Lecture 2 EE 421 / C\$ 425 Digital System Design

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Topics in Lecture 2

- Background to Digital Systems
- Useful number systems, conversion
- Introducing RTL Design
- The Language and Abstraction of Digital Systems
- Review Boolean Algebra
- Review K-Maps, 5 and 6 variables ??
- Logic Minimization and Synthesis
- Introducing WinLogiLab tool

Review material, use reference books:

- Digital Systems Principles and Applications, 12th edition, Widmer, Moss, Tocci
- 2. Digital Design, Mano and Ciletti



Useful Number Systems

- Initial electro-mechanical computers were based on analog and decimal numbers (base 10). The computers were difficult to scale, complex, and expensive.
- Binary number system (base 2) along with Boolean Algebra was found to be the most efficient (area ← speed ← power Δ) mechanism for computer implementation in microelectronics technology.
- Switching square waveforms in electronic circuits directly map to '1' and '0' in binary number system.
- In computer systems, Hex numbers (base 16) and 2's Complement numbers are used to represent information in a compact way.



Review of Useful Number Systems

Decimal to Binary

Eg. Convert 37 (base 10) to a binary number (base 2)

$$37 \div 2 = 18.5$$
 (whole number is 18, remainder is 0.5, so put a 1 here)
→ 1 (LSB)

 $18 \div 2 = 9$ (whole number is 9, remainder is 0 so put a zero here)
→ 0

 $9 \div 2 = 4.5$ (whole number is 4, remainder is 0.5, put a 1 here)
→ 1

 $4 \div 2 = 2$ (whole number is 2, remainder is 0, put a 0 here)
→ 0

 $2 \div 2 = 1$ (whole number is 1, remainder is 0, put a 0 here)
→ 0

 $1 \div 2 = 0.5$ (whole number is 0, remainder is 0.5, put a 1 here)
→ 1 (MSB)

Write MSB to LSB \rightarrow 100101

Hence $(37)_{10} = (100101)_2$

Binary to Decimal conversion

- Convert Binary number (1001101)₂ to Decimal number system
- Each bit position has a proportional weight in the power of 2

1	0	0	1	1	0	1
2 ⁶	2 ⁵	2 ⁴	2 ³	2 ²	2 ¹	2 ⁰

MSB

Add up all the numbers with proportional weights

$$= 1 \times 2^{6} + 0 \times 2^{5} + 0 \times 2^{4} + 1 \times 2^{3} + 1 \times 2^{2} + 0 \times 2^{1} + 1 \times 2^{0}$$

$$= 64 + 8 + 4 + 1 = (77)_{10}$$



Signed Binary Numbers – 2's Complement

• 2's complement of -(0110010)₂ Keep one or more extra '0' bit in MSB

Step 1: Take 1's complement \rightarrow (1001101)

Step 2: Add '+1' to the 1's Complement

10011101

· 1

1's Complement means invert all bits

 $(100111110)_2$

negative numbers will always have MSB = 1



Hexadecimal	Decimal	Binary
0	0	0000
1	1	0001
2	2	0010
3	3	0011
4	4	0100
5	5	0101
6	6	0110
7	7	0111
8	8	1000
9	9	1001
Α	10	1010
В	11	1011
С	12	1100
D	13	1101
E	14	1110
F	15	1111

Hexadecimal Number System (group of 4 bits)

Binary numbers up to 4 bits can be represented as single Hex digit 7-Segment Display can show Hex Digits



Converting from Hex Number System

Hex to Decimal Conversion:

$$(356)_{16} = 3x16^2 + 5x16^1 + 6x16^0$$

$$= 768 + 80 + 6$$

$$=(854)_{10}$$

Hex to Decimal Conversion:

$$(2AF)_{16} = 2x16^2 + 10x16^1 + 15x16^0$$

$$= 512 + 160 + 15$$

$$= (687)_{10}$$

Hex to Binary Conversion:

