

Lecture 25

EE 421 / CS 425

Digital System Design

Fall 2023

Shahid Masud

Topics

QUIZ NEXT WEEK

- Digital Scramblers
- Digital Un-Scramblers
- Scrambler Applications
- Scrambler Examples
- Binary Rate Multipliers

What is a Scrambler?

- Scramblers are circuits that **pseudo-randomly change the values of some bits** in a data block or stream with the purpose of '**whitening**' its spectrum so that **no strong spectral components** will exist.
- This will **reduce electromagnetic interference** and **introduce data security** as part of an encryption algorithm.
- In natural signals, e.g. voice, image, transducer, energy is concentrated in some bands. Scrambling helps **spread the energy to reduce interference and spikes**.
- An ALFSR forms backbone of scrambling operation.
- There are two main types:
 - **Additive Scrambler**
 - **Multiplicative Scrambler**

Applications of LFSR

Ref: from Wikipedia page

Digital broadcasting systems that use linear-feedback registers:

- [ATSC Standards](#) (digital TV transmission system – North America)
- [DAB](#) ([Digital Audio Broadcasting](#) system – for radio)
- [DVB-T](#) (digital TV transmission system – Europe, Australia, parts of Asia)
- [NICAM](#) (digital audio system for television)

Other digital communications systems using LFSRs:

- [INTELSAT](#) business service (IBS)
- Intermediate data rate (IDR)
- [HDMI 2.0](#)
- [SDI](#) (Serial Digital Interface transmission)
- Data transfer over [PSTN](#) (according to the [ITU-T](#) V-series recommendations)
- [CDMA](#) (Code Division Multiple Access) cellular telephony
- [100BASE-T2 "fast" Ethernet](#) scrambles bits using an LFSR
- [1000BASE-T Ethernet](#), the most common form of Gigabit Ethernet, scrambles bits using an LFSR
- [PCI Express](#)
- [SATA](#)^[13]
- [Serial attached SCSI](#) (SAS/SPL)
- [USB 3.0](#)
- [IEEE 802.11a](#) scrambles bits using an LFSR
- [Bluetooth Low Energy](#) Link Layer is making use of LFSR (referred to as whitening)
- [Satellite navigation systems](#) such as [GPS](#) and [GLONASS](#). All current systems use LFSR outputs to generate some or all of their ranging codes (as the chipping code for CDMA or DSSS) or to modulate the carrier without data (like GPS L2 CL ranging code). GLONASS also uses [frequency-division multiple access](#) combined with DSSS.

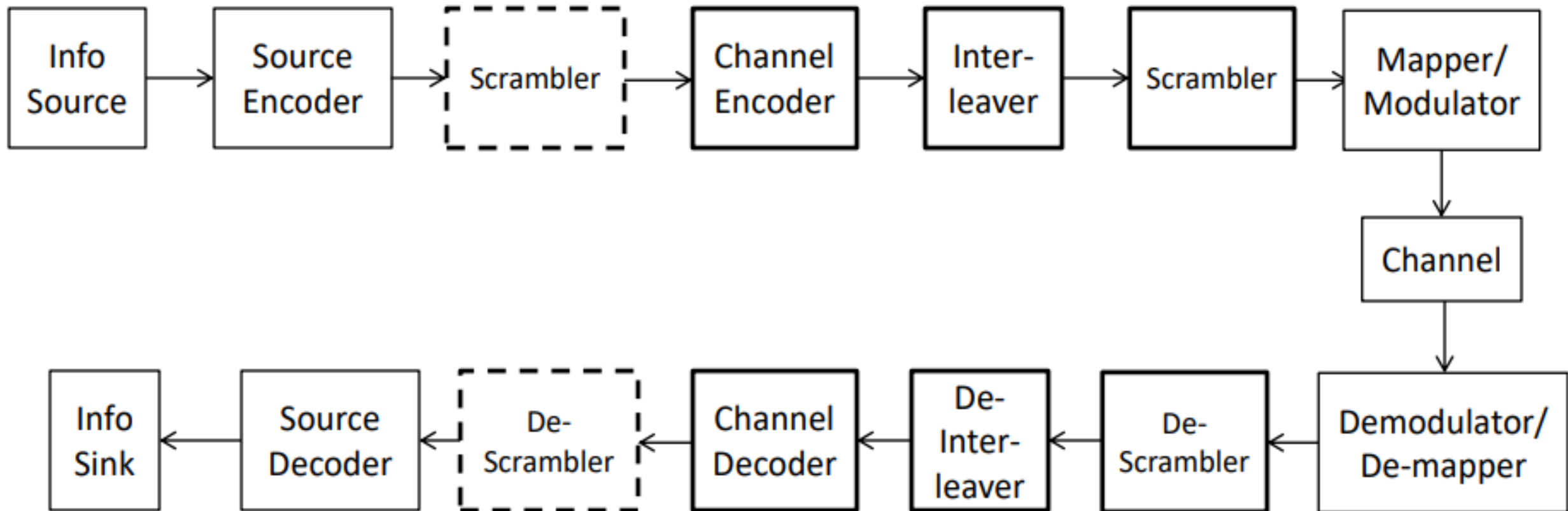
Other uses [[edit](#)]

Types of Scramblers

Standard	Scrambler type
V.34 (33 Kbps Modem)	Multiplicative
56 Kbps Modem	Multiplicative
DVB	Additive
LTE	Additive
IEEE 802.11 a	Additive
IEEE 802.16 a	Multiplicative

