

Tutorial sheet 3.

① $f(t) = e^{kt}$

$$\mathcal{L}\{e^{kt}\} = \int_0^{\infty} e^{-st} e^{kt} dt = \lim_{T \rightarrow \infty} \int_0^T e^{-(s-k)t} dt$$

$$= \lim_{T \rightarrow \infty} \frac{-1}{s-k} \left[e^{-(s-k)t} \right]_0^T = \frac{1}{s-k} \left(1 - \lim_{T \rightarrow \infty} e^{-(s-k)T} \right)$$

(thus $\mathcal{L}\{e^{kt}\} = \frac{1}{s-k}$ provided that, for real k ,

$\sigma = \text{Re}(s) > k$. If k is complex then we require $\text{Re}(s) > \text{Re}(k)$.)

Thus we have the Laplace transform pair

$$\left. \begin{aligned} f(t) &= e^{kt} \\ F(s) &= \frac{1}{s-k} \end{aligned} \right\} \text{Re}(s) > \text{Re}(k)$$

Since $\exp(jat) = \cos at + j \sin at$ we may write

$$f(t) = \sin at = \text{Im } e^{jat} \quad \& \quad g(t) = \cos at = \text{Re } e^{jat}$$

From the result of Q1 with $k = ja$ we have

$$\mathcal{L}\{e^{jat}\} = \frac{1}{s - ja} \quad \text{Re}(s) > 0$$

$$= \frac{s + ja}{s^2 + a^2} \quad \text{Re}(s) > 0$$

Thus equating real and imaginary parts and assuming that s is real we have

$$\mathcal{L}\{\sin at\} = \text{Im } \mathcal{L}\{e^{jat}\} = \frac{a}{s^2 + a^2}$$

$$\mathcal{L}\{\cos at\} = \text{Re } \mathcal{L}\{e^{jat}\} = \frac{s}{s^2 + a^2}$$

which also hold for complex s with $\text{Re}(s) > 0$.

$$\begin{aligned}
 (2) \quad \text{Since } \mathcal{L}\{\alpha f(t) + \beta g(t)\} &= \int_0^{\infty} [\alpha f(t) + \beta g(t)] e^{-st} dt \\
 &= \int_0^{\infty} \alpha f(t) e^{-st} dt + \int_0^{\infty} \beta g(t) e^{-st} dt \\
 &= \alpha \int_0^{\infty} f(t) e^{-st} dt + \beta \int_0^{\infty} g(t) e^{-st} dt
 \end{aligned}$$

we have $\mathcal{L}\{\alpha f(t) + \beta g(t)\} = \alpha \mathcal{L}\{f(t)\} + \beta \mathcal{L}\{g(t)\}$

If $f(t)$ & $g(t)$ have abscissae of convergence σ_f & σ_g respectively & $\sigma_1 > \sigma_f$, $\sigma_2 > \sigma_g$ then

$$|f(t)| < M_1 e^{\sigma_1 t}, \quad |g(t)| < M_2 e^{\sigma_2 t}$$

it follows that

$$\begin{aligned}
 |\alpha f(t) + \beta g(t)| &\leq |\alpha| |f(t)| + |\beta| |g(t)| \\
 &\leq |\alpha| M_1 e^{\sigma_1 t} + |\beta| M_2 e^{\sigma_2 t} \\
 &\leq (|\alpha| M_1 + |\beta| M_2) e^{\sigma t} \quad \text{where } \sigma = \max(\sigma_1, \sigma_2)
 \end{aligned}$$

Exploiting the linearity property we calculate:

$$\begin{aligned}
 \mathcal{L}\{t\} &= \int_0^{\infty} e^{-st} t dt = \lim_{T \rightarrow \infty} \int_0^T e^{-st} t dt \\
 &= \lim_{T \rightarrow \infty} \left[-\frac{t}{s} e^{-st} - \frac{e^{-st}}{s^2} \right]_0^T = \frac{1}{s^2} - \lim_{T \rightarrow \infty} \frac{T e^{-sT}}{s} - \lim_{T \rightarrow \infty} \frac{e^{-sT}}{s^2} \\
 &= \frac{1}{s^2} \quad \text{Re}(s) > 0
 \end{aligned}$$

