

## Mathematical Methods for Engineers (MathEng)

### EXAM

December 2023

Duration: 2 hrs, all documents and calculators permitted  
ATTEMPT ALL QUESTIONS – ANSWER IN ENGLISH

1. Using Euler's identity (or any other appropriate method), write down an expression for the complex Fourier series of the signal  $x(t)$ :

$$x(t) = 3 \cos(5t) + 4 \sin(10t)$$

[5 marks]

To find the complex Fourier series of  $x(t) = 3 \cos(5t) + 4 \sin(10t)$ ,

we use Euler's identity:  $\cos(\omega t) = \frac{e^{j\omega t} + e^{-j\omega t}}{2}$ ,  $\sin(\omega t) = \frac{e^{j\omega t} - e^{-j\omega t}}{2j}$ .

Step 1: Rewrite  $\cos(5t)$  and  $\sin(10t)$  using Euler's identity

$$1. \ 3 \cos(5t) \rightarrow 3 \left( \frac{e^{j5t} + e^{-j5t}}{2} \right) = \frac{3}{2} e^{j5t} + \frac{3}{2} e^{-j5t}$$

$$2. \ 4 \sin(10t) \rightarrow 4 \left( \frac{e^{j10t} - e^{-j10t}}{2j} \right) = \frac{4}{2j} (e^{j10t} - e^{-j10t})$$

$$\text{Recall: } \frac{1}{j} = \frac{j}{j^2} = \frac{j}{-1} = -j$$

$$\begin{aligned} \frac{4}{2j} (e^{j10t} - e^{-j10t}) &= \frac{2}{j} (e^{j10t} - e^{-j10t}) = \frac{2}{j} e^{j10t} - \frac{2}{j} e^{-j10t} \\ &= \frac{2j}{j^2} e^{j10t} - \frac{2j}{j^2} e^{-j10t} = \frac{2j}{-1} e^{j10t} - \frac{2j}{-1} e^{-j10t} = -2je^{j10t} + 2je^{-j10t} \end{aligned}$$

$$\text{Thus, } x(t) = \frac{3}{2} e^{j5t} + \frac{3}{2} e^{-j5t} - 2je^{j10t} + 2je^{-j10t}.$$

Step 2: Group the terms

The complex Fourier series representation of  $x(t)$  is:  $x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$ , where  $c_k$  are the complex Fourier coefficients.

Here,  $x(t)$  has terms at frequencies  $\pm 5$  and  $\pm 10$ . The coefficients  $c_k$  are:

- At  $k = 5$ :  $c_5 = \frac{3}{2}$ .
- At  $k = -5$ :  $c_{-5} = \frac{3}{2}$ .
- At  $k = 10$ :  $c_{10} = -2j$ .
- At  $k = -10$ :  $c_{-10} = 2j$ .
- All other  $c_k = 0$ .

Final Answer:

$$\text{The complex Fourier series of } x(t) \text{ is: } \boxed{x(t) = \frac{3}{2} e^{j5t} + \frac{3}{2} e^{-j5t} - 2je^{j10t} + 2je^{-j10t}}$$

2. Develop an expression for the Fourier Transform of the signal  $x(t)$  illustrated in Figure Q2 below:

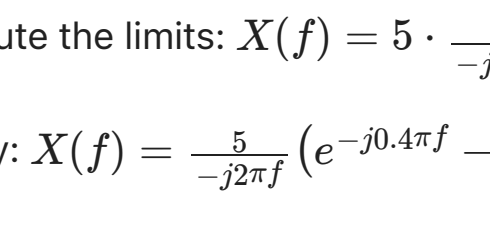


Figure Q2

[6 marks]

To develop the Fourier Transform  $X(f)$  of the signal  $x(t)$  illustrated in the figure, we follow the same steps for a rectangular pulse.

Step 1: Signal Description

The signal  $x(t)$  is defined as:  $x(t) = \begin{cases} 5, & 0 \leq t \leq 0.2 \\ 0, & \text{otherwise.} \end{cases}$

Step 2: Fourier Transform Definition

The Fourier Transform is given by:  $X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$ .

Since  $x(t)$  is nonzero only in the interval  $[0, 0.2]$ , the limits of integration reduce to  $[0, 0.2]$ :  $X(f) = \int_0^{0.2} 5e^{-j2\pi f t} dt$ .