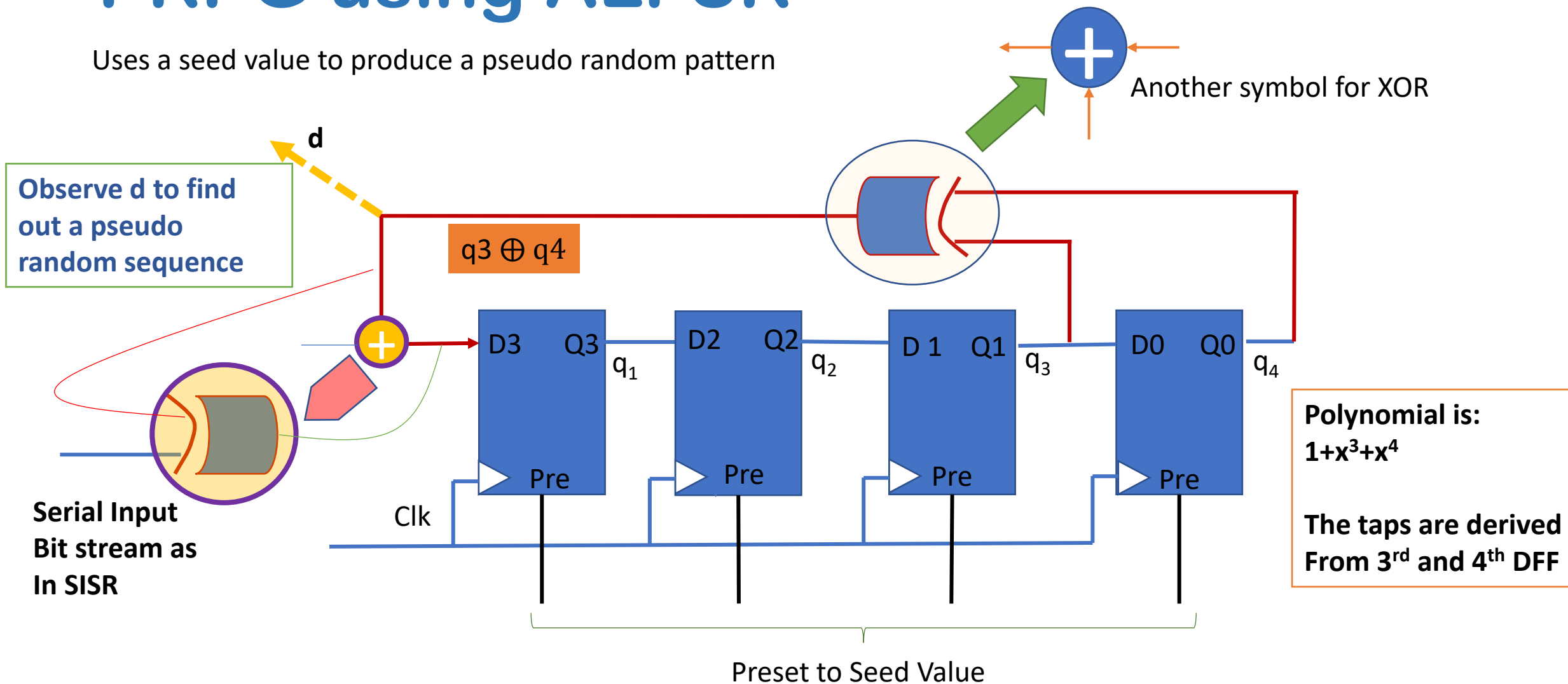
$1+x^4+X^7$ is polynomial for 802.11

Typical Communication System

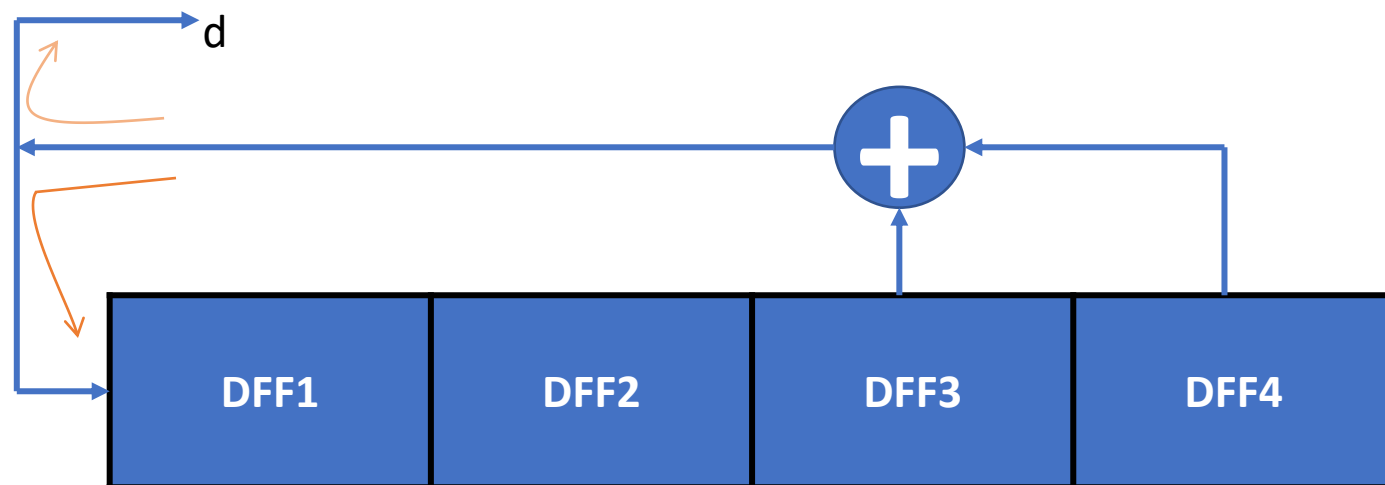


PRPG using ALFSR

Uses a seed value to produce a pseudo random pattern



Simplified Representation



Preset

Characteristic Polynomial
 $1+x^3+x^4$

The sequence observed at $q_3q_2q_1q_0 =$
 $\{15, 7, 3, 1, 8, 4, 2, 9, 12, 6, 11, 5, 10, 13, 14, 15, \dots\}$

q1	q2	q3	q4	d
1	1	1	1	0
0	1	1	1	0
0	0	1	1	0
0	0	0	1	1
1	0	0	0	0
0	1	0	0	0
0	0	1	0	1
1	0	0	1	1
1	1	0	0	0
0	1	1	0	1
1	0	1	1	0
0	1	0	1	1
1	0	1	0	1
1	1	0	1	1
1	1	1	0	1
1	1	1	1	0

Changing seed value with same polynomial

What happens when seed value is changed from '1111' to '1000'?

