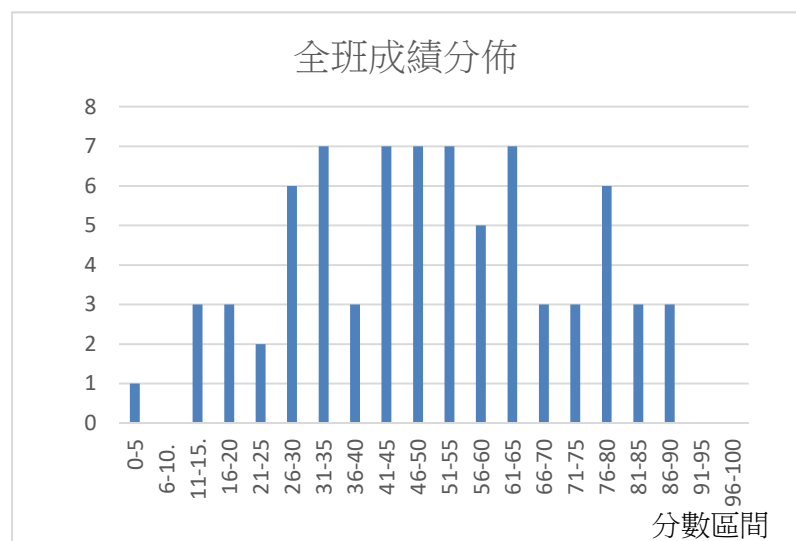


%%%%%%%%%

成績統計：

組距	人數
0-5	1
6-10.	0
11-15.	3
16-20	3
21-25	2
26-30	6
31-35	7
36-40	3
41-45	7
46-50	7
51-55	7
56-60	5
61-65	7
66-70	3
71-75	3
76-80	6
81-85	3
86-90	3
91-95	0
96-100	0
交卷人數	76
缺考人數	10
班平均	50.18



常見錯誤：

題號	錯誤	正確
Q1	(n) $w/2$ (m) 空白 (l) 空白 (o) Parseval	(n) 2W (m) $\phi(t) \ll 1$ (l) constant instantaneous power (o) Carson's rule
Q2	(a) 直接寫答案，沒有 \cos 角度變化過程 (c) 直接寫答案，沒有寫 $1/2$ 怎麼得到的	
Q3	$\mathcal{F}\{\cos(2\pi f_c t)\} = \delta(f - f_c) + \delta(f + f_c)$	$\mathcal{F}\{\cos(2\pi f_c t)\} = \frac{1}{2}[\delta(f - f_c) + \delta(f + f_c)]$
Q4	Index 標示有誤 fm-fc	Index 標示有誤 fc-fm
	message signal 為脈衝非三角波。 $\mathcal{F}\{\cos(2\pi f_c t)\} = \delta(f - f_c) + \delta(f + f_c)$ $\mathcal{F}\{\cos(2\pi f_c t)\} = \frac{1}{2}[\delta(f - f_c) + \delta(f + f_c)]$	
Q5	沒有寫單位 (rad)	
Q6	$a = \frac{1}{5} = 0.2$ $E = \frac{a \langle m_n^2(t) \rangle}{1 + a \langle m_n^2(t) \rangle} \times 100\%$	$a = \frac{10}{25} = 0.4$ $E = \frac{a^2 \langle m_n^2(t) \rangle}{1 + a^2 \langle m_n^2(t) \rangle} \times 100\%$

