



# **DESCRIPTION**

The MP6522 is an H-bridge motor driver that operates from a supply voltage of up to 35V and delivers a peak motor current of up to 3.2A. The MP6522 is typically used to drive DC brush motors.

An internal current-sensing circuit provides an output with a voltage proportional to the load current. Additionally, cycle-by-cycle current regulation and limiting is provided. These features do not require the use of a low-ohm shunt resistor.

Full protection features include over-current protection (OCP), input over-voltage protection (OVP), under-voltage lockout (UVLO), and thermal shutdown.

The MP6522 is available in a 24-pin QFN (5mmx5mm) package with an exposed thermal pad.

#### **FEATURES**

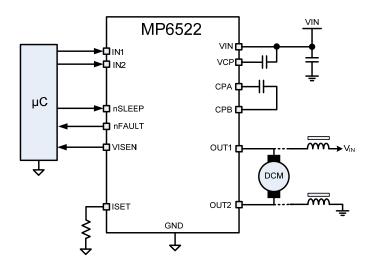
- Wide 5.4V to 35V Input Voltage Range
- 3.2A Maximum Output Current
- Internal, Full, H-Bridge Driver
- Cycle-by-Cycle Current Regulation and Limiting
- Low On Resistance (HS:  $300m\Omega$ ; LS:  $300m\Omega$ )
- Simple, Versatile Logic Interfaces
- 3.3V and 5V Compatible Logic Inputs
- Over-Current Protection (OCP)
- Over-Voltage Protection (OVP)
- Thermal Shutdown
- Under-Voltage Lockout (UVLO)
- Fault Indication Output
- Thermally Enhanced Surface-Mount Package
- Available in a QFN-24 (5mmx5mm) Package

#### **APPLICATIONS**

- Solenoid Drivers
- DC Brush Motor Drives

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#### TYPICAL APPLICATION





# ORDERING INFORMATION

Part Number*	Package	Top Marking
MP6522GU*	QFN-24 (5mmx5mm)	See Below

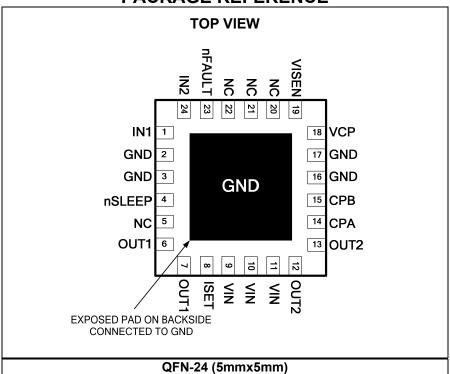
<sup>\*</sup> For Tape & Reel, add suffix -Z (e.g. MP6522GU-Z)

# **TOP MARKING**

MPSYYWW MP6522 LLLLLL

MPS: MPS prefix YY: Year code WW: Week code MP6522: Part number LLLLLL: Lot number

# PACKAGE REFERENCE





<b>ABSOLUTE MAXIMUM</b>	I RATINGS (1)
Supply voltage (VIN)	
OUTx voltage (V <sub>OUT1/2</sub> )	
VCP, CPB	
ISET	
All other pins to GND	
ESD rating (HBD)	
Continuous power dissipation	
	3.5W
Storage temperature	
Junction temperature	
Lead temperature (solder)	+260°C
Recommended Operating	Conditions (3)
Supply voltage (VIN)	
Peak output current (I <sub>OUT</sub> )	±3.2A
Load current (I <sub>VISEN</sub> )	±2mA
Operating junction temp. (T <sub>J</sub> )	40°C to +125°C

#### NOTES:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub> (MAX), the junction-to-ambient thermal resistance θ<sub>JA</sub>, and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub> (MAX) = (T<sub>J</sub> (MAX)-T<sub>A</sub>)/θ<sub>JA</sub>. Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



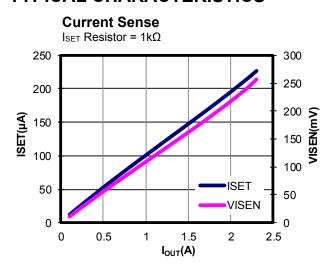
# **ELECTRICAL CHARACTERISTICS**

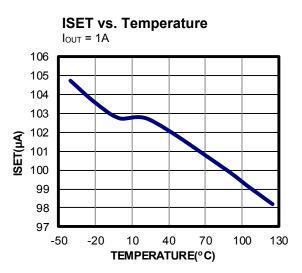
VIN = 24V,  $T_A$  = +25°C, unless otherwise noted.