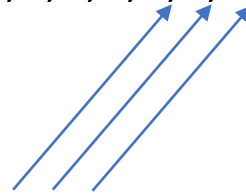
Answer:

With seed value of '1111'

The sequence observed at $q_3q_2q_1q_0 = \{15, 7, 3, 1, 8, 4, 2, 9, 12, 6, 11, 5, 10, 13, 14, 15, \dots\}$

With seed value of '1000'

The sequence observed at $q_3q_2q_1q_0 = \{8, 4, 2, 9, 12, 6, 11, 5, 10, 13, 14, 15, 7, 3, 1, 8, \dots\}$



Both sequences are equal in a circle, just the starting point is different

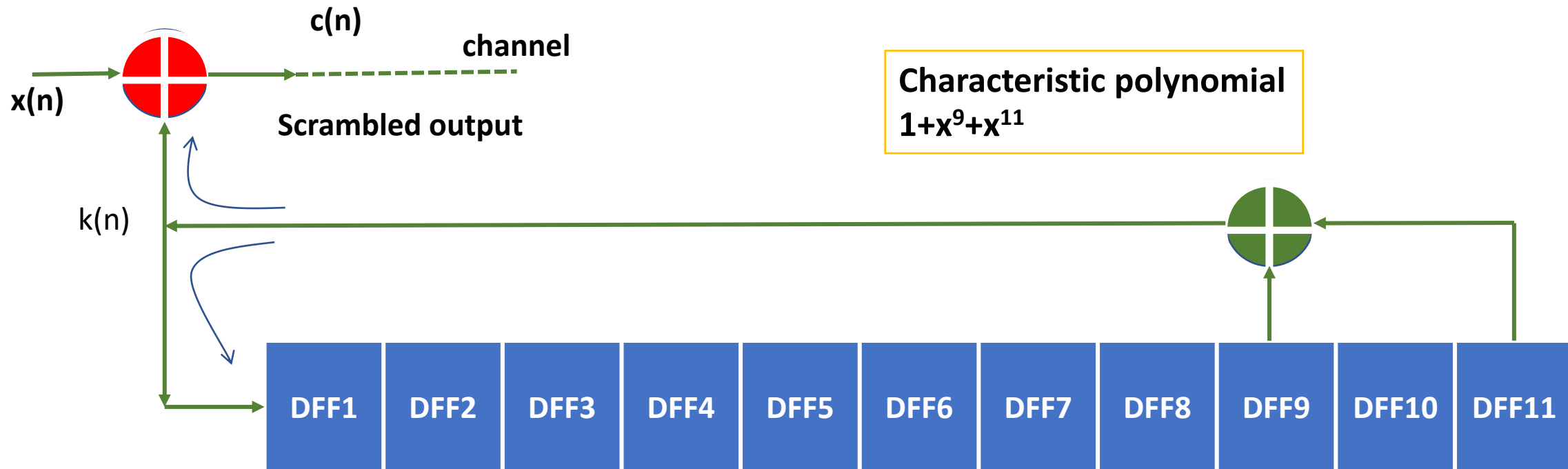
PRPG Cycle length vs Register length

# of Bits	Length of Loop	Taps
2	3 *	[0,1]
3	7 *	[0,2]
4	15	[0,3]
5	31 *	[1,4]
6	63	[0,5]
7	127 *	[0,6]
8	255	[1,2,3,7]
9	511	[3,8]
10	1,023	[2,9]
11	2,047	[1,10]
12	4,095	[0,3,5,11]
13	8,191 *	[0,2,3,12]
14	16,383	[0,2,4,13]
15	32,767	[0,14]
16	65,535	[1,2,4,15]
17	131,071 *	[2,16]
18	262,143	[6,17]
19	524,287 *	[0,1,4,18]
20	1,048,575	[2,19]
21	2,097,151	[1,20]
22	4,194,303	[0,21]
23	8,388,607	[4,22]
24	16,777,215	[0,2,3,23]
25	33,554,431	[2,24]
26	67,108,863	[0,1,5,25]
27	134,217,727	[0,1,4,26]
28	268,435,455	[2,27]
29	536,870,911	[1,28]
30	1,073,741,823	[0,3,5,29]
31	2,147,483,647 *	[2,30]
32	4,294,967,295	[1,5,6,31]