Step 3: Evaluate the Integral

Factor out the constant 5:  $X(f) = 5 \int_0^{0.2} e^{-j2\pi f t} dt$ .

The integral of  $e^{-j2\pi f t}$  is:  $\int e^{-j2\pi f t} dt = \frac{e^{-j2\pi f t}}{-j2\pi f}$ .

Apply the limits of integration:  $X(f) = 5 \left[ \frac{e^{-j2\pi f t}}{-j2\pi f} \right]_0^{0.2}$ .

Substitute the limits:  $X(f) = 5 \cdot \frac{1}{-j2\pi f} (e^{-j2\pi f (0.2)} - e^0)$ .

Simplify:  $X(f) = \frac{-5}{j2\pi f} (e^{-j0.4\pi f} - 1)$ .

Step 4: Simplify Further

Using the property  $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) e^{-j\frac{\theta}{2}}$  which is derived as follows:

1. Rewrite  $e^{-j\theta} - 1 = 1$ : Expand using Euler's formula:  $e^{-j\theta} - 1 = \cos(\theta) - j\sin(\theta) - 1 = (\cos(\theta) - 1) - j\sin(\theta)$

2. Factorize Trigonometric Terms: Use the half-angle identities:

- $\cos(\theta) - 1 = -2 \sin^2\left(\frac{\theta}{2}\right) \implies \cos(\theta) - 1 = -2 \sin^2\left(\frac{\theta}{2}\right)$
- $\sin(\theta) = 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)$ .

Substituting these:  $e^{-j\theta} - 1 = -2 \sin^2\left(\frac{\theta}{2}\right) - j 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)$ .

3. Factor Out Common Terms:

- Identify Common Factor:

Both terms contain  $-2j \sin\left(\frac{\theta}{2}\right)$  as a common factor:  $-2 \sin^2\left(\frac{\theta}{2}\right) - j 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right) = -2j \sin\left(\frac{\theta}{2}\right) \cdot \frac{\sin(\frac{\theta}{2})}{j}$ .

$-j 2 \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right) = -2j \sin\left(\frac{\theta}{2}\right) \cos\left(\frac{\theta}{2}\right)$ . This is already proportional to  $-2j \sin\left(\frac{\theta}{2}\right)$ .

Factorization:

Factor  $-2j \sin\left(\frac{\theta}{2}\right)$  out of the entire expression:  $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) \cdot \left( \frac{\sin(\frac{\theta}{2})}{j} + \cos\left(\frac{\theta}{2}\right) \right)$ .

Simplify the term:  $\frac{\sin(\frac{\theta}{2})}{j} = -j \sin\left(\frac{\theta}{2}\right)$ . Thus:  $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) \cdot (\cos\left(\frac{\theta}{2}\right) - j \sin\left(\frac{\theta}{2}\right))$ .

Recognize the Exponential Form: The term  $\cos\left(\frac{\theta}{2}\right) - j \sin\left(\frac{\theta}{2}\right)$  is equivalent to  $e^{-j\frac{\theta}{2}}$ , using Euler's formula.

5. Simplify: Recognize the term in parentheses as  $e^{-j\frac{\theta}{2}}$ :  $e^{-j\theta} - 1 = -2j \sin\left(\frac{\theta}{2}\right) e^{-j\frac{\theta}{2}}$ .

This compactly combines the amplitude term  $-2j \sin\left(\frac{\theta}{2}\right)$  and the phase shift  $e^{-j\frac{\theta}{2}}$ .

rewrite  $X(f)$ :

$$X(f) = \frac{-5}{j2\pi f} \cdot -2j \sin(0.2\pi f) e^{-j0.2\pi f}.$$

Cancel  $-j$  and simplify:  $X(f) = \frac{5 \cdot 2 \sin(0.2\pi f)}{2\pi f} e^{-j0.2\pi f}$ .

$$\text{Finally: } X(f) = \frac{5 \sin(0.2\pi f)}{\pi f} e^{-j0.2\pi f}.$$

Final Expression

$$\boxed{X(f) = \frac{5 \sin(0.2\pi f)}{\pi f} \cdot e^{-j0.2\pi f}} \text{ where the sinc function is defined as: } \text{sinc}(x) = \frac{\sin(\pi x)}{\pi x}.$$

