Performance-oriented (same 3-stage latency, 1 result/cycle)

Math: AvgTxLen_new = (AvgTxLen*InstExed + CurTxLen) / (InstExed+1) Replace ÷(InstExed+1) with × recipQ16(InstExed+1) and a right shift by 16 (with rounding).

Schematic (ASCII)

```
Stage 1 Stage 2 Stage 3

Avg \rightarrow 8x8 MUL \rightarrow prod[15:0] \rightarrow 17b ADD \rightarrow num[16:0] \rightarrow 17x16 MULT \rightarrow prodQ[32:0]

Inst \rightarrow add + round \rightarrow >>16 \rightarrow Avg_new[7:0]

\rightarrow +1 \rightarrow den[8:0] \rightarrow 256x16 ROM \rightarrow recipQ16[15:0] \rightarrow InstExed_new = den[7:0]

(1/den in Q16)

Regs between stages (all signals registered at stage boundaries)
```

Key blocks

- Stage 1: mul_s1 = Avg×Inst , den_s1 = Inst+1 , pipeline Cur.
- Stage 2: num = mul_s1 + Cur .
- Stage 3: q = ((num * recipQ16(den)) + 2^15) >> 16 (rounded), register outputs.

Why it helps: Pushing the slow, wide divider off the critical path eliminates the long carry-propagate network that caused GLS setup violations in the synthesized version with 100MHz Clock. In current design, we reduced the clock frequency to 25 MHz to meet timing. A 17×16 multiplier + add/shift maps much better to std-cell multipliers (or DW02_mult) and retimes cleanly. The 256×16 ROM is tiny (≈4 Kb) and becomes simple combinational logic. In practice on 45 nm cells, this swap typically boosts Fmax (often $1.3-2\times$ over a single-stage divider), reduces dynamic power (shorter chains toggle), and can even lower area because the divider is expensive while one small ROM + 1 extra multiplier is modest for 8-bit outputs.

The synthesized ALU's critical path is dominated by the Stage-3 integer divider, which is difficult to close at high frequencies in 45 nm std-cell flows. Replacing this block with a pipelined reciprocal×multiply (Q16 LUT + 17×16 multiplier + round/shift) preserves the arithmetic to 8-bit precision while transforming the critical path into multiplier and short adder logic that retimes and scales cleanly. The reciprocal LUT is tiny $(256 \times 16 \approx 4 \text{ Kb})$, the pipeline depth remains three stages with **one result per cycle**, and the divider's long carry chains—and their associated timing violations and power—are eliminated. This redesign typically increases Fmax, reduces toggling on long paths, and can even lower total area versus a wide combinational divider, delivering a better PPA point without changing the external interface or latency.

Area-oriented (multi-cycle divider, smaller area)

If throughput (1 result/cycle) isn't mandatory, keep the original math but replace DIV with a small **sequential divider** (e.g., restoring or DW_div_seq) and add a tiny valid/ready handshake.

Schematic (ASCII)

Why it helps: A small iterative divider plus a controller is much smaller than a fast combinational divider. You trade CPI (e.g., one result every ~8–9 cycles) for area/power. Great for resource-constrained designs or when this ALU isn't on the top-throughput path.