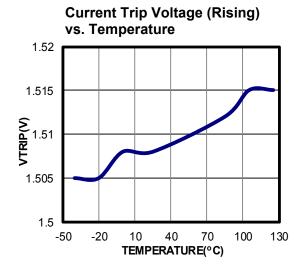
Parameter	Symbol	Condition	Min	Тур	Max	Units
Power Supply						
Input supply voltage	V <sub>IN</sub>		5.4	24	35	V
Quiescent current	IQ	VIN = 24V, nSLEEP = 1, no load current		1.5	2	mA
	I <sub>SLEEP</sub>	VIN = 24V, nSLEEP = 0			1	μA
Charge pump frequency	f <sub>CP</sub>			660		kHz
Internal MOSFETs				l .	L.	I.
	R	VIN = 24V, I <sub>OUT</sub> = 1A, T <sub>J</sub> = 25°C		0.3	0.45	Ω
Output on resistance	R <sub>HS</sub>	VIN = 24V, I <sub>OUT</sub> = 1A, T <sub>J</sub> = 85°C		0.35		Ω
Output on resistance	R <sub>LS</sub>	VIN = 24V, I <sub>OUT</sub> = 1A, T <sub>J</sub> = 25°C		0.3	0.45	Ω
		VIN = 24V, I <sub>OUT</sub> = 1A, T <sub>J</sub> = 85°C		0.35		Ω
Body diode forward voltage	$V_{F}$	I <sub>OUT</sub> = 1.5A			1.1	V
Control Logic						
Input logic low threshold	V <sub>IL</sub>				0.8	V
Input logic high threshold	V <sub>IH</sub>		2			V
Logic input current	I <sub>IN(H)</sub>	V <sub>IH</sub> = 5V V <sub>IL</sub> = 0.8V	-12 -2		12 2	μA μA
Internal pull-down resistance	R <sub>PD</sub>	VIE 0.0V		530		kΩ
nFault Output (Open-Drain Ou						1 1 1 1 1 1 1
Output low voltage	VoL	I <sub>O</sub> = 5mA			0.5	V
Output high leakage current	Іон	V <sub>O</sub> = 3.3V			1	μA
Protection Circuits	-	-	I	I		
UVLO rising threshold	V <sub>IN_RISE</sub>			5.1	5.3	V
UVLO hysteresis	V <sub>HYS</sub>			310	0.0	mV
Input OVP threshold	Vove		36	38	40	V
Input OVP hysteresis	$\Delta V_{OVP}$			2000		mV
•	I <sub>OCP1</sub>	Sinking	3.2	4	5.2	Α
Over-current trip level	I <sub>OCP2</sub>	Sourcing	3.2	4	5.2	Α
Over-current deglitch time	tocpd			500		ns
Over-current retry time	tocp			0.9		ms
Thermal shutdown	T <sub>TSD</sub>			165		°C
Thermal shutdown hysteresis	$\DeltaT_TSD$			30		°C
Current Control						
Off time	t <sub>ITRIP</sub>	After ITRIP		17		μs
ISET current	liset		90	100	110	μA/A
Current trip voltage (rising)	V <sub>ITRIP-R</sub>	At VISEN	1.44	1.5	1.56	· V
Current trip voltage (falling)	V <sub>ITRIP-F</sub>	At VISEN	1.15	1.2	1.25	V
VISEN Output		•	•			
Output voltage accuracy	$\Delta V_{VISEN}$	V <sub>ISET</sub> > 0.5V	-5		5	%