& from Q1 we have $\mathcal{L}\{e^{kt}\} = \frac{1}{s-k} \quad \text{Re}(s) > \text{Re}(k)$

Thus

$$\begin{aligned}
 \mathcal{L}\{3t + 2e^{3t}\} &= 3\mathcal{L}\{t\} + 2\mathcal{L}\{e^{3t}\} \\
 &= \frac{3}{s^2} + \frac{2}{s-3} \quad \text{Re}(s) > \max(0, 3) \\
 &= \frac{3}{s^2} + \frac{2}{s-3} \quad \text{Re}(s) > 3
 \end{aligned}$$

$$\textcircled{3} \quad \text{Since } \mathcal{L}\{e^{at}f(t)\} = \int_0^{\infty} e^{at}f(t)e^{-st}dt = \int_0^{\infty} f(t)e^{-(s-a)t}dt$$

$$\text{and } \mathcal{L}\{f(t)\} = \int_0^{\infty} f(t)e^{-st}dt = F(s) \quad \text{Re}(s) > \sigma_c$$

we see that

$$\mathcal{L}\{e^{at}f(t)\} = F(s-a) \quad \text{Re}(s-a) > \sigma_c$$

$$\text{or } \text{Re}(s) > \sigma_c + \text{Re}(a)$$

$$\therefore \text{ from Q2 we have that } \mathcal{L}\{t\} = \frac{1}{s^2} \quad \text{Re}(s) > 0$$

$$\text{so } \mathcal{L}\{te^{-2t}\} = F(s+2) = [F(s)]_{s \rightarrow s+2} \quad \text{Re}(s) > 0 - 2$$

$$= \frac{1}{(s+2)^2} \quad \text{Re}(s) > -2$$

$$\textcircled{4} \quad \mathcal{L}\{f(t)\} = F(s) = \int_0^{\infty} e^{-st}f(t)dt$$

$$\text{so that } \frac{d^n F(s)}{ds^n} = \frac{d^n}{ds^n} \int_0^{\infty} e^{-st}f(t)dt = \int_0^{\infty} \frac{\partial^n}{\partial s^n} [e^{-st}f(t)]dt$$

$$\left. \begin{array}{l} \text{For } \mathcal{L}\{t \sin 3t\} \text{ given that} \\ \mathcal{L}\{\sin 3t\} = F(s) = \frac{3}{s^2+9} \quad \text{Re}(s) > 0 \end{array} \right\} \begin{array}{l} = (-1)^n \int_0^{\infty} e^{-st} t^n f(t) dt \\ = (-1)^n \mathcal{L}\{t^n f(t)\} \quad \text{Re}(s) > \sigma_c \end{array}$$

we have that

$$\mathcal{L}\{t \sin 3t\} = -\frac{dF(s)}{ds} = \frac{6s}{(s^2+9)^2} \quad \text{Re}(s) > 0$$

$$\text{For } \mathcal{L}\{t^2 e^t\} \text{ given that } \mathcal{L}\{e^t\} = F(s) = \frac{1}{s-1} \quad \text{Re}(s) > 1$$

$$\mathcal{L}\{t^2 e^t\} = (-1)^2 \frac{d^2 F(s)}{ds^2} = (-1)^2 \frac{d^2}{ds^2} \left(\frac{1}{s-1} \right) = (-1) \frac{d}{ds} \left(\frac{1}{(s-1)^2} \right)$$

$$= \frac{2}{(s-1)^3} \quad \text{Re}(s) > 1$$

but we could have used the first shift theorem.

$$(5) \quad \mathcal{L}\left\{\frac{df}{dt}\right\} = \int_0^{\infty} e^{-st} \frac{df}{dt} dt$$

