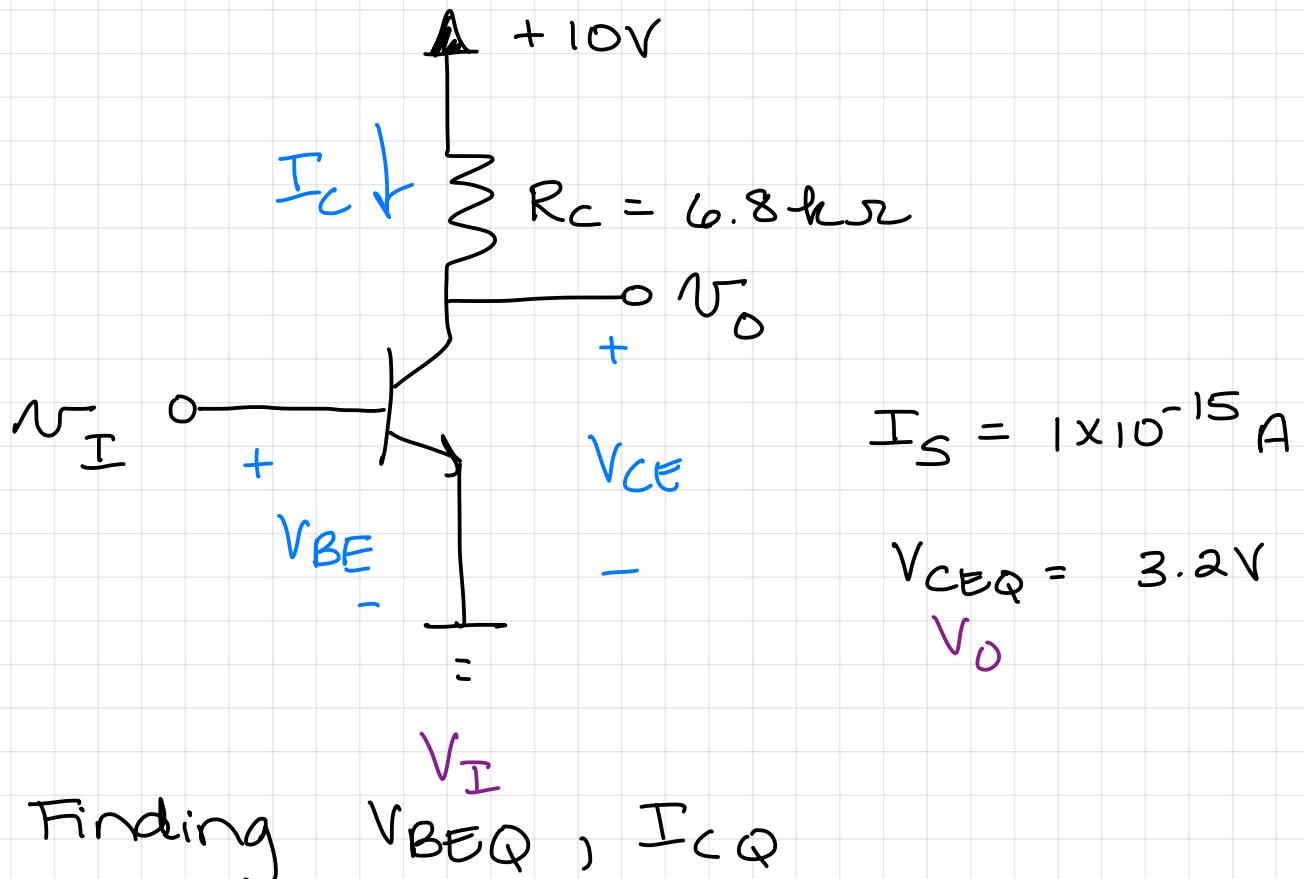
$$R_{in} = 101.5\text{k}\Omega$$

$$R_o = r_e = 5\Omega$$

$$G_V = \frac{(\beta + 1)(R_L)}{(1 + \beta)(r_e + R_L) + R_{sig}}$$

$$= \frac{(101)(1000)}{(101)(1005) + 10000}$$

$$G_V = 0.91 \text{ V/V}$$



a) Finding V_{BEQ} , I_{CQ}

$$I_{CQ} = \frac{10 - V_{CEQ}}{6.8} = \frac{10 - 3.2}{6.8} = 1 \text{ mA}$$

$$I_C = I_S \exp\left(\frac{V_{BEQ}}{V_T}\right)$$

$$1 \times 10^{-3} = 1 \times 10^{-15} \exp(V_{BEQ}/0.025)$$

$$V_{BEQ} = 0.691 \text{ V}$$

$$b) A_V = - \frac{I_{CQ} R_C}{V_T}$$

$\underbrace{V_T}_{g_m}$

$$A_V = \frac{-(1 \times 10^{-3})(6.8 \times 10^3)}{.025}$$

$$= -272 \text{ V/V}$$

c) let small signal of 5mV sine wave be superimposed on V_{BEQ}

$$V_{CE} \Rightarrow A_V = \frac{V_{CE}}{V_{BE}}$$

$$V_{CE} = A_V \cdot V_{BE}$$