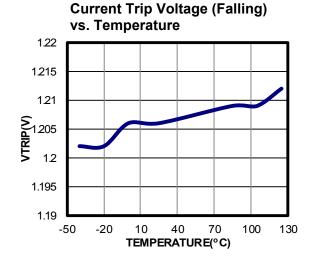


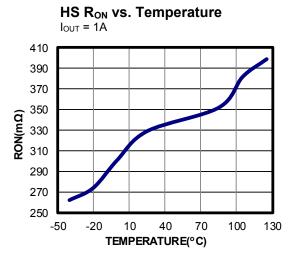
# TYPICAL CHARACTERISTICS

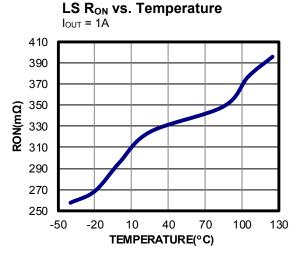








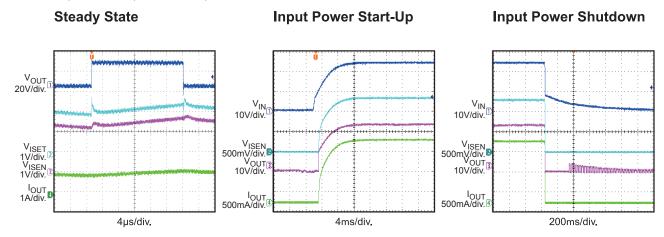


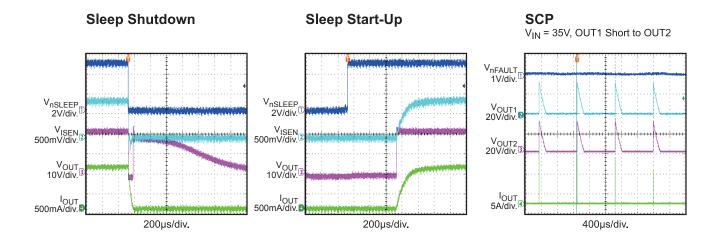


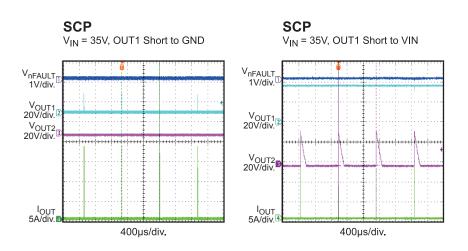


# TYPICAL PERFORMANCE CHARACTERISTICS

VIN = 24V, I<sub>OUT</sub> = 1A, T<sub>A</sub> = 25°C, unless otherwise noted.







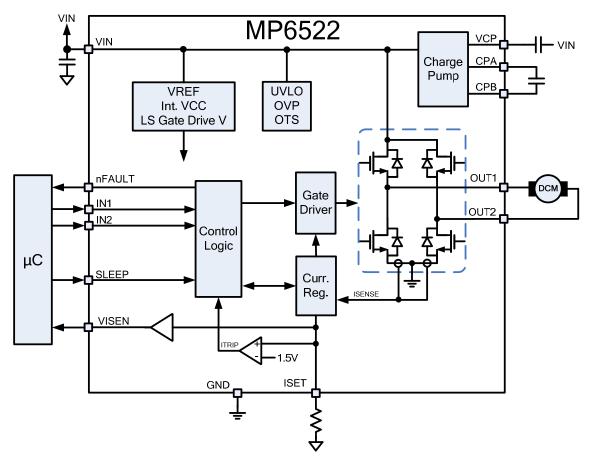


# **PIN FUNCTIONS**

Pin#	Name	Description		
1	IN1	Control inpute IN14 and IN12 have internal null down registers		
24	IN2	Control inputs. IN1 and IN2 have internal pull-down resistors.		
2, 3, 16, 17	GND	System ground connection.		
4	nSLEEP	<b>Sleep mode input.</b> Set nSLEEP to logic low to enter low-power sleep mode. nSLEEP has an internal pull-down resistor.		
5, 20, 21, 22	NC	No connection.		
6, 7	OUT1	Output terminal 1.		
8	ISET	<b>Current programming resistor.</b> Connect a resistor from ISET to ground to set the current limit and VISEN output voltage.		
9, 10, 11	VIN	<b>Input supply voltage.</b> Decouple VIN to GND with a ceramic capacitor ≥100nF to GND.		
12, 13	OUT2	Output terminal 2.		
14	CPA	Charge pump flying capacitor. Connect a 100nF ceramic capacitor between CPA		
15	CPB	and CPB.		
18	VCP	Charge pump output. Connect a 100nF capacitor from VCP to VIN.		
19	VISEN	Current sense output voltage.		
23	nFAULT	<b>Fault indication.</b> nFAULT is an open-drain output. nFAULT is at logic low when in a fault condition (OCP, OTP, OVP).		
EP	GND	Exposed pad. The exposed pad must be connected to ground.		



# **BLOCK DIAGRAM**



**Figure 1: Functional Block Diagram** 



#### **OPERATION**

The MP6522 is an H-bridge motor driver that integrates four N-channel power MOSFETs with 3.2A of peak current capability. The MP6522 operates over a wide 5.4V to 35V input voltage range and is designed to drive DC brush motors, solenoids, and other loads.

#### **Current Sensing**

The current flowing in the two low-side MOSFETs is sensed with an internal currentsensing circuit. A voltage proportional to the output current is sourced on VISEN.

The VISEN output voltage scaling is set by a resistor connected between ISET and ground. For 1A of output current,  $100\mu A$  of current is sourced into the resistor connected to ISET. For example, if a  $10k\Omega$  resistor is connected between ISET and ground, the output voltage on VISEN is 1V/A of output current.

The current is sensed when one of the low-side MOSFETs is turned on, including during slow decay (brake) mode.

The load current applied to VISEN should be kept below 2mA with no more than 500pF of capacitance.

#### **Current Limit and Regulation**

The current in the outputs is limited using constant-off-time pulse-width modulation (PWM) control circuitry.

