

Wide-Band E-Shaped Patch Antennas for Wireless Communications

Fan Yang, *Student Member, IEEE*, Xue-Xia Zhang, Xiaoning Ye, and Yahya Rahmat-Samii, *Fellow, IEEE*

Abstract—This paper presents a novel single-patch wide-band microstrip antenna: the E-shaped patch antenna. Two parallel slots are incorporated into the patch of a microstrip antenna to expand its bandwidth. The wide-band mechanism is explored by investigating the behavior of the currents on the patch. The slot length, width, and position are optimized to achieve a wide bandwidth. The validity of the design concept is demonstrated by two examples with 21.2% and 32.3% bandwidths. Finally, a 30.3% E-shaped patch antenna, resonating at wireless communication frequencies of 1.9 and 2.4 GHz, is designed, fabricated, and measured. The radiation pattern and directivity are also presented.

Index Terms—Dual parallel slots, E-shaped, patch antenna, wide band, wireless communications.

I. INTRODUCTION

MICROSTRIP patch antennas are widely used because of their many advantages, such as the low profile, light weight, and conformity. However, patch antennas have a main disadvantage: narrow bandwidth. Researchers have made many efforts to overcome this problem and many configurations have been presented to extend the bandwidth. The conventional method to increase the bandwidth is using parasitic patches. In [1], the authors presented a multiple resonator wide-band microstrip antenna. The parasitic patches are located on the same layer with the main patch. In [2], an aperture-coupled microstrip antenna is described with parasitic patches stacked on the top of the main patch. However, these methods typically enlarge the antenna size, either in the antenna plane or in the antenna height. With the rapid development of wireless communications, single-patch wide-band antennas have attracted many researchers' attention [3]–[5]. In [6], the authors presented a U-slot microstrip antenna and demonstrated that its bandwidth could exceed 30%.

In this paper, we present a novel single-patch wide-band microstrip antenna: the E-shaped patch antenna. When two parallel slots are incorporated into the antenna patch, the bandwidth increases above 30%. Compared to the U-slot microstrip patch antenna, the E-shaped patch antenna is simpler in construction. By only adjusting the length, width, and position of the slots, one can obtain satisfactory performances. Some experimental

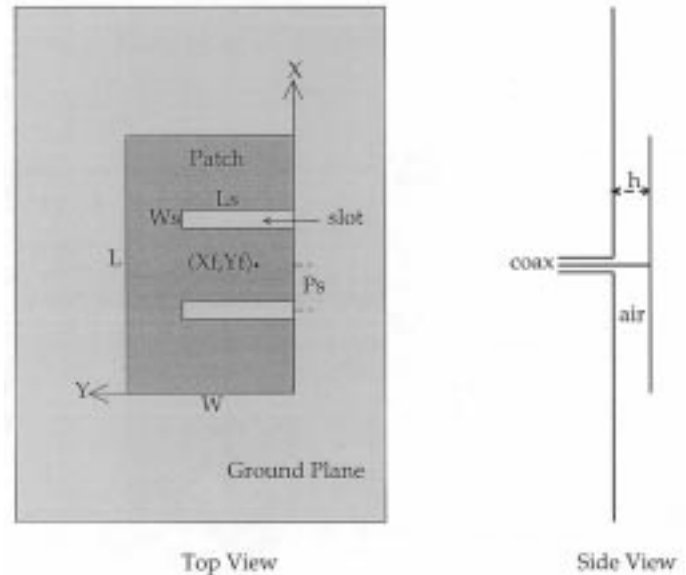


Fig. 1. Geometry of a wide-band E-shaped patch antenna consisting of two parallel slots in the patch.

results prove the validity of this design. The method of moments [7] with the vector triangular basis function [8] is used for analysis, as well as HP-HFSS software. The electric currents on the E-shaped patch are calculated and graphically presented to explain the wide-band mechanism. Subsequently, a wide-band E-shaped patch antenna with 30.3% bandwidth is designed to cover both 1.9 and 2.4 GHz. These ranges of frequencies are very desirable in modern wireless communications. Radiation patterns are also measured and compared with the numerical data.

II. PERFORMANCE FEATURES OF E-SHAPED PATCH ANTENNAS

The antenna geometry is shown in Fig. 1. The antenna has only one patch, which is simpler than traditional wide-band microstrip antennas. The patch size is characterized by (L, W, h) and it is fed by a coaxial probe at position (X_f, Y_f) . To expand the antenna bandwidth, two parallel slots are incorporated into this patch and positioned symmetrically with respect to the feed point. The topological shape of the patch resembles the letter "E," hence the name E-shaped patch antenna. The slot length (L_s) , width (W_s) , and position (P_s) are important parameters in controlling the achievable bandwidth.

Fig. 2 demonstrates the basic idea of the wide-band mechanism of the E-shaped patch antenna. The ordinary microstrip patch antenna can be modeled as a simple LC resonant circuit [Fig. 2(a)] [9]. Currents flow from the feeding point to the

Manuscript received July 10, 1999; revised August 6, 2000.

F. Yang and Y. Rahmat-Samii are with the Department of Electrical Engineering, University of California at Los Angeles, Los Angeles, CA 90095-1594 USA (e-mail: ygfn@ee.ucla.edu; rahmat@ee.ucla.edu).

X.-X. Zhang is with the Department of Electronic Engineering, Tsinghua University, Beijing 100084, China.

X. Ye is with the Department of Electrical and Computer Engineering, University of Missouri at Rolla, Rolla, MO 65401-0249 USA (e-mail: xiaoning@umr.edu).

