

# SEMICONDUCTOR DEVICES

## $p$ - $n$ Junctions: Part 2

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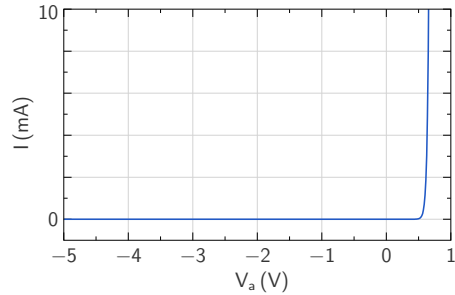
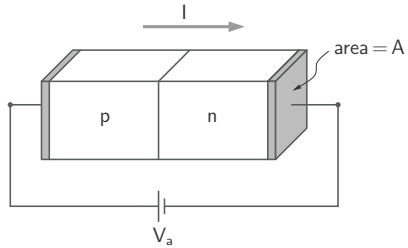
M. B. Patil

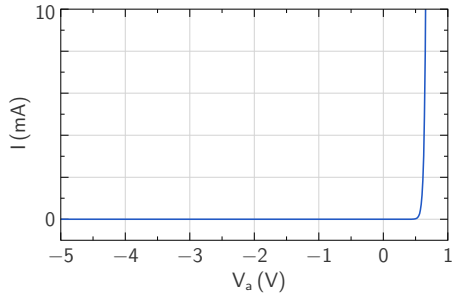
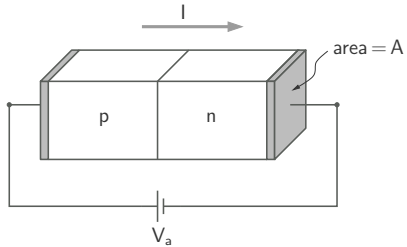
[mbpatil@ee.iitb.ac.in](mailto:mbpatil@ee.iitb.ac.in)

[www.ee.iitb.ac.in/~sequel](http://www.ee.iitb.ac.in/~sequel)

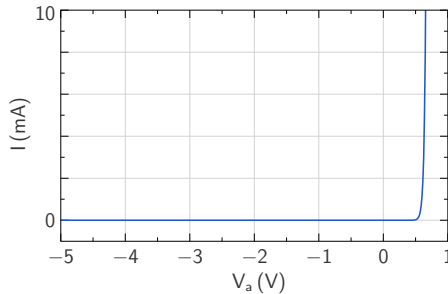
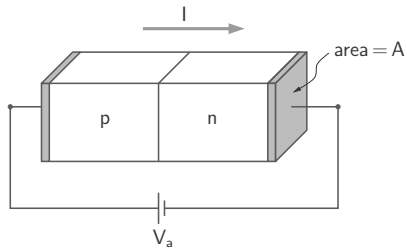
Department of Electrical Engineering  
Indian Institute of Technology Bombay

## $pn$ junction under bias

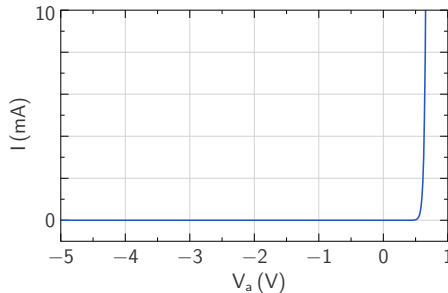
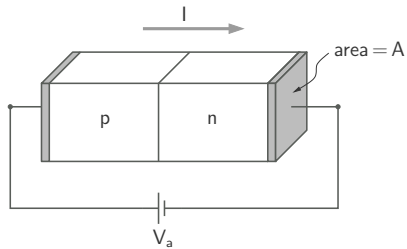




- \* With  $V_a \approx 0.6$  V a substantial current flows. When  $V_a$  is increased further, the current increases rapidly.

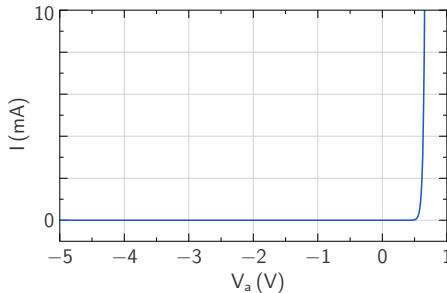
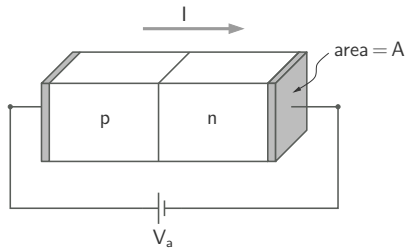


- \* With  $V_a \approx 0.6$  V a substantial current flows. When  $V_a$  is increased further, the current increases rapidly.
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We want to understand this “rectifying” behaviour.



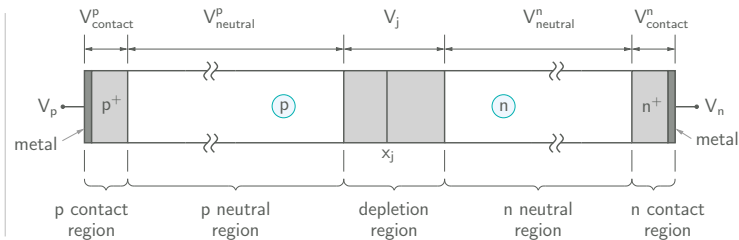
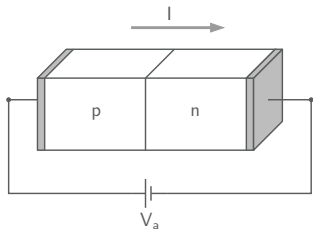
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We want to understand this “rectifying” behaviour.

- \* As we increase the forward bias, the current increases rapidly, and at some point, the device will get damaged because of overheating. For silicon diodes used in low-power applications, the forward voltage must be restricted to about 0.8 V.

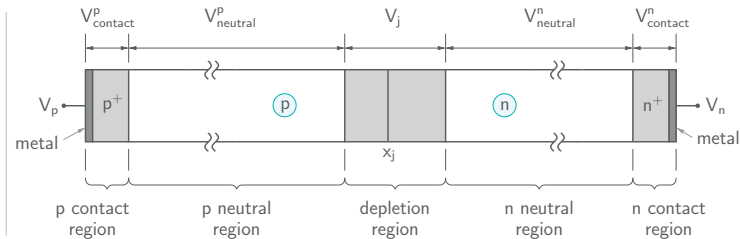
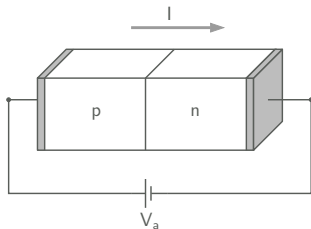
(Note: Although we will show an applied forward/reverse bias with a battery, in practice, a battery is generally not connected directly across a diode.)

## Where does the voltage drop?



Consider a *pn* junction in equilibrium ( $V_a = 0$  V).

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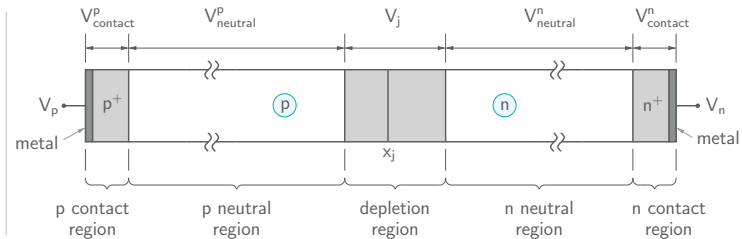
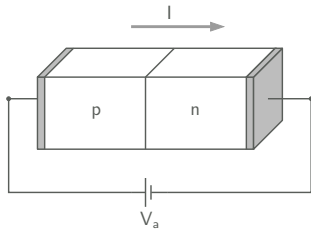


