

MathEng2223

December 9, 2024

Mathematical Methods for Engineers (MathEng)

EXAM

December 2023

Duration: 2 hrs, all documents and calculators permitted

ATTEMPT ALL QUESTIONS - ANSWER IN ENGLISH

- 1 Using Euler's identity (or any other appropriate method), write down an expression for the complex Fourier series of the signal $x(t)$:**

$$x(t) = 3 \cos(5t) + 4 \sin(10t)$$

[5 marks]

To find the complex Fourier series of $x(t) = 3 \cos(5t) + 4 \sin(10t)$, we use Euler's identity: $\cos(\omega t) = \frac{e^{j\omega t} + e^{-j\omega t}}{2}$, $\sin(\omega t) = \frac{e^{j\omega t} - e^{-j\omega t}}{2j}$.

Step 1: Rewrite $\cos(5t)$ and $\sin(10t)$ using Euler's identity

- $3 \cos(5t) \rightarrow 3 \left(\frac{e^{j5t} + e^{-j5t}}{2} \right) = \frac{3}{2} e^{j5t} + \frac{3}{2} e^{-j5t}$
- $4 \sin(10t) \rightarrow 4 \left(\frac{e^{j10t} - e^{-j10t}}{2j} \right) = \frac{4}{2j} (e^{j10t} - e^{-j10t})$

$$\text{Recall: } \frac{1}{j} = \frac{j}{j^2} = \frac{j}{-1} = -j$$

$$= \frac{4}{2j} (e^{j10t} - e^{-j10t}) = \frac{4}{2j} e^{j10t} - \frac{4}{2j} e^{-j10t} = \frac{4}{2j} e^{j10t} - \frac{4}{2j} e^{-j10t} = \frac{4}{2j} e^{j10t} - \frac{4}{2j} e^{-j10t}$$

$$\text{Thus, } x(t) = \frac{3}{2} e^{j5t} + \frac{3}{2} e^{-j5t} - 2j e^{j10t} + 2j e^{-j10t}.$$

Step 2: Group the terms The complex Fourier series representation of $x(t)$ is: $x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$, where c_k are the complex Fourier coefficients.

Here, $x(t)$ has terms at frequencies ± 5 and ± 10 . The coefficients c_k are:

- At $k = 5$: $c_5 = \frac{3}{2}$,

- At $k = -5$: $c_{-5} = \frac{3}{2}$,
- At $k = 10$: $c_{10} = -2j$,
- At $k = -10$: $c_{-10} = 2j$,
- All other $c_k = 0$.

Final Answer: The complex Fourier series of $x(t)$ is:

$$x(t) = \frac{3}{2}e^{j5t} + \frac{3}{2}e^{-j5t} - 2je^{j10t} + 2je^{-j10t}$$

2 Develop an expression for the Fourier Transform of the signal $x(t)$ illustrated in Figure Q2 below:

[6 marks]

To develop the Fourier Transform $X(f)$ of the signal $x(t)$ illustrated in the figure, we follow the same steps for a rectangular pulse.

Step 1: Signal Description The signal $x(t)$ is defined as: $x(t) =$

$$\begin{cases} 5, & 0 \leq t \leq 0.2, \\ 0, & \text{otherwise.} \end{cases}$$

Step 2: Fourier Transform Definition The Fourier Transform is given by: $X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$.

Since $x(t)$ is nonzero only in the interval $[0, 0.2]$, the limits of integration reduce to $[0, 0.2]$: $X(f) = \int_0^{0.2} 5 e^{-j2\pi ft} dt$.

Step 3: Evaluate the Integral Factor out the constant 5: $X(f) = 5 \int_0^{0.2} e^{-j2\pi ft} dt$.

The integral of $e^{-j2\pi ft}$ is: $\int e^{-j2\pi ft} dt = e^{-j2\pi ft} \frac{1}{-j2\pi f}$.

Apply the limits of integration: $X(f) = 5 \left[\frac{e^{-j2\pi ft}}{-j2\pi f} \right]_0^{0.2}$.