Interpretation

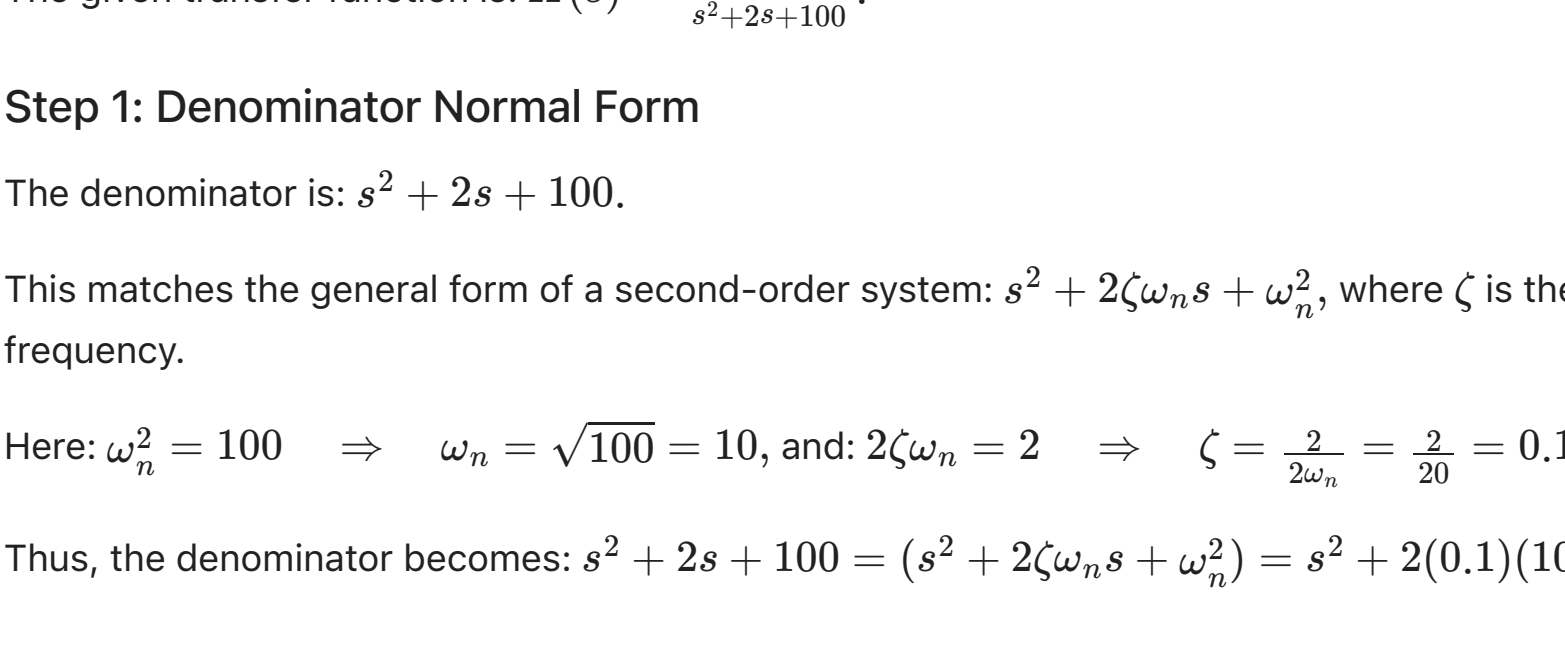
- $\frac{\sin(0.2\pi f)}{\pi f}$ : This is the sinc function, representing the frequency-domain shape of the rectangular pulse.
- $e^{-j0.2\pi f}$ : This is a phase shift due to the non-centered nature of the pulse (starting at  $t = 0$ ).

In [1]: using FFTW, LinearAlgebra, Plots, LaTeXStrings

In [2]: include("modules/operations.jl");

In [3]:  
# Define the unscaled sinc function  
sinc\_unscaled(x::Real) = x == 0 ? 1.0 : sin(pi \* x) / (pi \* x)  
  
# Define the polymorphic sinc function with a normalization option  
sinc(x::Real; normalized::Bool = true) = normalized ? sinc\_unscaled(x / pi) : sinc\_unscaled(x)  
  
# Frequency range  
f = range(-40, 40, length=1000)  
  
# Function components  
A = 5 .\* sinc.(0.2 .\* f) # Amplitude of the signal  
phi = pi .\* (-f) .\* 0.2 # Phase shift  
X = A .\* phi # Combined function  
  
