

BLG 231E

(Digital Circuits)

analog signals : continuous, infinite resolution

digital signals : discrete, two possible values
(1-0, true-false, on-off, high-low)

Digital Coding

- n bits $\rightarrow 2^n$ different "things"
- Binary coded decimal (BCD) $\rightarrow (0-9 \rightarrow$ four bit)

$$805 = 1000 \ 0000 \ 0101$$

- Positional (weighted) coding

$$\begin{array}{r} 11010 \\ \downarrow \\ \text{least significant bit (LSB)} \\ \text{most significant bit (MSB)} \end{array}$$
$$11010 = 1 \cdot 2^4 + 1 \cdot 2^3 + 0 \cdot 2^2 + 1 \cdot 2^1 + 0 \cdot 2^0 = 26$$

- Haming distance : number of bit positions in which they are different. (011 and $101 \Rightarrow h.d. = 2$)
- Adjacent codes : a pair codes that haming d. = 1

▷ if the hamming distance between the first and last code word is 1 in a coding system then this code is called circular.

Gray Code : binary, non-redundant and circular
(n bits - 2^n things)

- integers are represented in computers by "natural binary weighted (positional) coding".

▷ repr. of signed \Rightarrow MSB

- 0 \Rightarrow positive
- 1 \Rightarrow negative

↓
2's complement system
(invert the number and add 1)

extension

- unsigned \Rightarrow 4-bit $3_{10} = 0011 \rightarrow$ 8-bit $3_{10} = 0000\ 0011$

- signed \Rightarrow 4-bit $-7_{10} = 1001 \rightarrow$ 8-bit $-7_{10} = \underbrace{1111\ 1001}_{\text{filled with the sign of MSB}} \text{ (sign extension)}$

Hexadecimal Numbers

- (0 - 9) and (A - F) \rightarrow base 16 system
- $01011101_2 = 5D_{16} = (5 \cdot 16) + 13 = 93_{10}$
 \rightarrow (usually #5D or 5Dh)

Arithmetical Operations

- addition of two n-bits can be $(n+1)$ -bits then
 $(n+1)^{th}$ bit \rightarrow carry ($=0 \Rightarrow$ no carry)
- addition of two n-bits signed if $(n+1)^{th}$ bit arises
it should be ignored.
- in signed integers if:
 $pos + pos \rightarrow neg$ and $neg + neg \rightarrow pos$
 \Rightarrow there is an overflow and $(n+1)^{th}$ bit is ignored.
- subtraction of two n-bit signed if $(n+1)^{th}$ bit is 0
the first operand is smaller than the second one and
there is borrow (no carry)
- $pos - neg \rightarrow neg$ or $neg - pos \rightarrow pos \Rightarrow$ overflow

Boolean Algebra

* the dual of $a + a \cdot b$ is $a \cdot (a+b)$ and they are equal.

Logical Functions

1) simple (basic) functions → multiple inputs, single output

→ for n binary variables there are $2^{(2^n)}$ possible basic logical functions.

2) general functions → multiple inputs, multiple outputs

3) incompletely specified functions

→ they have unspecified outputs for some input combinations

Representation of Logical Functions

1) truth table representation

2) numbered (indexed) representation

truth table:

row num	input $x_1 x_2$	output y
0	0 0	1
1	0 1	0
2	1 0	1
3	1 1	0

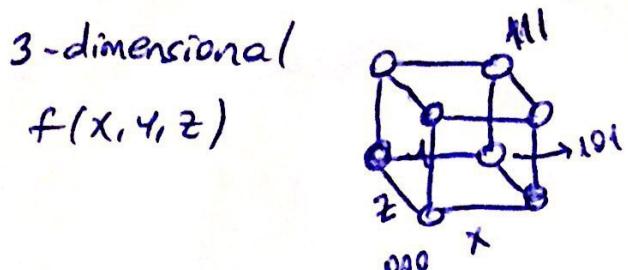
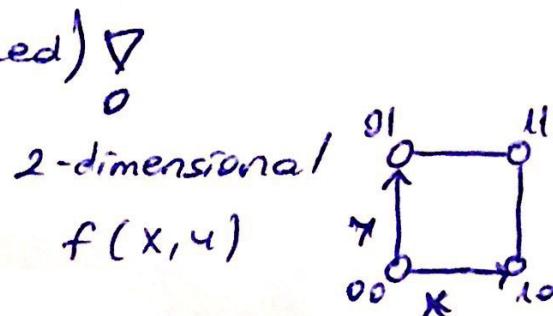
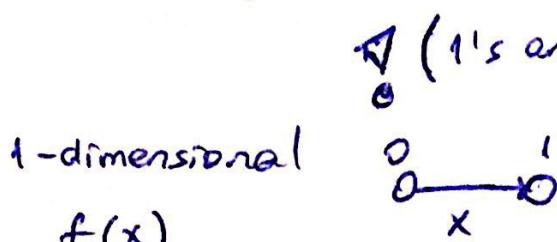
numbered (indexed):