Substitute the limits: $X(f) = 5 \frac{1}{-j2\pi f} (e^{-j2\pi f(0.2)} - e^{-j2\pi f(0)})$.

Simplify: $X(f) = 5 \frac{1}{-j2\pi f} (e^{-j0.4\pi f} - 1)$.

Step 4: Simplify Further

Using the property $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) e^{-j\frac{\theta}{2}}$ which is derived as follows:

1. **Rewrite $e^{-j\theta} - 1$: ***Expand using Euler's formula*: $e^{-j\theta} - 1 = \cos(\theta) - j\sin(\theta) - 1 = (\cos(\theta) - 1) - j\sin(\theta)$ **Factorize Trigonometric Terms:** Use the half-angle identities:

2. • $\cos(\theta) = 1 - 2 \sin^2\left(\frac{\theta}{2}\right) \implies \cos(\theta) - 1 = -2 \sin^2\left(\frac{\theta}{2}\right)$ $\sin(\theta) = 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)$.

Substituting these: $e^{-j\theta} - 1 = \{-2 \sin^2\left(\frac{\theta}{2}\right)\} - j\{2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)\}$. **Factor Out Common Terms:**

3. Identify Common Factor:

Both terms contain $-2j \sin\left(\frac{\theta}{2}\right)$ as a common factor : $1 - 2 \sin^2\left(\frac{\theta}{2}\right) = -2j \sin\left(\frac{\theta}{2}\right) \cdot \frac{\sin\left(\frac{\theta}{2}\right)}{j}$ $-j \cdot 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right) = -2j \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)$. **Factorization:**

Factor $-2j \sin\left(\frac{\theta}{2}\right)$ out of the entire expression : $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) \cdot \left(\frac{\sin\left(\frac{\theta}{2}\right)}{j} + \cos\left(\frac{\theta}{2}\right)\right)$.

Simplify the term : $\sin\left(\frac{\theta}{2}\right) \frac{1}{j - j \sin\left(\frac{\theta}{2}\right)}$. *Thus: $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) \cdot (\cos\left(\frac{\theta}{2}\right) - j \sin\left(\frac{\theta}{2}\right))$.*

- **Recognize the Exponential Form:** The term $\cos\left(\frac{\theta}{2}\right) - j \sin\left(\frac{\theta}{2}\right)$ is equivalent to $e^{-j\frac{\theta}{2}}$, using Euler's formula.

5. **Simplify:** Recognize the term in parentheses as $e^{-j\frac{\theta}{2}}$: $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) e^{-j\frac{\theta}{2}}$.

This compactly combines the amplitude term $-2j \sin\left(\frac{\theta}{2}\right)$ and the phase shift $e^{-j\frac{\theta}{2}}$.

****rewrite $X(f)$ **** : $X(f) = 5 \frac{1 - e^{-j2\pi f \cdot 2j \sin(0.2\pi f)} e^{-j0.2\pi f}}{-j2\pi f \cdot -2j \sin(0.2\pi f)}$

Cancel $-j$ and simplify: $X(f) = 5 \cdot 2 \sin(0.2\pi f) \frac{e^{-j0.2\pi f}}{2\pi f e^{-j0.2\pi f}}$

Finally: $X(f) = 5 \sin(0.2\pi f) \frac{1}{\pi f}$

Final Expression \square where the **sinc function** is defined as: $\text{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$

Interpretation

- $\frac{\sin(0.2\pi f)}{\pi f}$: This is the sinc function, representing the frequency-domain shape of the rectangular pulse.
- $e^{-j0.2\pi f}$: This is a phase shift due to the non-centered nature of the pulse (starting at $t = 0$).