Additive Scrambler and Descrambler

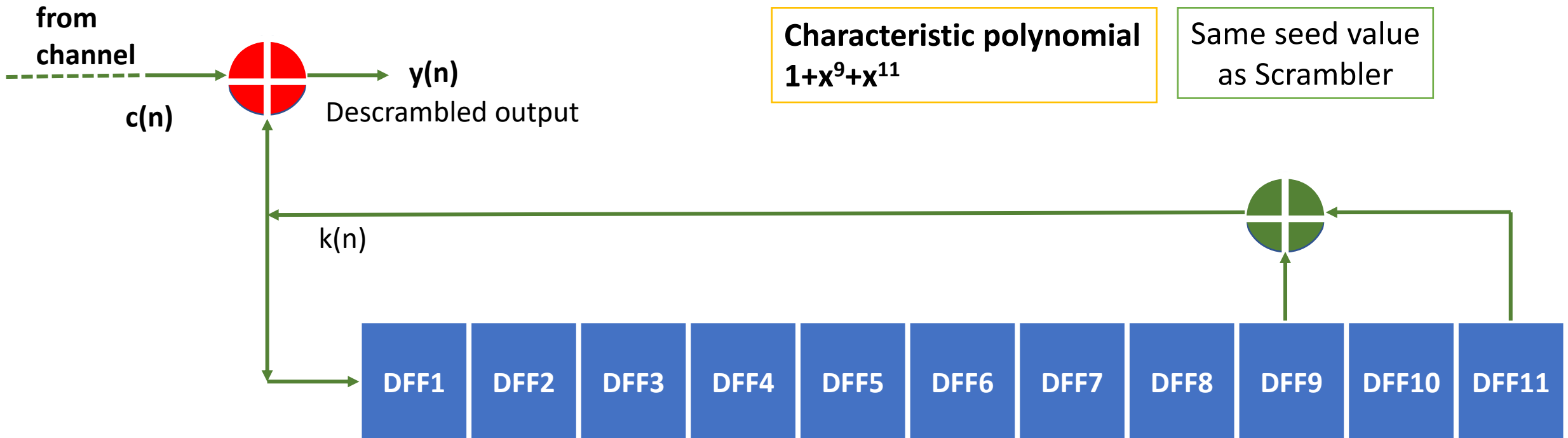
- They are synchronous scramblers because they require the initial state of the scrambler and descrambler to be the same
- They are non-recursive because they do not have any feedback loop in the process
- LFSR is connected to the data stream by means of an additional modulo-2 adder (XOR) gate
- Used in 100Base-TX interface which repeats its sequence after $2^N - 1 = 2047$ bits; with $N=11$
- Usually employed where a fixed size frame is to be transmitted

Additive Scrambler Example



$x(n)$ represents the data to be scrambled at time n
 $k(n)$ represents the 'key' produced by the LFSR
 $c(n)$ represents the scrambled code word

Corresponding Additive Descrambler



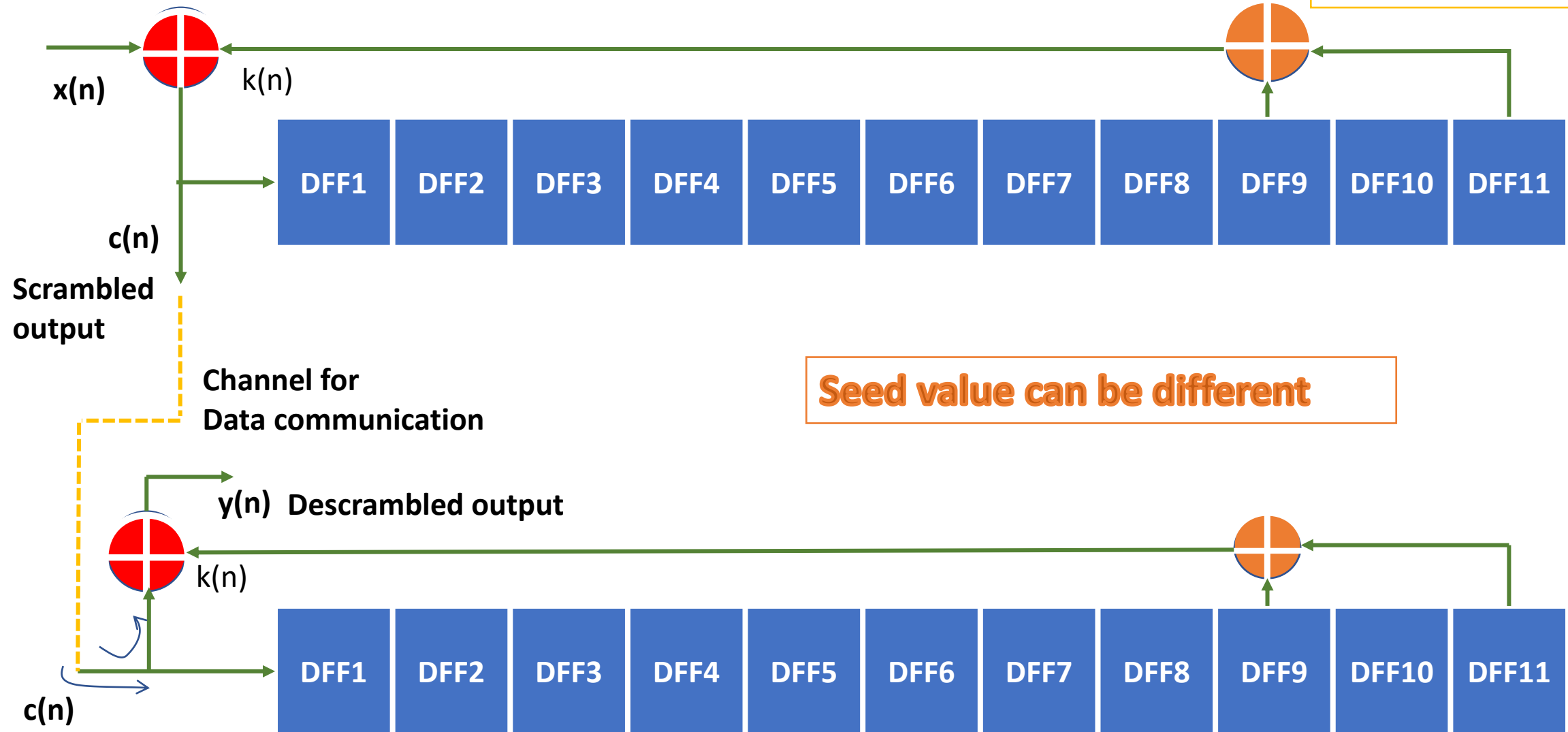
$x(n)$ represents the data to be scrambled at time n
 $k(n)$ represents the 'key' produced by the LFSR
 $c(n)$ represents the scrambled code word