Publisher Item Identifier S 0018-926X(01)05657-5.

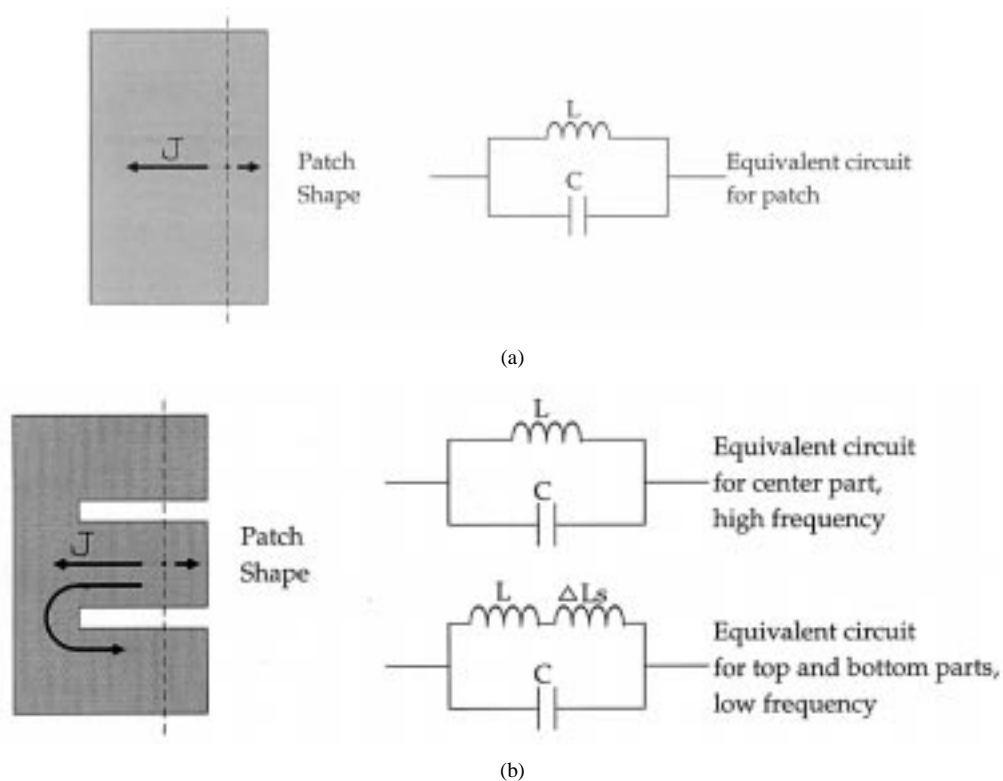


Fig. 2. Dual resonance: the wide-band mechanism of E-shaped patch antennas. (a) The ordinary microstrip patch antenna. (b) The E-shaped patch antenna.

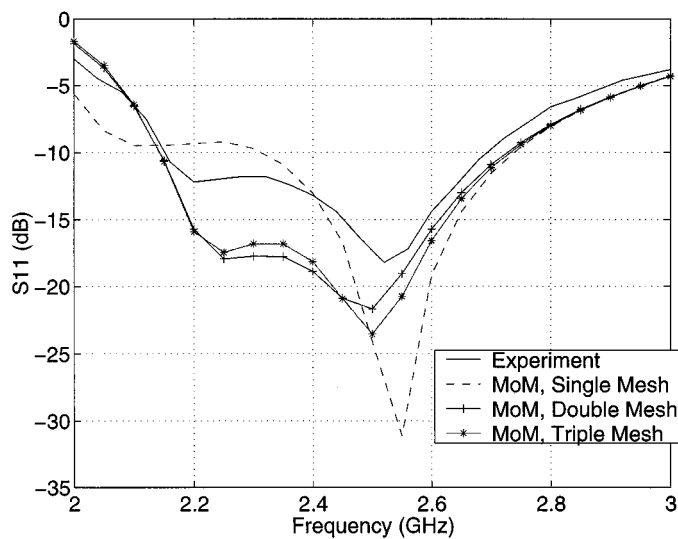


Fig. 3. Measured and calculated S_{11} of a 21.2% bandwidth E-shaped patch antenna.

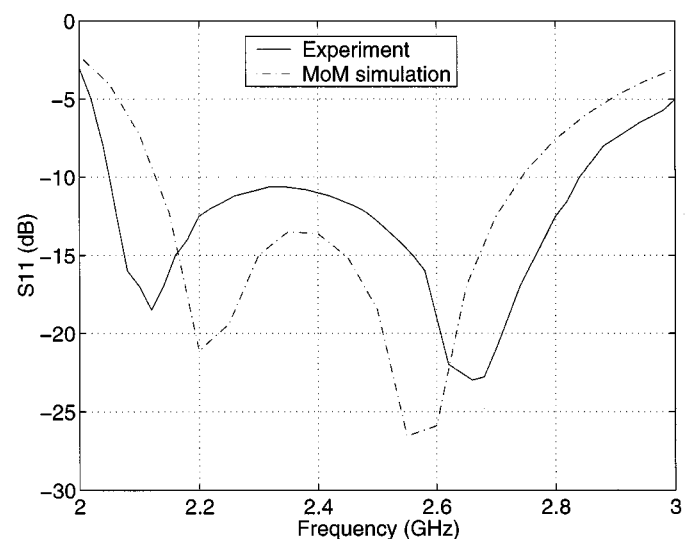
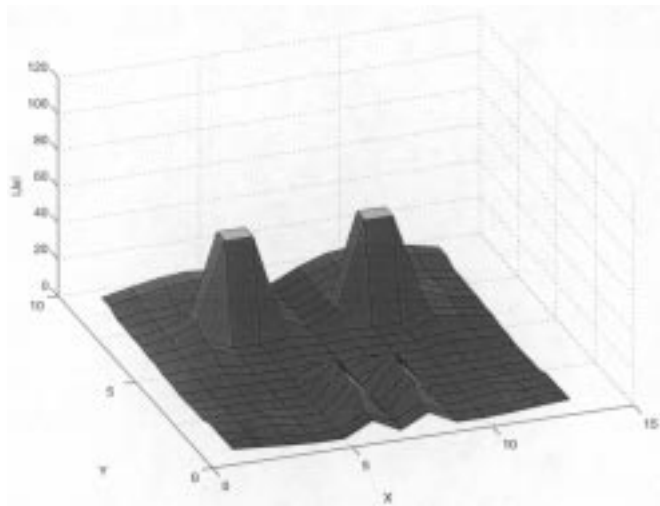