[1]: `using FFTW, LinearAlgebra, Plots, LaTeXStrings`

[2]: `include("modules/operations.jl");`

```
[3]: # Define the unscaled sinc function
sinc_unscaled(x::Real) = x == 0 ? 1.0 : sin(pi * x) / (pi * x)

# Define the polymorphic sinc function with a normalization option
sinc(x::Real; normalized::Bool = true) = normalized ? sinc_unscaled(x / ) : sinc_unscaled(x)

# Frequency range
f = range(-40, 40, length=1000)

# Function components
```

```

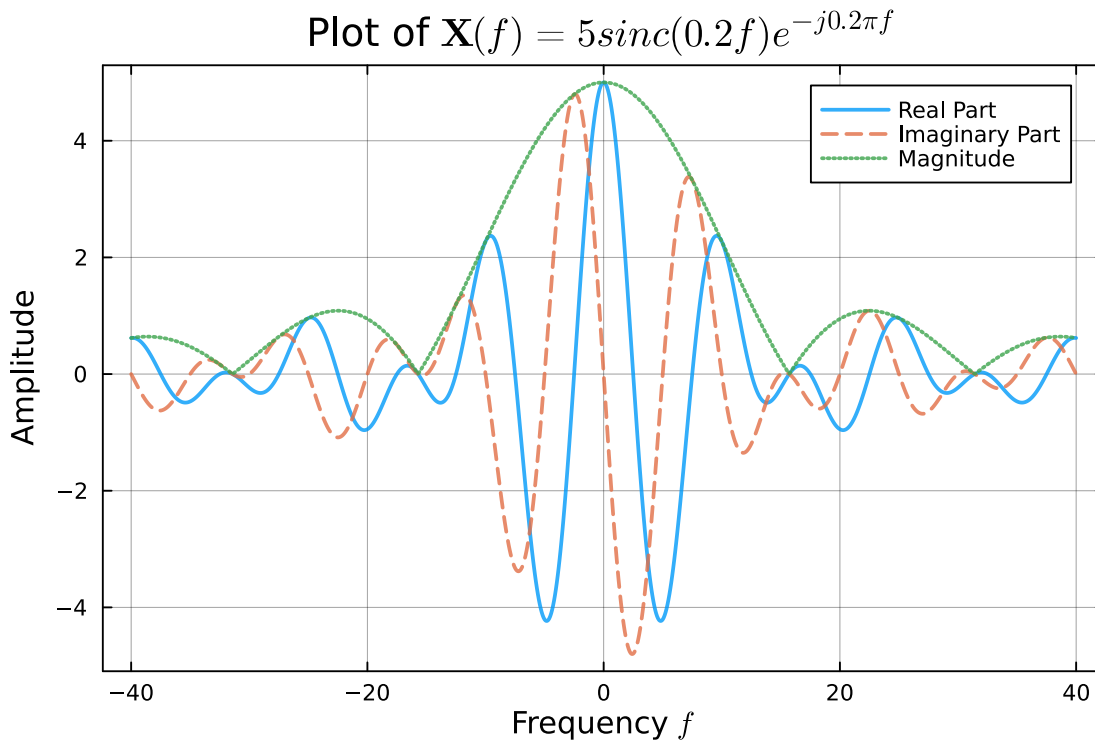
= 5 .* sinc(0.2 .* f) # Amplitude of the signal
= .^ (-j .* 0.2 .* f) # Phase shift
= .* # Combined function

# Plot with title, labels, and semi-transparent grid
plot(f, real.( ),
     , label="Real Part", linestyle=:solid, linewidth=2, alpha=0.8, size =_
↪(600,400)
     , xlabel="Frequency " * L"f", ylabel="Amplitude"
     , title="Plot of " * L"\mathbf{X}(f) = 5 sinc(0.2 f) e^{-j 0.2 f}"
     , grid=true, gridalpha=0.2 # Enable grid and set transparency
     , framestyle=:box
)

# Overlay additional lines
plot!(f, imag.( ), label="Imaginary Part", linestyle=:dash, linewidth=2,_
↪alpha=0.8)
plot!(f, abs.( ), label="Magnitude", linestyle=:dot, linewidth=2, alpha=0.8)

```

[3]:



3 A linear, time-invariant system has the following transfer function:

$$H(s) = 10(s + 100) \frac{1}{s^2 + 2s + 100}$$

- (a) Derive an expression for $H(s)$ in the usual, normal form.
- (b) Determine the frequency-invariant gain K and the position of any poles and zeros.
- (c) Sketch a Bode plot of the magnitude-frequency response.
- (d) Sketch a Bode plot of the phase-frequency response.