Straightforward

$$(9F2)_{16} = 9 F 2$$

$$= (1001 \ 1111 \ 0010)_{2}$$

Converting into Hex Number System

Binary to Hex Conversion: Straightforward

$$(1110100110)_2 = 0011 1010 0110$$

= 3 A 6

Answer = $(3A6)_{16}$

Decimal to Hex Conversion:

$$(378)_{10} =$$

$$378 \div 16 = 23 + \text{remainder of } 10_{10} = \mathbf{A_{16}}$$

$$23 \div 16 = 1 + remainder of 7$$

$$1 \div 16 = 0 + remainder of 1$$

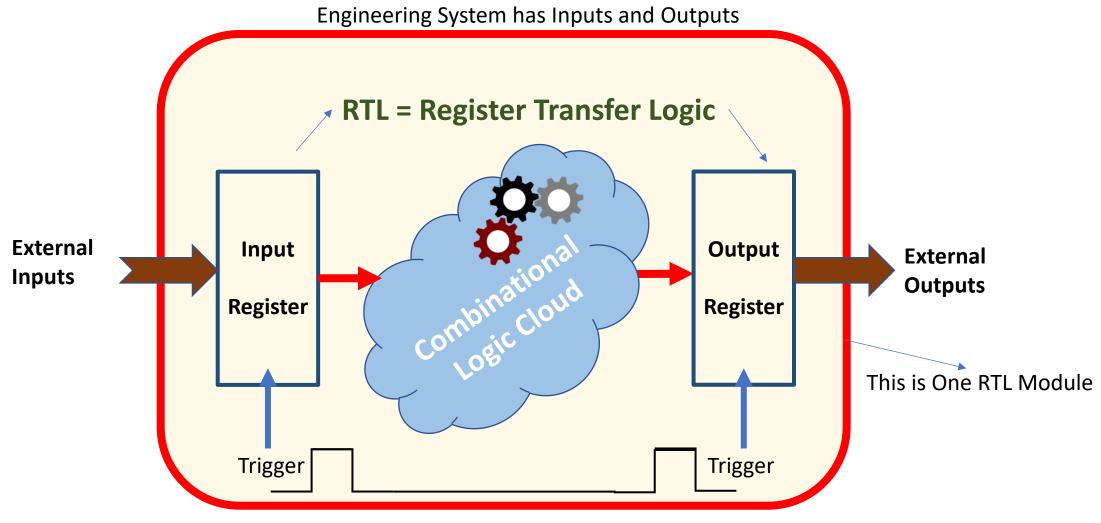
Answer =
$$(17A)_{16}$$

In Binary form:

$$(17A)_{16} = (0001 \ 0111 \ 1010)_2$$



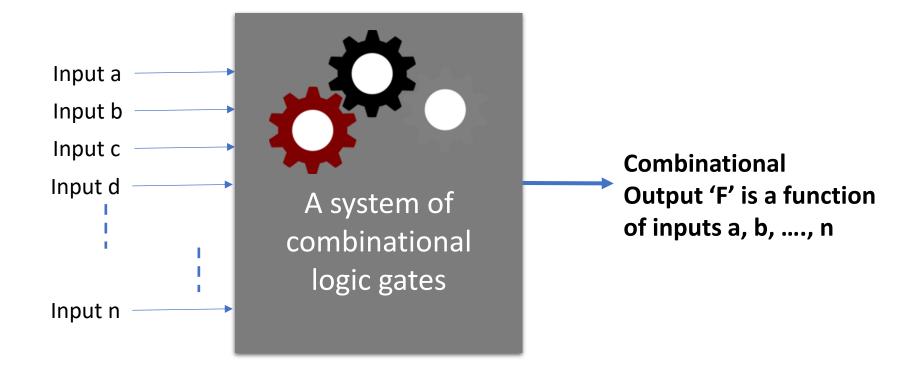
RTL View of Digital System Design



A Complex Digital System has many RTL Modules



Combinational Logic Circuits





The Language of Digital Systems

- Minterms and Maxterms
- Truth Table
- Boolean Algebra
- K-Maps
- State Tables
- State Machines and Diagrams
- Schematic Diagram
- HDL Code

Combinational Logic

Sequential Logic

All types of complex

Logic Functions





Truth table, minterms, SOP



P	Q	R	S output	Expression	Minterms
0	0	0	1	$ar{P}ar{Q}ar{R}$	m0
0	0	1	1	$ar{P}ar{Q}$ R	m1
0	1	0	1	$\bar{P}Q\bar{R}$	m2
0	1	1	1	$\bar{P}QR$	m3
1	0	0	0	0	
1	0	1	0	0	
1	1	0	0	0	
1	1	1	1	PQR	m7

$$S = \bar{P}\bar{Q}\bar{R} + \bar{P}\bar{Q}R + \bar{P}Q\bar{R} + \bar{P}QR + PQR \qquad \text{Sum of Products Expression}$$

$$S = \sum (m0, m1, m1, m3, m7)$$
 Minterms Expression



Truth table, Maxterms, POS



P	Q	R	F output (POS)	Expression	Maxterms
0	0	0	1		
0	0	1	1		
0	1	0	1		
0	1	1	1		
1	0	0	0	$\bar{P} + Q + R$	M4
1	0	1	0	$\bar{P} + Q + \bar{R}$	M5
1	1	0	0	$\bar{P} + \bar{Q} + R$	M6
1	1	1	1		

$$F_{POS} = \overline{(P+Q+R)}.(\overline{P}+Q+\overline{R}).(\overline{P}+\overline{Q}+R)$$

