Mobile Communication Techniques Petros Elia, elia@eurecom.fr

Final Exam February 10 2010 Time: 09:00-11:00

Instructions

- Exercises fall in categories of 1-point, 2-point and 4-point exercises.
- Total of $10 \times 1 + 6 \times 2 + 2 \times 4 = 30$ points.
- Each answer should be clearly written, and the solution should be developed in detail.
- Mathematical derivations need to show all steps that lead to the answer.
- Complete as many exercises as you can. Don't spend too much time on an individual question.
- Partial credit will be given for incomplete solutions.
- There is NO penalty for incorrect solutions.
- If in certain cases you are unable to provide rigorous mathematical proofs, go ahead and provide intuitive justification of your answers. Partial credit will be given.
- Calculators are allowed. Feel free to provide close form expressions in their simplest form.
- Open book and open class notes are allowed. No other notes are allowed.

Hints - equations - conventions:

Notation

- R represents the rate of communication in bits per channel use (b.p.c.u),
- ρ represents the SNR (signal to noise ratio),
- w will denote additive noise which will be distributed as a circularly symmetric Gaussian random variable $\mathbb{C}\mathcal{N}(0, N_0)$. If N_0 is not specified, then set $N_0 = 1$,
- h_i will denote *independent* fading scalar coefficients which will be distributed as circularly symmetric Gaussian random variables $\mathbb{C}\mathcal{N}(0,1)$

• Useful equations

- For small ϵ , and for the statistical distributions of interest, the following holds:

$$P(|h_i|^2 < \epsilon) = \int_0^{\epsilon} p(|h_i|^2) d|h_i|^2 \approx \max_{[0,\epsilon]} \left\{ p(|h_i|^2 = \epsilon) \right\}$$

- For small ϵ , and for the statistical distributions of interest, the following holds:

$$P(|h_i|^2 + |h_j|^2 < \epsilon) \approx p\bigg(\{|h_i|^2 < \epsilon\} \text{ or } \{|h_j|^2 < \epsilon\}\bigg)$$

- For a given probability of error P(error), the diversity order or diversity gain is defined as

$$d := -\lim_{\rho \to \infty} \frac{\log P(\text{error})}{\log \rho}.$$

- The multiplexing gain is defined as $r := R/\log \rho$
- The diversity multiplexing tradeoff (DMT) is defined as a diversity gain achieved, given a specific multiplexing gain
- GOOD LUCK!!

EXAM PROBLEMS

- 1) (1 point). What is the maximum and minimum wavelengths corresponding to a 25MHz signal band centered around 5.8 GHz?
- 2) (1 point). Consider a transmitter and a receiver which receives from different paths.
 - What is the maximum length difference between any two paths so that the delay spread is approximately $3\mu s$?
- 3) (1 point). At 1.9 GHz, a narrowband channel induces approximately how big of a coherence distance?
- 4) (1 point). What is the capacity of the AWGN channel when using power equal to the average power of 4-QAM, over noise $\sim \mathbb{C}\mathcal{N}(0,2)$?
- 5) (1 point). Define mathematically the concept of deep fade.
- 6) (1 point). What are some ways that allow for the negative effect of deep fade be reduced?
- 7) (1 point). What is the average probability of error (averaged over the fading realizations), in a $1 \times n_r$ SIMO Rayleigh fading channel (high snr approximation)?
- 8) (1 point). Describe the rate of the Alamouti code carrying QPSK elements.
- 9) (1 point). How many degrees of freedom can you have in a
 - SISO channel
 - $n_t \times 1$ MISO channel
 - $n_t \times n_r$ MIMO channel
- 10) (1 point). Describe the relation between the cardinality of the code $|\mathcal{X}|$, the rate of transmission R and the coding duration T
- 11) (2 points). Consider the general noiseless MIMO channel model

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \underbrace{\begin{bmatrix} \underline{h}_1 & \underline{h}_2 & \underline{h}_3 \end{bmatrix}}_{H} \underbrace{\begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}}_{x}$$

- What is a necessary condition on H such that decoding of all the components of \underline{x} is always possible?
- What is a necessary condition on H such that decoding of any of the components of \underline{x} is always possible?
- 12) (2 points). Consider the space-time (ST) code that takes the form

$$\begin{bmatrix} x_0 & x_1 & x_2 \\ x_4 & x_0 & x_2 \\ x_1 & x_3 & x_5 \end{bmatrix}$$

where the x_i are drawn independently from a 16-QAM constellation.

- What is the rate of the code, in bits per channel use (bpcu)?
- 13) (2 points). Consider the following two channel models:

$$y = h_1 x + w$$
, and $y = h_1 h_2 h_3 x + w$

where $h_1, h_2, h_3 \sim \text{i.i.d } \mathbb{C}\mathcal{N}(0, 1)$.

• What is the diversity order

$$d_{\text{out}} = -\lim_{\rho \to \infty} \frac{\log P(\text{outage})}{\log \rho}$$

for the two cases? (here you can simply state the answer and the explanation - no derivation necessary)

• What is the time-diversity corresponding to channel

$$[y_1 \quad y_2 \quad y_3] = [h_1h_2x_1 \quad h_1h_2h_3x_2 \quad h_1h_3x_3] + [w_1 \quad w_2 \quad w_3]$$

for x_i that optimize diversity

14) (2 points). Consider a space time code that takes the form

$$\begin{bmatrix} x_0 & x_1 & x_2 \\ x_4 & x_5 & x_6 \\ x_7 & x_8 & x_9 \end{bmatrix}$$

where the x_i are drawn independently from a QAM constellation. Describe the optimal diversity gain $-\lim_{\rho \to \infty} \frac{\log P(\text{error})}{\log \rho}$ provided by the code over the

- 3×1 MISO channel $\underline{h} = [h_1 \ h_2 \ h_3]$
- 3×5 MIMO channel.
- 15) (2 points). Let C_1 be the capacity of the AWGN channel with SNR $\rho_1=e^{3n}$, and let C_2 be the capacity of the AWGN channel with SNR $\rho_2=e^{-5n}$. Let n>>1.
 - What is the ratio $\frac{C_1}{C_2}$. (NOTE!: the final answer should NOT include logarithms!!)
- 16) (2 points). Consider the MISO 3×1 channel

$$y[m] = h_1 x_1[m] + h_2 x_2[m] + h_3 x_3[m] + w[m]$$

where the h_i coefficients are constant at $h_1 = 1, h_2 = 0.5, h_3 = 0.25$.

- What is the optimal encoding-decoding policy, and what channel state information is required at the receiver and at the transmitter, so that the corresponding capacity is achieved?
- What is the corresponding capacity?
- 17) (4 points). For the 1×3 SIMO Rayleigh fading channel

$$y = \underline{h}x + \underline{w}$$

calculate

$$\lim_{\rho \to \infty} \frac{\log P(\mathsf{outage})}{\log \rho}$$

as a function of the multiplexing gain

$$r := R/\log \rho$$
.

- 18) (4 points). Consider the quasi-static Rayleigh fading channel. Let SNR = 10^{-2} and $\epsilon = 10^{-3}$.
 - What is the approximate $\epsilon\text{-outage}$ capacity C_ϵ for SISO?
 - Show the gains in ϵ -outage capacity, brought about by adding an extra receive antenna.