[8 marks]

(a) Derive an expression for $H(s)$ in the usual, normal form. To derive the transfer function $H(s)$ in the usual, **normal form**, we factorize the numerator and denominator in terms of their natural frequencies and damping ratios.

The given transfer function is: $H(s) = 10(s + 100) \frac{1}{s^2 + 2s + 100}$.

Step 1: Denominator Normal Form The denominator is: $s^2 + 2s + 100$.

This matches the general form of a second-order system: $s^2 + 2\zeta\omega_n s + \omega_n^2$, where ζ is the damping ratio and ω_n is the natural frequency.

Here: $\omega_n^2 = 100 \Rightarrow \omega_n = \sqrt{100} = 10$, and $2\zeta\omega_n = 2 \Rightarrow \zeta = \frac{2}{2\omega_n} = \frac{2}{20} = 0.1$.

Thus, the denominator becomes: $s^2 + 2s + 100 = (s^2 + 2\zeta\omega_n s + \omega_n^2) = s^2 + 2(0.1)(10)s + 10^2$.

Step 2: Numerator Normal Form The numerator is: $10(s + 100)$.

Factor out 100 to normalize: $10(s + 100) = 10 \cdot 100 \left(\frac{s}{100} + 1\right) = 1000 \left(\frac{s}{100} + 1\right)$.

Step 3: Rewrite in Normal Form Substitute the factored numerator and denominator into $H(s)$: $H(s) = 1000 \left(\frac{s}{100} + 1\right) \frac{1}{s^2 + 2(0.1)(10)s + 10^2}$.

Simplify: $H(s) = 1000 \frac{\left(\frac{s}{100} + 1\right)}{100 \cdot \frac{s^2 + 2(0.1)(10)s + 10^2}{100}}$.

After normalization: $H(s) = 10 \left(\frac{s}{100} + 1\right) \frac{1}{\frac{s^2}{100} + \frac{2s}{10} + 1}$.

Alternatively: $H(s) = 10 \left(\frac{s}{100} + 1\right) \frac{1}{\frac{s^2}{100} + \frac{s}{5} + 1}$.

This is the normalized form of $H(s)$.

(b): Determine the Frequency-Invariant Gain K and the Positions of Poles and Zeros

1. Transfer Function The given transfer function is: $H(s) = 10(s + 100) \frac{1}{s^2 + 2s + 100}$.

2. Frequency-Invariant Gain \$ K \$ The frequency-invariant gain is the gain of the system as \$ s \rightarrow 0 \$. This is determined by evaluating the transfer function at \$ s = 0 \$: \$ K = H(0) = \frac{10(0 + 100)}{(0)^2 + 2(0) + 100} \$

Simplify: \$ K = \frac{10 \cdot 100}{100 = 10} \$

Thus, the frequency-invariant gain is: \$ K = 10 \$.

3. Poles The poles are the roots of the denominator \$ s^2 + 2s + 100 = 0 \$: \$ s^2 + 2s + 100 = 0 \$.

Solve using the quadratic formula: \$ s = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a} \$, where \$ a=1, b=2, \$ and \$ c=100 \$. Substituting: \$ s = \frac{-2 \pm \sqrt{2^2 - 4(1)(100)}}{2(1) = \frac{-2 \pm \sqrt{4 - 400}}{2}} \$.

Simplify: \$ s = \frac{-2 \pm \sqrt{-396}}{2} \$.

The roots are: \$ s = -1 \pm j\sqrt{99} \$.

Thus, the poles are: \$ s = -1 + j\sqrt{99}, \quad s = -1 - j\sqrt{99} \$.