Multiplicative Scramblers

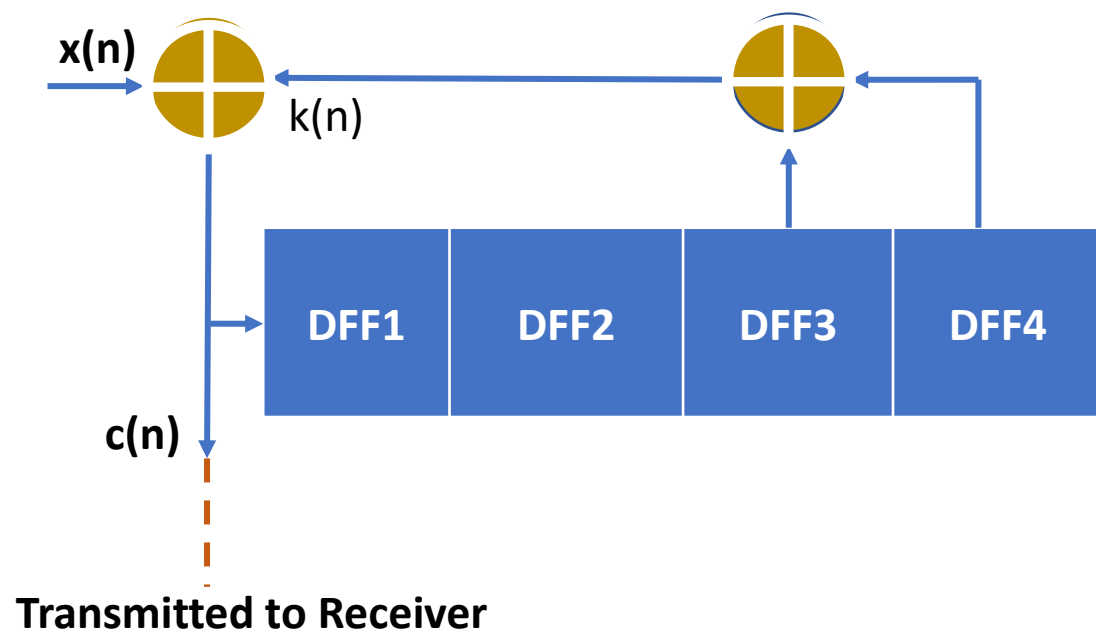
- They are **asynchronous** because they do not require LFSR synchronization.
- They are **recursive as they have feedback loops**.
- The **pair of scrambler-descrambler is self-synchronizing**. They do not need to start from same initial value.
- The **self-synchronizing process could take up to N bits** (or N clock cycles) so the first N values of $y(n)$ should be discarded.
- With this approach, the transmission errors are multiplied by $T+1$, where T is the number of taps used in LFSR. Thus if one bit is flipped due to some error, the descrambler will result in 3 bit error.

Multiplicative Scrambler - Descrambler

Characteristic polynomial
 $1+x^9+x^{11}$



Example – Multiplicative Scrambler



LFSR Using polynomial $1 + x^3 + x^4$

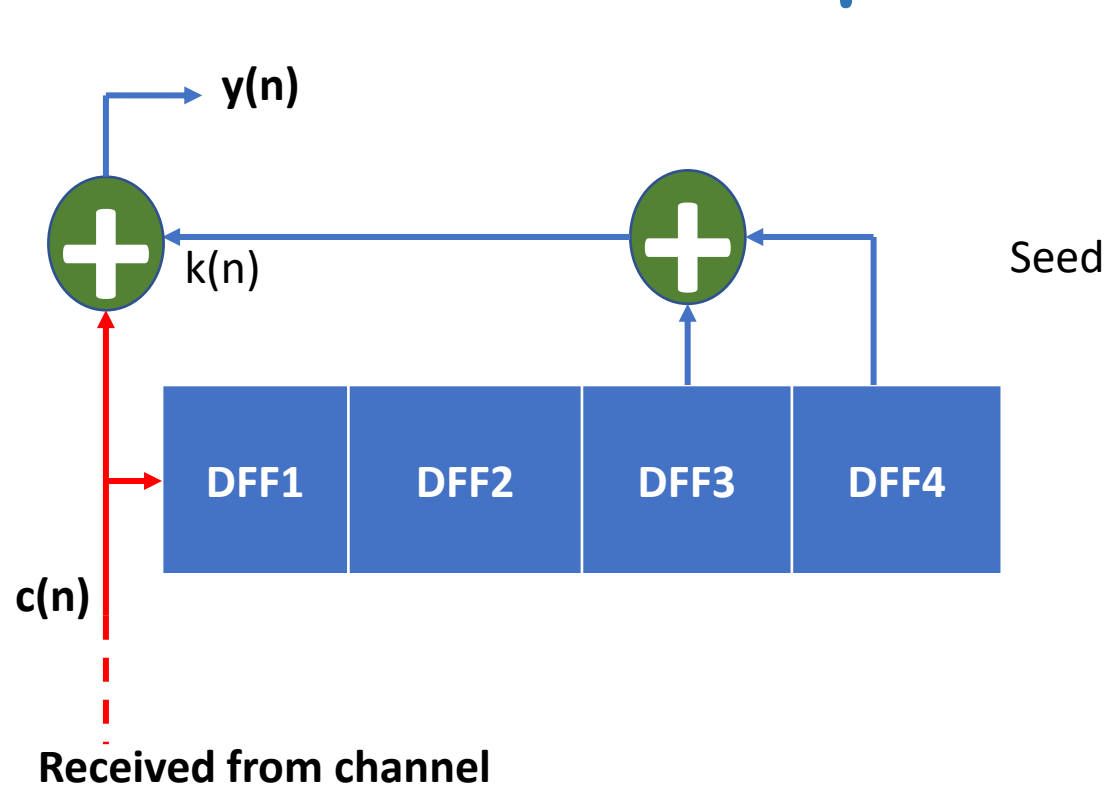
Scrambler Seed Value "0000"

Input Data stream:

"101100010110"

