ECE 3640 - Discrete-Time Signals and Systems Final Exam - Spring 2022

Name: Zach Wilcox Due: Friday, 29 April 2022 at 12 AM (midnight).

Instructions:

- 1. You may use: your text book, your homework, your notes, or resources available from the course web page.
- 2. Do not talk to anyone about this exam or get help from any source on this exam.
- 3. Write your answers in the spaces provided.
- 4. By signing in the name box above, you verify that you have complied with these instructions.
- 1. The continuous-time signal $x(t) = \exp(j2\pi F_0 t)$, where $F_0 = 29655$ Hz, is sampled at a rate of $F_s = 900$ samples per second. What is the frequency $f_0 \in [-\frac{1}{2}, +\frac{1}{2})$ (in cycles/sample) of the sampled sequence x[n]?

 $f_0 = -0.05$ cycles/sample

2. A continuous-time signal x(t) is reconstructed from a discrete-time sequence $x[n] = \cos(2\pi f_0 n)$, where $f_0 = 0.0833$ cycles/sample, assuming a reconstruction rate of $F_s = 72,000$ sample/second. What is the frequency F_0 (in Hz) of the reconstructed signal?

 $F_0 = 5.997.6$

3. What is the minimum sampling rate F_s (in samples/second) for signal $x(t) = \frac{\sin(21352t)}{t}$ if aliasing is to be avoided?

 $F_s = 6797$ samples/second

4. A signal x(t) with highest frequency 7832 Hz is going to be sampled at a rate of 8820 samples/second. To avoid aliasing, the signal x(t) is filtered prior to sampling through an ideal low pass filter having frequency response

$$H(F) = \begin{cases} 1, & |F| \le F_0, \\ 0, & |F| > F_0. \end{cases}$$

What is the largest frequency F_0 (in Hz) that will prevent aliasing in the sampled signal?

 $F_0 = Hz$

5. The sample rate of a real discrete-time signal x[n] with highest frequency 0.4 cycles/samples to be adjusted by a rational factor by processing as shown below.

$$x[n] \xrightarrow{} \uparrow U \xrightarrow{} \texttt{LPF} \xrightarrow{} \downarrow D \xrightarrow{} y[n]$$

What upsampling and downsampling factors U and D will produce a sampled signal that occupies the full bandwidth from -1/2 to +1/2? Use the smallest possible values of U and D.

U =

D = 5

$$F_0 = \frac{29655 \text{ cycles}}{1 \text{ Sec}}$$
 $\frac{F_0}{F_5} = 72.95 \frac{\text{cycles}}{\text{Sample}}$
 $F_5 = \frac{900 \text{ Samples}}{1 \text{ sec}}$ $-33 = -0.05$

P2:

P3:

$$\frac{21352}{2\pi} = 3399.29$$

$$\times 2$$

$$6796.55$$

$$\vdots \quad f_3 \ge 6796.55$$

$$.4\frac{0}{0} = .5$$

 $\frac{0}{0} = 1.25 = \frac{5}{4}$

6. If an ideal low pass filter is used in the system below, what is the cut-off frequency $f_{\text{cut-off}}$ (in cycles/sample) for the filter?



 $f_{\text{cut-off}} = 0.0714$

cycles/sample

7. If the signal x[n] is sampled at $F_{s,\text{input}} = 8000$ sample/second, what is the sample rate $F_{s,\text{output}}$ (in samples/second) of y[n]?

$$x[n] \longrightarrow \uparrow 7 \longrightarrow \text{LPF} \longrightarrow \downarrow 3 \longrightarrow y[n]$$

$$F_{s,\text{output}} = 18666.667$$

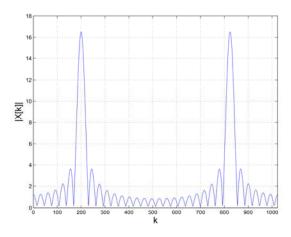
samples/second

8. If the signal $x[n] = \cos(2\pi 0.38n)$ is processed as shown below, then $y[n] = \cos(2\pi f_0 n)$. What is f_0 (in cycles/sample)?