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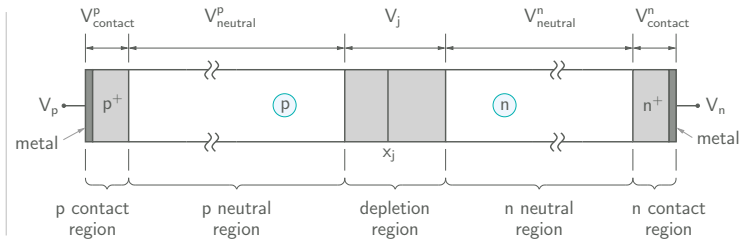
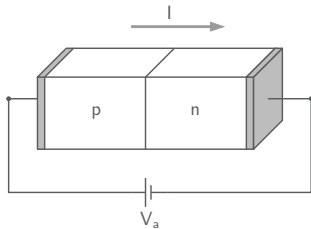
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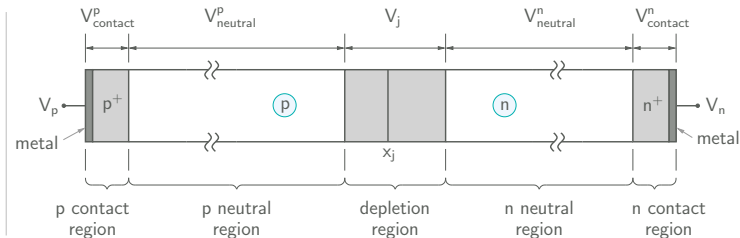
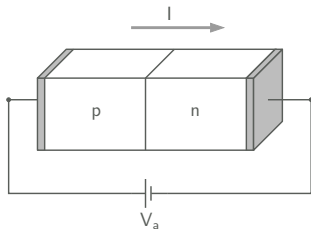
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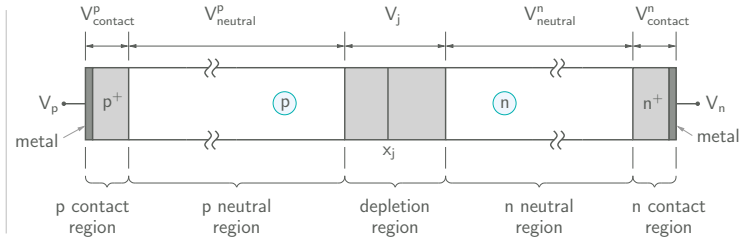
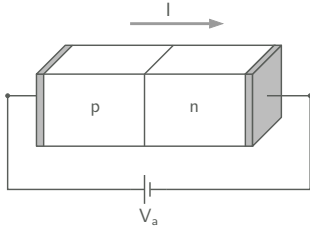
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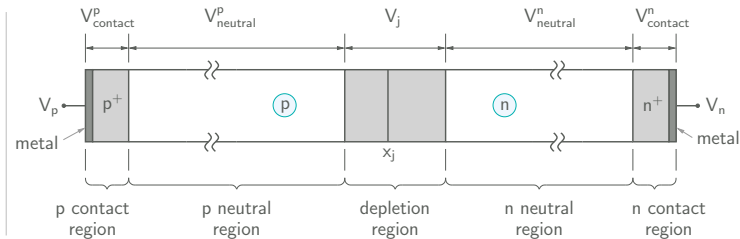
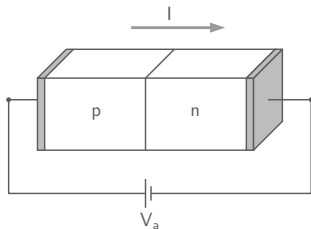
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- \* Even with current flow,  $V_{\text{neutral}}^p$  and  $V_{\text{neutral}}^n$  remain negligibly small since a very small electric field is sufficient to create the required  $J_p^{\text{drift}} = q\mu_p \mathcal{E}$  or  $J_n^{\text{drift}} = q\mu_n \mathcal{E}$  (note that *p* and *n* in these equations represent the *majority* carrier densities).

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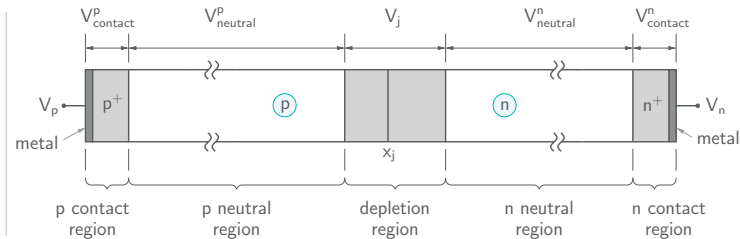
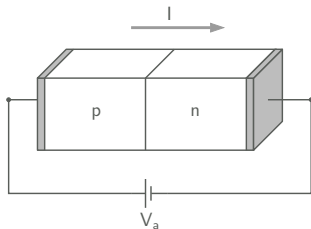


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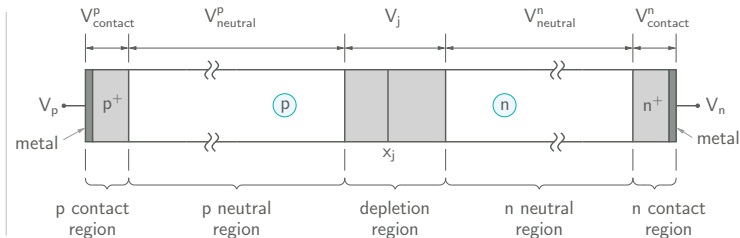
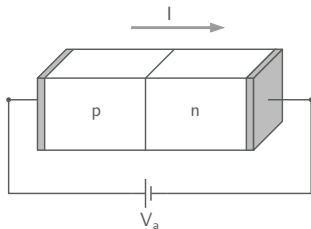
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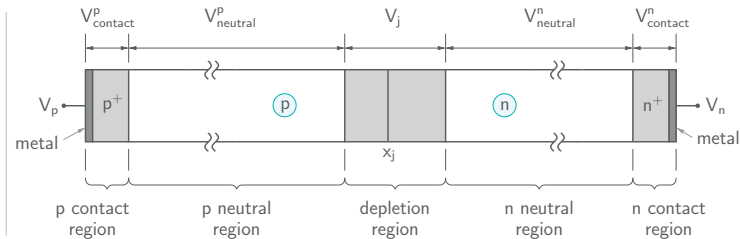
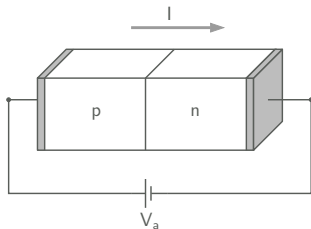
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- \* In equilibrium,  $V_p = V_n$ , and we get  
 (1):  $V_{\text{contact}}^p - V_{bi} + V_{\text{contact}}^n = 0$ , taking voltage drop as positive.  
 (We assume that the signs of  $V_{\text{contact}}^p$  and  $V_{\text{contact}}^n$  are taken into account.)

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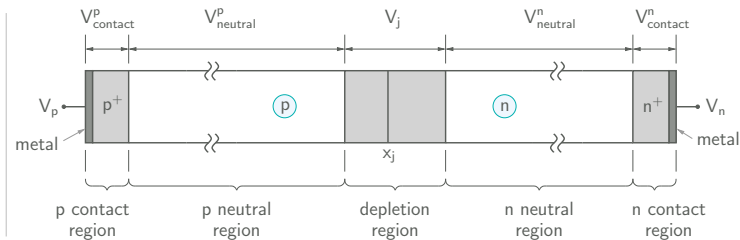
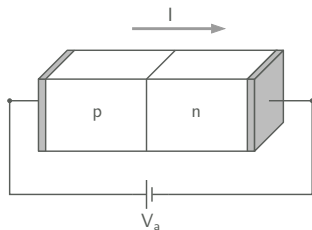
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- \* (1) – (2) gives  $-V_{bi} + V_j = -V_a$ , i.e.,  $V_j = V_{bi} - V_a$

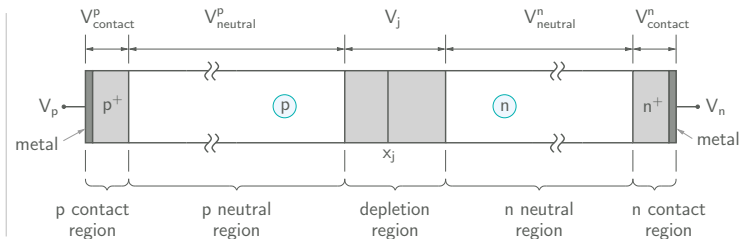
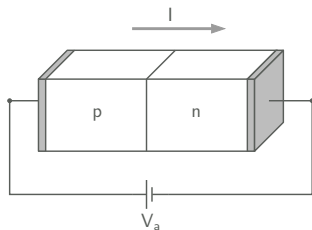


## Example: forward bias



For an abrupt silicon  $pn$  junction, the built-in voltage is  $V_{bi} = 0.85 \text{ V}$ . Let  $W_0$  and  $W_1$  denote the depletion widths for  $V_a = 0 \text{ V}$  and  $V_a = 0.6 \text{ V}$ , respectively. What is  $W_1/W_0$ ?