$$y = f(x_1 x_2) = \bigcup_{i=1}^3 \{ (x_1, x_2) \}$$

$\bigcup_{i=1}^3$ → "union" or "set of"
 $\{ (x_1, x_2) \}$ → set of 1-generating points

Graphical Representation

n -bits input \rightarrow n -dimensional cube



* The variables and outputs transferred into a table that is in a matrix form and rows and columns labeled according to the Gray Code property. \rightarrow Karnaugh Maps

F		00	01	11	10	Gray code
AB		00	01	11	10	
		00	01	11	10	
		0	1	3	2	
		4	5	7	6	
		12	13	15	14	
		8	9	11	10	



absorption $\rightarrow E + E \cdot F = E$, $E \cdot (E + F) = E$

$E + \bar{E} \cdot F = E + F$, $E \cdot (\bar{E} + F) = E \cdot F$

consensus $\rightarrow x_1 E_1 + \bar{x}_1 E_2 + E_1 E_2 = x_1 E_1 + \bar{x}_1 E_2$

$(x_1 + E_1) \cdot (\bar{x}_1 + E_2) \cdot (E_1 + E_2) = (x_1 + E_1) \cdot (\bar{x}_1 + E_2)$



Canonical Forms

1st) Sum of Products $\rightarrow SOP (\Sigma \pi)$

\rightarrow sum of products, each of which corresponds to a " 1 " - generating combination.

Minterm : a product of n literals (in a n -variables function) in which each variable appears exactly once (a or a').

$\rightarrow abc, a'b'c, a'b'c', ab'c' \rightarrow 3$ variables
4 minterms

* indexing minterms \rightarrow we assign the indexes of minterms like $m_0, m_1, m_2 \dots$ and we can represent the function:

$$F(A, B, C) = \sum m(1, 3, 5, 6, 7) \text{ for example.}$$

2nd) Products of Sum $\rightarrow POS (\Pi \Sigma)$

\rightarrow products of sum, each of which corresponds to a " 0 " - generating combination.

Maxterm : the difference of minterm is each maxterm has a value of " 0 ".

! while finding maxterms \rightarrow we substitute variable for zeros and complements of variables for ones. ($A \equiv 0, A' \equiv 1$)

* to indexing maxterms we use M and represent like $F(A, B, C) = \prod M(0, 2, 4)$

Logic Gates

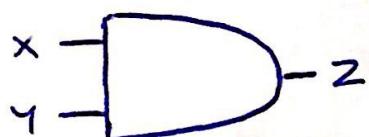
Buffer ($Y = X$)



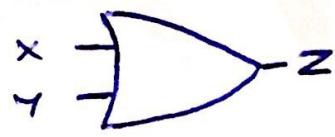
Inverter ($Y = \bar{X}$)



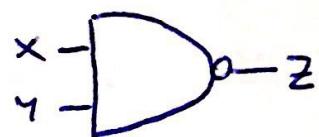
And ($Z = X \cdot Y$)



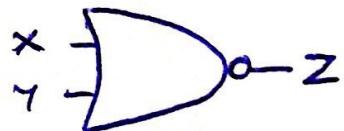
Or ($Z = X + Y$)



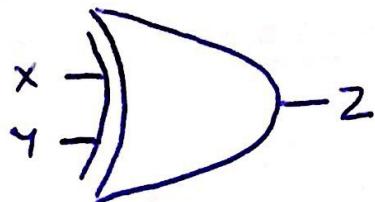
Not And (NAND) ($Z = (\overline{X} \cdot \overline{Y})$)