Registers output "q1q2q3q4"				k	x	c
0	0	0	0	0	1	1
1	0	0	0	0	0	0
0	1	0	0	0	1	1
1	0	1	0	1	1	0
0	1	0	1	1	0	1
1	0	1	0	1	0	1
1	1	0	1	1	0	1
1	1	1	0	1	1	0
0	1	1	1	0	0	0
0	0	1	1	0	1	1
1	0	0	1	1	1	0
0	1	0	0	0	0	0

Continue to – Multiplicative De-Scrambler



Seed

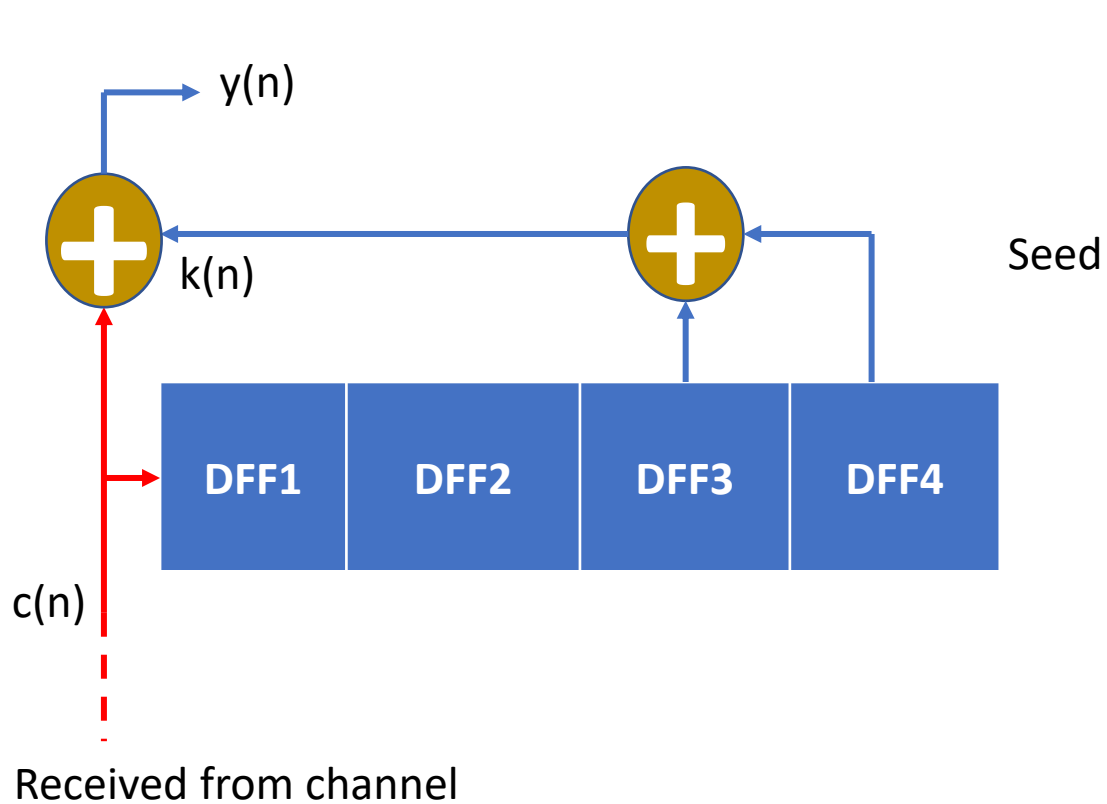
Registers output "q1q2q3q4"				k	c	y
1	1	1	1	0	1	1
1	1	1	1	0	0	0
0	1	1	1	0	1	1
1	0	1	1	0	0	0
0	1	0	1	1	1	0
1	0	1	0	1	1	0
1	1	0	1	1	1	0
1	1	1	0	1	0	1
0	1	1	1	0	0	0
0	0	1	1	0	1	1
1	0	0	1	1	0	1
0	1	0	0	0	0	0

N=4

y=x
Input String
From here

LFSR Using polynomial $1 + x^3 + x^4$
 Scrambler Seed Value "0000"
 De-Scrambler Seed Value "1111"
 Input Data stream:
 "101100010110"

Multiplicative De-Scrambler with Error Detection



LFSR Using polynomial $1 + x^3 + x^4$
 Scrambler Seed Value "0000"
 De-Scrambler Seed Value "1111"
 Input Data stream:
 "101100010110"

T is the number of TAPS used in ALFSR

Registers output "q1q2q3q4"				k	c	y
1	1	1	1	0	1	1
1	1	1	1	0	0	0
0	1	1	1	0	1	1
1	0	1	1	0	0	0
0	1	0	1	1	1	0
1	0	1	0	1	0	1
1	1	0	1	1	1	0
1	1	1	0	1	0	1
0	1	1	1	0	0	1
0	0	1	1	0	1	0
1	0	0	1	1	0	1
0	1	0	0	0	0	0

N=4

y=x
Input String From here

One error in blue was introduced in c(n)

This error resulted in T+1 = 3 errors in y(n) shown in red

Case Study: Binary Rate Multipliers



Binary Rate Multipliers - BRM

Definition: A circuit module that transforms a stream of input clock pulses into another stream of output clock pulses

Let N_i = no. of input pulses for a particular time period

Let N_o = no. of output clock pulses for the same time period

$$\text{Binary Fractional Rate Multiplier} \Rightarrow N_o = \frac{B}{2^n} N_i$$

B = Rate Constant Input to module

$$B = (B_{n-1}, B_{n-2}, \dots, B_2, B_1, B_0)_2$$

n = no. of binary counter stages controlling the module

The clock input drives an n -bit binary counter whose outputs are labelled $(X_n, X_{n-1}, \dots, X_2, X_1)$.

Digital Fractional Rate Multiplier - Principle

The counter outputs (X_n to X_1 = MSB to LSB) are ANDed with the incoming Clock signals to form intermediate pulse trains $P_i(i=1,n)$:

$$P_n = X_1 \cdot \text{Clock}$$

$$P_{n-1} = \overline{X_1} \cdot X_2 \cdot \text{Clock}$$

$$P_{n-2} = \overline{X_1} \cdot \overline{X_2} \cdot X_3 \cdot \text{Clock}$$

.....