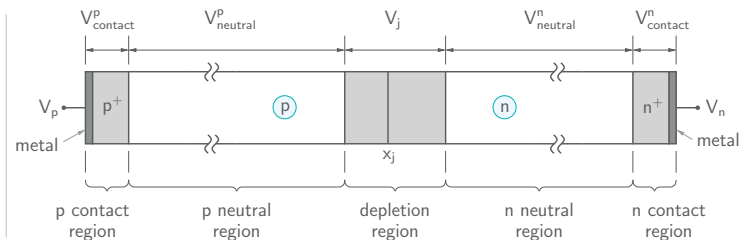
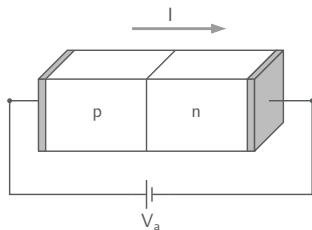
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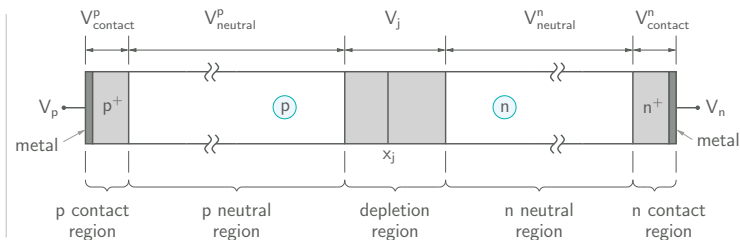
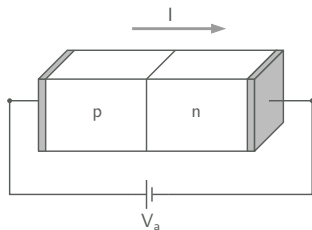


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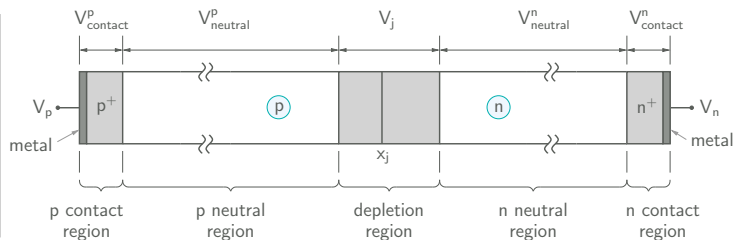
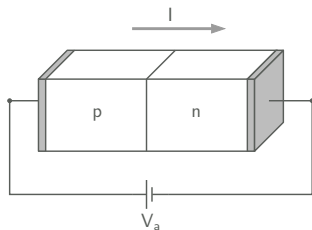
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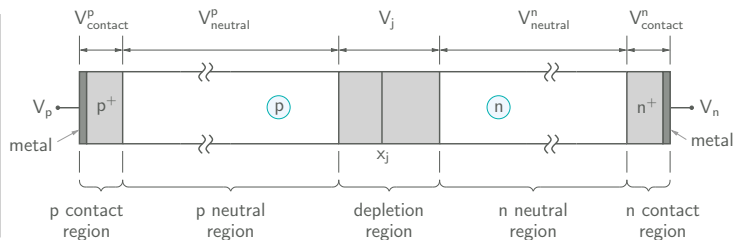
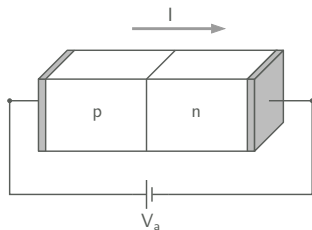
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$$V_{bi} - 0.6 \text{ V} = kW_1^2 \quad (\text{for } V_a = 0.6 \text{ V})$$

$$\rightarrow \frac{0.85 - 0.6}{0.85} = \left( \frac{W_1}{W_0} \right)^2 \rightarrow \frac{W_1}{W_0} = 0.54.$$

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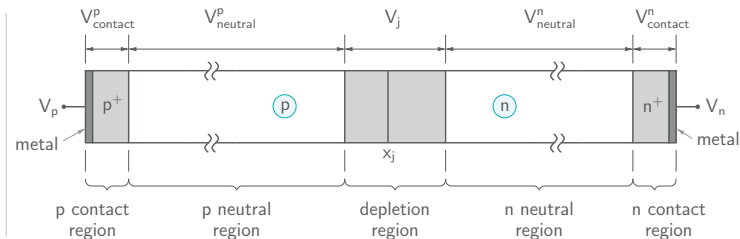
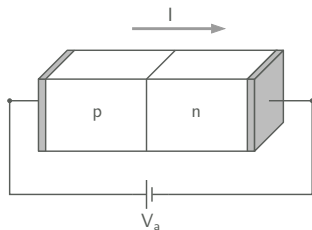
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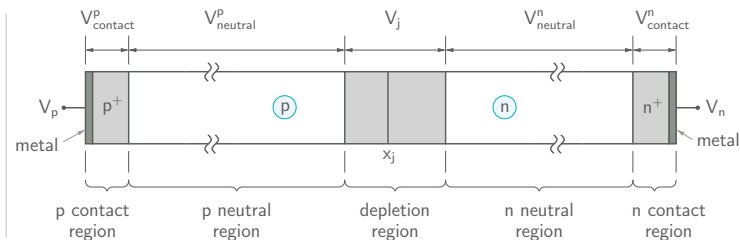
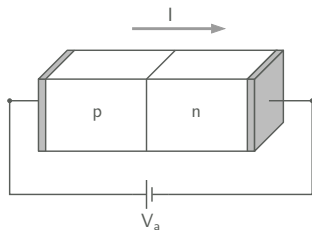
Application of a forward bias of  $0.6 \text{ V}$  causes the depletion region to shrink by a factor 0.54.

## Example: reverse bias



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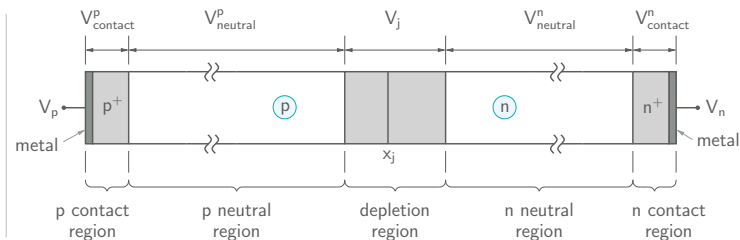
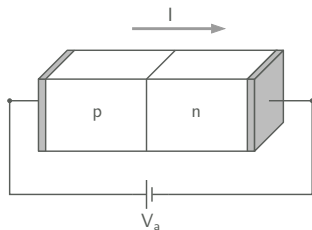


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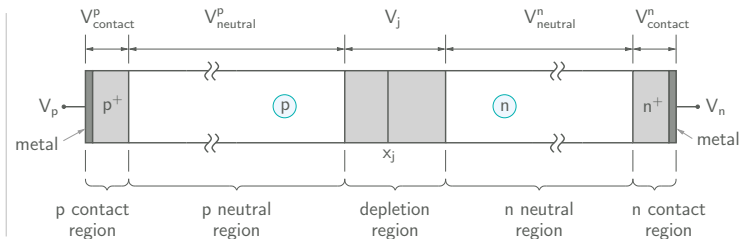
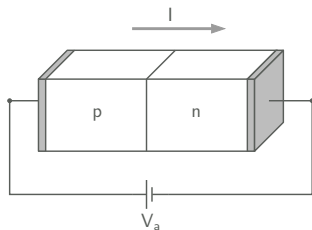


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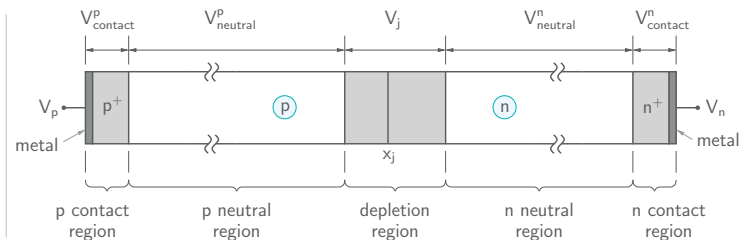
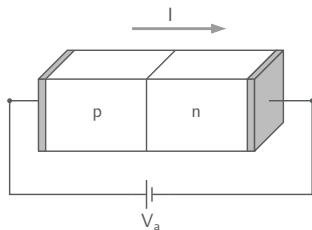
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## Example: reverse bias



For an abrupt silicon  $pn$  junction, the built-in voltage is  $V_{bi} = 0.85 \text{ V}$ . Let  $W_0$  and  $W_1$  denote the depletion widths for  $V_a = 0 \text{ V}$  and  $V_a = -2 \text{ V}$  (i.e., a reverse bias  $V_R$  of  $2 \text{ V}$ ), respectively. What is  $W_1/W_0$ ?

