

# RTL Design

*This lecture describes an approach to logic design called Register Transfer Level (RTL) or dataflow design. This is the method currently used for the design of complex logic circuits such as microprocessors.*

*The steps in RTL design are: (1) determine the number and sizes of registers needed to hold the data used by the device, (2) determine the logic and arithmetic operations that need to be performed on these register contents, and (3) design a state machine whose outputs determine how these register contents are updated with the results of those operations in order to obtain the desired results.*

*After this lecture you should be able to:*

- *classify a VHDL description as a behavioral, structural, or dataflow (RTL) description*
- *identify the registers and logic/arithmetic functions required to implement a particular algorithm*
- *partition this algorithm into a sequence of these operations and register transfers*
- *write synthesizable VHDL RTL code to implement the algorithm*

## Design Strategies

There are a number of strategies that are useful when designing complex logic circuits. You may recognize that similar strategies are used in computer programming.

One strategy is to design at the most abstract (“highest”) level possible with the tools available. For example, using a behavioral design style with VHDL instead of a structural style (e.g. schematics) will make it easier to write, read, document, and debug your design.

Another design strategy is hierarchical decomposition. The device being designed should be decomposed into a number of modules (represented as VHDL entities) that interface through well-defined interfaces (VHDL ports). The internal structure of these modules should not be visible from outside the module. Each of these modules should then be further subdivided into other modules. The decomposition process should be repeated until the remaining modules are simple enough to be easily written and tested. This decomposition makes it easy to test the modules individually, allows modules to be re-used and allows more than one person to work on the same project at the same time.

It’s also a good idea to keep the design as portable as possible. Avoid using language features that are specific to a particular manufacturer or target technology unless they are necessary to meet other re-

quirements. This will make it possible to use different manufacturing processes and different devices with a minimum of redesign.

## Structural Design

Structural design is the oldest digital logic design method. In this method the designer does all the work. The designer selects the low-level components and decides exactly how they are to be connected. The parity generator described in the previous lecture is an example of structural design.

A structural design can be represented as a parts list and a list of the connections between the pins on the components (for example: “pin 12 on chip 3 is connected to pin 5 on chip 7”). This representation of a circuit is called a *netlist*.

*Schematic capture* is the most common structural design method. The designer works with a program similar to a drawing program that allows components to be inserted into the design and connected to other components.

Exercise: What would be the most common type of statement in a structural VHDL description?

## Behavioral Design

At the other extreme, a behavioral design is meant to demonstrate the functional behaviour of a device

without concerning itself about implementation details. Thus a behavioral design may include operations such as integer division or behaviour such as propagation delays that would be difficult or impossible to synthesize.

However, every design should start with a behavioral description. The behavioral description can be simulated and used to verify that all of the required aspects of the design have been identified. Often the output of a behavioral description can be compared to the output of a structural or RTL description to check for errors.

Exercise: A VHDL description contains non-synthesizeable constructs. Is it a behavioural or structural description?

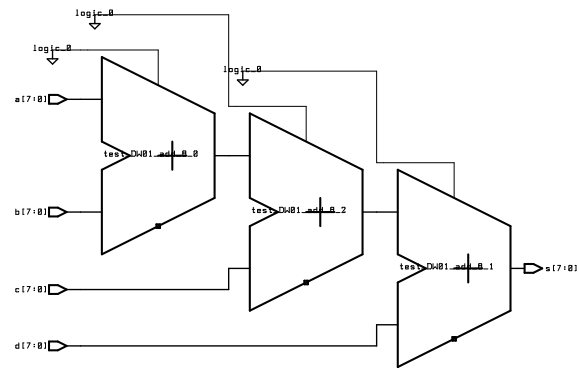
## RTL Design

Register Transfer Level, or RTL<sup>1</sup> design lies between a purely behavioral description of the desired circuit and a purely structural one. An RTL description describes a circuit's registers and the sequence of transfers between these registers but does not describe the hardware used to carry out these operations.

As a simple example, consider a device that needs to add four numbers. In VHDL, given signals of the correct type, we can simply write:

```
s <= a + b + c + d ;
```

This particular description is simple enough that it can be synthesized. However, the resulting circuit will be a fairly large combinational circuit comprising three adder circuits as follows:



A behavioral description, not being concerned with implementation details would be complete at this point.