Product of Sum Expression

$$F_{POS} = \boxed{(M4. M5. M6)}$$
 Maxterms Expression

Output is Zero at F_{POS}



Boolean Algebra theorems

$$x + y = y + x$$

$$x \cdot y = y \cdot x$$

$$x + (y + z) = (x + y) + z = x + y + z$$

$$x(yz) = (xy)z = xyz$$

$$x(y + z) = xy + xz$$

$$(w + x)(y + z) = wy + xy + wz + xz$$

$$x + xy = x$$

$$x + \overline{x}y = x + y$$

$$\overline{x} + xy = \overline{x} + y$$

$$\overline{(x + y)} = \overline{x} \cdot \overline{y}$$

$$\overline{(x \cdot y)} = \overline{x} + \overline{y}$$
DeMorgan's Laws



Logic Minimization – using Boolean Algebra

From truth table example:
$$S = \bar{P}\bar{Q}\bar{R} + \bar{P}\bar{Q}R + \bar{P}Q\bar{R} + \bar{P}QR + PQR$$

Identify and Eliminate Common terms

$$S = \overline{P}\overline{Q}(\overline{R} + R) + \overline{P}Q(\overline{R} + R) + PQR$$

$$S = \overline{P}\overline{Q} + \overline{P}Q + PQR$$

$$S = \overline{P}(\overline{Q} + Q) + PQR$$

$$S = \overline{P} + PQR$$

Use theorem

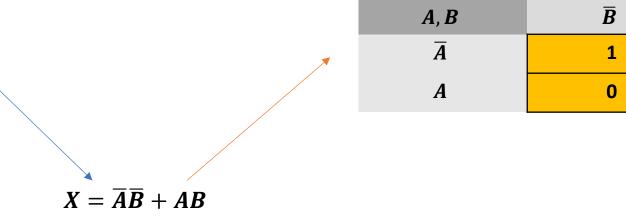
$$S = \overline{P} + QR$$

Final simplified expression

Logic Simplification – using Karnaugh K-Maps

2 Variable K-maps

A input	B input	X output
0	0	1
0	1	0
1	0	0
1	1	1



B, A	\overline{A}	A
\overline{B}	1	0
В	0	1



B

0

1

Three Variable K-Maps

A input	B input	C input	X output
0	0	0	1
0	0	1	1
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	1
1	1	1	0

AB,C	\overline{C}	C
$\overline{A}\overline{B}$	1	1
$\overline{A}B$	1	0
AB	1	0
$A\overline{B}$	0	0

A,BC	B̄C	БС	ВС	B₹
\overline{A}	1	1	0	1
A	0	0	0	1

$$X = \overline{A}\overline{B}\overline{C} + \overline{A}\overline{B}C + \overline{A}B\overline{C} + AB\overline{C}$$



4 Variable K-Maps

A input	B input	C input	D input	X output
0	0	0	0	0
0	0	0	1	1
0	0	1	0	0
0	0	1	1	0
0	1	0	0	0
0	1	0	1	1
0	1	1	0	0
0	1	1	1	0
1	0	0	0	0
1	0	0	1	0
1	0	1	0	0
1	0	1	1	0
1	1	0	0	0
1	1	0	1	1
1	1	1	0	0
1	1	1	1	1

$$X = \overline{A}\overline{B}\overline{C}D + \overline{A}B\overline{C}D + AB\overline{C}D + ABCD$$

AB, CD	<u></u> C D	C D	CD	CD
$\overline{A}\overline{B}$	0	1	0	0
$\overline{A}B$	0	1	0	0
AB	0	1	1	0
$A\overline{B}$	0	0	0	0

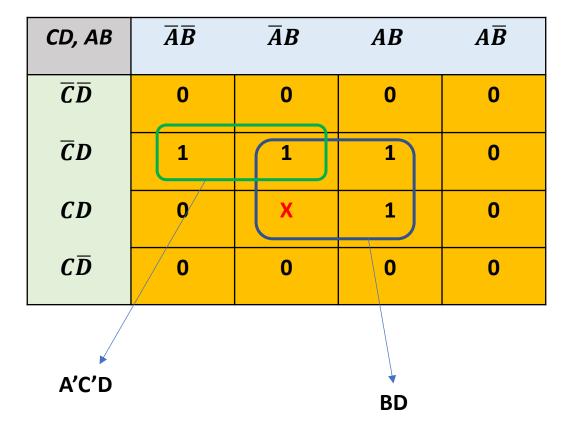
Different orientation - 4 Variable K-Maps