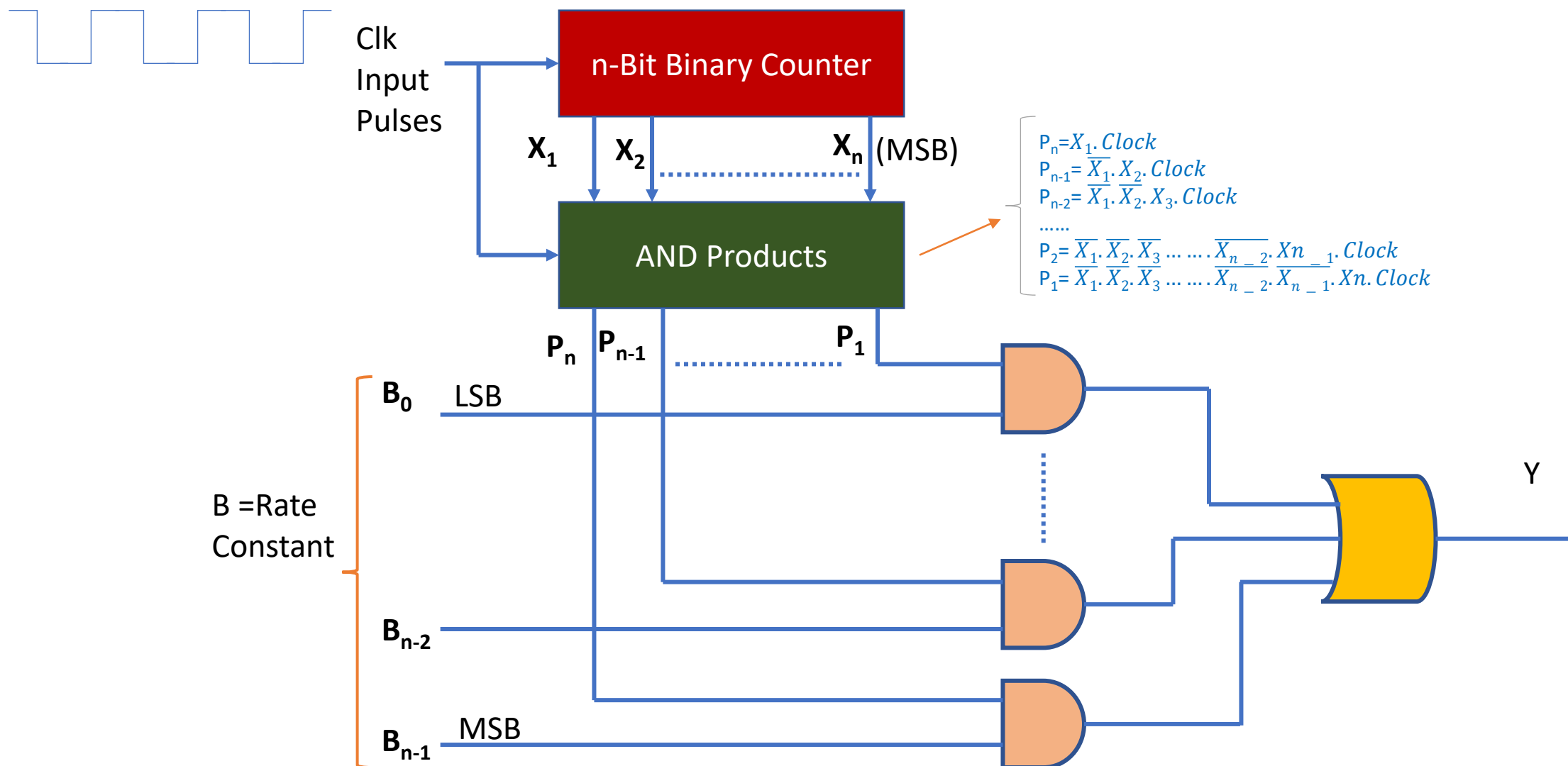
$$P_2 = \overline{X_1} \cdot \overline{X_2} \cdot \overline{X_3} \dots \dots \overline{X_{n-2}} \cdot X_{n-1} \cdot \text{Clock}$$

$$P_1 = \overline{X_1} \cdot \overline{X_2} \cdot \overline{X_3} \dots \dots \overline{X_{n-2}} \cdot \overline{X_{n-1}} \cdot X_n \cdot \text{Clock}$$

The logic output Y uses the rate constant M and the pulse train signals P_i to implement the equation $Y = \sum B_{i-1} \cdot P_i$

This output configuration will deliver the proper number of output pulses (N_o) as specified by rate constant B

Basic Circuit Diagram – DFRM



Example and timing diagram of BRM

- BRM using a 3-Bit counter
- Rate Constant $B = 7_{10}$
- Counter generates 8 pulses per time period. Due to $B=7$, 7 pulses appear at the output and one pulse is eliminated

Analysis of timing diagram 3 bit BRM, B=7

- $Y = B_2 \cdot P_3 + B_1 \cdot P_2 + B_0 \cdot P_1$
- Where
- $P_3 = X_1 \cdot \text{Clock}$
- $P_2 = \overline{X_1} \cdot X_2 \cdot \text{Clock}$
- $P_1 = \overline{X_1} \cdot \overline{X_2} \cdot X_3 \cdot \text{Clock}$
- And $B = (B_2 \cdot B_1 \cdot B_0) = (1.1.1)_2$
- Thus $Y = (X_1 + \overline{X_1} X_2 + \overline{X_1} \cdot \overline{X_2} \cdot X_3) \cdot \text{Clock}$
- In terms of counter output $X = (X_3 \cdot X_2 \cdot X_1)$
- Product term $X_1 \Rightarrow \sum m(1, 3, 5, 7)$
- Product term $X_2 \cdot \overline{X_1} \Rightarrow \sum m(2, 6)$
- Product term $X_3 \cdot \overline{X_2} \cdot \overline{X_1} \Rightarrow \sum m(4)$
- So that $Y(X_3 \cdot X_2 \cdot X_1) = (\sum m(1, 2, 3, 4, 5, 6, 7)) \cdot \text{Clock}$
- As m_0 is missing from this list so first clock from each sequence of 8 pulses will be eliminated

How many pulses are produced?