Not Or (NOR) ($Z = (\overline{X} + \overline{Y})$)

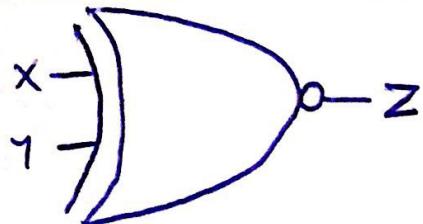


Difference (XOR) ($Z = X\bar{Y} + \bar{X}Y$)



$$\rightarrow Z = X \oplus Y$$

Equality (XNOR) ($Z = XY + \bar{X}\bar{Y}$)



$$\rightarrow Z = X \odot Y$$

- ▷ 1 = high value and 0 = low value \Rightarrow positive logic
- 1 = low value and 0 = high value \Rightarrow negative logic

- ▷ NAND and NOR gates are called universal logic gates.

Simplification of Logic Functions

- * Objectives of minimization are decreasing
 - the size of circuit
 - power consumption
 - the delay (increasing the speed)
 - the cost
- * prime implicant: implicants that cannot be simplified any further.
- * the set of all prime implicants: all possible prime implicants, groups of all 1's.
- * sufficient base (sufficient covering sum): the minimum set of prime implicants that covers all 1's.
- * distinguished points: 1's that may be covered only by a single prime implicant.
- * essential prime implicant: prime implicant that covers a distinguished point. (must be included in the minimal covering sum)

Simplification

- 1) find the set of all prime implicants
- 2) select a subset with minimum cost that covers the function

* Prime Implicant Chart

- 1) assign simple symbols to each prime implicants (A,B,C)
- 2) calculate the cost of each prime implicant with the criteria
- 3) symbols → rows → put * on intersection
1 points → columns } make a matrix
costs → last column }
- 4) simplificate the prime implicant chart
 - 1) single * in a column → distinguished point
the row of this point and columns that are covered are removed.
 - 2) if row i covers row j and the cost at i is smaller or equal to the cost at j , then row j is removed.
 - 3) if column a covers column b , then column a is removed.



when we select the don't care values (Ξ)

→ we can take the values as 1 in Karnaugh Map.

→ we can take the values as 0 in the prime implicant chart.

0

Tabular (Quine-McCluskey) method

- * we put true points and don't cares in a truth table.
if only one variable changes between two minterm, we merge them.
we do that until there is no minterms that can be merged.
therefore, prime implicants are the minterms that can't merged with the other ones.

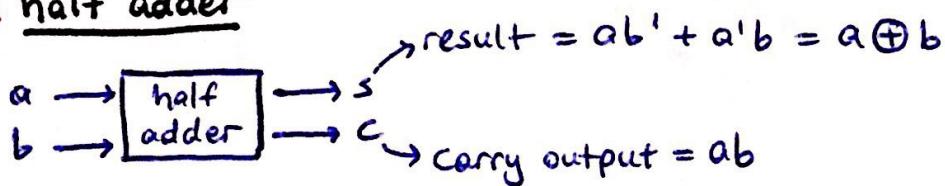
Simplification in the Pos form

- * we use the same process with SoP simplification but there is only one difference : we group zeros
end of dc 04 —————— end of dc04

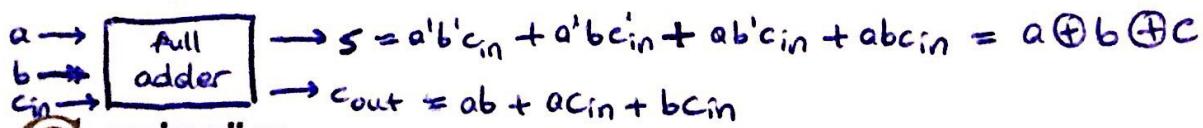
- * adders, multiplexers, decoders are common structures that used building blocks in larger systems. they are manufactured and sold as integrated circuits (ICs).