& on integrating by parts we have

$$\begin{aligned} \mathcal{L}\left\{\frac{df}{dt}\right\} &= [e^{-st} f(t)]_0^{\infty} + s \int_0^{\infty} e^{-st} f(t) dt \\ &= -f(0) + s F(s) \end{aligned}$$

$$\text{so } \mathcal{L}\left\{\frac{df}{dt}\right\} = s F(s) - f(0)$$

$$\begin{aligned} \mathcal{L}\left\{\frac{d^2f}{dt^2}\right\} &= \int_0^{\infty} e^{-st} \frac{d^2f}{dt^2} dt = [e^{-st} \frac{df}{dt}]_0^{\infty} + s \int_0^{\infty} e^{-st} \frac{df}{dt} dt = -\left[\frac{df}{dt}\right]_{t=0} + s \mathcal{L}\left\{\frac{df}{dt}\right\} \\ &= -\left[\frac{df}{dt}\right]_{t=0} + s [s F(s) - f(0)] \end{aligned}$$

$$\text{thus } \mathcal{L}\left\{\frac{d^2f}{dt^2}\right\} = s^2 F(s) - s f(0) - \left[\frac{df}{dt}\right]_{t=0} = s^2 F(s) - s f(0) - f'(0)$$

$$\text{& in general } \mathcal{L}\{f^{(n)}(t)\} = s^n F(s) - \sum_{i=1}^n s^{n-i} f^{(i-1)}(0)$$

$$(6) \quad \text{Writing } g(t) = \int_0^t f(\tau) d\tau \quad \text{we have } \frac{dg}{dt} = f(t) \quad g(0) = 0$$

taking Laplace transforms $\mathcal{L}\left\{\frac{dg}{dt}\right\} = \mathcal{L}\{f(t)\}$ which from above

$$\text{gives } s G(s) = F(s) \text{ or } \mathcal{L}\{g(t)\} = G(s) = \frac{1}{s} F(s) = \frac{1}{s} \mathcal{L}\{f(t)\}$$

which gives the required result.

$$\text{Starting with } f(t) = t^3 + \sin 2t$$

$$F(s) = \mathcal{L}\{t^3\} + \mathcal{L}\{\sin 2t\}$$

$$= \frac{6}{s^4} + \frac{2}{s^2 + 4}$$

$$\text{Thus } \mathcal{L}\left\{\int_0^t (\tau^3 + \sin 2\tau) d\tau\right\} = \frac{1}{s} F(s) = \frac{6}{s^5} + \frac{2}{s(s^2 + 4)}$$

(7)

(a) Resolving into partial fractions

$$\frac{1}{(s+3)(s-2)} = \frac{-1/5}{s+3} + \frac{1/5}{s-2}$$

Using the result $\mathcal{L}^{-1}\{1/(s+a)\} = e^{-at}$ & the linearity property

$$\begin{aligned}\mathcal{L}^{-1}\left\{\frac{1}{(s+3)(s-2)}\right\} &= -\frac{1}{5}\mathcal{L}^{-1}\left\{\frac{1}{s+3}\right\} + \frac{1}{5}\mathcal{L}^{-1}\left\{\frac{1}{s-2}\right\} \\ &= -\frac{1}{5}e^{-3t} + \frac{1}{5}e^{2t}\end{aligned}$$

(b) Resolving into partial fractions and results in the table of Laplace transforms gives

$$\begin{aligned}\mathcal{L}^{-1}\left\{\frac{s+1}{s^2(s^2+9)}\right\} &= \mathcal{L}^{-1}\left\{\frac{1/9}{s} + \frac{1/9}{s^2} - \frac{1/9(s+1)}{s^2+9}\right\} \\ &= \mathcal{L}^{-1}\left\{\frac{1/9}{s} + \frac{1/9}{s^2} - \frac{1/9 s}{s^2+3^2} - \frac{1/27 \times 3}{s^2+3^2}\right\} \\ &= \frac{1}{9} + \frac{1}{9}t - \frac{1}{9}\cos 3t - \frac{1}{27}\sin 3t\end{aligned}$$