系所	學號	姓名	HW#2
電機系 4B	402415078	沈上荏	O
電機系 4A	405220003	林柏漢	V
電機系 4A	405220012	吳定濬	V
電機系 4A	405415007	陳億賢	X
電機系 4B	405415010	劉鎮宇	X
電機系 4A	405415029	呂羿葦	V
電機系 4B	405415048	鍾文宏	V
電機系 4B	405415052	李曜任	V
電機系 4A	405415075	林詩婷	V
電機系 4B	405415076	李承侑	V
電機系 4A	405430008	蔡秉欣	V
電機系 3A	406415001	張言睿	V
電機系 3A	406415005	賴欣儀	V
電機系 3B	406415010	陳信宏	V
電機系 3A	406415011	張峻祥	V
電機系 3A	406415013	連奕丞	X
電機系 3B	406415014	許博惟	V
電機系 3B	406415018	蔡培鑑	V
電機系 3B	406415020	楊博禕	V
電機系 3A	406415025	劉彥廷	X
電機系 3A	406415027	鄭筠賡	V
電機系 3B	406415028	黃柏瑜	V
電機系 3B	406415030	陳昱凱	V
電機系 3B	406415032	姚松伯	V
電機系 3A	406415033	詹佰晉	V
電機系 3A	406415035	陳彥邦	V
電機系 3B	406415036	吳祝樟	V
電機系 3B	406415038	黃柏軒	X
電機系 3A	406415049	郭 恕	O
電機系 3A	406415055	朱冠誌	V
電機系 3B	406415058	黃彥勛	V
電機系 3A	406415061	楊詠舜	V
電機系 3B	406415064	莊博傑	V
電機系 3A	406415071	詹育平	V
電機系 3B	406415082	陳妍臻	V

電機系 3B	406420062	黃議賢	X
電機系 3A	406530020	洪澤廷	V
N/A	408415901	張國璇 (GUOXUAN ZHANG)	V
N/A	408415904	孫道源 (DAOYUAN SUN)	V
N/A	408420904	吳俊宏 (JUNHONG WU)	V
N/A	408420906	王崧年 (SONGNIAN WANG)	V
N/A	408420908	江育聰 (YUCONG JIANG)	V

系所	學號	姓名	HW#2
通訊系 3A	404430045	王振宇	X
通訊系 4A	405430007	林郁鈞	X
通訊系 4A	405430013	黃信豪	V
通訊系 4A	405430032	區雅婷	O
通訊系 4A	405430039	林佑宸	V
通訊系 3A	406430001	陳有朋	V
通訊系 3A	406430002	鄭佳宣	V
通訊系 3A	406430003	丁凱文	V
通訊系 3A	406430004	姜昱丞	V
通訊系 3A	406430006	李柏豫	V
通訊系 3A	406430007	王繼賢	V
通訊系 3A	406430009	曾國珩	V
通訊系 3A	406430010	鄭宇倫	V
通訊系 3A	406430011	邱靖博	V
通訊系 3A	406430012	李逸帆	V
通訊系 3A	406430013	陳美瑜	O
通訊系 3A	406430014	王信荃	V
通訊系 3A	406430015	簡敬倫	V
通訊系 3A	406430016	陳姿妤	V
通訊系 3A	406430017	鍾念慈	V
通訊系 3A	406430018	詹哲嘉	V
通訊系 3A	406430019	黃宇傑	X
通訊系 3A	406430020	陳昱瑋	V
通訊系 3A	406430022	徐子翔	V
通訊系 3A	406430023	林宥均	V

通訊系 3A	406430024	曹智捷	V
通訊系 3A	406430026	楊鎧安	V
通訊系 3A	406430028	李翊嘉	X
通訊系 3A	406430029	廖辰偉	V
通訊系 3A	406430030	廖岳軒	V
通訊系 3A	406430033	李孟寰	V
通訊系 3A	406430034	鄭又華	V
通訊系 3A	406430035	陳御任	V
通訊系 3A	406430036	陳奕禹	X
通訊系 3A	406430037	郭慶安	X
通訊系 3A	406430038	林顯庭	V
通訊系 3A	406430039	江昱瑩	V
通訊系 3A	406430040	翁浩均	V
通訊系 3A	406430041	鄭博駿	V
通訊系 3A	406430042	陳彥邑	V
通訊系 3A	406430043	楊士緯	V
通訊系 3A	406430044	郭信俞	V
通訊系 3A	406430045	廖章甫	X
通訊系 4A	405430034	徐國越	V

1. 每題 2 分，共 30 分

(a) $A_c m(t) \cos(2\pi f_c t)$

(b) $\frac{A_c}{2} [M(f - f_c) + M(f + f_c)]$

(c) $A_c [1 + am_n(t)] \cos(2\pi f_c t)$

(d) $A_c [1 + am_n(t)]$ or $1 + am_n(t)$

(e)

(f) $x_{SSB}(t) = \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \mp \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t).$

