An Investigation into Low Power Synthesis Techniques.

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***Abstract*—This paper explores a number of low power techniques used in digital systems. Each technique is explained, giving the underlying theory of operation, and the synthesis techniques used to realise the technique. This report investigates power gating, power scaling, frequency domains, frequency scaling and clock gating,**

***Index Terms*—Low Power, Clock gating, power gating, frequency domains, frequency scaling, power scaling.**

I. Introduction

THE desire for low power devices has been driven by the mobile age. Companies are competing on battery life of portable devices, such as smartphones and tablets. This drive for low power has resulted in different synthesis techniques. This paper will review and explore some of the techniques used to help reduce the power consumption of a circuit, with particular interest on the synthesis techniques used to do so.

There are two main categories of power consumption - dynamic and leakage. Dynamic power is the power consumed when the circuit is enabled and functioning. Power is used here through the charging and discharging of the internal capacitors of the module. Leakage power is due to the non-ideal characteristics of sub-micron CMOS transistors. Leakage power mainly occurs when the module is in a idle, non-switching state [1] and can contribute around 22% of the total power consumption of a $90nm$ FPGA [2]

The report begins with a review of different techniques in turn. This includes a brief introduction to the theory and a discussion of synthesis problems and solutions. The report concludes by reviewing all techniques and their relevant advantages and disadvantages.

II. Power Techniques

A. Power Gating

Leakage currents begin to occur in sub-micron technologies, typically lower than 100nm [1], [3]. With each technology generation, gate leakage increases 30 fold [4]. Leakage currents can are identified as two sources - sub-threshold currents and gate tunnelling. Sub-threshold currents occur when the voltage applied to the gate is lower than the transistor threshold voltage. The channel of the transistor during weak inversion still allows some current to flow from source to drain. This current occurs when the transistor is off and is also exponentially dependant on V_{th} [5]. Gate tunnelling currents increase with thinner oxide layer [6]. A thin oxide results in electrons being able to pass through the gate into the substrate of the transistor. Gate tunnelling occurs when the transistor is active

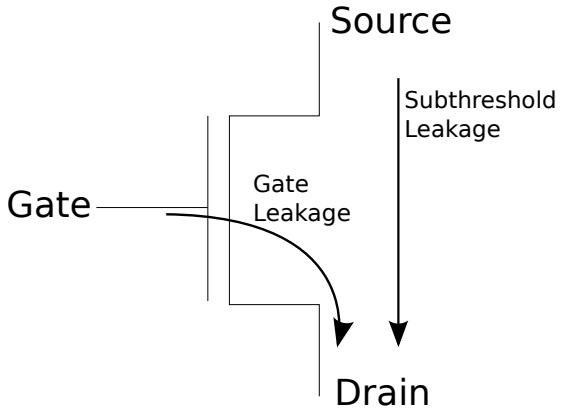

Fig. 1. The locations of gate and subthreshold leakage on an NMOS transistor

[3]. Figure 1 shows where these leakage currents occur on an NMOS transistor.

1) Theory: Power gating, in principle, is where the power to a module is switched off when not in use. By doing this, the module does not consume any power. It can be achieved by using either a header or a footer switch to disable either the supply connection or the ground connection. Figure 2 shows the realisation of the power gating circuits.

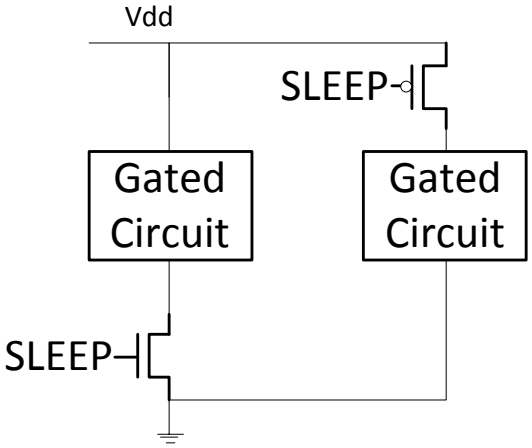

Fig. 2. Circuit diagrams showing the use of footer switches (left) or header switches (right).