(c) Similarly

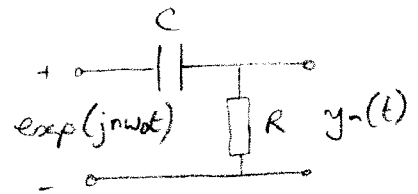
$$\mathcal{L}^{-1}\left\{\frac{2}{s^2+6s+13}\right\} = \mathcal{L}^{-1}\left\{\frac{2}{(s+3)^2+4}\right\} = \left[\frac{2}{s^2+2^2}\right]_{s \rightarrow s+3}$$

& since $2/(s^2+2^2) = \mathcal{L}\{\sin 2t\}$ the shift theorem gives

$$\mathcal{L}^{-1}\left\{\frac{2}{s^2+6s+13}\right\} = e^{-3t}\sin 2t$$

⑧ Using Kirchhoff's current law

$$\text{current} = C \frac{d(\exp(j\omega_0 t) - y)}{dt} = y/R$$



$$\therefore C \exp(j\omega_0 t) j\omega_0 - C \frac{dy}{dt} = y/R$$

$$\Rightarrow y + RC \frac{dy}{dt} = RC j\omega_0 \exp(j\omega_0 t)$$

& assuming a solution of $y(t) = K \exp(j\omega_0 t)$

$$K \exp(j\omega_0 t) + RC K j\omega_0 \exp(j\omega_0 t) = RC j\omega_0 \exp(j\omega_0 t)$$

& solving for K

$$K + RC K j\omega_0 = RC j\omega_0 \Rightarrow K = \frac{RC j\omega_0}{1 + RC j\omega_0}$$

$$\text{From } C \frac{d}{dt} (x(t) - y(t)) = \frac{y}{R}$$

$$RC \frac{dx}{dt} = RC \frac{dy}{dt} + y$$

Taking Laplace transforms & assuming zero initial conditions

$$RC s X(s) = RC s Y(s) + Y(s)$$

$$H(s) = \frac{RC s}{RC s + 1}$$

⑨ Applying Kirchhoff's second law to the circuit gives

$$Ri + L \frac{di}{dt} + \frac{1}{C} \int i dt = e(t)$$

or, using $i = \frac{dq}{dt}$

$$L \frac{d^2 q}{dt^2} + R \frac{dq}{dt} + \frac{1}{C} q = e(t)$$

& substituting the given values for R, C, L & $e(t)$

$$\frac{d^2 q}{dt^2} + 160 \frac{dq}{dt} + 10^4 q = 20$$

Taking Laplace transforms

$$(s^2 + 160s + 10^4) Q(s) = [sq(0) + \dot{q}(0)] + 160q(0) + \frac{20}{s}$$

where $Q(s)$ is the transform of $q(t)$

We are told that $q(0) = 0$ & $\dot{q}(0) = i(0) = 0$ so
the equation is reduced to

$$(s^2 + 160s + 10^4) Q(s) = \frac{20}{s}$$

that is

$$\begin{aligned} Q(s) &= \frac{20}{s(s^2 + 160s + 10^4)} = \frac{1/500}{s} - \frac{1}{500} \frac{s + 160}{s^2 + 160s + 10^4} \\ &= \frac{1}{500} \left[\frac{1}{s} - \frac{(s + 80) + 4/3(60)}{(s + 80)^2 + (60)^2} \right] = \frac{1}{500} \left[\frac{1}{s} - \left[\frac{s + 4/3 \times 60}{s^2 + 60^2} \right]_{s \rightarrow s+80} \right] \end{aligned}$$