However, if we were concerned about the cost of the implementation we might decide to break down the computation into a sequence of steps, each one involving only a single addition:

```
s = 0
s = s + a
s = s + b
s = s + c
s = s + d
```

where each operation is executed sequentially. The logic required is now one adder, a register to hold the value of *s* in-between operations, a multiplexer to select the input to be added on, and a circuit to clear *s* at the start of the computation.

Although this approach only needs one adder, the process requires more steps and will take longer. Circuits that divide up a computation into a sequence of arithmetic and logic operations are quite common and this type of design is called Register Transfer Level (RTL) or “dataflow” design.

An RTL design is composed of (1) registers and combinational function blocks (e.g. adders and multiplexers) called the *datapath* and (2) a finite state machine, called the *controller* that controls the transfer of data through the function blocks and between the registers.

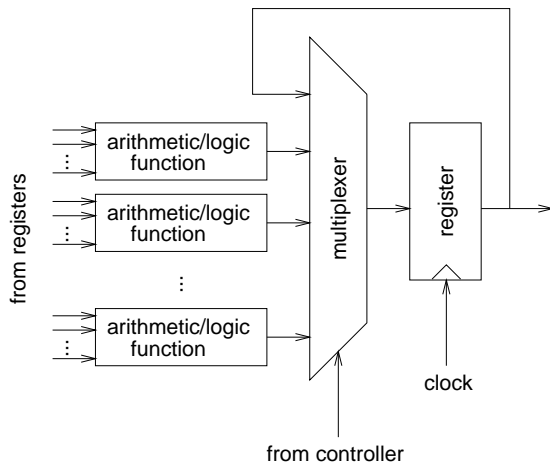
In VHDL RTL design the gate-level design and optimization of the datapath (registers, multiplexers, and combinational functions) is done by the synthesizer. However, the designer must design the state

<sup>1</sup>The “L” in RTL sometimes stands for “Language” or “Logic” – all refer to the same method of designing complex logic circuits.

machine and decide which register transfers are performed in which state.

The RTL designer can trade off datapath complexity (e.g. using more adders and thus using more chip area) against speed (e.g. having more adders means fewer steps are required to obtain the result). RTL design is well suited for the design of microprocessors and special-purpose processors such as disk drive controllers, video display cards, network adapter cards, etc. It gives the designer great flexibility in choosing between processing speed and circuit complexity.

The diagram below shows a generic component in the datapath. Each RTL design will be composed of one of the following building blocks for each register. The structure allows the contents of each register to be updated at the end of each clock period with a value selected by the controller. The widths of the registers, the types of combinational functions and their inputs will be determined by the application. A typical design will include many of these components.

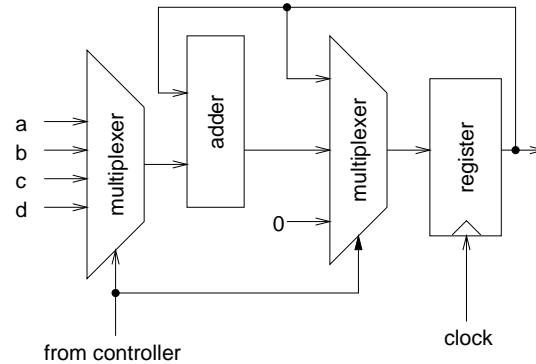


## RTL Design Example

To show how an RTL design is described in VHDL and to clarify the concepts involved, we will design a four-input adder. This design will also demonstrate how to create packages of components that can be re-used.

The datapath shown below can load the register at the start of each clock cycle with zero, the current value of the register, or the sum of the register and one of the four inputs. It includes one 8-bit register,

an 8-bit adder and a multiplexer that selects one of the four inputs as the value to be added to the current value of the register.



Exercise: Other datapaths could compute the same result. Draw the block diagram of a datapath capable of computing the sum of the four numbers in three clock cycles.

The first design unit is a package that defines a new type, `num`, for eight-bit unsigned numbers and an enumerated type, `states`, with six possible values. `nums` are defined as a subtype of the unsigned type.

```
-- RTL design of 4-input summer

-- subtype used in design

library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_arith.all ;

package averager_types is
    subtype num is unsigned (7 downto 0) ;
    type states is (clr, add_a, add_b, add_c,
        add_d, hold) ;
end averager_types ;
```

The first entity defines the datapath. In this case the four numbers to be added are available as inputs to the entity and there is one output for the current sum.