- From the timing diagram, each of the pulse trains P_i generates 2^{i-1} pulses during one counter sequence period (2^n clock pulses)
- Thus P_1 generates 1 pulse, P_2 generates 2 pulses, P_3 generates 4 pulses, and so on
- Note that pulses do not overlap in time
- We may OR the respective pulses to produce desired output stream
- The output pulses can be irregularly spaced
- The output pulses are synchronized with input clock pulses

Serially Cascaded Binary Rate Multiplier BRM

Design a BRM to implement $N_{out} = \frac{63}{320} \cdot N_{in}$

Break into binary powers ($63 \times 2 = 126$; and $320 \times 2 = 640$) :

$$N_{out} = \frac{(7 \times 18)}{(10 \times 64)} \cdot N_{in}$$

$$N_{out} = \left(\frac{7}{10}\right) \cdot \left(\frac{18}{64}\right) \cdot N_{in}$$

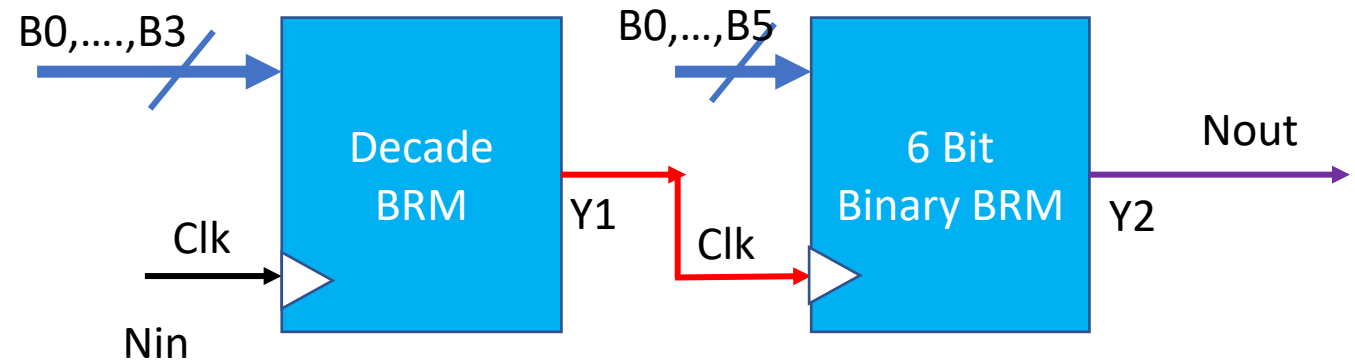
Use:

1 x Decade Counter BRM

1 x 6-Bit Binary Counter BRM

Set B for Decade Counter = 0111

Set B for 6-Bit BRM = 10010



Example 2: Serially Cascaded BRM

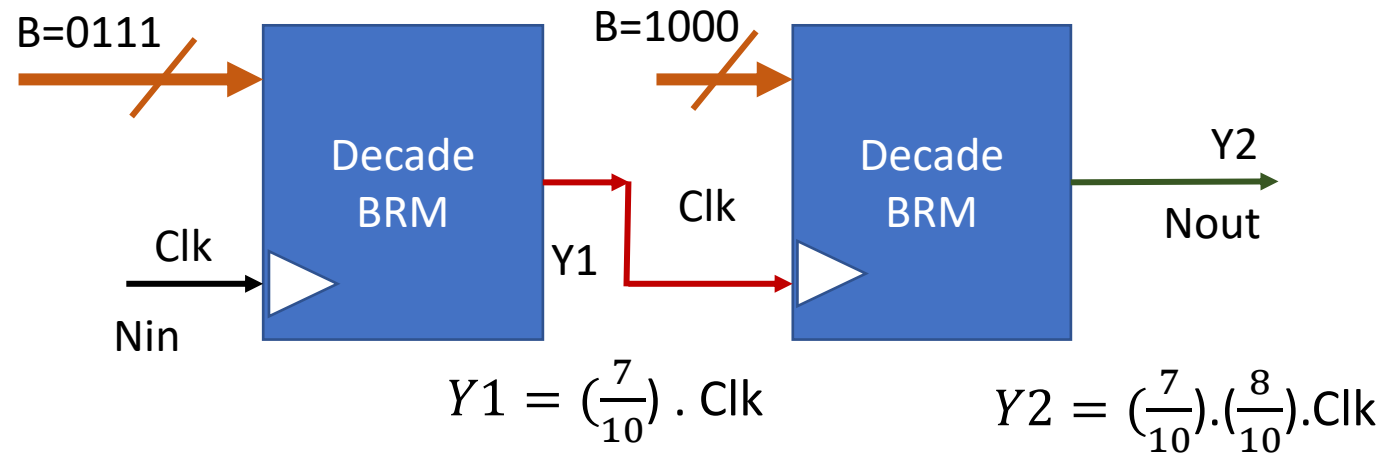
Design a BRM circuit that produces: 0.56 of input clock rate

We use two Decade BRMs with

Value of B is set as 7 and 8 respectively

$$N_{out} = \frac{56}{100} \cdot N_{in}$$

$$N = \frac{7}{10} \cdot \frac{8}{10} \cdot N_{in}$$



Eg 3: Parallel Cascaded Binary Rate Multiplier

Design a BRM circuit to produce $N_{out} = (0.297)N_{in}$