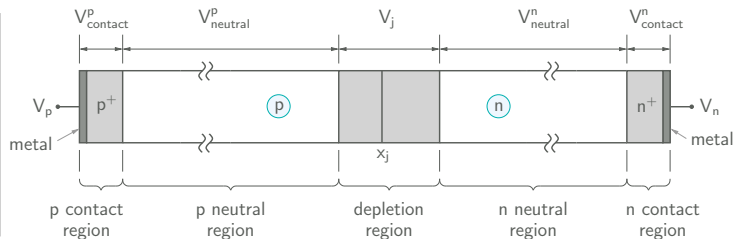
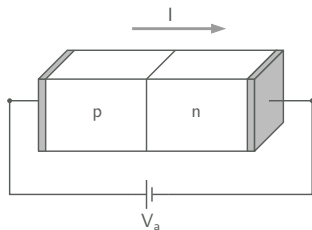
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$$\rightarrow \frac{0.85 + 2}{0.85} = \left( \frac{W_1}{W_0} \right)^2 \rightarrow \frac{W_1}{W_0} = 1.83.$$

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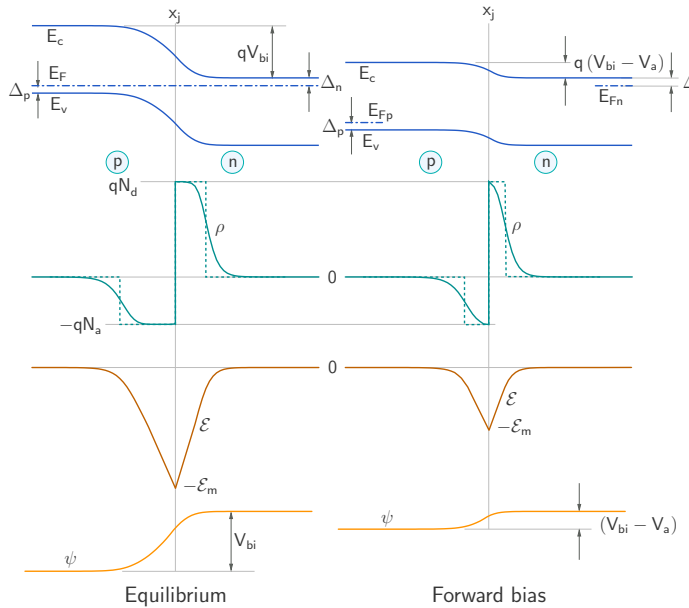
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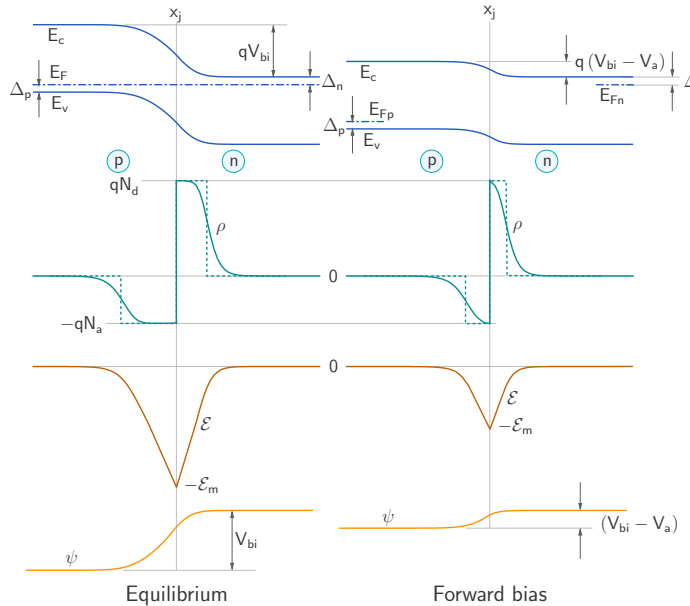
Application of a reverse bias of  $2 \text{ V}$  causes the depletion region to expand by a factor  $1.83$ .

# Forward bias



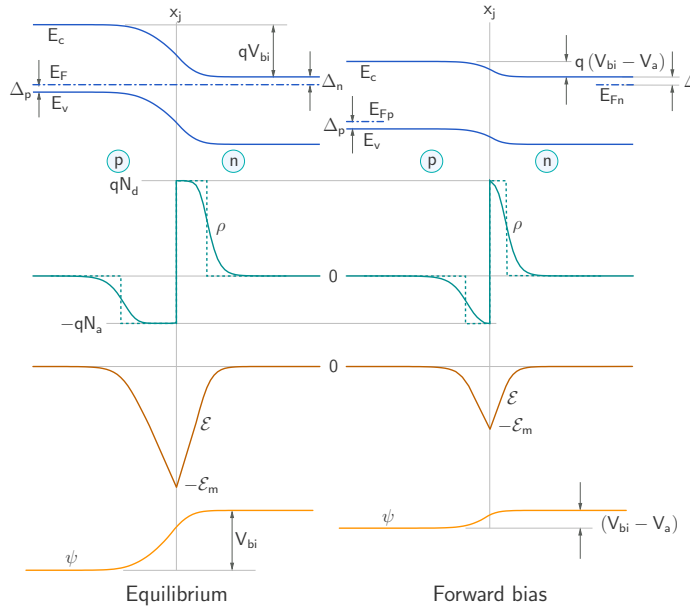
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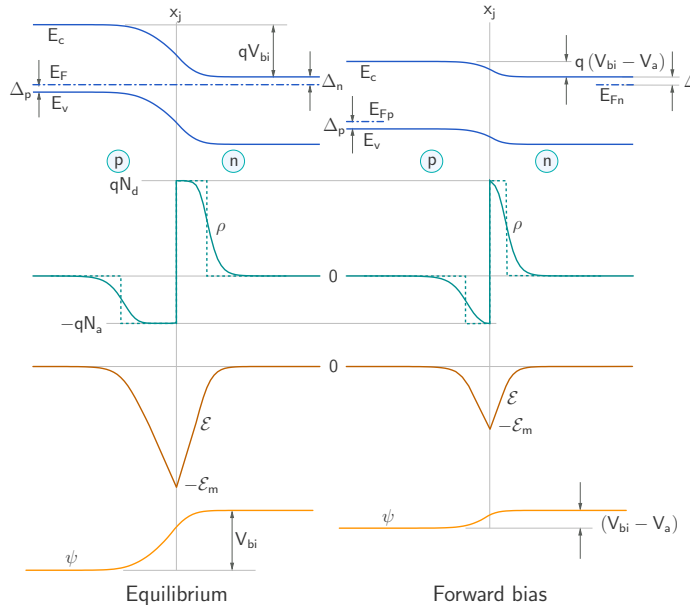


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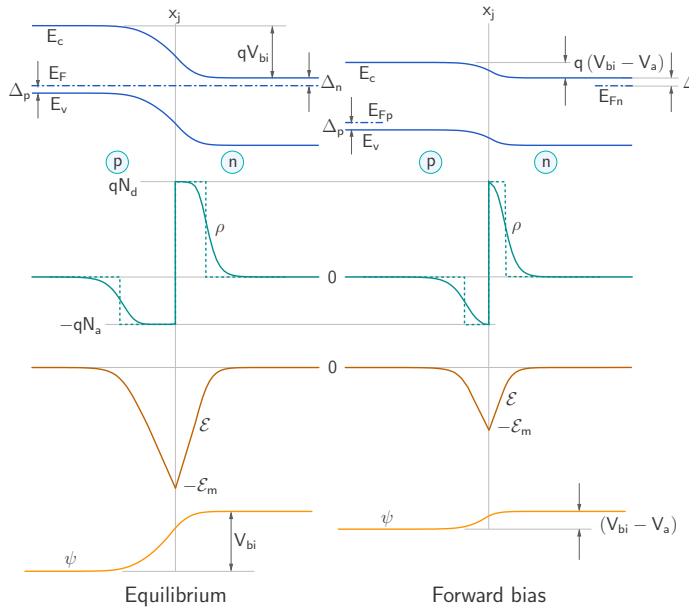
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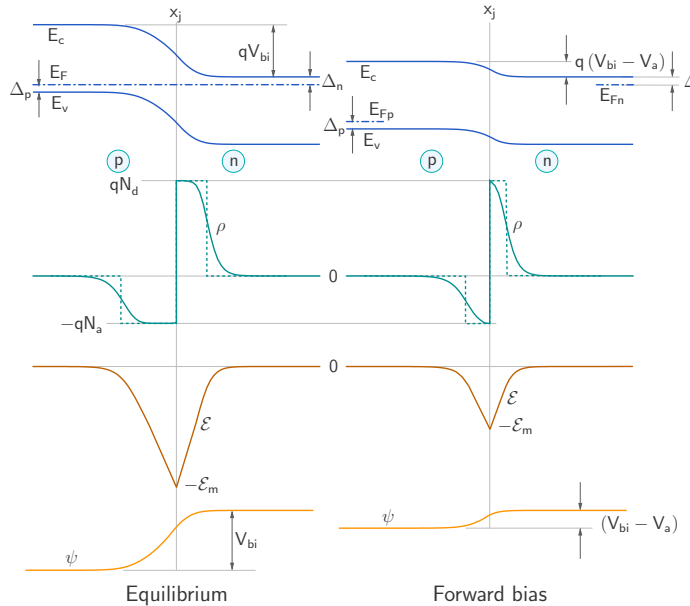