$$x[n] \longrightarrow \uparrow 7 \longrightarrow \text{LPF} \longrightarrow \downarrow 3 \longrightarrow y[n]$$

cycles/sample

9. Aliasing is avoided when a real continuous-time signal x(t) is sampled at the rate $F_s = 36000$ samples/second producing the discrete-time signal x[n]. A rectangular window of length N=32 samples is applied to the sampled sequence x[n] and the result is zero padded to length 1024. A 1024-point FFT is computed on the zero padded sequence and the magnitude |X[k]| of the result is plotted in the figure below.



What is the frequency F_0 (in Hz) of the input signal x(t)?

$$F_0 = 180$$

 $_{\mathrm{Hz}}$

$$\begin{cases}
f = min\left(\frac{1}{27}, \frac{1}{23}\right) \\
= \frac{1}{14} = .0714
\end{cases}$$

10. The signal $x(t) = \exp(j2\pi 35000t) + \exp(j2\pi 37000t)$ is sampled at a rate $F_s = 9000$ samples/second and an FFT is applied to N = 16384 samples of the signal to compute the DFT X[k]. In what bins k_1 and k_2 of the DFT do we expect to see peaks?

$$k_1 = 63715.5$$

$$k_2 = 67356$$

11. In Matlab, N-point circular convolution may be computed efficiently by ifft(fft(x,N).*fft(h,N)). Suppose this method is used to perform linear convolution of the signal $x[n], n = 0, 1, \cdots, 7999$ with the filter impulse response $h[n], n = 0, 1, \cdots, 80$. What is the smallest value of N that causes circular convolution to compute linear convolution. (Hint, this is a zero padding question.)

$$N = \mathcal{GO}_{\mathcal{O}} \mathcal{GO}$$

12. A finite length L=100 sequence x[n] has DTFT given by $X(f)=\sqrt{1+(f/0.1)^4}$ for $-0.5 \le f < 0.5$. A length N=800 point DFT X[k] is computed on x[n]. What are X[197] and X[711]?

$$X[197] = 6.146$$

$$X[711] = 78.994$$

- 13. A 1024-point DFT is used to analyze a signal x[n] after applying a window w[n] of length $N \leq 1024$. Which of the following windows is most likely able to resolve two equal amplitude sinusoids with nearly equal frequencies? (circle one)
 - (a) 256-point rectangular window
 - (b) 512-point rectangular window
 - (c) 1024-point rectangular window
 - (d) 256-point Hann window
 - (e) 512-point Hann window
 - (f) 1024-point Hann window
- 14. An 1024-point DFT is used to analyze a signal x[n] after applying a window w[n] of length $N \leq 1024$. Which of the following windows is most likely able to resolve a small amplitude sinusoid in the presence of a large amplitude sinusoid? (circle one)
 - (a) 256-point rectangular window
 - (b) 512-point rectangular window
 - (c) 1024-point rectangular window
 - (d) 256-point Hann window
 - (e) 512-point Hann window
 - (f) 1024-point Hann window
- 15. The everlasting sinusoid $x[n] = 7\cos(2\pi(1/3)n + 9.7\pi)$ is input to a system with transfer function $H(z) = 1/(1 + 0.7z^{-1} + 0.9z^{-2})$. What is the output y[n]?

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 + .7z' + qz^2} = 2 = e^{j\frac{2\pi}{3}}$$

$$35000\left(\frac{16384}{9000}\right) = 63715.5$$
 $37000\left(\frac{16384}{9000}\right) = 67356$

$$\begin{array}{c}
X\left(\frac{K}{N}\right) \\
-X\left(\frac{197}{500}\right) = \\
X\left(\frac{711}{400}\right) = \\
\end{array}$$

 $y[n] = A\cos(2\pi f_0 n + \phi)$

A = 3.779

 $f_0 = \frac{1}{3}$

 $\phi = -40.99$

16. The everlasting complex exponential $x[n] = \exp(j2\pi(1/5)n)$ is input to an LTI system described by the difference equation $y[n] = 0.9 \exp(j2\pi/5)y[n-1] + x[n]$. What is the output y[n]?