(g) $[H(f - f_c) + H(f + f_c)] = \text{constant}, |f| \leq W$

(h) $\cos(2\pi(f_1 \pm f_2)t)$

(i) $A_c \cos(2\pi f_c t + k_p m(t))$

(j) $A_c \cos(2\pi f_c t + \int_{-\infty}^t 2\pi f_d m(\alpha) d\alpha)$

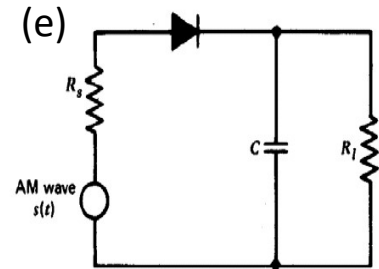
(k) $J_n(\beta) \triangleq \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{-[jnx - \beta \sin(x)]} dx$

(l) Constant instantaneous power or robust to additive noise

(m) $\phi(t)$ is small or $\phi(t) \ll 1$

(n) $2W$

(o) Carson's rule



2. 每題 5 分，共 15 分

- a) Assume that a DSB signal $x_c(t) = A_c m(t) \cos(2\pi f_c t + \phi_0)$ is

demodulated using the demodulation carrier $2 \cos[2\pi f_c t + \theta(t)]$.

Determine, in general, the demodulated output $y_D(t)$.

Sol:

The demodulated output, in general, is

$$y_D(t) = \text{Lp}\{x_c(t) 2 \cos[\omega_c t + \theta(t)]\}$$

where $\text{Lp}\{\bullet\}$ denotes the lowpass portion of the argument. With

$$x_c(t) = A_c m(t) \cos[\omega_c t + \phi_0]$$

the demodulated output becomes

$$\begin{aligned} y_D(t) &= \text{Lp}\{2A_c m(t) \cos[\omega_c t + \phi_0] \cos[\omega_c t + \theta(t)]\} \\ &= \text{Lp}\left\{2A_c m(t) \frac{1}{2} (\cos[2\omega_c t + \theta(t) + \phi_0] + \cos[\theta(t) - \phi_0])\right\} \end{aligned}$$

Performing the indicated multiplication and taking the lowpass portion yields

$$y_D(t) = A_c m(t) \cos[\theta(t) - \phi_0]$$

- b) Let $A_c = 1$ and $\theta(t) = \theta_0$, where θ_0 is a constant, and determine the mean-square error between $m(t)$ and the demodulated output as a function of ϕ_0 and θ_0 .

Sol:

If $\theta(t) = \theta_0$ (a constant), the demodulated output becomes

$$y_D(t) = A_c m(t) \cos(\theta_0 - \phi_0)$$

Letting $A_c = 1$ gives the error

$$\varepsilon(t) \triangleq m(t) - y_D(t) = m(t) [1 - \cos(\theta_0 - \phi_0)]$$

The mean-square error is

$$\langle \varepsilon^2(t) \rangle = \langle m^2(t) [1 - \cos(\theta_0 - \phi_0)]^2 \rangle$$

where $\langle \bullet \rangle$ denotes the time-average value. Since the term $[1 - \cos(\theta_0 - \phi_0)]$

is a constant, we have $\langle \varepsilon^2(t) \rangle = \langle m^2(t) \rangle [1 - \cos(\theta_0 - \phi_0)]^2$.

Note that for $\theta_0 = \phi_0$, the demodulation carrier is phase coherent with the original modulation carrier, and the mean-squared error $\langle \varepsilon^2(t) \rangle$ is zero.

- c) For $\theta(t) = \omega_0 t$, $A_c = 1$, we have the demodulated output for convenience

$$y_D(t) = m(t) \cos(\omega_0 t - \phi_0),$$

give the error

$$\varepsilon(t) = m(t) [1 - \cos(\omega_0 t - \phi_0)],$$

giving the mean-square error

$$\langle \varepsilon^2(t) \rangle = \langle m^2(t) [1 - \cos(\omega_0 t - \phi_0)]^2 \rangle$$