& taking inverse transforms & making use of the shift theorem

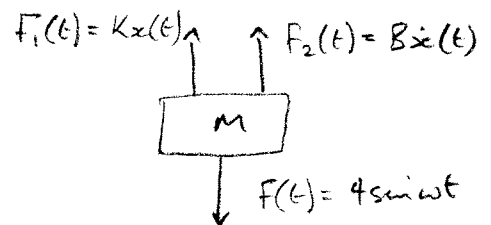
$$q(t) = \frac{1}{500} \left(1 - e^{-80t} \cos 60t - \frac{4}{3} e^{-80t} \sin 60t \right)$$

& the resulting current is given by

$$i(t) = \frac{dq}{dt} = \frac{1}{3} e^{-80t} \sin 60t$$

(10) By Newton's Law

$$M\ddot{x}(t) = F(t) - F_1(t) - F_2(t)$$



giving

$$\ddot{x}(t) + 6\dot{x}(t) + 25x(t) = 4\sin \omega t$$

Taking Laplace transforms gives

$$(s^2 + 6s + 25)X(s) = [sx(0) + \dot{x}(0)] + 6x(0) + \frac{4\omega}{s^2 + \omega^2}$$

$$\Rightarrow X(s) = \frac{4\omega}{(s^2 + \omega^2)(s^2 + 6s + 25)}$$

For (a) with $\omega = 2$ $X(s) = \frac{8}{(s^2 + 4)(s^2 + 6s + 25)}$

$$= \frac{4}{195} \frac{-4s + 14}{s^2 + 4} + \frac{2}{195} \frac{8s + 20}{s^2 + 6s + 25}$$

$$= \frac{4}{195} \frac{-4s + 14}{s^2 + 4} + \frac{2}{195} \frac{8(s+3) - 4}{(s+3)^2 + 16}$$

& taking inverse Laplace transforms gives

$$x(t) = \frac{4}{195} (7\sin 2t - 4\cos 2t) + \frac{2}{195} e^{-3t} (8\cos 4t - \sin 4t)$$

For (b) with $\omega = 5$ $X(s) = \frac{20}{(s^2 + 25)(s^2 + 6s + 25)}$

$$= \frac{-2/15 s}{s^2 + 25} + \frac{1}{15} \frac{2(s+3) + 6}{(s+3)^2 + 16}$$

& upon taking inverse Laplace transforms

$$x(t) = -\frac{2}{15} \cos 5t + \frac{1}{15} e^{-3t} (2\cos 4t + \frac{3}{2} \sin 4t)$$

If the damping term were missing we have $X(s) = \frac{20}{(s^2 + 25)^2}$

& from the derivative of transform property

$$\mathcal{L}\{t \cos st\} = -\frac{d}{ds} \mathcal{L}\{\cos st\} = -\frac{d}{ds} \left(\frac{s}{s^2 + 25} \right)$$

$$\text{that is } \mathcal{L}\{t \cos st\} = -\frac{1}{s^2 + 25} + \frac{2s^2}{(s^2 + 25)^2}$$

$$= \frac{1}{s^2 + 25} - \frac{50}{(s^2 + 25)^2}$$

$$= \frac{1}{5} \mathcal{L}\{\sin 5t\} - \frac{50}{(s^2 + 25)^2}$$

& by the linearity property

$$\mathcal{L}\left\{\frac{1}{5} \sin 5t - t \cos 5t\right\} = \frac{50}{(s^2 + 25)^2}$$

so taking inverse Laplace transforms of $X(s) = \frac{20}{(s^2 + 25)^2}$

$$\text{gives } x(t) = \frac{2}{25} (\sin 5t - 5t \cos 5t)$$

Because of the term $t \cos 5t$ the response is unbounded as $t \rightarrow \infty$. This arises because in this case the applied force $F(t) = 4 \sin 5t$ is in resonance with the system.