A input	B input	C input	D input	X output
0	0	0	0	0
0	0	0	1	1
0	0	1	0	0
0	0	1	1	0
0	1	0	0	0
0	1	0	1	1
0	1	1	0	0
0	1	1	1	0
1	0	0	0	0
1	0	0	1	0
1	0	1	0	0
1	0	1	1	0
1	1	0	0	0
1	1	0	1	1
1	1	1	0	0
1	1	1	1	1

<u> </u>			
$\mathbf{W} = \mathbf{A} \mathbf{D} \mathbf{C} \mathbf{D}$	ADOD	ADOD	ADOD
X = ABCD	$\perp \Delta R(1) \perp$	- <i>AR('I)</i> _	$\Delta R(1)$
$\Lambda - \Lambda D C D$		ADCD	IDCD
The second secon	,		

CD, AB	$\overline{A}\overline{B}$	$\overline{A}B$	AB	$A\overline{B}$
$\overline{C}\overline{D}$	0	0	0	0
C D	1	1	1	0
CD	0	0	1	0
$C\overline{D}$	0	0	0	0

K-Maps with Don't Care Literals





Logic Minimization – using Karnaugh K-Maps

Procedure:

Find biggest groups of adjacent '1's Edges can be folded The literals that remain fixed in the group are written down The literals that change, eg A to A' are removed All 1's to be covered at least once

2 Variable K-maps

Example 1

A, B	\overline{B}	В
\overline{A}	1	0
A	1	0

A and A' are literals that change, so removed B' remains common in the group The minimized expression is B'

> A and A' are literals that change, hence dropped B remains common in the group, retained The minimized expression is **B**

Example 2

B, A	\overline{A}	A
$ar{B}$	0	0
В	1	1



Logic Minimization – 3 variable K-Maps

Example 3-1

AB,C	$\overline{\it C}$	С
$\overline{A}\overline{B}$	1	1
$\overline{A}B$	1	0
AB	1	0
$A\overline{B}$	0	0

The 1 at A'BC' is covered in two groups, only one is sufficient So minimization will be on two non-redundant groups Minimized Expression is A'B' (where C and C' are removed) And BC' (where A and A' are removed).

Hence Answer is A'B' + BC' (in the form of Minterms)

A,BC	$\overline{B}\overline{C}$	$\overline{B}C$	BC	$\mathbf{B}\overline{\mathbf{C}}$
\overline{A}	1	1	0	1
A	0	0	0	1

Example 3-2

23

Three groups containing two 1's each are identified One group A'B'C' and A'BC' is redundant and can be Left out

C and C' can be removed to give A'B' as one expression And A and A' can be removed to give BC' as the other Final Minterms expression is **A'B' + BC'**



More Examples – 3 variable K-Maps

Example 3-3

AB,C	$\overline{\it C}$	С
$\overline{A}\overline{B}$	1	1
$\overline{A}B$	1	1
AB	0	0
$A\overline{B}$	0	0

Group of 4 1's
C' and B' change and can be removed
The answer is only **A'**

Example 3-4

AB,C	\overline{c}	<i>C</i>
$\overline{A}\overline{B}$	1	0
$\overline{A}B$	1	0
AB	1	0
$A\overline{B}$	1	0

Another group of 4 1's
A and A' can be removed
B and B' can be removed
The answer is only **C'**

Example 3-5

AB,C	<u></u> C	<i>C</i>		
$\overline{A}\overline{B}$	1	1		
$\overline{A}B$	0	0		
AB	0	0		
$A\overline{B}$	1	1		

Another group of 4 1's with folded edges A and A' change so can be removed C and C' change so can be removed The answer is only **B'**



Logic Minimization – 4 Variable K-maps

Example 4-1

AB, CD	$\overline{C}\overline{D}$	C D	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	0	1	1	0
AB	0	1	1	0
$A\overline{B}$	0	0	0	0

Group of 4 1's
A' changes along vertical axis
C' changes along horizontal axis
Answer is remaining terms = **BD**

Example 4-2

AB, CD	<u></u> C D	C D	CD	CD
$\overline{A}\overline{B}$ _	1	0	0	1
ĀB	0	0	0	0
AB	0	0	0	0
$A\overline{B}$	1	0	0	1