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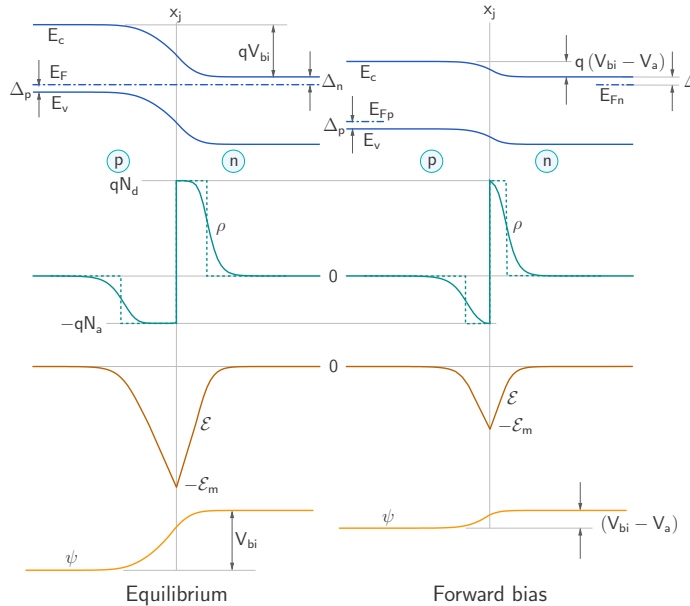
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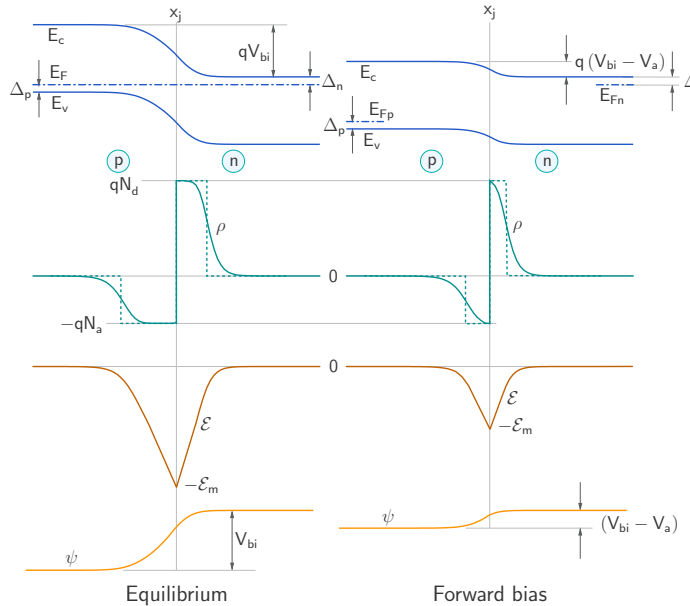
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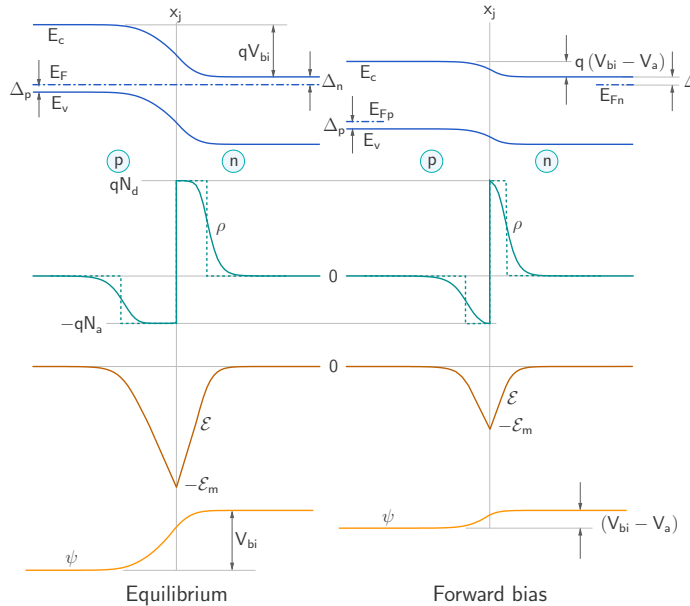
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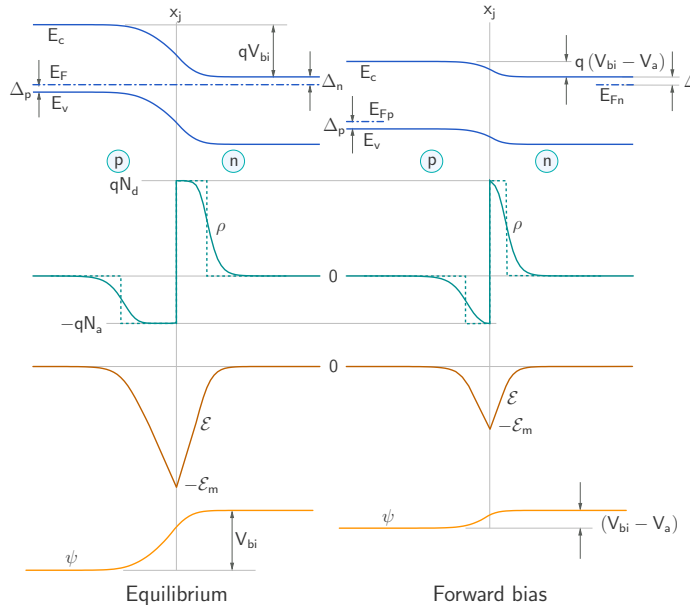
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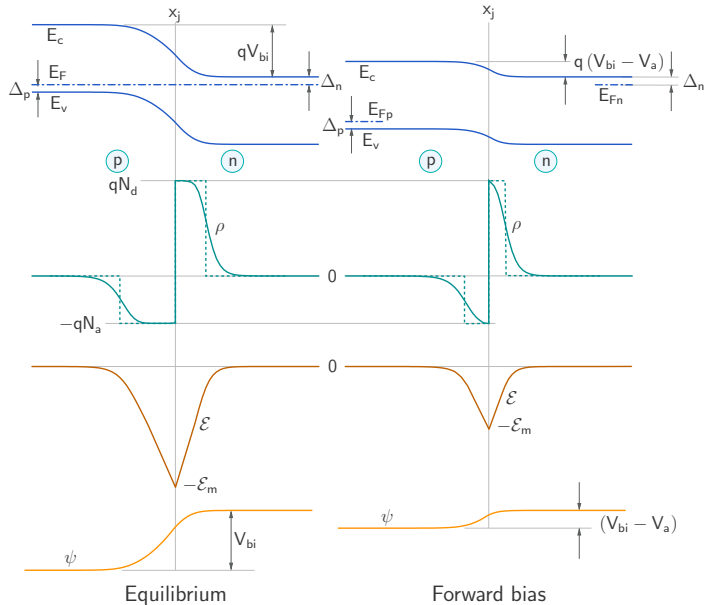
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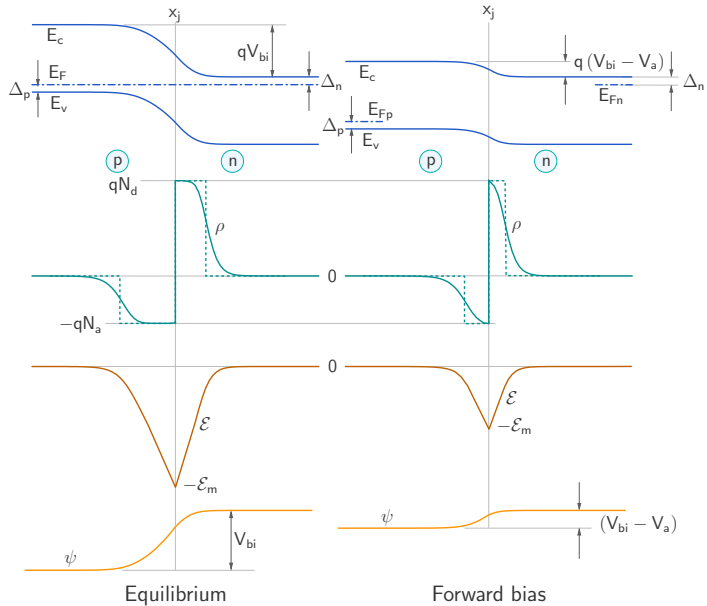
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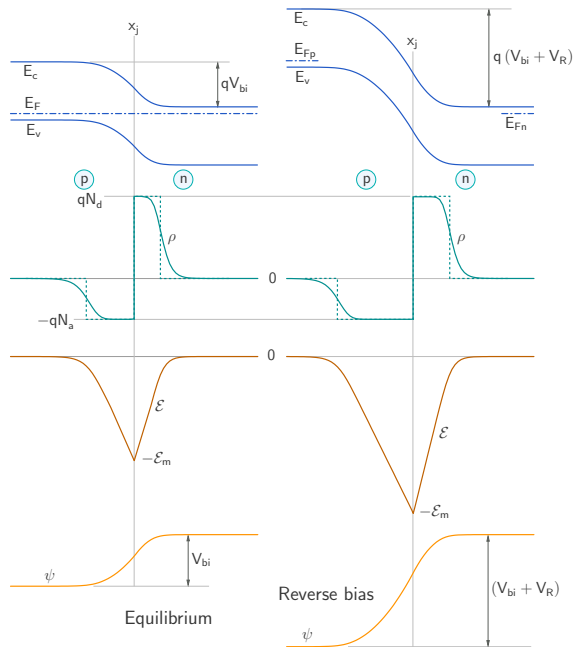
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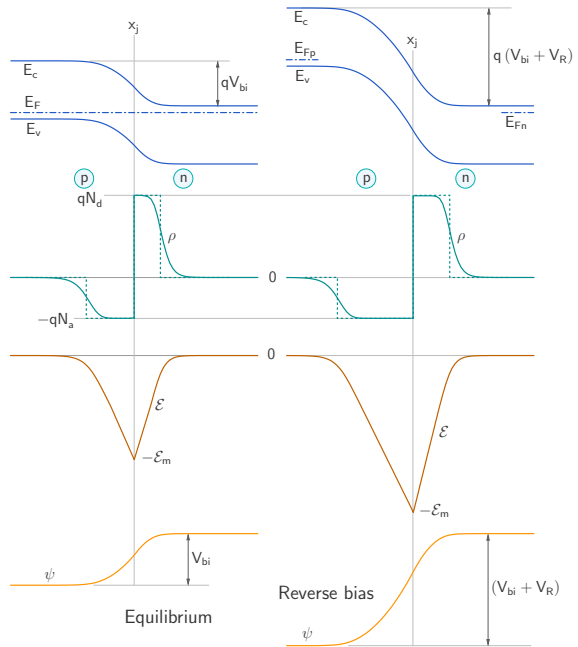