(11)

(a) There are three major regions

region 1: $0 \leq t < 1$

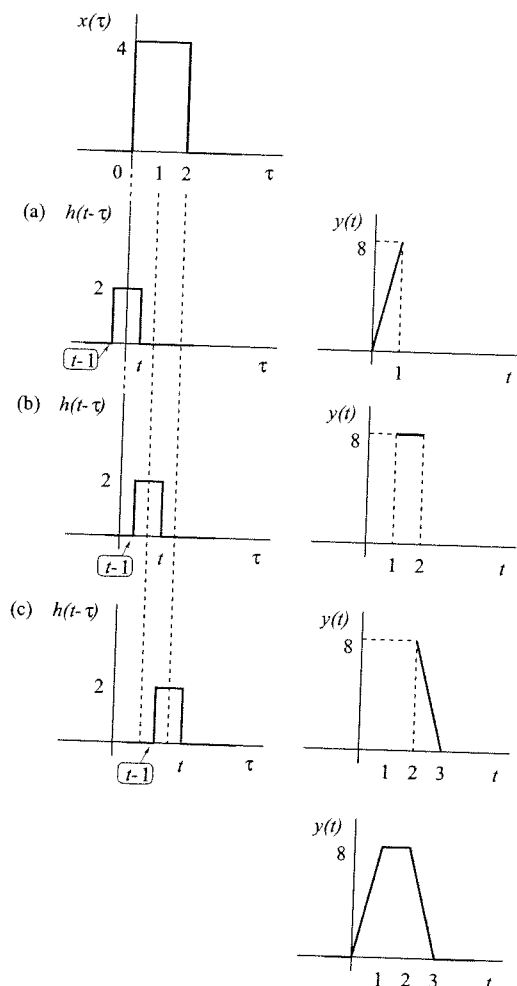
$$y(t) = \int_0^t (4)(2) dt = 8t$$

region 2: $1 \leq t < 2$

$$y(t) = \int_{t-1}^t (4)(2) dt = 8$$

region 3: $2 \leq t < 3$

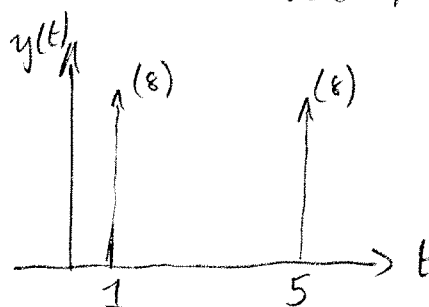
$$y(t) = \int_{t-1}^2 (4)(2) dt = 8(3-t)$$

(b) the output $y(t) = h(t) * x(t)$

$$= 4\delta(t-1) * (2\delta(t) + 2\delta(t-4))$$

$$= 8\delta(t-1) * \delta(t) + 8\delta(t-1) * \delta(t-4)$$

$$= 8\delta(t-1) + 8\delta(t-5)$$



(c) There are five major regions

region 1: $0 \leq t < 1$ $y(t) = \int_0^t h(\tau) x(t-\tau) d\tau = \text{shaded area in (a)}$

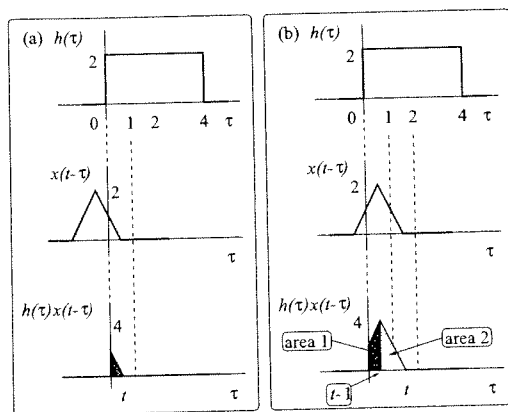
$$= \frac{1}{2}(t)(4t) = 4t^2$$

region 2: $1 \leq t < 2$ $y(t) = \int_0^t h(\tau) x(t-\tau) d\tau = \text{area 1 + area 2 in (b)}$