- * generations of ICs according to integration scale → 5.1

* half adder

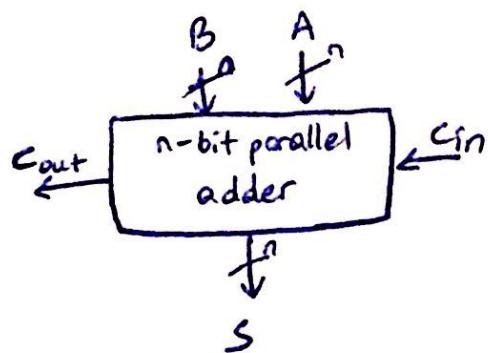


* full adder



* n-bit binary parallel adder

→ it adds two n-bit binary number

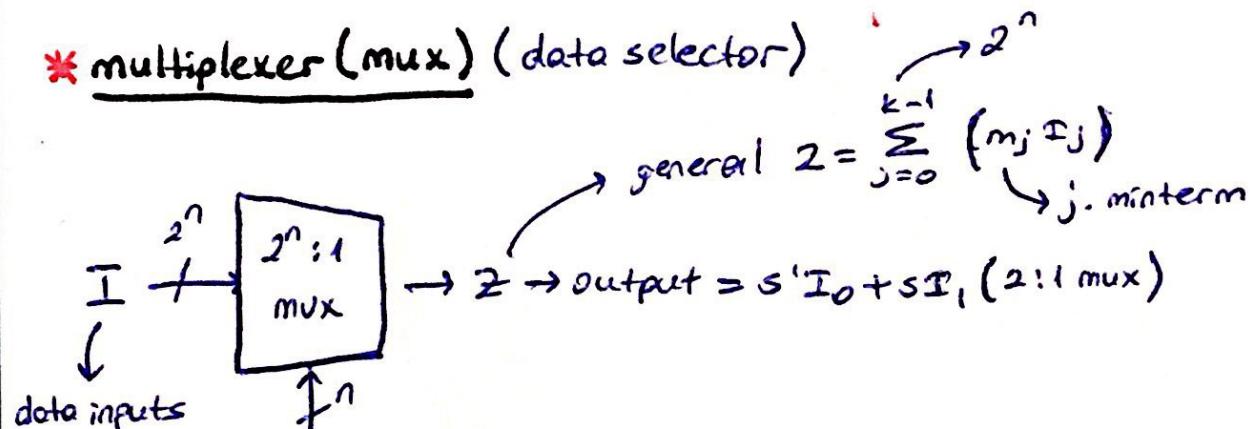


Ex

$$\begin{aligned}
 1. \text{num} &= 0110 \rightarrow 6 &> 18 \\
 2. \text{num} &= 1100 \rightarrow 12 &> 18 \\
 \text{result} &= 0010 \rightarrow 2 &> 18 \\
 \text{carry} &= 1 \rightarrow 16
 \end{aligned}$$

! a 4-bit subtraction circuit = $S = \underline{A - B} = A + 2\text{'s com}(B) = \underline{A + (\bar{B} + 1)}$

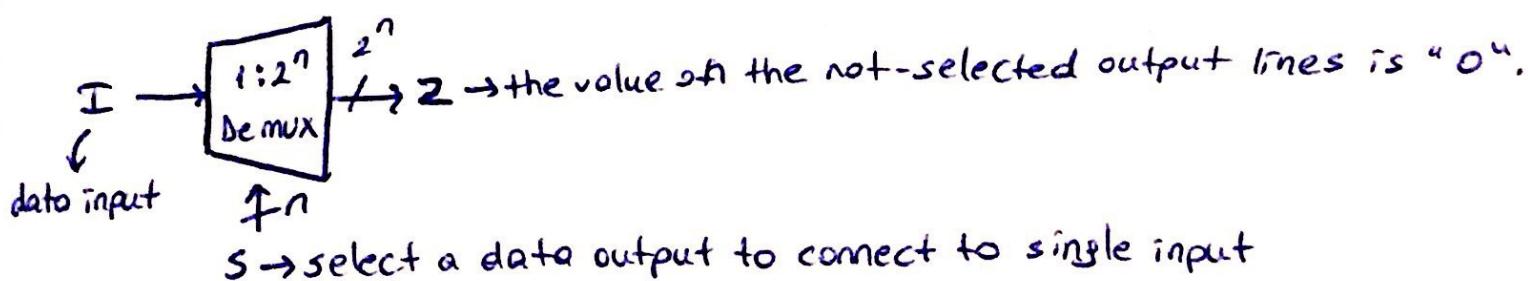
* multiplexer (mux) (data selector)