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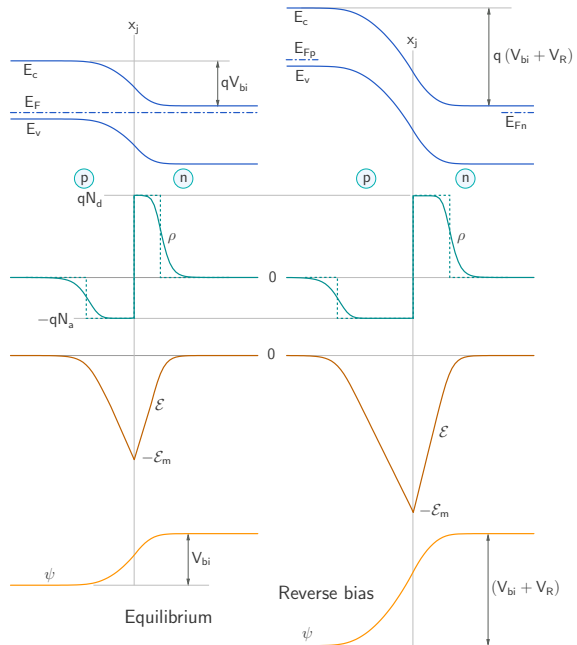
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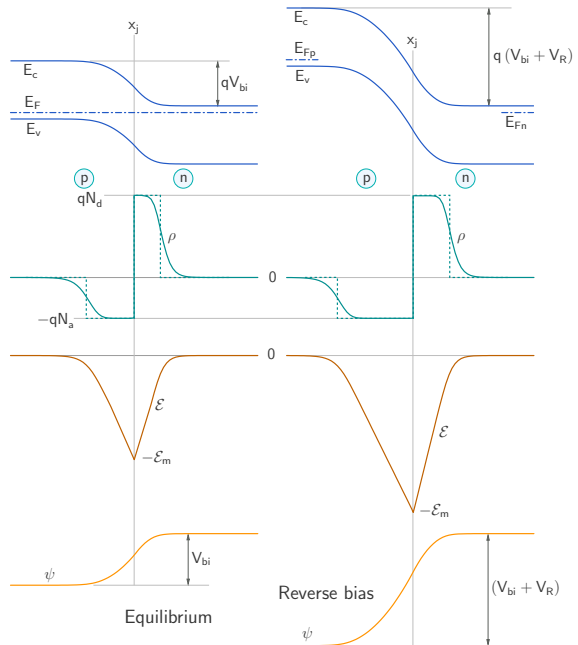


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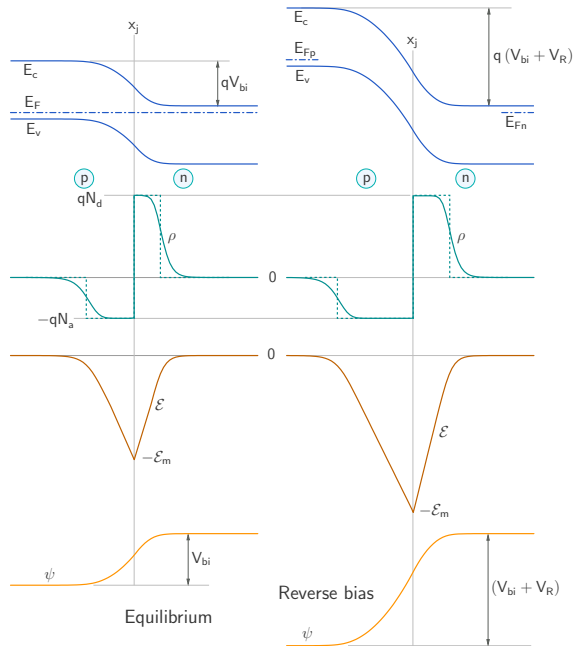
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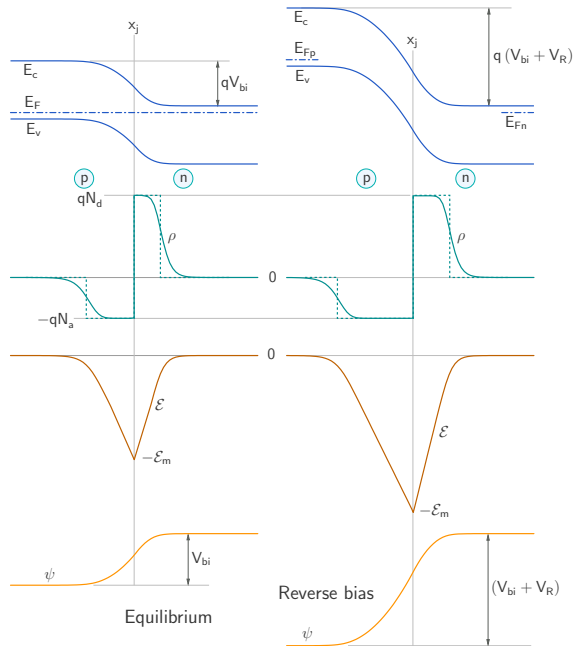