 $y[n] = A \exp(j[2\pi f_0 n + \phi])$

A = 10

 $f_0 = \frac{1}{5}$

 $\phi = 0$

17. Let $H(z) = 1/(1-z^{-1}-z^{-2})$. Let h[n] be the impulse response of a causal realization of this system. What is h[8]?

h[8] = 34

18. Let $H(z) = 1/(1-z^{-1}-z^{-2})$. Let h[n] be the impulse response of a stable realization of this system. What is h[8]?

h[8] = 34

19. Let $H(z) = 1/(1-z^{-1}-z^{-2})$. Let h[n] be the impulse response of an anti-causal realization of this system. What is h[-8]?

h[-8] = 13

20. Let

$$H(z) = \frac{5 + 3z^{-1} + 2z^{-2} + 5z^{-3}}{1 + 6z^{-1} + 3z^{-2} + 2z^{-3} + 5z^{-4} + z^{-5}}.$$
 (1)

How many different regions of convergence does this H(z) have?

Number of regions = 5

21. For H(z) defined in the previous problem, how many of the regions of convergence lead to a stable h[n].

Number of regions = 2

$$H(z) = \frac{1}{1 - .9e^{3\frac{2\pi}{3}}z^{-1}}$$
 $\left|H\left(\frac{2\pi}{3}\right)\right|^{2} = \left|\frac{1}{1 - .9(1)}\right|^{2} = \frac{1}{.1} = 0$
 $L(H\left(\frac{2\pi}{3}\right)) = 0$

22. Find the impulse response h[n] corresponding to the causal system described by the difference equation

$$y[n] = 0.9863y[n-1] - 0.4494y[n-2] + 0.1093x[n] + 0.1942x[n-1] + 0.1093x[n-2].$$

Express your answer in the form $Ap^n \cos(\omega n + \varphi)u[n]$.

$$h[n] = [003 + .188^{n} \cos(n - \frac{3}{4}) M[n]$$

23. For the system in problem 22, find the output y[n] when the input is $x[n] = 0.8^n u[n]$. Assume the system is initially at rest. Where possible combine complex conjugate pole pairs to form $\cos(\omega n + \varphi)u[n]$ terms.

$$y[n] = \left[-9^{n} + .56(.67)^{n} (05(\frac{3n}{4} - .00)) \right] u[n]$$

- 24. Using the Kaiser window method, design a linear phase low pass filter that meets the following specification.
 - Pass band: $0 \le f \le f_p = 0.2$ cycles/sample, maximum ripple $A_p = 0.05$ dB
 - Stop band: $f_s = 0.25 \le f \le 0.5$ cycles/sample, maximum ripple $A_s = 55$ dB
 - (a) What is the minimum filter length L that meets the specification?
 - (b) What are the filter order M = L 1 and group delay g = M/2?
 - (c) What type (I, II, III, IV) of filter did you use?

B= 5.1

M= 66 L=67

M=94 9= 47

L= 95 type II

Attach a magnitude frequency response plot of the filter you designed. Include a "zoom" of the entire pass band to show that your design meets the specification in that band. Do the same for the stop band. Plot the impulse response.