The inputs to the datapath from the controller are a 2-bit selector for the multiplexer and two control signals to load or clear (set to 0) the register.

```
-- datapath

library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_arith.all ;
use work.averager_types.all ;

entity datapath is
    port (
        a, b, c, d : in num ;
```

```

    sum : out num ;
    sel : in std_logic_vector (1 downto 0) ;
    load, clear, clk : in std_logic
  ) ;
end datapath ;

architecture rtl of datapath is
  signal mux_out, sum_reg, next_sum_reg : num ;
  constant sum_zero : num :=
    conv_unsigned(0,next_sum_reg'length) ;
begin

  -- mux to select input to add
  with sel select mux_out <=
    a when "00",
    b when "01",
    c when "10",
    d when others ;

  -- mux to select register input
  next_sum_reg <=
    sum_reg + mux_out when load = '1' else
    sum_zero when clear = '1' else
    sum_reg ;

  -- register sum
  process(clk)
  begin
    if clk'event and clk = '1' then
      sum_reg <= next_sum_reg ;
    end if ;
  end process ;

  -- entity output is register output
  sum <= sum_reg ;

end rtl ;

```

Exercise: Label the block diagram above with the bus widths and signal names used in the entity.

What would happen if both clear and load inputs were asserted? Why do we need to define both sum\_reg and sum signals?

How many operations will it take to compute the sum of the four inputs?

The RTL design's controller is a state machine whose outputs control the multiplexers in the datapath. The controller's inputs are signals that control the controller's state transitions. In this case the only input is an update signal that tells our device to recompute the sum (presumably because one or more of the inputs has changed).

This particular state machine sits at the "hold" state until the update signal is true. It then sequences through the other five states and then stops at the hold state again. The other five states are used to clear the register and to add the four inputs to the current value of the register.

```
-- controller
```

```

library ieee ;
use ieee.std_logic_1164.all ;
use work.averager_types.all ;

entity controller is
  port (
    update : in std_logic ;
    sel : out std_logic_vector (1 downto 0) ;
    load, clear : out std_logic ;
    clk : in std_logic
  ) ;
end controller ;

architecture rtl of controller is
  signal s, holdns, ns : states ;
  signal tmp : std_logic_vector (3 downto 0) ;
begin

  -- select next state
  with s select ns <=
    add_a when clr,
    add_b when add_a,
    add_c when add_b,
    add_d when add_c,
    hold when add_d,
    holdns when others ; -- hold

  -- next state if in hold state
  holdns <=
    clr when update = '1' else
    hold ;

  -- state register
  process(ns,clk)
  begin
    if clk'event and clk = '1' then
      s <= ns ;
    end if ;
  end process ;

  -- controller outputs
  with s select sel <=
    "00" when add_a,
    "01" when add_b,
    "10" when add_c,
    "11" when others ;

  load <= '0' when s = clr or s = hold else '1' ;

  clear <= '1' when s = clr else '0' ;

end rtl ;

```

```
-- package for datapath and controller
```

```

library ieee ;
use ieee.std_logic_1164.all ;
use work.averager_types.all ;

package averager_components is

component datapath
  port (
    a, b, c, d : in num ;
    sum : out num ;
    sel : in std_logic_vector (1 downto 0) ;
    load, clear, clk : in std_logic
  ) ;
end component ;

component controller
  port (
    update : in std_logic ;
    sel : out std_logic_vector (1 downto 0) ;
    load, clear : out std_logic ;
    clk : in std_logic
  ) ;
end component ;

end averager_components ;