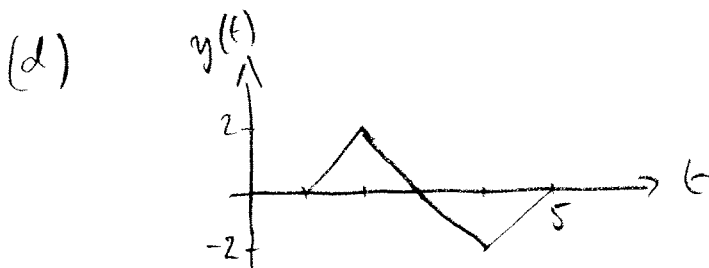
$$= \frac{t-1}{2}(-4(t-2)+4) + \frac{1}{2}(1)(4)$$

i.e. the area of a trapezium + area of a triangle

$$= 2(t-1)(3-t) + 2$$



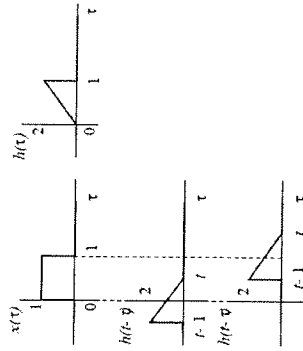
The rest of the solution proceeds in the same manner.



Using equation (2.8)

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$

Drawing diagrams:



From which, $h(t-\tau) = 2(t-\tau)$ provided $t-1 < \tau < t$ - the equation of a straight line. The convolution in the region $0 \leq t < 1$:

$$\begin{aligned} y(t) &= \int_0^t (1) 2(t-\tau) d\tau \\ &= 2 \int_0^t (t-\tau) d\tau \\ &= t^2 \end{aligned}$$

The convolution in the region $1 \leq t < 2$:

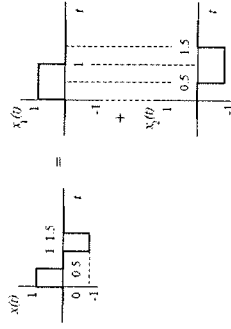
$$\begin{aligned} y(t) &= \int_{t-1}^1 (1) 2(t-\tau) d\tau \\ &= 2 \int_{t-1}^1 (t-\tau) d\tau \\ &= t(2-t) \end{aligned}$$

Solution:



New input $x(t)$. The new input can be described as the addition of the old input (which we shall call $x_1(t)$) and a delayed and inverted version of the old input which we shall call $x_2(t)$ as illustrated

below:



Thus

$$x(t) = x_1(t) + x_2(t)$$

and

$$\begin{aligned} y(t) &= h(t) * x(t) \\ &= h(t) * x_1(t) + h(t) * x_2(t) \\ &= y_1(t) + y_2(t) \end{aligned}$$

$y_1(t)$ has already been evaluated and in the first part of the question. $x_2(t)$ is a delayed and inverted version of $x_1(t)$

$$x_2(t) = -x_1(t-1/2)$$

Thus $y_2(t)$ is a delayed and inverted version of $y_1(t)$

$$y_2(t) = -y_1(t-1/2)$$

This gives the result that:

$$y(t) = y_1(t) - y_1(t-1/2)$$

Thus for $1/2 \leq t < 3/2$

$$y_2(t) = -(t-1/2)^2$$

and for $3/2 \leq t < 5/2$

$$y_2(t) = -(t-1/2)(2-(t-1/2))$$

elsewhere $y_2(t) = 0$ The complete solution for $y(t)$ becomes:

for $0 \leq t < 1/2$

$$y(t) = t^2$$

for $1/2 \leq t < 1$

$$y(t) = t^2 - (t-1/2)^2$$

for $1 \leq t < 3/2$

$$y(t) = t(2-t) - (t-1/2)^2$$

for $3/2 \leq t < 2$

$$y(t) = t(2-t) - (t-1/2)(2-(t-1/2))$$

for $2 \leq t < 5/2$

$$y(t) = -(t-1/2)(2-(t-1/2)) \text{ for } 5/2 < t$$

$$y(t) = 0$$