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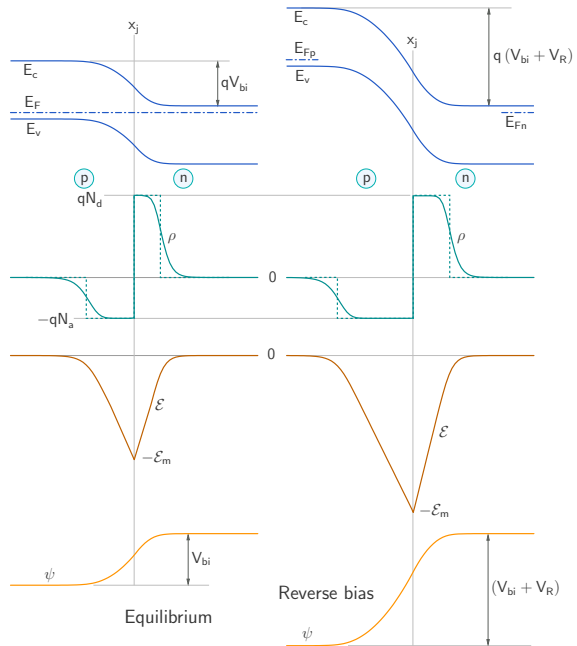
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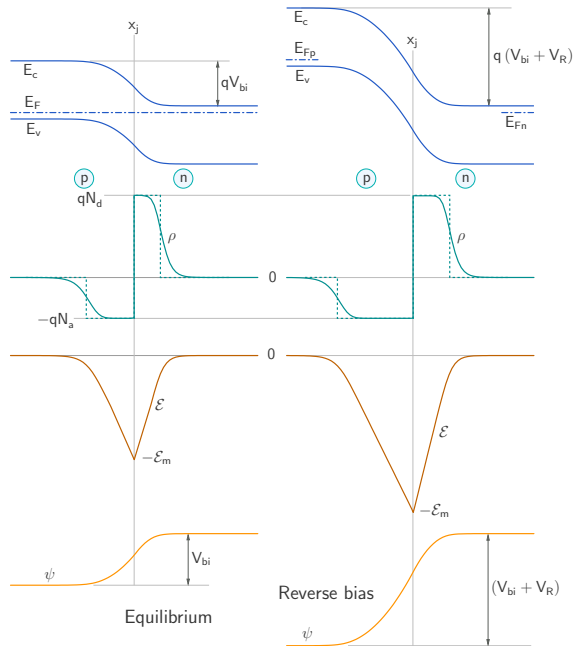
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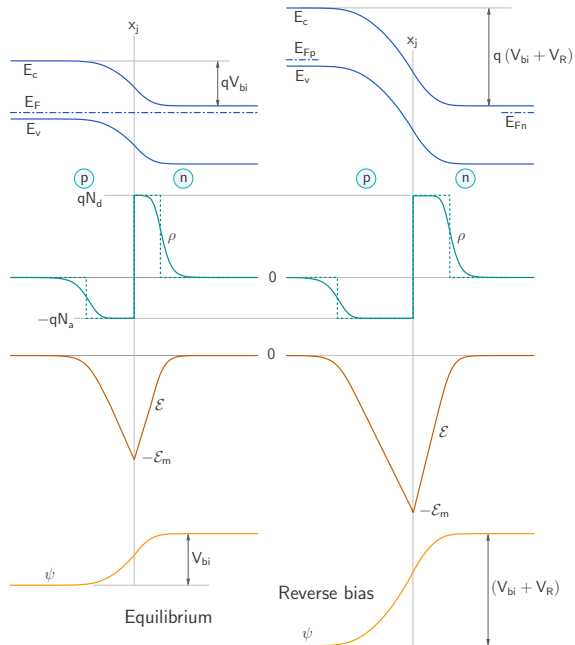
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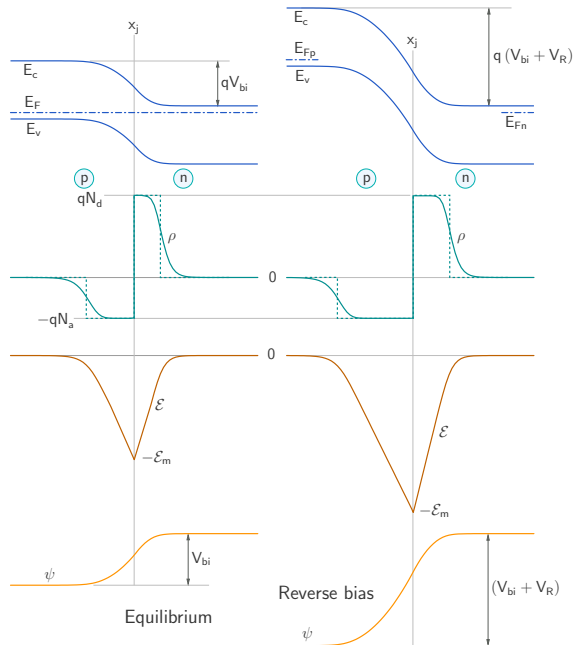
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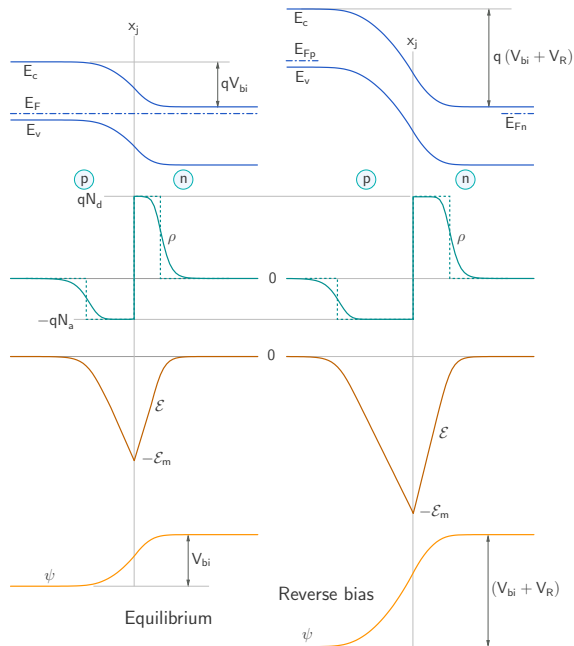
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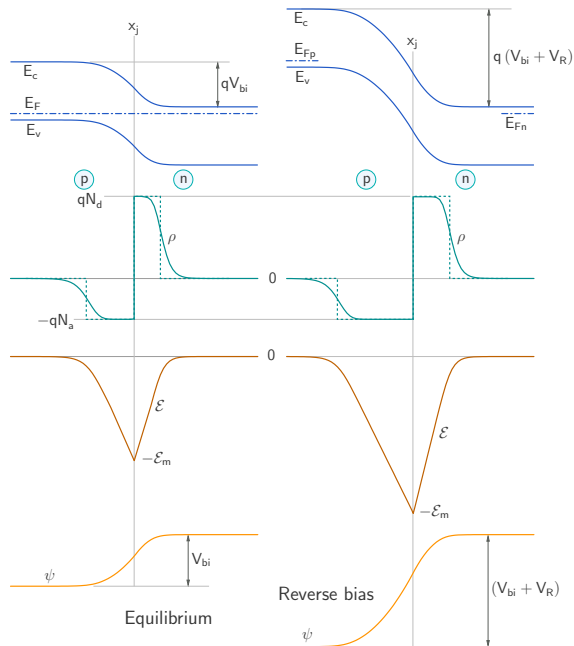
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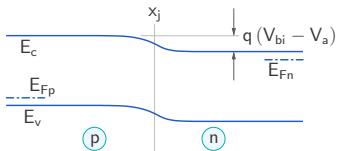
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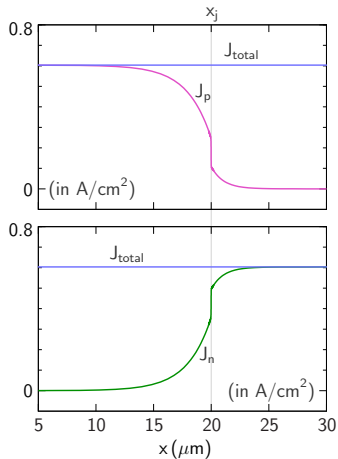
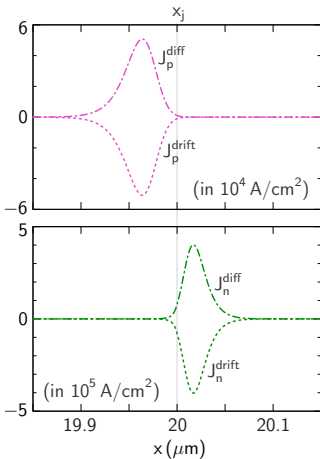
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# Current densities in forward bias

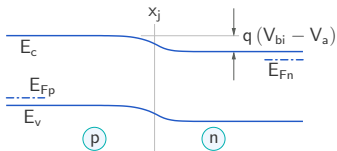


$N_d = 2 \times 10^{17} \text{ cm}^{-3}$	$\tau_n = 1 \text{ ns}$
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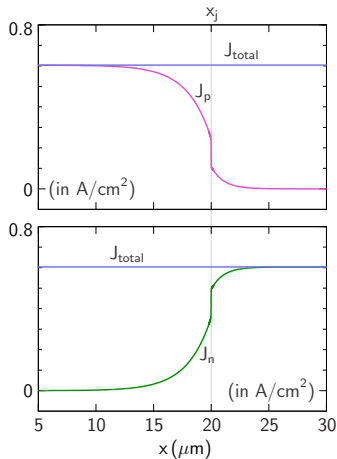
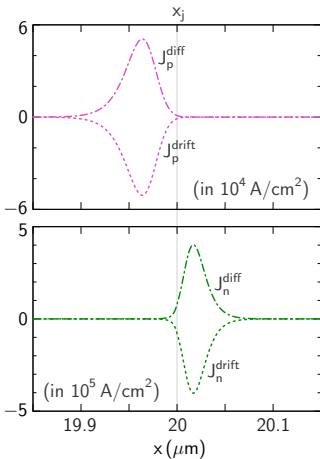