```

The top-level averager entity instantiates the two components and interconnects them.

```

-- averager

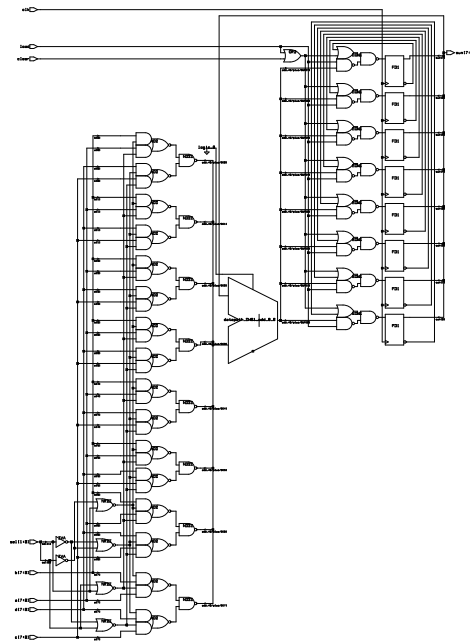
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_arith.all ;
use work.averager_types.all ;
use work.averager_components.all ;

entity averager is port (
  a, b, c, d : in num ;
  sum : out num ;
  update, clk : in std_logic ) ;
end averager ;

architecture rtl of averager is
  signal sel : std_logic_vector (1 downto 0) ;
  signal load, clear : std_logic ;
  -- other declarations (e.g. components) here
begin
  d1: datapath port map ( a, b, c, d, sum, sel, load,
                        clear, clk ) ;
  c1: controller port map ( update, sel, load,
                          clear, clk ) ;
end rtl ;

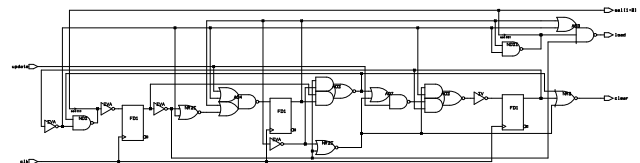
```

The result of the synthesizing the datapath is:

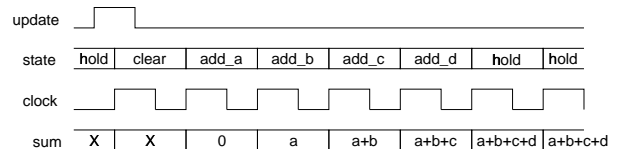


The register flip-flops are at the upper right, the adder is in the middle and the input multiplexer is at the lower left.

The result of the synthesizing the controller is:



The following timing diagram shows the datapath output and the controller state over one computation. Note that the state and output transitions take place on the rising edge of the clock. Also note that the output is updated at the *end* of the state in which a particular operation is performed.



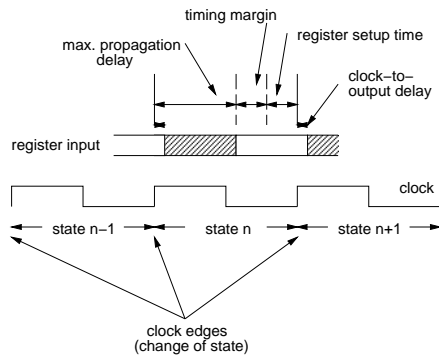
## RTL Timing Analysis

The datapath is a synchronous sequential circuit that uses the same clock for all registers and all register contents thus change at the same time. The controller uses the same clock as the datapath. Each datapath register loads the values “computed” during one state

at the end of that state (which is also the start of the next state).

We can guarantee that the correct results will be loaded into registers if the worst-case propagation delay ( $t_{PD}$ ) through any path of multiplexers and combinational function blocks is less than the clock period ( $t_{\text{clock}}$ ) minus the registers' setup time ( $t_s$ ) and clock-to-output ( $t_{CO}$ ) delays:

$$t_{PD} < t_{\text{clock}} - t_s - t_{CO}$$



Using a single clock means we only need to compute the delay through *combinational* logic blocks which is much simpler than having to deal with asynchronous clocks. This is why almost all large-scale digital circuits are synchronous designs.

Synthesis tools can be asked to synthesize logic that operates at a particular clock period. The synthesizer is supplied with the propagation delay specifications for the combinational logic components available in the particular technology being used and it will then try to arrange the logic so that the propagation delay from any input or register output to the inputs of all registers is less than the clock period. This ensures that the circuit will work properly at the specified clock rate.

## Behavioural Synthesis

It is possible to work at even higher levels of abstraction than RTL when design time is more important than cost. Advanced synthesis programs (for example, Synopsys' Behavioral Compiler) can convert a behavioral description of an algorithm into an RTL description. The compiler does this by automatically allocating registers and partitioning the processing over as many clock cycles as are required to meet high-level processing time requirements.