# Plot with title, labels, and semi-transparent grid  
plot(f, real(X))  
label="Real Part", linestyle=:solid, linewidth=2, alpha=0.8, size = (600, 400)  
xlabel="Frequency" \* "\u03c0 f", ylabel="Amplitude"  
title="Plot of " \* LaTeXStrings{X}(f) \* " = 5 sinc(0.2 f) e^{-j 0.2 \pi f}"  
grid=true, gridalpha=0.2 # Enable grid and set transparency  
framestyle=:box  
  
# Overlay additional lines  
plot!(f, imag(X), label="Imaginary Part", linestyle=:dash, linewidth=2, alpha=0.8)  
plot!(f, abs.(X), label="Magnitude", linestyle=:dot, linewidth=2, alpha=0.8)

Out [3]:



3. A linear, time-invariant system has the following transfer function:

$$H(s) = \frac{10(s+100)}{s^2+2s+100}$$

(a) Derive an expression for  $H(s)$  in the usual, normal form.

(b) Determine the frequency-invariant gain  $K$  and the position of any poles and zeros.

(c) Sketch a Bode plot of the magnitude-frequency response.

(d) Sketch a Bode plot of the phase-frequency response.

[8 marks]

(a) Derive an expression for  $H(s)$  in the usual, normal form.

To derive the transfer function  $H(s)$  in the usual, **normal form**, we factorize the numerator and denominator in terms of their natural frequencies and damping ratios.

$$\text{The given transfer function is: } H(s) = \frac{10(s+100)}{s^2+2s+100}$$

Step 1: Denominator Normal Form

The denominator is:  $s^2 + 2s + 100$ .

This matches the general form of a second-order system:  $s^2 + 2\zeta\omega_n s + \omega_n^2$ , where  $\zeta$  is the damping ratio and  $\omega_n$  is the natural frequency.

$$\text{Here: } \omega_n^2 = 100 \implies \omega_n = \sqrt{100} = 10, \text{ and: } 2\zeta\omega_n = 2 \implies \zeta = \frac{2}{2\omega_n} = \frac{2}{20} = 0.1.$$

$$\text{Thus, the denominator becomes: } s^2 + 2s + 100 = (s^2 + 2\zeta\omega_n s + \omega_n^2) = s^2 + 2(0.1)(10)s + 10^2.$$

Step 2: Numerator Normal Form

The numerator is:  $10(s + 100)$ .

$$\text{Factor out 100 to normalize: } 10(s + 100) = 10 \cdot 100 \left( \frac{s}{100} + 1 \right) = 1000 \left( \frac{s}{100} + 1 \right).$$

Step 3: Rewrite in Normal Form

$$\text{Substitute the factored numerator and denominator into } H(s): H(s) = \frac{1000 \left( \frac{s}{100} + 1 \right)}{s^2 + 2(0.1)(10)s + 10^2}.$$

$$\text{Simplify: } H(s) = \frac{1000}{100} \cdot \frac{\left( \frac{s}{100} + 1 \right)}{\frac{s^2}{100} + \frac{2s}{100} + \frac{100}{100}}.$$

$$\text{After normalization: } H(s) = \frac{10 \left( \frac{s}{100} + 1 \right)}{\frac{s^2}{100} + \frac{2s}{100} + 1}.$$

$$\text{Alternatively: } H(s) = \frac{10 \left( \frac{s}{100} + 1 \right)}{\frac{s^2}{100} + \frac{2s}{100} + 1}.$$

This is the normalized form of  $H(s)$ .

(b): Determine the Frequency-Invariant Gain  $K$  and the Positions of Poles and Zeros

1. Transfer Function

$$\text{The given transfer function is: } H(s) = \frac{10(s+100)}{s^2+2s+100}.$$

2. Frequency-Invariant Gain  $K$

The frequency-invariant gain is the gain of the system as  $s \rightarrow 0$ . This is determined by evaluating the transfer function at  $s = 0$ :

$$K = H(0) = \frac{1000 - 100}{(0)^2 + 2(0) + 100}.$$

$$\text{Simplify: } K = \frac{900}{100} = 10.$$

Thus, the frequency-invariant gain is:  $K = 10$ .

3. Poles

The poles are the roots of the denominator  $s^2 + 2s + 100 = 0$ :  $s^2 + 2s + 100 = 0$ .