4. Zeros The zero is the root of the numerator \$ 10(s + 100) = 0 \$: \$ s + 100 = 0 \quad s = -100 \$.

Thus, there is one zero at: \$ s = -100 \$.

Final Results:

- **Frequency-Invariant Gain \$ K \$:** \$ K = 10 \$.
- **Poles:** \$ s = -1 + j\sqrt{99}, \quad s = -1 - j\sqrt{99} \$. **Zero:** \$ s = -100 \$.

```
[4]: using FFTW, LinearAlgebra
      include("modules/operations.jl");
```

```
[5]: # Frequency range (logarithmic scale)
      = 10 .^ range(-1, 3, length=500) # Frequencies from 0.1 to 1000 (log scale)

      # Define the transfer function H(s)
      function H()
          numerator = 10 .* (j .* . + 100) # Element-wise addition
          denominator = (j .* ).^2 .+ 2 .* (j .* ) .+ 100 # Element-wise operations
          return numerator ./ denominator # Element-wise division
      end
```

```
[5]: H (generic function with 1 method)
```

```
[6]: using Plots
      using Printf
      using Measures

      # Magnitude response in dB
```

```

magnitude_dB = 20 .* log10.(abs.(H.( ))) # Broadcasting applied to H, abs, and
↳ log10

# Plot the Bode magnitude plot
p1 = plot( , magnitude_dB
    , xscale=:log10
    , xlabel="Frequency (rad/s)", ylabel="Magnitude (dB)"
    , title="Bode Magnitude Plot", legend=false, grid=true
    , margin = 5mm
)

# Phase response in degrees
phase_deg = angle.(H.( )) .* (180 / ) # Convert phase from radians to degrees

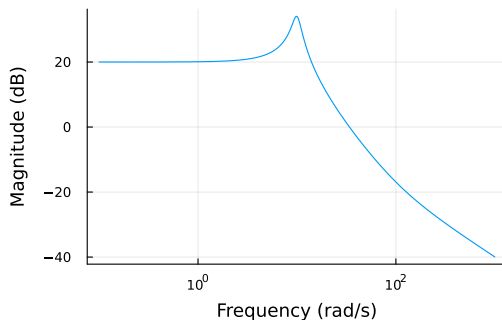
# Plot the Bode phase plot
p2 = plot( , phase_deg
    , xscale=:log10
    , xlabel="Frequency (rad/s)", ylabel="Phase (degrees)"
    , title="Bode Phase Plot"
    , legend=false, grid=true
    , left_margin=10mm, right_margin=10mm, top_margin=15mm, bottom_margin=15mm
)

plot(p1, p2, layout = (1, 2), size = (1000, 400))

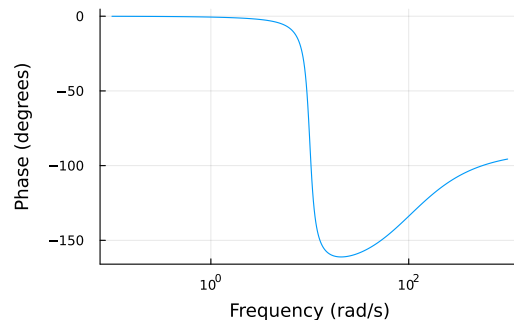
```

[6]:

Bode Magnitude Plot



Bode Phase Plot



3.1 4. Sketch magnitude and phase responses for a sampled data system with a pair of complex conjugate zeros and two poles at the origin.

[4 marks]

[]:

- 4 A random variable X is uniformly distributed between $x = 0$ and $x = 1$. Via any appropriate method, determine the expected value $E[Y]$ of $Y = \exp(X)$.

[4 marks]

Given $Y = \exp(X)$ and $X \sim U(0, 1)$,

1. Expected Value Formula The expected value of a random variable Y is given by: $E[Y] = \int_{-\infty}^{\infty} y f_Y(y) dy$.