Although the theory of operation is simple, the technique

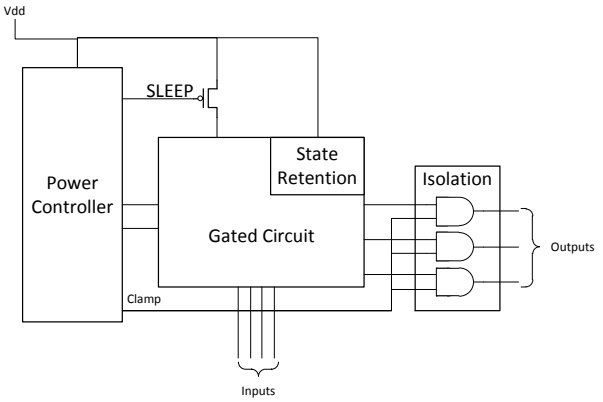

Fig. 3. The realisation of a power gated module. State retention, isolation and a power manager are all needed

poses many issues in the implementation. Firstly, as the module is floating, the outputs are undefined. This can cause the gates in another module to short circuit - at an input voltage of $V_{dd}/2$, both the NMOS and PMOS transistors will be on, resulting in a direct line from supply to ground. The solution to this is is to add logic gates with a 'clamp' signal. This could be either an `AND` gate, of which the clamp signal is active low, or an `OR` gate with an active high clamp. This added logic is needed per output signal of the power gated module.

The second issue that power gating raises is when the gated module contains sequential logic. By removing the power to the sequential logic, the state is lost. This can then cause issues as state retention may be necessary, as well as low power. The solution is to use a state retention register. There is a timing overhead involved to store the register before putting the device to sleep, disabling the majority of the circuit. State retention registers also require an ungated power supply, meaning all power gated modules require two supplies; one which is gated and one constant supply.

The final aspect of power gating that needs addressing is the power management. The power manager is responsible for when to gate the module.As well as the control of the sleep transistors, the power manager is also responsible for the stopping the clock, clamping the outputs and saving the state to the state retention mechanism.

There is a large overhead in the additional circuitry needed for power gating. The main increase are the state retention registers as these can require 20% more area in silicon [7]. Other techniques exist to avoid this, such as the use of the scan path to quickly clock the state in and out to/from memory. Digital signal processing, however, would not need this as it is data dependent. Processors with cache have a large amount of state data, so state retention registers are needed here.

When footer switches are used, the voltage that at the source of the NMOS is known as the virtual ground (VGND). When the footer is active, VGND $\approx$ GND. When off, VGND $\approx V_{dd}$. A full realisation of a power gated circuit is shown in figure 3.

TABLE I
POWER SAVING MODES OF [11]

State Name	NMOS State	PMOS State	Virtual Ground Voltage (V)
RUN	ON	OFF	GND
COLD	OFF	OFF	V_{DD}
PARK	OFF	ON	V_{th} PMOS

2) Synthesis Techniques: Power gating can be split into two techniques, fine and coarse grain. Fine grain gating is where individual cells (e.g. in ASIC) have their own gating transistor. Coarse grain techniques are where gating transistors are applied on a larger module scale. [3] compares the fine grain and coarse grain methods on a 65nm FPGA technology. Here, a SRAM block is used as the test circuit. Fine grain, in this case, is where the memory of each cell is individually gated, and coarse is where a 16×1 SRAM module is gated. It was conclusively shown that coarse grain gating reduced power consumption more than fine grain due to being able to turn off the read/write circuitry.

When deciding to sleep a module, there are a number of overheads. Due to the charging / discharging of the gated circuit (depending on the use of header or footer switches), when the circuit is re-enabled, there is an amount of time needed to allow the capacitors to return to normal [8]. This is referred to as the wake-up time and can also include the time to restore state if applicable. The (dis)charging can also poses ground bounce issues to the device due to the sudden draw of current, [9], [10].

This wake up time can then impact the performance of the device - if more energy is consumed waiting for the disabled module than it would have consumed being awake, the overall power is not reduced. [11] discusses a method where an intermediate power saving mode is introduced. This is done by using a standard footer power gating NMOS. An extra PMOS is then added in parallel with the power gating switch. Table I summarises the states of the transistors in the circuit in figure 4

The RUN and COLD states operate as previously discussed. The PARK state, however, is where the virtual ground voltage does not rise to V_{dd} due to the PMOS conducting. The virtual ground is at the threshold voltage of the PMOS, so the leakage current is less than in the RUN state, but not so high that state retention is needed. This drastically reduces the wake up overhead of the circuit.