Solve using the quadratic formula:  $s = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$ , where  $a = 1$ ,  $b = 2$ , and  $c = 100$ . Substituting:

$$s = \frac{-2 \pm \sqrt{2^2 - 4(1)(100)}}{2} = \frac{-2 \pm \sqrt{4 - 400}}{2}.$$

$$\text{Simplify: } s = \frac{-2 \pm \sqrt{-396}}{2}.$$

The roots are:  $s = -1 \pm j\sqrt{99}$ .

$$\text{Thus, the poles are: } s = -1 + j\sqrt{99}, \quad s = -1 - j\sqrt{99}.$$

4. Zeros

The zero is the root of the numerator  $10(s + 100) = 0$ :  $s + 100 = 0 \implies s = -100$ .

Thus, there is one zero at:  $s = -100$ .

Final Results:

- Frequency-Invariant Gain  $K$ :  $K = 10$ .

- Poles:  $s = -1 + j\sqrt{99}$ ,  $s = -1 - j\sqrt{99}$ .

- Zero:  $s = -100$ .

In [4]: using FFTW, LinearAlgebra

In [5]: include("modules/operations.jl");

Out [5]: H (generic function with 1 method)

In [6]:

```
using Plots
using Printf
using Measures

# Magnitude response in dB
magnitude_db = 20 .* log10.(abs.(H.(w))) # Broadcasting applied to H, abs, and log10

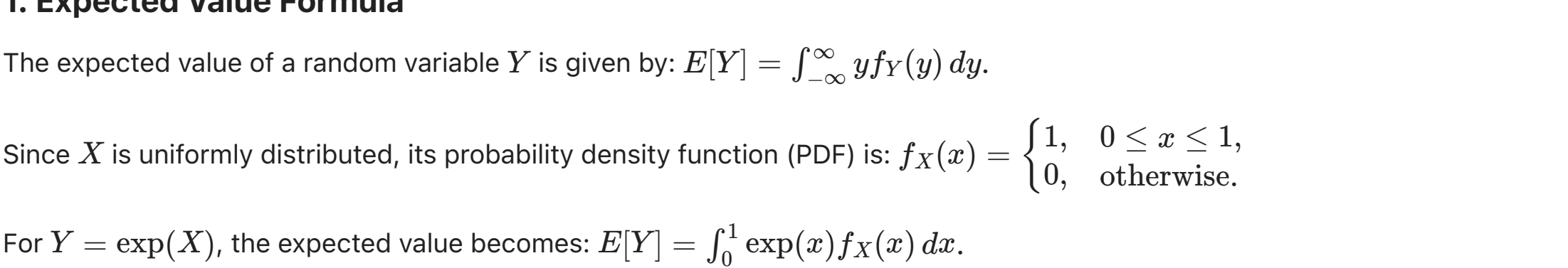
# Plot the Bode magnitude plot
p1 = plot(w, magnitude_db,
    , xscale=:log10
    , xlabel="Frequency (rad/s)", ylabel="Magnitude (dB)"
    , title="Bode Magnitude Plot", legend=false, grid=true
    , margin = 5mm
)

# Phase response in degrees
phase_deg = angle.(H.(w)) .* (180 / pi) # Convert phase from radians to degrees

# Plot the Bode phase plot
p2 = plot(w, phase_deg,
    , xscale=:log10
    , xlabel="Frequency (rad/s)", ylabel="Phase (degrees)"
    , title="Bode Phase Plot"
    , legend=false, grid=true
    , left_margin=10mm, right_margin=10mm, top_margin=15mm, bottom_margin=15mm
)

plot(p1, p2, layout = (1, 2), size = (1000, 400))
```

Out [6]:



4. Sketch magnitude and phase responses for a sampled data system with a pair of complex conjugate zeros and two poles at the origin.

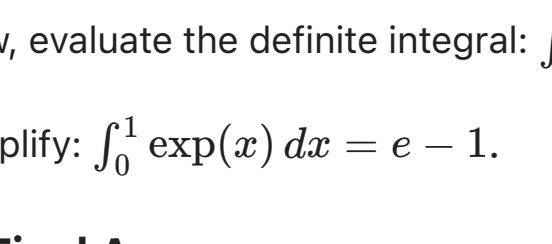


Figure Q4

[4 marks]

In [ ]:

5. A random variable  $X$  is uniformly distributed between  $x = 0$  and  $x = 1$ . Via any appropriate method, determine the expected value  $E[Y]$  of  $Y = \exp(X)$ .