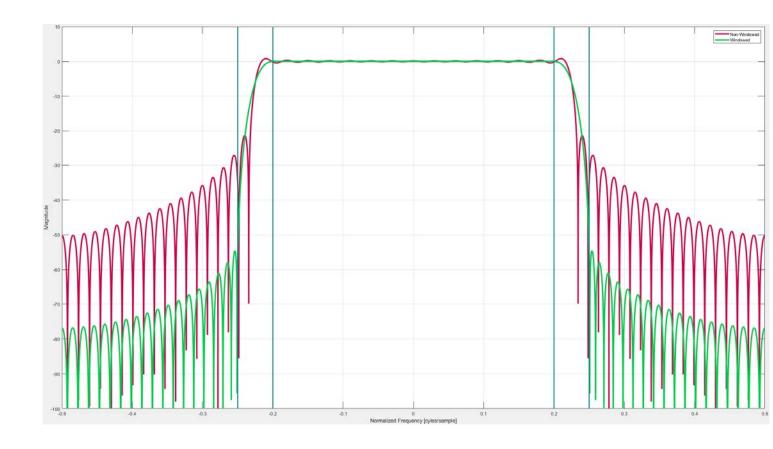
- 25. Using the Kaiser window method, design a linear phase band stop filter that meets the following specification.
 - Stop band: $f_{s,1}=0.15 \le f \le f_{s,2}=0.3$ cycles/sample, maximum ripple $A_p=0.05$ dB Pass band 2: $f_{p,2}=0.4 < f < 0.5$ cycles/sample, maximum ripple $A_s=75$ dB

• Pass band 2:
$$f_{p,2}=0.4 \leq f \leq 0.5$$
cycles/sample, maximum ripple $A_p=0.05~\mathrm{dB}$

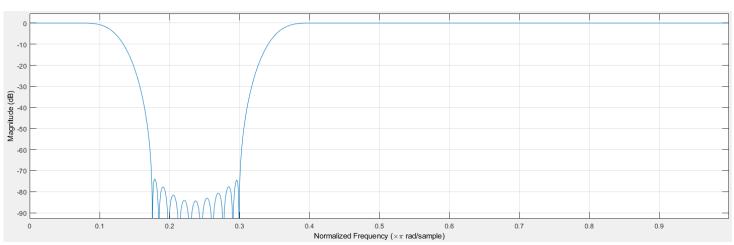
- (a) What is the minimum filter length L that meets the specification?
- (b) What are the filter order M = L 1 and group delay g = M/2?
- (c) What type (I, II, III, IV) of filter did you use?

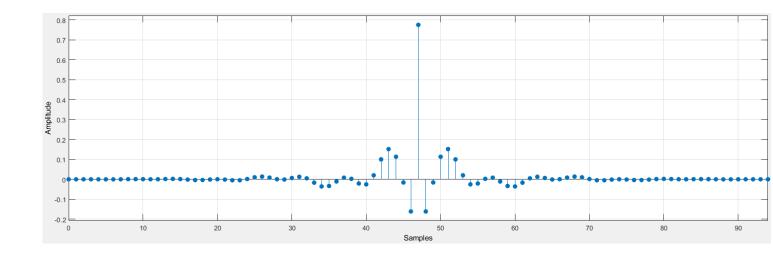
Attach a magnitude frequency response plot of the filter you designed. Include a "zoom" of the entire pass band to show that your design meets the specification in that band. Do the same for the stop band. Plot the impulse response.

Q24



Q25





$$H(z) = \frac{.1043 + .1942z^{2} + .1943z^{2}}{1 - .4463z^{2} + .4444z^{2}}$$

$$= .1043 + \frac{.151^{2} \cdot .2303z}{2 - (.44 + .45z)} + \frac{.151 + .2303z}{2 - (.44 - .45z)}$$

$$= .1043 + \frac{.27e^{j.44}}{2 - .67e^{j.75}} + \frac{.27e^{j.44}}{2 - .67e^{-j.75}}$$

$$= .10435[n] + .188^{1} \cdot cos(n - \frac{3}{4}) \cdot u[n]$$

$$\frac{.10432^{2} + .14422^{2} + .1043}{2^{2} - .4632^{2} + .14444} \cdot \frac{2}{2 - .4}$$

$$\frac{.10432^{3} + .14422^{2} + .10432}{2^{3} - 1.772^{2} + 1.242 - .36}$$

$$\frac{.9^{4} + .28e^{j.49} (.67e^{j.71})^{4} + .29e^{j.49} (.67e^{-j.73})^{4}}{(.67e^{-j.73})^{4} + .56(.67)^{4} (.65(\frac{34}{4} - .44)) u[4]}$$