A further improvement to this single extra state is proposed in [12]. Here, multiple intermediate power states are implemented by using different sub-threshold gate voltages on the NMOS footer switch. The compromise of wake up time against leakage reduction can then be more finely controlled to best suit the circuit. In general, the closer the virtual ground is to ground, the more leakage current occurs, but a faster wake up is available.

B. Power Scaling

1) Theory:

2) Synthesis Techniques:

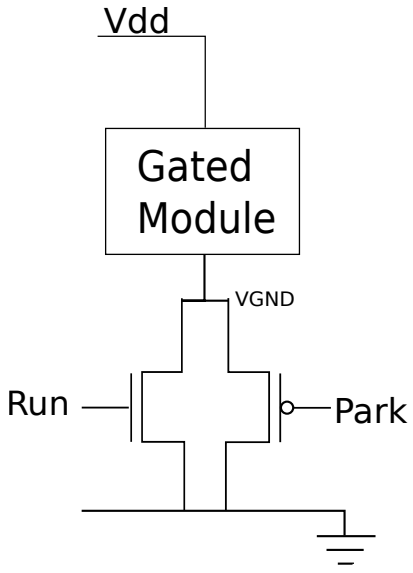

Fig. 4. Middle power state implementation used in [11]

III. Clock Techniques

A. Frequency Domains

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B. Frequency Scaling

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C. Clock Gating

1) Theory: The clock in a sequential circuit can consume 15-45% of the power [13]. Therefore it is a large area of potential power saving. Clock gating is an approach of controlling the clock to individual modules of a design by either stopping or slowing down the clock with respect to a master clock [14]. An approach, seen in [15], involves stopping the clock to unused modules.

This method can be realised using two simple circuits seen in figures 5 and 6. Although figure 5 is functionally complete, in reality, a latch is needed to remove any glitches in the circuit. These are fundamentally different to load-enable registers, where the input is multiplexed between the current value or the input. The load-enable registers are still clocked at the master clock frequency.

2) Synthesis Techniques: A gating function is typically defined by the designer within the RTL design stage. However, a more common approach is to allow the synthesis tool to obtain the gating functions from a gate-level netlist [16], [17].

The general outline for the synthesis is to find the clock gating function for each flip-flop. The flip-flops are then grouped so that they are driven by the same function. The problem of simplifying the gating function is looked at in [18]. Here, an algorithm is suggested where the gating function is shared by existing combinational logic. This was shown to reduce the logic added by introducing clock gating.

However, sometimes the addition of clock gating is not advantageous. The gating function can be large and therefore can cause timing violations, resulting in an unsuitable synthesis. If the gating function is large enough, it can also consume more power than it saves. Both of these issues are addressed in [19]. Here, the author proposed solutions to large gating functions by reducing the depth of logic by an approximation. The approximation is made such that the resulting logic does not gate the clock more than the original function. It results in the flip-flop being clocked more often, but reduces the logic so that it can be utilised, thereby saving some power. [19] also proposes the use of a clustering algorithm that looks at grouping similar gating functions. This can then reduce the overall logic needed to implement the clock gating and maximise the energy saved.

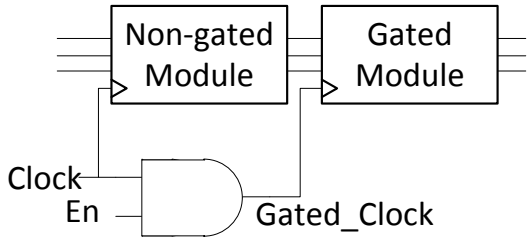

Fig. 5. Clock gating circuit using an AND gate

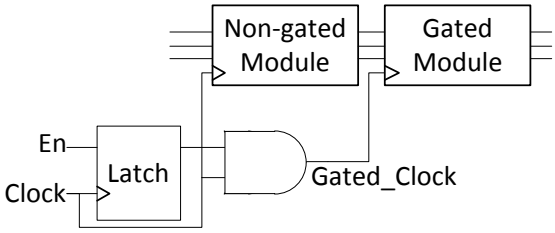

Fig. 6. Clock gating circuit using an AND gate and a latch

3) Conclusion: Clock gating is a simple principle to implement on a small scale. The underlying theory is to disable a module when it is not in use. This is done by identifying a gating function which disallows clock propagation if the function is asserted.

However, clock gating can produce large gating functions which violate the timing constraints of the circuit, or even consume more power than they save. This results in two problems that need solving - group formation and simplification [18], [20].

IV. Other Techniques

V. Conclusion

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