The MSE

$$\begin{aligned} \langle \varepsilon^2(t) \rangle &= \langle m^2(t) [1 - \cos(\omega_0 t - \phi_0)]^2 \rangle \\ &= \left\langle m^2(t) \cdot \left[1 - 2 \cos(\omega_0 t - \phi_0) + \frac{1}{2} (1 + \cos(2\omega_0 t - 2\phi_0)) \right] \right\rangle \\ &= \frac{3}{2} \langle m^2(t) \rangle + \left\langle m^2(t) \cdot \left[-2 \cos(\omega_0 t - \phi_0) + \frac{1}{2} \cos(2\omega_0 t - 2\phi_0) \right] \right\rangle \\ &\approx \frac{3}{2} \langle m^2(t) \rangle + \left\langle m^2(t) \right\rangle \left\langle \left[-2 \cos(\omega_0 t - \phi_0) + \frac{1}{2} \cos(2\omega_0 t - 2\phi_0) \right] \right\rangle, \\ &\quad f_0 \text{ far larger than } W \\ &= \frac{3}{2} \langle m^2(t) \rangle \end{aligned}$$

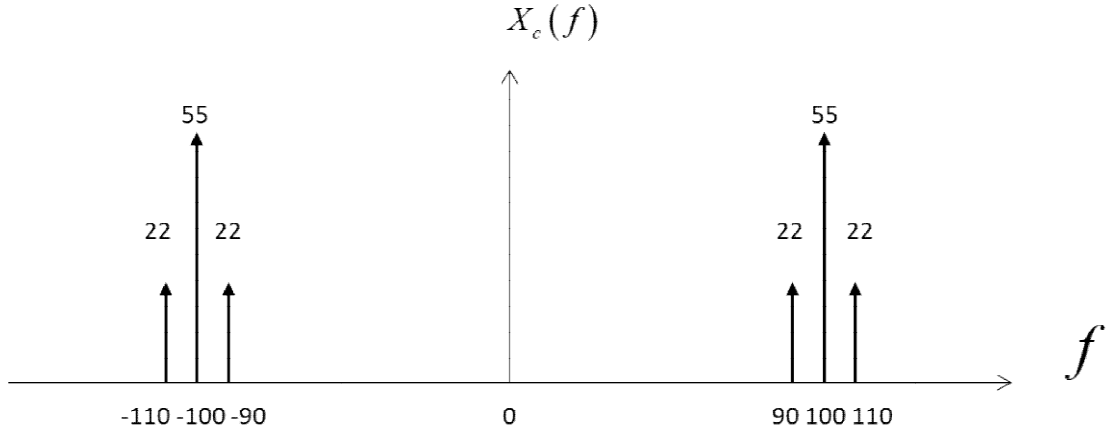
3. 每題 2/2/3/3 分，共 10 分

$$(a) m_n(t) \triangleq \frac{m(t)}{\max |m(t)|} = \frac{9 \cos(20\pi t)}{9} = \cos(20\pi t)$$

$$(b) \langle m_n^2(t) \rangle = \langle \cos^2(20\pi t) \rangle = \left\langle \frac{1}{2} \langle 1 + \cos(40\pi t) \rangle \right\rangle = \frac{1}{2} = 0.5$$

$$(c) E = \frac{a^2 \langle m_n^2(t) \rangle}{1 + a^2 \langle m_n^2(t) \rangle} = \frac{[(0.8)^2 \cdot 0.5]}{[1 + (0.8)^2 \cdot 0.5]} = 0.2424 = 24.24\%$$

$$\begin{aligned} (d) x_c(t) &= 110 [1 + 0.8 \cos(20\pi t)] \cos(200\pi t) \\ &= 110 \cos(200\pi t) + 44 \cos(180\pi t) + 44 \cos(220\pi t) \end{aligned}$$



4. 每題 10 分，共 10 分

$$x_c(t) = \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \pm \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t)$$

$$m(t) = 4 \cos(2\pi f_m t) + \cos(4\pi f_m t)$$

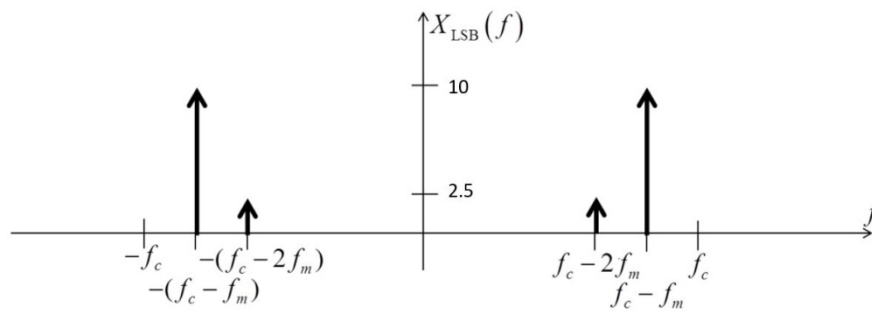
$$\text{Hilbert Transform : } \hat{m}(t) = 4 \sin(2\pi f_m t) + \sin(4\pi f_m t)$$