Group of 4 1's with folded edges
C and C' change along horizontal axis
A and A' change along vertical axis
Answer is terms that do not change = **B'D'**



Logic Minimization – 4 Variable K-maps

Example 4-3

AB, CD	$\overline{C}\overline{D}$	C D	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	1	1	1	1
AB	1	1	1	1
$A\overline{B}$	0	0	0	0

Group of 8 1's

A and A' change in vertical axis
All variables change in horizontal axis
Answer = **B**

Example 4-4

AB, CD	$\overline{C}\overline{D}$	C D	CD	$C\overline{D}$
$\overline{A}\overline{B}$	1	0	0	1
$\overline{A}B$	1	0	0	1
AB	1	0	0	1
$A\overline{B}$	1	0	0	1

Group of 8 1's

Folded along horizontal edges for biggest group All variables change along vertical axis C and C' change along horizontal axis Answer = **D'**



K-Maps for 5 variables ??

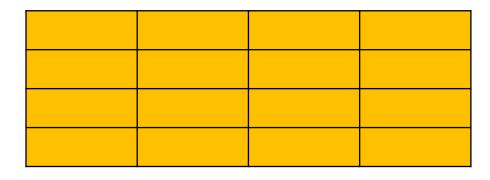
Starts to look complicated for manual working For any function X=f{A,B,C,D,E} Visualize two 4 input truth tables for {B,C,D,E}, First truth table is when A=0, second is when A=1

When A=0

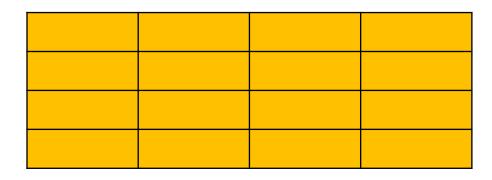
A can be eliminated if is common for any Literal in both K-maps

When A=1

K-map for {B,C,D,E}



K-map for {B,C,D,E}





K-Maps for 6 or more variables

Generally computer programs are used to model multivariable K-Maps through multidimensional arrays Conceptually, we can continue to extend 4-variable K-Maps

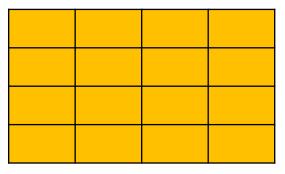
Eg for $X=f\{A,B,C,D,E,F\}$:

Consider there are four K-Maps of 4 variables {C,D,E,F}

There is one dedicated K-Map each for AB=00, AB=01, AB=10 and AB=11

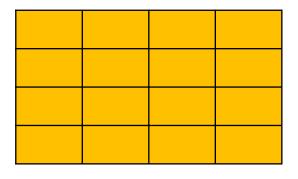
After simplification in K-Maps, further simplification is done through Boolean Algebra

K-map for {C,D,E,F}



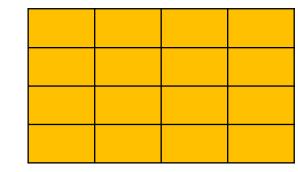
When AB=00

K-map for {C,D,E,F}



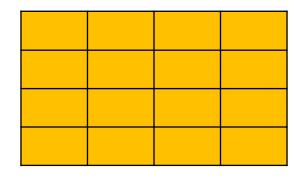
When AB=01

K-map for {C,D,E,F} K-map for {C,D,E,F}



When AB=10

K-map for {C,D,E,F}



When AB=11

Higher level Logic Minimization

Quine McCluskey Tabular Method is a famous algorithm that is used in many computer programs Logic Minimizer https://www.logicminimizer.com/ (up to 8 variables, tabular method)

QMC Logic Minimizer https://sourceforge.net/projects/qmclm/ (up to 16 variables, based on QM)

Logic Friday https://logic-friday.software.informer.com/ (up to 16 variables, based on Espresso)

Mini Log https://softwiki.net/Logic_Minimizer.html (up to 40 variables, based on Espresso)

Espresso uses Graph theory and Cubes

https://ptolemy.berkeley.edu/projects/embedded/pubs/downloads/espresso/index.htm

An overview document by digilent https://learn.digilentinc.com/Documents/412

WinLogiLab program can minimize using both QM and Espresso algorithms

All this leads to:

LOGIC SYNTHESIS Synopsys, Cadence (discuss further in labs)
This is built into modern design automation tools for digital design and implementation on FPGA and ASIC

UC Berkeley is considered as the source of many simulation and synthesis tools, Eg SPICE, ESPRESSO and others



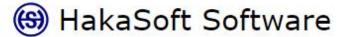






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