$S \rightarrow$ control inputs → select a data to connect to the output

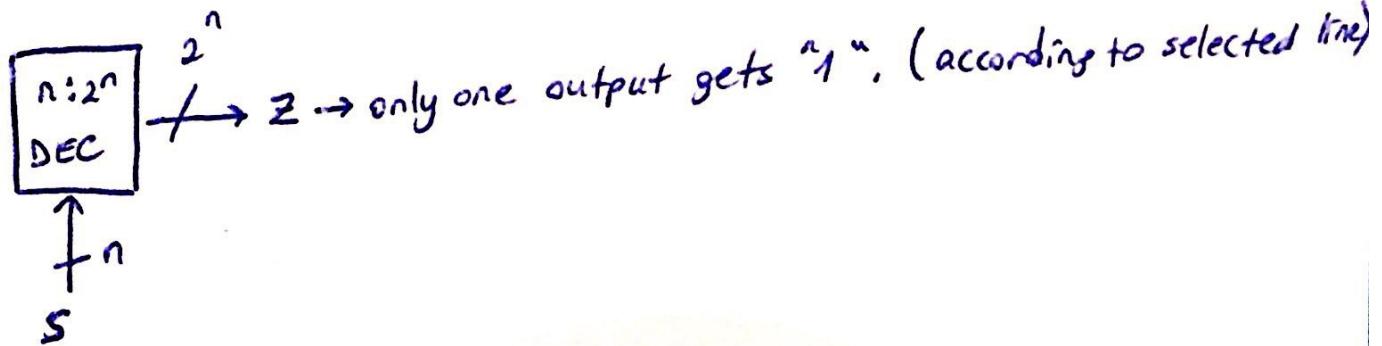
→ logic circuits designs with multiplexers → 5.13, 5.14, 5.15

* demultiplexer



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* decoder (minterm generator)



- decoders may have "enable" (EN) input.
if EN input = 1 \Rightarrow normal
 $= 0 \Rightarrow$ all outputs = "0"

Programmable Logic Device (PLD)

- * some PLD's also include memory units (flip-flops)
- * there are different kinds of PLD's:
 - programmable logic array (PLA)
 - programmable array logic (PAL)
 - generic array logic (GAL)
 - complex PLD (CPLD)
 - field-programmable gate array (FPGA)

} - early versions
- bipolar transistors
- one-time programmable (OTP)

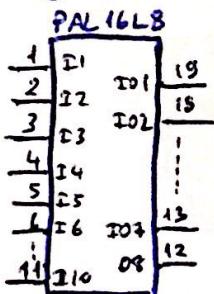
} - todays devices
- Cmos transistors, Memory
- can be reprogrammed
- * there are various Hardware Description Languages (HDL) and programming devices to program PLDs.
 - PALASM
 - ABEL
 - Verilog
 - VHDL (very high speed integrated circuits HDL)

Programmable Logic Array - PLA

- * AND (product) units → in the input layer
- OR (sum) units → in the output layer
- ! Both AND, OR arrays are flexible programmable.
- * they called as "n × m PLA with p products"
 - input
 - output
 - AND gates
- * simple representation in 5.29 (X's)

Programmable Array Logic - PAL

- * inputs of AND gates can be flexible programmed as in PLA's. But inputs of OR gates are fixed. However, PAL's can be easily programmed.



Generic Array Logic (GAL)

- * its logical properties are similar to PAL,
- * it is made of CMOS transistors → can be erased and programmed many.

Complex PLD (CPLD)

- * IC that contains several PLDs (macro cell)
- * internal structures of macro cells and connections between them can be programmed.

Field - Programmable Gate Array (FPGA)

- * they contain many logical blocks and interconnections between these blocks.
- * can be used implement complex digital circuits (special purpose microprocessors)
- * more flexible and can implement more circuits than CPLDs but their delay cost is higher.