Fig. 4. Measured and calculated S_{11} of a 32.3% bandwidth E-shaped patch antenna.

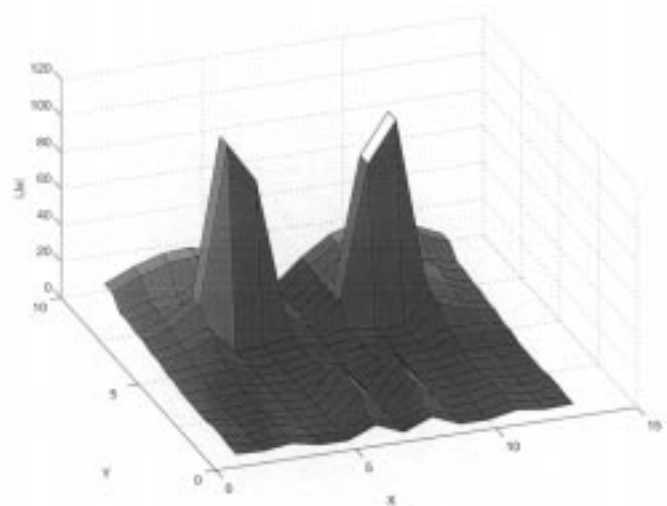
top and bottom edges. L and C values are determined by these currents path length. When two slots are incorporated into the patch, the resonant feature changes, as shown in Fig. 2(b). In the middle part of the patch, the current flows like normal patch. It represents the initial LC circuit and resonates at the initial frequency. However, at the edge part of the patch, the current has to flow around the slots and the length of the current path is increased. This effect can be modeled as an additional series

inductance ΔL_s [10]. So the equivalent circuit of the edge part resonates at a lower frequency. Therefore, the antenna changes from a single LC resonant circuit to a dual resonant circuit. These two resonant circuits couple together and form a wide bandwidth.

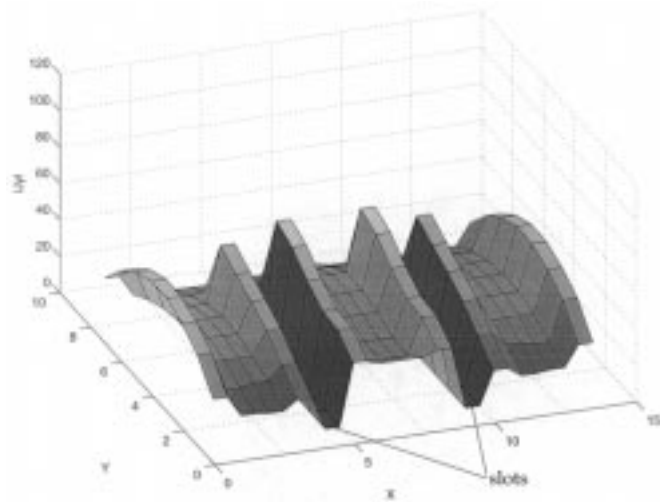
Some experiments and calculations were carried out to demonstrate the performance of this wide-band configuration. The method of moments (MoM) was used for analysis and the



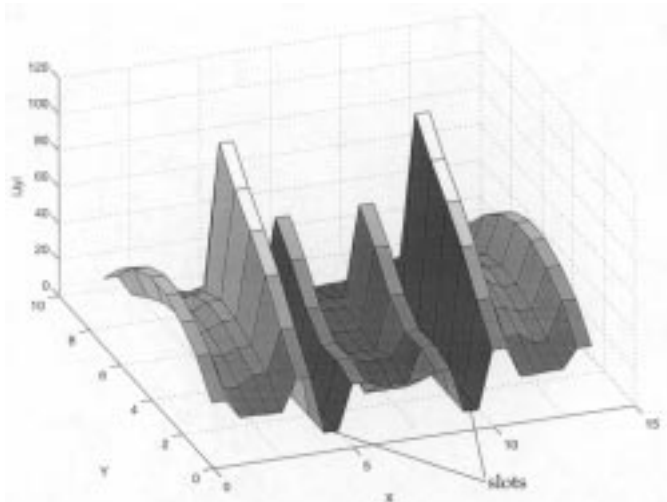
(a)



(a)



(b)



(b)

Fig. 5. The currents on the patch at high resonant frequency (a) J_x and (b) J_y . X, Y denote the number of the cell in MoM, while each cell is $5 \text{ mm} \times 5 \text{ mm}$. The unit for current is amperes/square meter while the excitation voltage is 1 V.