$$\begin{aligned} x_c(t) &= \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \pm \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t) \\ &= \frac{1}{2} A_c (4 \cos(2\pi f_m t) + \cos(4\pi f_m t)) \cos(2\pi f_c t) \pm \frac{1}{2} A_c (4 \sin(2\pi f_m t) + \sin(4\pi f_m t)) \sin(2\pi f_c t) \\ &= A_c \cos[2\pi(f_c + f_m)t] + A_c \cos[2\pi(f_c - f_m)t] + \frac{1}{4} A_c \cos[2\pi(f_c + 2f_m)t] + \frac{1}{4} A_c \cos[2\pi(f_c - 2f_m)t] \\ &\quad \pm \{-A_c \cos[2\pi(f_c + f_m)t] + A_c \cos[2\pi(f_c - f_m)t]\} + \{-\frac{1}{4} A_c \cos[2\pi(f_c + 2f_m)t] + \frac{1}{4} A_c \cos[2\pi(f_c - 2f_m)t]\} \end{aligned}$$

Lower –sideband:

$$\begin{aligned}
 x_{LSB}(t) &= \frac{1}{2} A_c m(t) \cos(2\pi f_c t) - \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t) \\
 &= 2 A_c \cos[2\pi(f_c - f_m)t] + \frac{1}{2} A_c \cos[2\pi(f_c - 2f_m)t], \\
 X_{LSB}(f) &= A_c \delta[f - (f_c - f_m)] + A_c \delta[f + (f_c - f_m)] \\
 &+ \frac{1}{4} A_c \delta[f - (f_c - 2f_m)] + \frac{1}{4} A_c \delta[f + (f_c - 2f_m)],
 \end{aligned}$$

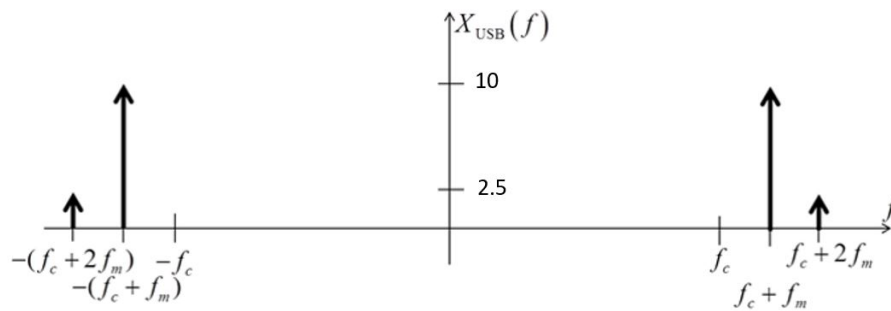
Assume $A_c = 10$.



Upper –sideband:

$$\begin{aligned}
 x_{USB}(t) &= \frac{1}{2} A_c m(t) \cos(2\pi f_c t) + \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t) \\
 &= 2 A_c \cos[2\pi(f_c + f_m)t] + \frac{1}{2} A_c \cos[2\pi(f_c + 2f_m)t], \\
 X_{LSB}(f) &= A_c \delta[f - (f_c + f_m)] + A_c \delta[f + (f_c + f_m)] \\
 &+ \frac{1}{4} A_c \delta[f - (f_c + 2f_m)] + \frac{1}{4} A_c \delta[f + (f_c + 2f_m)],
 \end{aligned}$$

Assume $A_c = 10$.