$$= -272 (.005)$$

$$V_{CE} = 1.36 \text{ V peak sine wave}$$

d) Drive Q pt to edge of saturation

$$V_{CE} = 0.3V$$

what is the change in V_{BE} to do so?

$$i_d = \frac{10 - 0.3}{6.8} = 1.43 \text{ mA}$$

$$\underbrace{V_{BE2} - V_{BE1}}_{\Delta V_{BE}} = V_T \ln \left(\frac{i_{d2}}{i_{d1}} \right)$$

$$\Delta V_{BE} = (0.025) \ln \left(\frac{1.43}{1} \right)$$

$$\Delta V_{BE} = 8.94 \text{ mV}$$

Example 2

NPN transistor

g_m , r_π , r_e at

Q point

$$V_T = 25 \text{ mV}$$

$$\beta = 100$$

$$I_{CQ} = 1 \text{ mA}$$

$$g_m = \frac{I_{CQ}}{V_T} = \frac{1 \text{ mA}}{0.025 \times 10^{-3}} = \frac{40 \text{ mA}}{V}$$

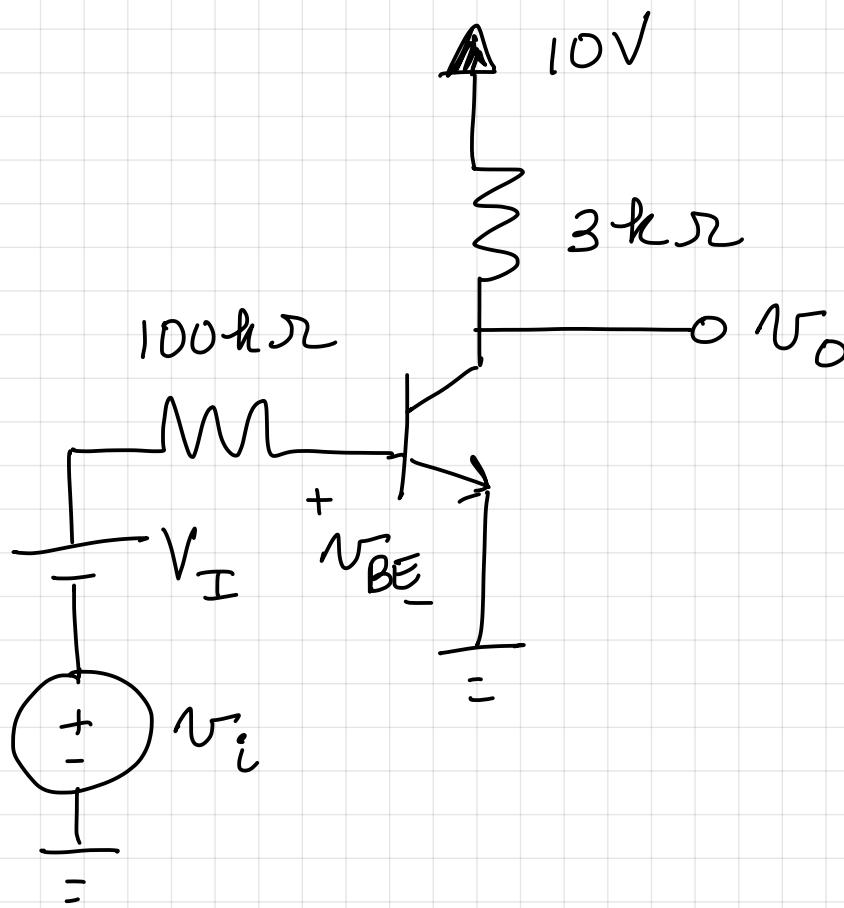
$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{40 \text{ mA/V}} = 2.5 \text{ k}\Omega$$