Fig. 6. The currents on the patch at low resonant frequency (a) J_x and (b) J_y . X, Y denote the number of the cell in MoM, while each cell is $5 \text{ mm} \times 5 \text{ mm}$. The unit for current is amperes/square meter while the excitation voltage is 1 V.

vector triangular function was chosen as the basis function. Fig. 3 gives the return loss of a 21.2% bandwidth (at -10 dB) for an E-shaped patch antenna example. The antenna parameters are listed below (in millimeters):

$$(L, W, h) = (70, 45, 10), \quad (Xf, Yf) = (35, 10) \\ L_s = 30, \quad W_s = 5, \quad P_s = 12.5.$$

The S_{11} is measured on an HP-8510 network analyzer. From the figure, it can be observed that the antenna has clearly two resonant frequencies: 2.2 and 2.52 GHz. It agrees well with the explanation given above. The antenna frequency band with -10-dB return loss covers the frequency range of 2.15–2.66 GHz. It has a bandwidth of 21.2% with the center frequency 2.4 GHz. Fig. 3 also shows the MoM simulation with different mesh densities. The patch was divided into nine subsections along the y direction (Ny), which is about 20 cells per wavelength at the high-frequency end, 3 GHz. The initial number of subsections along the x direction (Nx) was 14 so that the subsection size

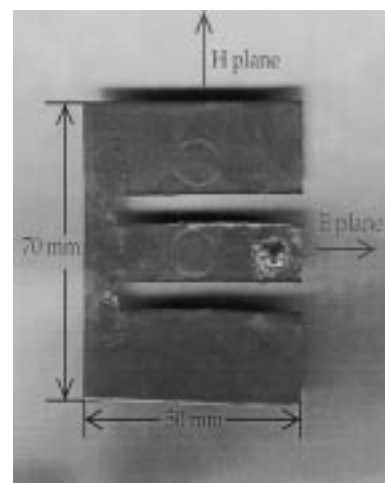


Fig. 7. A photo of an E-shaped patch antenna resonating at wireless communication frequencies of 1.9 and 2.4 GHz.

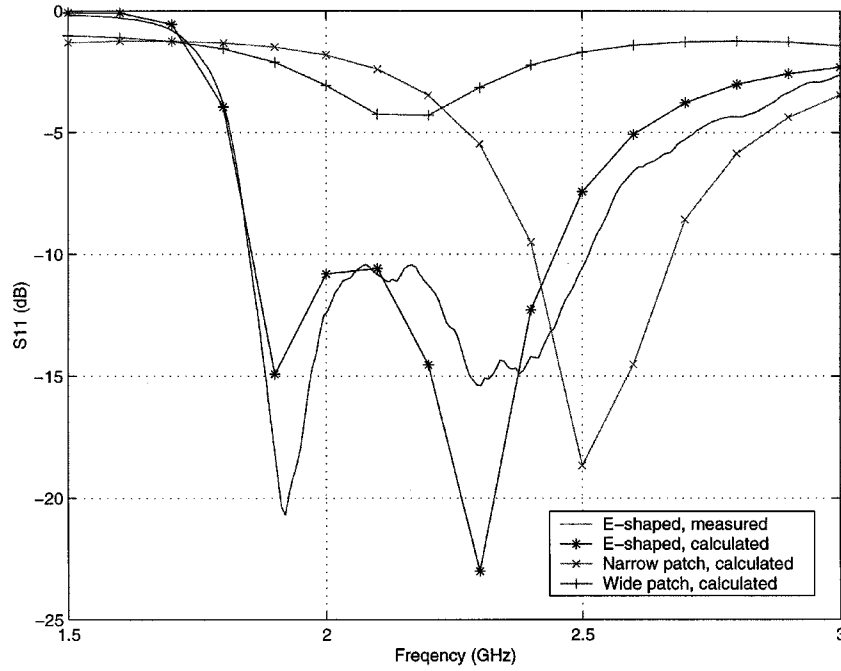


Fig. 8. S_{11} of the E-shaped patch antenna for wireless communications (measured and calculated), compared with simple patch antennas without slots.

was the same in both directions. However, the computation result of this single-density mesh was not good. The subsections number was increased to obtain an accurate result. It was found that the number of subsections along the x direction was not enough while it was sufficient along the y direction. The reason was that the currents had more variations along the x direction due to the two slots. The double-mesh ($N_x = 28, N_y = 9$) and triple-mesh ($N_x = 42, N_y = 9$) results are also presented. The converged results agree well with the experimental result.

By properly adjusting the parameters of slots and the position of the feeding point, a 32.3% antenna bandwidth is achieved. An elegant approach for the parameter selection would be the utilization of modern genetic algorithms [11]. Fig. 4 gives the return loss of this 32.3% E-shaped patch antenna. The antenna parameters are listed below (in millimeters):

$$(L, W, h) = (70, 45, 10), \quad (X_f, Y_f) = (35, 7) \\ L_s = 35, \quad W_s = 4, \quad P_s = 9.$$

This antenna also has two distinct resonant frequencies. One is 2.12 GHz and the other is 2.66 GHz. The antenna frequency band covers the range of 2.05–2.64 GHz and achieves a bandwidth of 32.3%.