Since X is uniformly distributed, its probability density function (PDF) is: $f_X(x) =$

$$\begin{cases} 1, & 0 \leq x \leq 1, \\ 0, & \text{otherwise.} \end{cases}$$

For $Y = \exp(X)$, the expected value becomes: $E[Y] = \int_0^1 \exp(x) f_X(x) dx$.

Because $f_X(x) = 1$ for $0 \leq x \leq 1$, this simplifies to: $E[Y] = \int_0^1 \exp(x) dx$.

2. Solve the Integral The integral of $\exp(x)$ is: $\int \exp(x) dx = \exp(x) + C$.

Now, evaluate the definite integral: $\int_0^1 \exp(x) dx = [\exp(x)]_0^1 = \exp(1) - \exp(0)$.

Simplify: $\int_0^1 \exp(x) dx = e - 1$.

3. Final Answer The expected value is: $E[Y] = e - 1$.

- 5 Identify the pivots and free variables of the following two matrices A and B . Following the method which we studied in class, find the special solution corresponding to each free variable and, by combining the special solutions, describe every solution to $Ax = 0$ and $Bx = 0$.

$$A = \begin{bmatrix} 1 & 2 & 2 & 4 & 6 \\ 1 & 2 & 3 & 6 & 9 \\ 0 & 0 & 1 & 2 & 3 \end{bmatrix} \quad \text{and} \quad B = \begin{bmatrix} 2 & 4 & 2 \\ 0 & 4 & 4 \\ 0 & 8 & 8 \end{bmatrix}.$$

[7 marks]

[]:

6 For a projection matrix $P = A(A^T A)^{-1} A^T$, show that $P^2 = P$ and then explain, in terms of the column space of P , why projections P_b and $P(P_b)$ give identical results.

[5 marks]

1. Show that $P^2 = P$ The projection matrix P is defined as: $P = A (A^T A)^{-1} A^T$, where A is a matrix with linearly independent columns.

****Compute P^2 : **** *We want to show : $P^2 = P$.*

Start with $P^2 : P^2 = P \cdot P = (A(A^T A)^{-1} A^T) \cdot (A(A^T A)^{-1} A^T)$.

Expand the multiplication: $P^2 = A (A^T A)^{-1} A^T A (A^T A)^{-1} A^T$.

Since $A^T A$ is invertible, $A^T A (A^T A)^{-1} = I$ (identity matrix). So: $P^2 = A (A^T A)^{-1} (I) A^T = A (A^T A)^{-1} A^T$.

This simplifies to: $P^2 = P$.

2. Projections P_b and $P(P_b)$ Give Identical Results

Interpretation of P : The projection matrix P projects any vector b onto the **column space** of A , denoted as $\text{Col}(A)$.

****Explain P_b : **** $P_b = P b = A (A^T A)^{-1} A^T b$. This gives the projection of b onto $\text{Col}(A)$.

****Explain $P(P_b)$: **** $P(P_b) = P(P b)$. Substitute P_b into $P(P_b) : P(P_b) = P \cdot P b$. Since we showed that $P^2 = P$, this becomes: $P(P_b) = P b$.

Why Are P_b and $P(P_b)$ Identical?

- $P_b = P b$ is already the projection of b onto $\text{Col}(A)$.
- Applying P again to P_b does not change it, because projecting a vector already in the subspace $\text{Col}(A)$ onto the same subspace leaves it unchanged.
- Hence: $P(P_b) = P_b$.

3. Column Space Perspective In terms of the column space of P : 1. The column space of P (and thus P_b) is the **same as $\text{Col}(A)$** . 2. Applying P to P_b projects P_b onto $\text{Col}(A)$, but since $P_b \in \text{Col}(A)$, the result is unchanged.

Thus, projections P_b and $P(P_b)$ are identical because projecting a vector already in the column space does nothing.

Conclusion

- **Projection matrix property:** $P^2 = P$.
- **Projections:** Pb and $P(Pb)$ are identical because Pb lies in the column space, and re-projecting it does not alter it.
- **Idempotence:** P an idempotent matrix, which is a key characteristic of projection matrices.

[]: