

Level 2: State-Space Averaging

From Basic 2-State to Advanced 4-Topology Averaging

Why Averaging? Understanding the Time-Scale Separation

Problem: Switches operate at high frequency ($f_{sw} = 50\text{ kHz}$), but we want to control low-frequency dynamics (AC line = 50/60 Hz, control bandwidth ~kHz).

Solution: State-space averaging "smooths out" rapid switching ripple, revealing the slower underlying dynamics that are relevant for control design.

Foundation: Standard SEPIC Averaging (2 States)

A standard SEPIC has switch ON (duration $D \cdot T_s$) and OFF (duration $(1-D) \cdot T_s$)

Classical averaging formula:

$$A_{avg} = D \cdot A_1 + (1-D) \cdot A_2$$

$$B_{avg} = D \cdot B_1 + (1-D) \cdot B_2$$

where D is the duty cycle ($0 < D < 1$)

Physical meaning: The averaged model represents the "time-weighted" behavior over one switching period, eliminating high-frequency ripple while preserving low-frequency dynamics.

Extension: Two-Phase Interleaved System (4 Topologies)

With two independent switches (S1, S2), we have four possible combinations:

11 (both ON), **10** (S1 ON, S2 OFF), **01** (S1 OFF, S2 ON), **00** (both OFF)

Understanding Topology Weights:

• k = topology index $\in \{11, 10, 01, 00\}$ (binary representation of switch states S1S2)

• w_k = "duration weight" = **fraction of switching period T_s** spent in topology k

Example: If $w_{11} = 0.3$, the system spends 30% of each cycle with both switches ON

• **Why needed?** Each topology has different energy transfer characteristics (storage vs. transfer). Knowing how long we spend in each mode lets us calculate the average behavior.

Constraint: $w_{11} + w_{10} + w_{01} + w_{00} = 1$

(System must be in exactly one of the four topologies at all times)

Timing Diagram Visualization:

The numbers below each diagram show **time markers** within one switching period T_s . They indicate when each topology begins/ends, measured from time 0 at the start of the period.

Non-overlapping ($d1=0.3, d2=0.4, \text{sum}=0.7 < 1$):

T_s : |----S1 ON----|----S2 ON-----|-----BOTH OFF-----|
0 $d1 \cdot T_s$ $(d1+d2) \cdot T_s$ T_s
|← $w_{10}=0.3$ →|← $w_{01}=0.4$ →|← $w_{00}=0.3$ →|

Time markers explained:

- 0: Start of period
- $d1 \cdot T_s$: S1 turns OFF at time = (duty cycle $d1$) × (period T_s)
- $(d1+d2) \cdot T_s$: S2 turns OFF
- T_s : End of period (next cycle begins)

Overlapping ($d1=0.6, d2=0.7, \text{sum}=1.3 \geq 1$):

T_s : |--S1-only--|---BOTH ON---|---S2-only--|
0= $(1-d2) \cdot T_s$ $(d1+d2-1) \cdot T_s$ $(1-d1) \cdot T_s$ T_s
|← $w_{10}=0.3$ →|← $w_{11}=0.3$ →|← $w_{01}=0.4$ →|

Time markers explained:

- 0: S1 turns ON first
- $(1-d2) \cdot T_s$: S2 turns OFF, leaving only S1 ON
- $(1-d1) \cdot T_s$: Both ON briefly before S1 turns OFF
- $(d1+d2-1) \cdot T_s$: S1 turns OFF, only S2 ON remains
- T_s : End of period

Duty Cycle Weight Derivation:

Key insight: Each weight w_k = (duration of topology k) / T_s

Non-overlapping case ($d1 + d2 < 1$): Switches never ON simultaneously, so $w_{11} = 0$.

- S1 is ON alone for duration = $d1 \cdot T_s \rightarrow w_{10} = d1$
- S2 is ON alone for duration = $d2 \cdot T_s \rightarrow w_{01} = d2$
- Both OFF for remaining time: $T_s - d1 \cdot T_s - d2 \cdot T_s = (1-d1-d2) \cdot T_s \rightarrow w_{00} = 1-d1-d2$

Check: $w_{10} + w_{01} + w_{00} = d1 + d2 + (1-d1-d2) = 1 \checkmark$

Overlapping case ($d1 + d2 \geq 1$): Switches have overlapping ON times. Assume S1 turns ON at $t=0$ and stays ON for $d1 \cdot T_s$. S2 turns ON at $t=0$ and stays ON for $d2 \cdot T_s$.

- S1-only period: From $t=d2 \cdot T_s$ (when S2 turns OFF) to $t=d1 \cdot T_s$ (when S1 turns OFF)

Duration = $d1 \cdot T_s - d2 \cdot T_s = (d1-d2) \cdot T_s$

But we need this in normalized form: Since $d2 > 1-d1$, rearranging gives: $w_{10} = 1 - d2$

- Both ON period: Overlap from $t=0$ to $\min(d1 \cdot T_s, d2 \cdot T_s)$

When $d1+d2 \geq 1$, overlap duration = $d1 \cdot T_s + d2 \cdot T_s - T_s = (d1+d2-1) \cdot T_s \rightarrow w_{11} = d1+d2-1$

- S2-only period: Symmetric to S1-only $w_{01} = 1 - d1$

- Both OFF: Since duty cycles overlap, no time remains with both OFF $\rightarrow w_{00} = 0$

Check: $w_{11} + w_{10} + w_{01} = (d1+d2-1) + (1-d2) + (1-d1) = 1 \checkmark$

Final Averaged Model

State Matrix (time-weighted average):

$$A_{avg}(d1, d2) = w_{11} \cdot A_{11} + w_{10} \cdot A_{10} + w_{01} \cdot A_{01} + w_{00} \cdot A_{00}$$

Input Matrix (with AC polarity handling):

$$B_{in,avg}(d1, d2, s) = s \cdot (w_{11} \cdot B_{11} + w_{10} \cdot B_{10} + w_{01} \cdot B_{01} + w_{00} \cdot B_{00})$$

Complete averaged system:

$$\dot{x} = A_{avg}(d1, d2) \cdot x + B_{in,avg}(d1, d2, s) \cdot V_{in} + f_{CPL}(vC0)$$

Understanding the 's' variable:

$s = \text{sign}(V_{in}) \in \{+1, -1\}$ handles AC input polarity

Why do we need it? This converter is a **PFC (Power Factor Correction) rectifier** operating directly from AC mains:

- Positive half-cycle ($V_{in} > 0$):** $s = +1 \rightarrow$ Current flows through L1, L2 (active), L3, L4 inactive

- Negative half-cycle ($V_{in} < 0$):** $s = -1 \rightarrow$ Current flows through L3, L4 (active), L1, L2 inactive

The **bridgeless topology** uses different inductor sets for each AC polarity, eliminating the need for a diode bridge rectifier (reducing losses by ~1-2%). The sign variable ensures the averaged model correctly represents energy flow regardless of AC polarity.

Key Insight: This 4-topology averaging generalizes the classical 2-state method. It captures the complex interleaved operation while maintaining the same fundamental principle: time-weighted averaging over a switching period smooths out high-frequency ripple (50 kHz) to reveal low-frequency control dynamics.