To thoroughly comprehend the effect of the slots, Figs. 5 and 6 present the currents at two resonant frequencies. First, one can see that, in both frequencies, currents J_y are greater than J_x . This is reasonable because the basic cavity mode under the patch is in the TM_{01} mode. Second, there are strong currents flowing around the slots. The amplitudes of the J_y currents become larger along the slots. At the ends of the slots, the amplitude of J_y is the largest. Here, J_x with large amplitude also appears. The amplitude of J_x is almost the same as the amplitude of J_y . Thus, it guarantees the continuity of the currents around the slots. Most importantly, the amplitudes of currents around the slots are different at low resonant frequencies and high res-

onant frequencies. It means that the effects of the slots at these two resonant frequencies are different. This is the key reason why the slots can extend the bandwidth. At the high frequency, the amplitudes of the currents around the slots are almost the same as those at the left and right edges of the patch. The effect of the slots are not significant. The patch works like ordinary patch. Therefore, the high resonant frequency is mainly determined by the patch width W , less affected by the slots. While at the low frequency, the amplitudes of the currents around slots are greater than those at high frequency. The slots congregate the currents and this effect can be modeled as an inductance. Due to this additional inductance effect, it resonates at a low frequency. Thus, this lower resonant frequency is mainly characterized by the slots. This phenomena agrees well with our previous explanation. Now it can be concluded that the antenna width W controls the higher resonant frequency while the slots control the lower resonant frequency. Because of the dual resonant character, this kind of microstrip antenna can achieve a wide bandwidth.

III. E-SHAPED PATCH ANTENNA DESIGN FOR 1.9–2.4 GHz WIRELESS COMMUNICATIONS

In this section, a wide-band E-shaped patch antenna for wireless communications is characterized in detail. A photo of the antenna is shown in Fig. 7. The antenna parameters are listed below (in millimeters):

$$(L, W) = (70, 50), \quad h = 15, \quad (X_f, Y_f) = (35, 6) \\ L_s = 40, \quad W_s = 6, \quad P_s = 10.$$

Fig. 8 shows the S_{11} results of this E-shaped patch antenna. The S_{11} is calculated by HP-HFSS software and measured on an HP-8510 network analyzer. From the figure, one can observe that the E-shaped patch antenna resonates at 1.9 and 2.4 GHz. These frequencies are chosen because they are useful frequen-

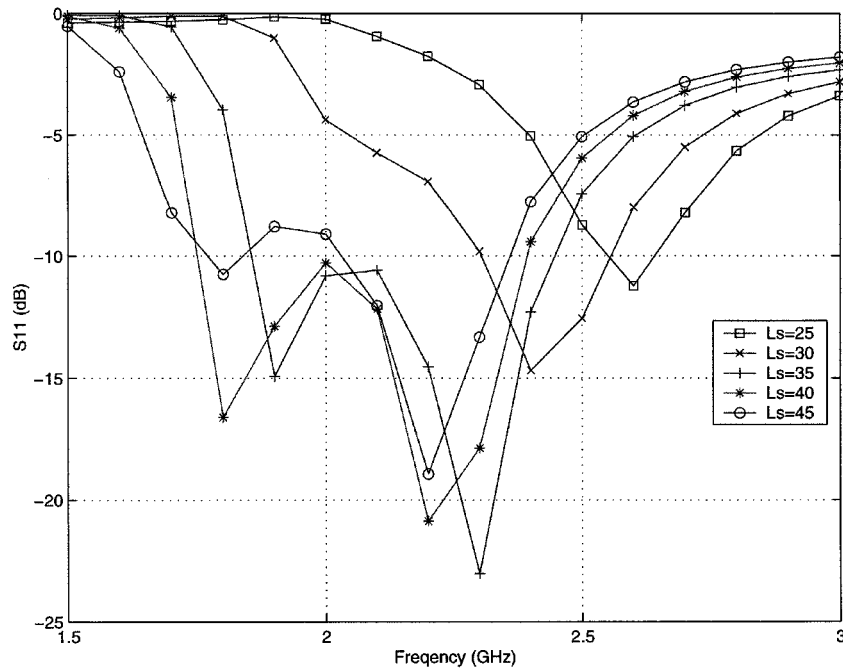


Fig. 9. Calculated S_{11} of E-shaped patch antennas with different slot lengths.

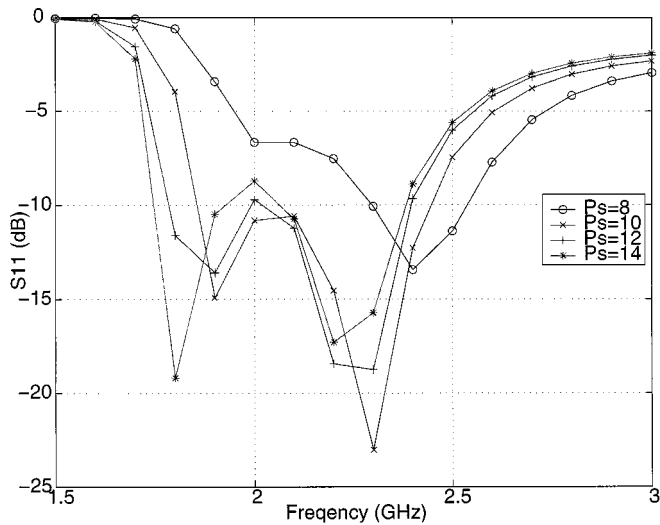


Fig. 10. Calculated S_{11} of E-shaped patch antennas with different slot positions.