5. 每題 5 分，共 15 分

- (a) Since the carrier frequency is 1000 Hertz, the general form of $x_c(t)$ is

$$x_c(t) = A_c \cos[2\pi(1000)t + 40\sin(5t^2)] = A_c \cos[2\pi(1000)t + \phi(t)]$$

The phase deviation, $\phi(t) = 40\sin(5t^2)$ rad

The frequency deviation is

$$\begin{aligned} \frac{d\phi(t)}{dt} &= 400t \cos(5t^2) \text{ rad/sec} \\ \text{or } \frac{1}{2\pi} \frac{d\phi(t)}{dt} &= \frac{200}{\pi} t \cos(5t^2) \text{ Hz} \end{aligned}$$

- (b) Since the carrier frequency is 1000 Hertz, the general form of $x_c(t)$ is

$$x_c(t) = A_c \cos[2\pi(1000)t - 2\pi(400)t] = A_c \cos[2\pi(1000)t + \phi(t)]$$

The phase deviation is $\phi(t) = -2\pi(400)t$ rad

(Note that we subtracted the phase of the unmodulated carrier from the instantaneous carrier.) The frequency deviation is

$$\begin{aligned} \frac{d\phi}{dt} &= -2\pi(400) \text{ rad/sec} \\ \text{or } \frac{1}{2\pi} \frac{d\phi}{dt} &= -400 \text{ Hz} \end{aligned}$$

- (c) Since the carrier frequency is 1000 Hertz, the general form of $x_c(t)$ is

$$x_c(t) = A_c \cos[2\pi(1000)t - 2\pi(100)t + 10\sqrt{t}] = A_c \cos[2\pi(1000)t + \phi(t)]$$

The phase deviation is

$$\phi(t) = -2\pi(100)t + 10\sqrt{t} \text{ rad}$$

And the frequency deviation is

$$\frac{d\phi}{dt} = -2\pi(100) + \frac{1}{2}(10)t^{-\frac{1}{2}} = -2\pi(100) + \frac{5}{\sqrt{t}} \text{ rad/sec}$$

$$\text{or } \frac{1}{2\pi} \frac{d\phi}{dt} = -100 + \frac{5}{2\pi\sqrt{t}} \text{ Hz}$$

6. 每題 10 分，共 10 分

$$\begin{aligned} x_c(t) &= 25 \cos[2\pi(150)t] + 5 \cos[2\pi(150+10)t] + 5 \cos[2\pi(150-10)t] \\ &= 25 \cos[2\pi(150)t] + 10 \cos[2\pi(10)t] \cos[2\pi(150)t] \\ &= 25 \left\{ 1 + \frac{10}{25} \cos[(2\pi(10)t)] \right\} \cos[2\pi(150)t] \end{aligned}$$

$$a = \frac{10}{25} = 0.4$$

$$m_n(t) = \cos(2\pi \cdot 10t)$$

$$\begin{aligned} E &= \frac{a^2 \langle m_n^2(t) \rangle}{1 + a^2 \langle m_n^2(t) \rangle} \times 100\% \\ \therefore &= \frac{0.4^2 \langle \cos^2(2\pi \cdot 10t) \rangle}{1 + 0.4^2 \langle \cos^2(2\pi \cdot 10t) \rangle} \times 100\% \\ &= \frac{0.4^2 \cdot 0.5}{1 + 0.4^2 \cdot 0.5} \times 100\% \\ &= 0.0741 \end{aligned}$$

7. 每題 5 分，共 10 分

(a) $\phi_1(t) = \cos[2\pi f_c t + 10K_p \cos(5\pi t)]$

(b) $\phi_2(t) = \cos(2\pi f_c t + K_p[10\cos(5\pi t) + 2\sin(7\pi t)])$

$$\begin{aligned}\sin(\alpha)\sin(\beta) &= \frac{1}{2}[\cos(\alpha - \beta) - \cos(\alpha + \beta)] \\ \cos(\alpha)\cos(\beta) &= \frac{1}{2}[\cos(\alpha - \beta) + \cos(\alpha + \beta)] \\ \sin(\alpha)\cos(\beta) &= \frac{1}{2}[\sin(\alpha - \beta) + \sin(\alpha + \beta)] \\ \sin(\alpha \pm \beta) &= \sin(\alpha)\cos(\beta) \pm \cos(\alpha)\sin(\beta) \\ \cos(\alpha \pm \beta) &= \cos(\alpha)\cos(\beta) \mp \sin(\alpha)\sin(\beta)\end{aligned}$$