$$r_e \approx \frac{1}{g_m} = \frac{1}{40 \text{ mA/V}} = 25 \Omega$$

$$r_e = \frac{r_{\pi}}{(1 + \beta)} = 24.75 \Omega$$

$$\beta \rightarrow \text{larger} \quad \beta + 1 \approx \beta$$
$$\alpha \approx 1$$

Example 3

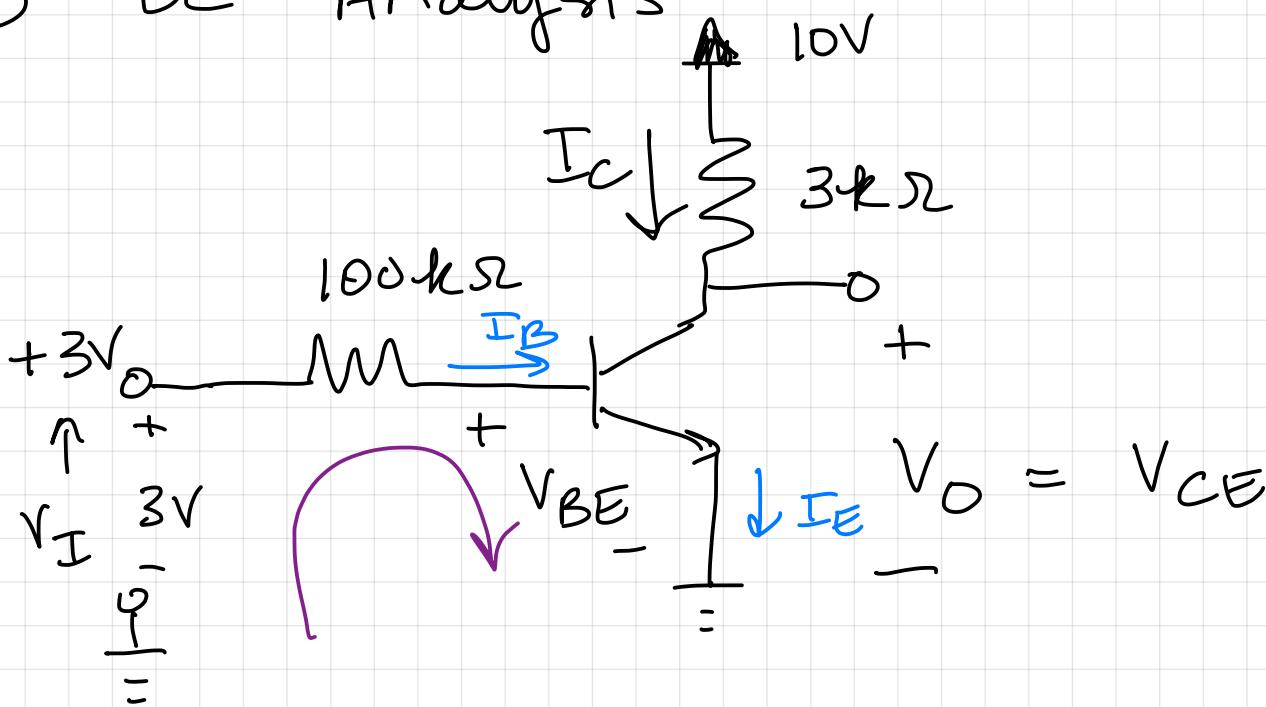


$$\beta = 100$$

$$V_{BE} = 0.7V$$

$$V_I = 3V$$

① DC Analysis



$$\text{by KVL: } 3 - 100(I_B) - V_{BE} = 0$$

$$I_B = \frac{3 - 0.7}{100} = 0.023 \text{ mA}$$

$$I_C = \beta I_B = 100 (0.023 \text{ mA})$$

$I_{CQ} = 2.3 \text{ mA}$

$$V_{CEQ} = V_{OQ} = 10 - I_{CQ} R_C$$

$$= 10 - (2.3)(3)$$

$$V_C > V_B$$

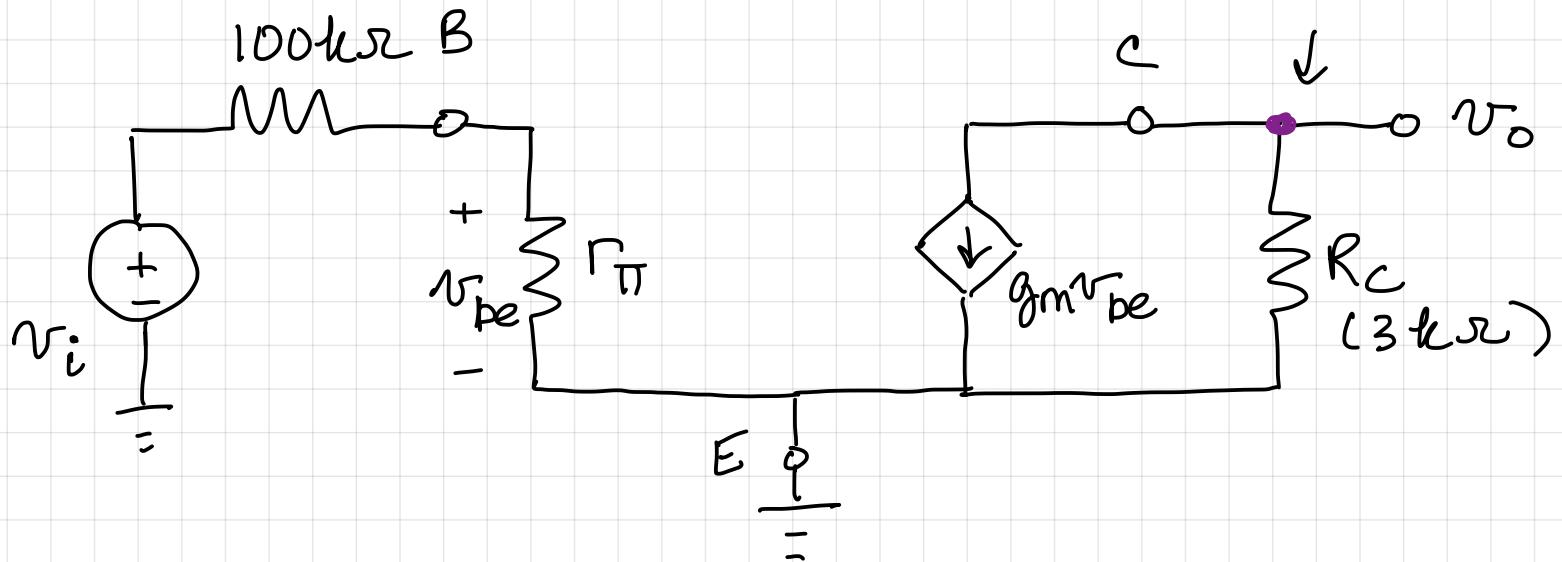
are in active

$$\Leftrightarrow V_{OQ} = 3.1 \text{ V}$$

② small signal

$$g_m = \frac{I_{CQ}}{V_T} = \frac{2.3}{0.025} = \frac{92 \text{ mA}}{\text{V}}$$

$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{92 \text{ mA/V}} = 1.09 \text{ k}\Omega$$



$$\frac{v_o}{v_i} \Rightarrow g_m v_{be} + \frac{v_o}{R_C} = 0$$

$$\frac{v_o}{R_C} = -g_m v_{be}$$

$$\frac{v_o}{v_{be}} = -g_m R_C$$

$$g_m = \frac{I_C}{V_T}$$

$$v_{be} = v_i \left(\frac{r_{\pi}}{r_{\pi} + 100k\Omega} \right)$$

$$\frac{v_{be}}{v_i} = \frac{(1.09)}{(1.09 + 100)}$$

$$\frac{v_o}{v_{be}} \cdot \frac{v_{be}}{v_i} = \frac{v_o}{v_i} = -92(3) \left(\frac{1.09}{100 + 1.09} \right)$$

$$\frac{V_0}{V_i} = -2.98 \text{ V/V}$$