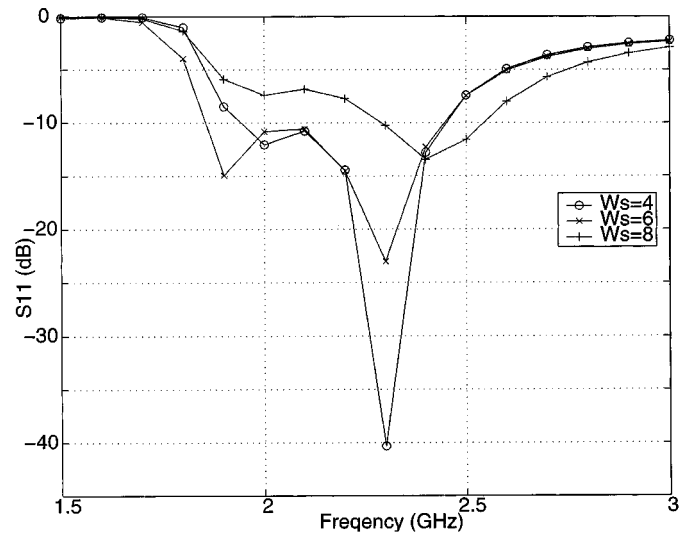


Fig. 11. Calculated S_{11} of E-shaped patch antennas with different slot widths.

cies in modern wireless communications. The E-shaped patch antenna has a wide bandwidth of 30.3%. The simple patch antennas without slots are also simulated for comparison. They have the same height and width as this E-shaped patch antenna. The narrow patch antenna, which has the same length as the middle part of the E-shaped patch antenna, has a bandwidth of 10% while the wide patch antenna with the same length as the E-shaped patch antenna doesn't match to 50 Ω .

Slots play an important role to control the wide-band behavior of the E-shaped patch antenna. There are three parameters to characterize the slots, namely slot length, slot position, and slot width. Fig. 9 shows the effect of the slot length (L_s) on the antenna. When the slot length is small, the antenna only has

one resonant frequency. When the slot length increases, another lower resonant frequency appears. The longer the slot length, the lower the second resonant frequency. In brief, the slot length is an important parameter to characterize the resonant frequencies of the E-shaped patch antenna. The slot position (P_s) is presented in Fig. 10. When P_s is small, the S_{11} at lower frequencies does not match well. When P_s becomes larger, the two resonant frequencies become distinct and a wide-band match is obtained. However, when P_s becomes even larger, the S_{11} between two resonant frequencies is larger than -10 dB. The antenna does not perform as a wide-band one but rather as a dual-frequency one. Therefore, P_s is a useful parameter to adjust the matching to 50 Ω . Fig. 11 details the importance of the slot width. The two

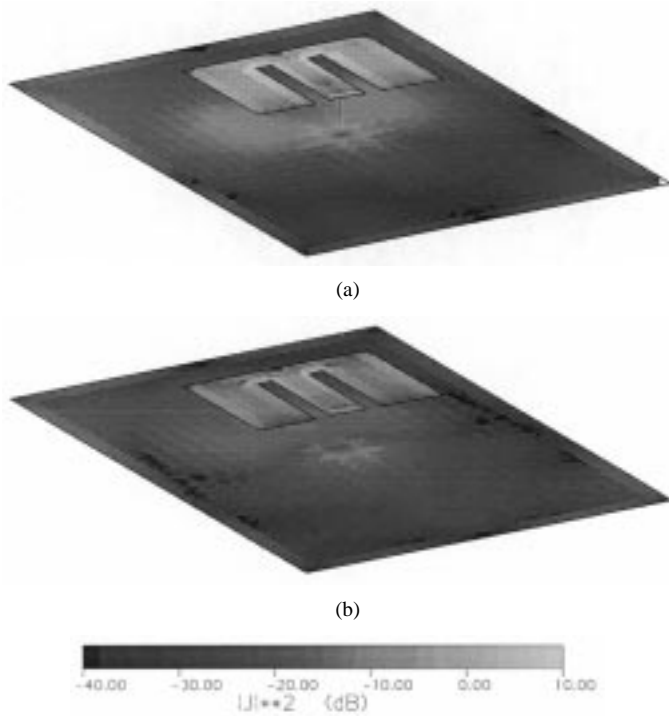


Fig. 12. Currents of the E-shaped patch antenna with finite ground plane at (a) 1.9 GHz and (b) 2.4 GHz.

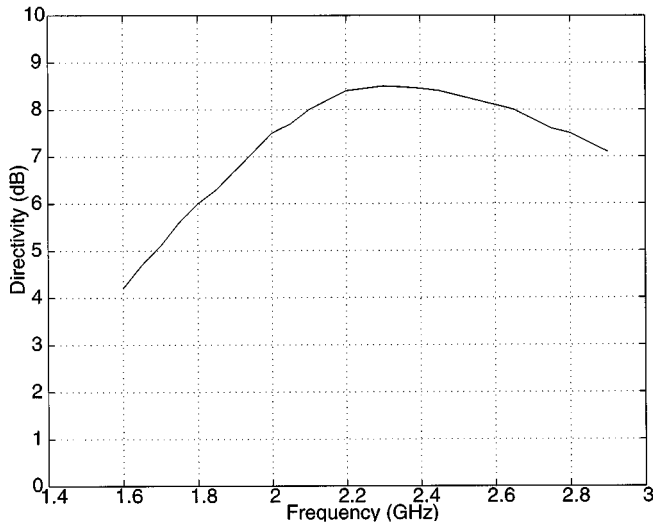
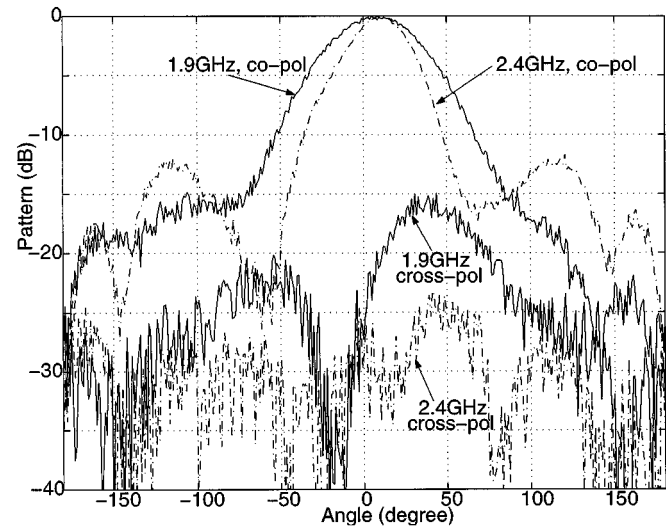