In PWM, a diagonal pair of MOSFETs is turned on initially and drives current through the load. The current increases in the load, which is sensed by the internal current-sense circuit. If the load current reaches the current-trip threshold, the H-bridge switches to slow decay mode with the two low-side MOSFETs turned on.

After a fixed off-time ( $t_{\rm ITRIP}$ ), if the load current falls at least 20% below the current-limit threshold, the MOSFETs are re-enabled and the cycle repeats. If the current is still higher than this level, the off time is extended until the current falls to 20% below the current limit threshold.

The current-limit threshold is reached when VISEN reaches 1.5V. For example, with a  $10k\Omega$  resistor from ISET to ground, the VISEN voltage is 1V/A of output current. Therefore, when the current reaches 1.5A, the VISET voltage reaches 1.5V, and a current trip occurs.

During current regulation, nFAULT is not activated.

# **Blanking Time**

There is often a current spike during start-up due to the body diode's reverse-recovery current or the shunt capacitance of the load. This current spike requires filtering to prevent it from erroneously shutting down the high-side MOSFET. An internal fixed blanking time ( $t_{\text{OCPD}}$ ) blanks the output of the current-sense comparator when the outputs are switched. This blanking time also sets the minimum on time for the high-side MOSFET.

#### Input Logic

For the MP6522, control of the outputs is accomplished through IN1 and IN2 (see Table 1).

**Table 1: Output Control Pins** 

IN1	IN2	OUT1	OUT2	Function
0	0	Ζ	Ζ	Coast
0	1	L	Н	Forward
1	0	Н	L	Reverse
1	1	L	L	Brake (low)

### **nSLEEP Operation**

Driving nSLEEP low puts the device into a low-power sleep state. In this state, all internal circuits including the gate drive charge pump are disabled, and the H-bridge outputs are turned off. All inputs are ignored when nSLEEP is active low. When waking up from sleep mode, approximately 1ms of time must pass before the outputs can operate.

#### Faul

The MP6522 provides an nFAULT pin, which is driven active low if any protection circuits (over-current, over-voltage, over-temperature) are activated. nFAULT is not driven low when a current-limit trip occurs. nFAULT is an open-drain output and requires an external pull-up resistor. When the fault condition is removed, nFAULT is pulled inactive high by the pull-up resistor.



# **Over-Current Protection (OCP)**

The over-current protection (OCP) circuit protects the device from damaging high currents at the outputs. If the current through a MOSFET exceeds the over-current threshold for longer than the over-current deglitch time, all MOSFETs in the H-bridge are disabled, and nFAULT is driven low. The driver remains disabled for the over-current retry time (t<sub>OCP</sub>), at the driver re-enabled which point is automatically. Over-current conditions are sensed on both high- and low-side devices (i.e.: a short to ground, supply, or across the motor winding results in an over-current shutdown. Note that OCP does not use the current-sense circuitry used for PWM current control and is independent of the ISET resistor value.

#### **Over-Voltage Protection (OVP)**

If the input voltage applied to VIN is higher than the over-voltage protection (OVP) threshold, the H-bridge output is disabled, and nFAULT is driven low. This protection is released when VIN drops to a safe level.

#### Input **Under-Voltage** Lockout (UVLO) **Protection**

If at any time the voltage on VIN falls below the under-voltage lockout (UVLO) threshold, all circuitry in the device is disabled, and the internal logic is reset. Operation resumes when VIN rises above the UVLO threshold.

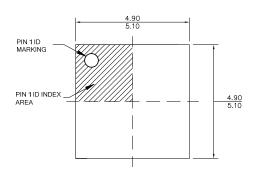
#### Thermal Shutdown

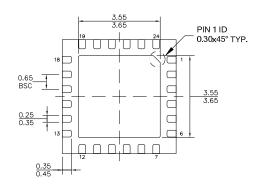
If the die temperature exceeds the safe limits, all MOSFETs in the H-bridge are disabled, and nFAULT is driven low. Once the temperature has fallen to a safe level, operation resumes automatically.



# **PACKAGE INFORMATION**

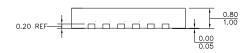
# QFN-24 (5mmx5mm)



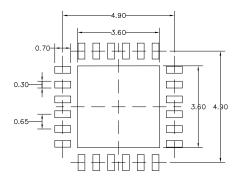


**TOP VIEW** 

**BOTTOM VIEW** 



SIDE VIEW



### RECOMMENDED LAND PATTERN

# NOTE:

1) ALL DIMENSIONS ARE IN MILLIMETERS.
2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX.
4) DRAWING CONFIRMS TO JEDEC MO-220.

5) DRAWING IS NOT TO SCALE.

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