[4 marks]

Given  $Y = \exp(X)$  and  $X \sim U(0, 1)$ .

1. Expected Value Formula

The expected value of a random variable  $Y$  is given by:  $E[Y] = \int_{-\infty}^{\infty} y f_Y(y) dy$ .

Since  $X$  is uniformly distributed, its probability density function (PDF) is:  $f_X(x) = \begin{cases} 1, & 0 \leq x \leq 1, \\ 0, & \text{otherwise.} \end{cases}$

For  $Y = \exp(X)$ , the expected value becomes:  $E[Y] = \int_0^1 \exp(x) f_X(x) dx$ .

Because  $f_X(x) = 1$  for  $0 \leq x \leq 1$ , this simplifies to:  $E[Y] = \int_0^1 \exp(x) dx$ .

2. Solve the Integral

The integral of  $\exp(x)$  is:  $\int \exp(x) dx = \exp(x) + C$ .

Now, evaluate the definite integral:  $\int_0^1 \exp(x) dx = [\exp(x)]_0^1 = \exp(1) - \exp(0)$ .

$$\text{Simplify: } \int_0^1 \exp(x) dx = e - 1.$$

3. Final Answer

$$\text{The expected value is: } \boxed{E[Y] = e - 1}$$

6. Identify the pivots and free variables of the following two matrices  $A$  and  $B$ . Following the method which we studied in class, find the special solution corresponding to each free variable and, by combining the special solutions, describe every solution to  $A\mathbf{x} = \mathbf{0}$  and  $B\mathbf{x} = \mathbf{0}$ .

$$A = \begin{bmatrix} 1 & 2 & 2 & 4 & 6 \\ 1 & 2 & 3 & 6 & 9 \\ 0 & 0 & 1 & 2 & 3 \end{bmatrix} \quad \text{and} \quad B = \begin{bmatrix} 2 & 4 & 2 \\ 0 & 4 & 4 \\ 0 & 8 & 8 \end{bmatrix}.$$

[7 marks]

In [ ]:

7. For a projection matrix  $P = A(A^T A)^{-1} A^T$ , show that  $P^2 = P$  and then explain, in terms of the column space of  $P$ , why projections  $P_b$  and  $P(P_b)$  give identical results.

[5 marks]

1. Show that  $P^2 = P$

The projection matrix  $P$  is defined as:  $P = A(A^T A)^{-1} A^T$ , where  $A$  is a matrix with linearly independent columns.

Compute  $P^2$ :

We want to show:  $P^2 = P$ .

$$\text{Start with } P^2: P^2 = P \cdot P = A(A^T A)^{-1} A^T \cdot (A(A^T A)^{-1} A^T).$$

$$\text{Expand the multiplication: } P^2 = A(A^T A)^{-1} A^T A(A^T A)^{-1} A^T.$$

Since  $A^T A$  is invertible,  $A^T A(A^T A)^{-1} = I$  (identity matrix). So:  $P^2 = A(A^T A)^{-1} I A^T = A(A^T A)^{-1} A^T$ .

This simplifies to:  $P^2 = P$ .

2. Projections  $P_b$  and  $P(P_b)$  Give Identical Results

Interpretation of  $P$ :

The projection matrix  $P$  projects any vector  $\mathbf{b}$  onto the **column space** of  $A$ , denoted as  $\text{Col}(A)$ .

Explain  $P_b$ :

$P_b = P\mathbf{b} = A(A^T A)^{-1} A^T \mathbf{b}$ . This gives the projection of  $\mathbf{b}$  onto  $\text{Col}(A)$ .

Explain  $P(P_b)$ :

$P(P_b)$  is the projection of  $P_b$  onto  $\text{Col}(P_b)$ :  $P(P_b) = P \cdot P_b$ . Since we showed that  $P^2 = P$ , this becomes:  $P(P_b) = P_b$ .

Why Are  $P_b$  and  $P(P_b)$  Identical?

- $P_b = P\mathbf{b}$  is already the projection of  $\mathbf{b}$  onto  $\text{Col}(A)$ .
- Applying  $P$  again to  $P_b$  does not change it, because projecting a vector already in the subspace  $\text{Col}(A)$  onto the same subspace leaves it unchanged.
- Hence:  $P(P_b) = P_b$