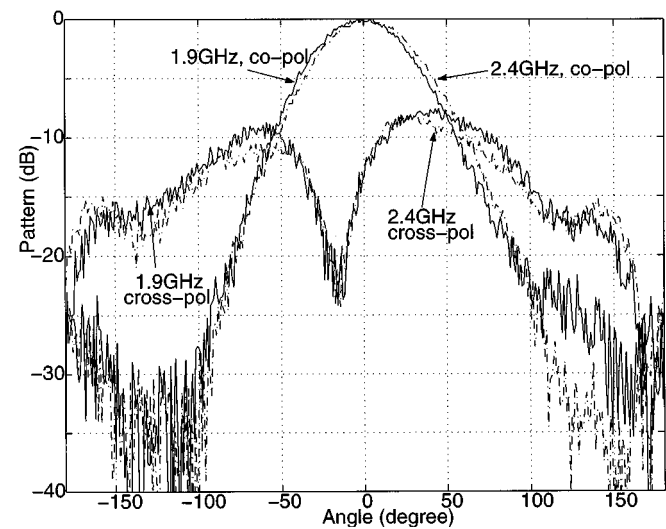
Fig. 13. Calculated directivity of the E-shaped patch antenna versus frequency.

resonant frequencies exist in all three cases, but the best match can be obtained only when $Ps = 6$ mm, which means that slot width is useful to adjust coupling and achieve good match.

Usually the ground plane effect is a critical factor in communication applications. A large finite ground plane is analyzed and the currents on the ground plane are calculated to understand which areas are effective for the antenna operation. The ground size is chosen to be $14 \text{ cm} \times 21 \text{ cm}$ which is about 8 times that of the patch size using the technique developed in [12]. Fig. 12 shows the simulated currents of the antenna structure with finite ground plane. One can observe that most currents concentrate under the patch and more areas are needed for low frequency than for high frequency. In other words, the



(a)



(b)

Fig. 14. Measured co-pol and cross-pol radiation patterns at two resonant frequencies of 1.9 GHz and 2.4 GHz. (a) *E* plane pattern. (b) *H* plane pattern.

ground plane size is determined by the low frequency. If -20 dB is chosen as the criterion for the relative level of ground plane current amplitude, $1\lambda \times 1\lambda$ (1.9 GHz) ground plane area is sufficient for the proper operation of the antenna.

Fig. 13 presents the calculated directivity of the antenna. It is 8.5 dB at 2.4 GHz and 6.7 dB at 1.9 GHz. The frequency range from 1.9 to 2.4 GHz is inside the 3-dB directivity band. Since the antenna matches well in this frequency range, it should have the similar level of gain. The radiation pattern of the E-shaped patch antenna is measured in the far-field chamber located at the UCLA antenna lab and shown in Fig. 14. They are measured at two resonant frequencies: 1.9 and 2.4 GHz. The experimental results agree well with the numerical results which are not shown here. In the *E* plane, the 3-dB beamwidth is 42 degrees at 2.4 GHz and 63 degrees at 1.9 GHz. The peak cross-pol at 1.9 GHz is -15 dB, which is higher than -25 dB at 2.4 GHz. This is due to different current distributions at these two frequencies. In the *H* plane, the radiation pattern is similar at 1.9 GHz and 2.4 GHz. The 3-dB beamwidth is 60 degrees at both

frequencies. The peak cross-pol is -7 dB at 50 degrees. This high cross-pol is generated by the leaky radiation of the slots. However, it's still acceptable for some communication applications.

IV. CONCLUSION

The E-shaped patch antenna with wide bandwidth is presented in this paper. Compared to conventional wide-band microstrip patch antennas, it has the attractive features of simplicity and small size. The electric currents on the patch are determined and the wide-band mechanism is discussed in depth. Finally, a 30.3% bandwidth E-shaped patch antenna, applicable to modern wireless communication frequencies of 1.9 to 2.4 GHz, is designed, measured, and characterized in detail.

ACKNOWLEDGMENT

The authors would like to thank M. Djordjevic for his calculation of the currents on the ground plane and the antenna directivity.

REFERENCES

- [1] G. Kumar and K. C. Gupta, "Directly coupled multiple resonator wide-band microstrip antenna," *IEEE Trans. Antennas Propagat.*, vol. AP-33, pp. 588–593, June 1985.
- [2] D. M. Pozar, "Microstrip antenna coupled to a microstrip-line," *Electron. Lett.*, vol. 21, no. 2, pp. 49–50, Jan. 1985.
- [3] K. L. Virga and Y. Rahmat-Samii, "Low profile enhanced-bandwidth PIFA antennas for wireless communications packaging," *IEEE Trans. Microwave Theory Tech.*, vol. 45, pp. 1879–1888, Oct. 1997.
- [4] I. Papapolymerous, R. F. Drayton, and L. P. B. Katehi, "Micromachined patch antennas," *IEEE Trans. Antennas Propagat.*, vol. 46, pp. 275–283, Feb. 1998.
- [5] N. Herscovici, "New considerations in the design of microstrip antennas," *IEEE Trans. Antennas Propagat.*, vol. 46, pp. 807–812, June 1998.
- [6] T. Huynh and K. F. Lee, "Single-layer single-patch wideband microstrip antenna," *Electron. Lett.*, vol. 31, no. 16, pp. 1310–1312, Aug. 1995.
- [7] E. H. Newman and P. Tulyathan, "Analysis of microstrip antennas using moment methods," *IEEE Trans. Antennas Propagat.*, vol. AP-29, pp. 47–53, Jan. 1981.
- [8] S. M. Rao, D. R. Wilton, and A. W. Glisson, "Electromagnetic scattering by surfaces of arbitrary shape," *IEEE Trans. Antennas Propagat.*, vol. AP-30, pp. 409–418, May 1982.
- [9] K. C. Gupta and A. B. Norwood, *Microstrip Antenna Design*. Norwood, MA: Artech House, 1988.
- [10] X.-X. Zhang and F. Yang, "The study of slit cut on the microstrip antenna and its applications," *Microwave Opt. Technol. Lett.*, vol. 18, no. 4, pp. 297–300, July 1998.
- [11] Y. Rahmat-Samii and E. Michielssen, *Electromagnetic Optimization by Genetic Algorithms*. New York, NY: Wiley, 1999.
- [12] R. E. Hodges and Y. Rahmat-Samii, "An iterative current-based hybrid method for complex structures," *IEEE Trans. Antennas Propagat.*, vol. 45, pp. 265–276, Feb. 1997.



Fan Yang (S'96) received the B.S. and M.S. degrees in electronic engineering from Tsinghua University, Beijing, China, in 1997 and 1999, respectively. He is currently working toward the Ph.D. degree at the University of California, Los Angeles.

From 1995 to 1999, he was a Research Assistant in the State Key Laboratory of Microwave & Digital Communications, Tsinghua University, China. Since September 1999, he has been working as a graduate research assistant in the UCLA Antenna Laboratory.

His research interests are microstrip antenna design, computational electromagnetics and photonic bandgap surfaces.



Xue-Xia Zhang Graduated from Tsinghua University, Beijing, China, in 1958.

From 1959 to 1961, she was a Visiting Scholar with Moscow Power Institute, U.S.S.R. Currently, she is a Professor with the Department of Electronic Engineering, Tsinghua University and a Fellow of the Chinese Institute of Electronics. Her research interests include antennas, electromagnetic field and microwave technology.

Xiaoning Ye was born in China in 1973. He received the B.S. and M.S. degrees in electronics engineering from Tsinghua University, Beijing, China, in 1995, and 1998, respectively.

Since 1997, he has studied and worked in the Electromagnetic Compatibility Laboratory at University of Missouri-Rolla. His research interests include numerical and experimental study of electromagnetic compatibility and signal integrity problems, and microstrip patch antennas.



Yahya Rahmat-Samii (F'85) received the M.S. and Ph.D. degrees in electrical engineering from the University of Illinois, Urbana-Champaign.

Currently, he is a Professor and Chairman of the Department of Electrical Engineering at the University of California, Los Angeles (UCLA). He was a Senior Research Scientist with NASA's Jet Propulsion Laboratory/California Institute of Technology, Pasadena, before joining UCLA. In the summer of 1986, he was a Guest Professor with the Technical University of Denmark (TUD). He

has also served as consultant to many aerospace companies. He has been the guest and plenary session speaker, as well as Chairman and Co-Chairman at many national and international symposia. He has authored and coauthored over 500 technical journal articles and conference papers, and has written fourteen book chapters. He is coauthor of, *Electromagnetic Optimization by Genetic Algorithms*, and *Impedance Boundary Conditions in Electromagnetics* published in 1999 and 1995, respectively. He holds several patents and has had pioneering research contributions in diverse areas of electromagnetics, antennas, measurement and diagnostics techniques, numerical and asymptotic methods, satellite and personal communications and human/antenna interactions, etc. (visit <http://www.ee.ucla.edu>).

Dr. Rahmat-Samii is a member of Commissions A, B, J, and K of USNC/URSI, AMTA, Sigma Xi, Eta Kappa Nu, and the Electromagnetics Academy. He was elected a Fellow of the IAE in 1986. He was a Director and Vice President of the Antennas Measurement Techniques Association (AMTA) for three years. He was a member of UCLA's Graduate council for three years. For his contributions, he has received numerous NASA and JPL Certificates of Recognition. In 1984, he received the prestigious Henry Booker Award of URSI. In 1999, he received the University of Illinois ECE Distinguished Alumni Award. In 2001, he received an Honorary Doctorate from the University of Santiago de Compostela, Spain. In 1993, 1994 and 1995, three of his Ph.D. students were named the Most Outstanding Ph.D. Students at UCLA's School of Engineering and Applied Science. Eight others received various Student Paper Awards at the 1993, 1996, 1997, 1998, 1999, and 2000 IEEE AP-S/URSI Symposiums. He was the 1995 President and 1994 Vice-President of the IEEE Antennas and Propagation Society. He was appointed an IEEE Antennas and Propagation Society Distinguished Lecturer and presented lectures internationally. He was a member of the Strategic Planning and Review Committee (SPARC) of IEEE. In 1992 and 1995, he received the Best Application Paper Award (Wheeler Award) for papers published in the 1991 and 1993 IEEE AP-S Transactions. In 2000, he was selected as the recipient of IEEE Third Millennium Medal. He is listed in *Who's Who in America*, *Who's Who in Frontiers of Science and Technology*, and *Who's Who in Engineering*.