Near the junction,

# Current densities in forward bias



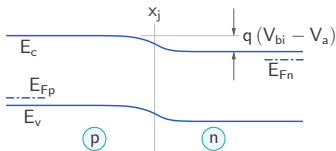
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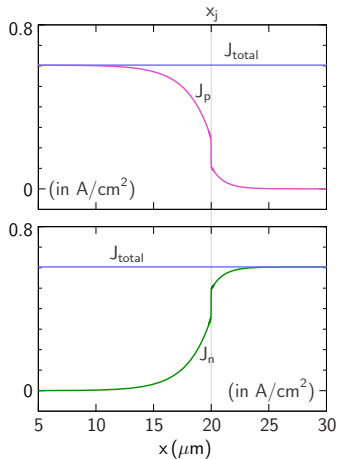
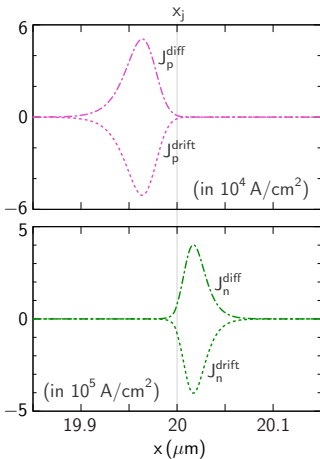
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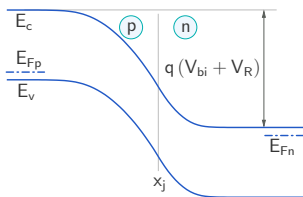
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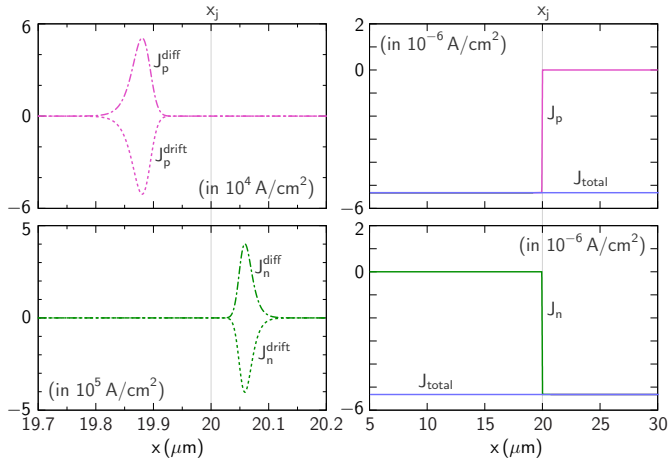
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- \* The net current densities  $J_n$  and  $J_p$  are much smaller than the drift and diffusion components.

# Current densities in reverse bias



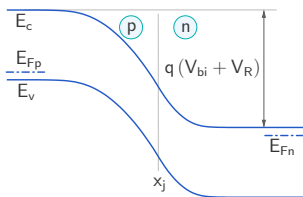
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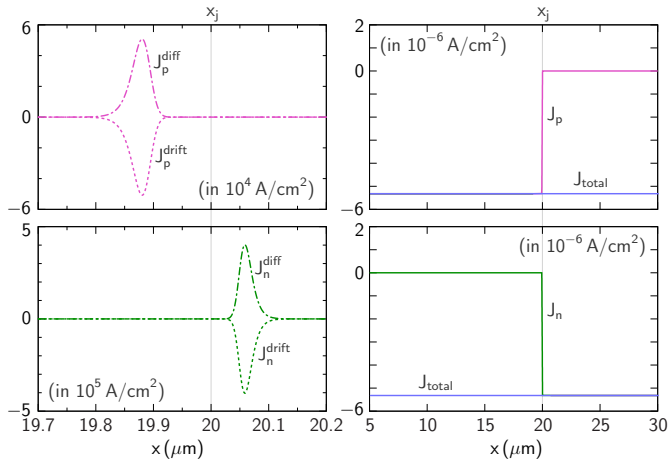




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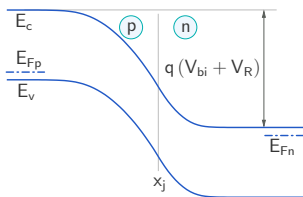
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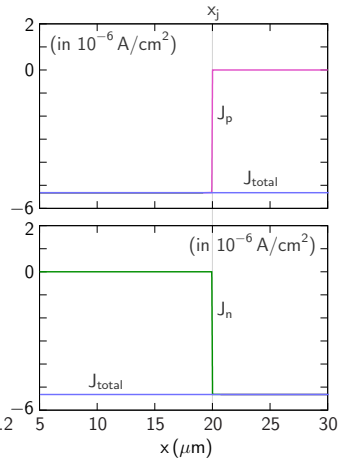
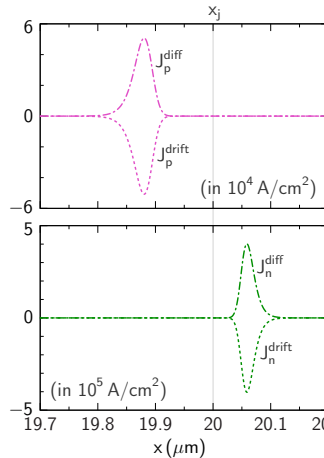
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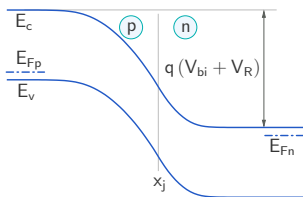
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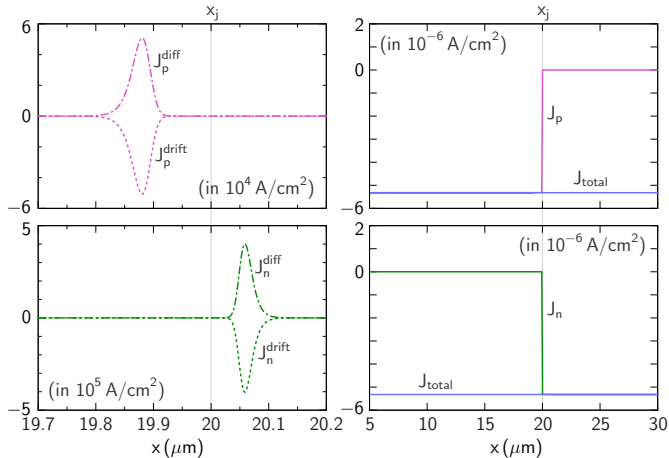
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- \* Although the equilibrium condition is disturbed, we still have  $J_p^{\text{diff}} \approx -J_p^{\text{drift}}$ , and  $J_n^{\text{diff}} \approx -J_n^{\text{drift}}$ .
- \* The net current densities  $J_n$  and  $J_p$  are much smaller than the drift and diffusion components.

# Current densities in reverse bias



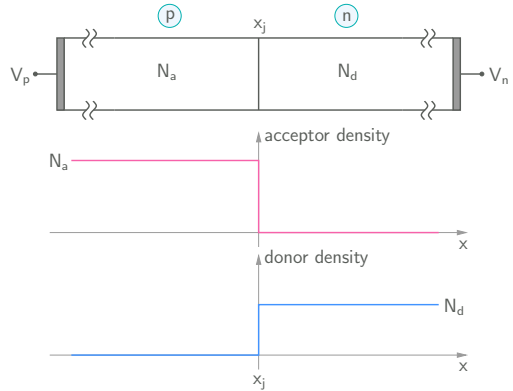
$N_d = 2 \times 10^{17} \text{ cm}^{-3}$	$\tau_n = 1 \text{ ns}$
$N_a = 10^{17} \text{ cm}^{-3}$	$\tau_p = 1 \text{ ns}$
$\mu_n = 1400 \text{ cm}^2/\text{V-s}$	$T = 300 \text{ K}$
$\mu_p = 500 \text{ cm}^2/\text{V-s}$	$V_a = -1 \text{ V}$



Near the junction,

- \* Although the equilibrium condition is disturbed, we still have  $J_p^{\text{diff}} \approx -J_p^{\text{drift}}$ , and  $J_n^{\text{diff}} \approx -J_n^{\text{drift}}$ .
- \* The net current densities  $J_n$  and  $J_p$  are much smaller than the drift and diffusion components.
- \* Note that  $J_{\text{total}}$  in reverse bias is negligibly small compared to the forward bias case ( $0.7 \text{ A/cm}^2$  for  $V_a = 0.6 \text{ V}$ ). For all practical purposes, we can say that the current is zero for reverse bias.

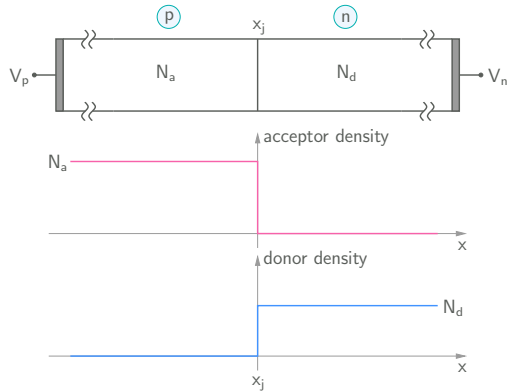
Definitions:



## $pn$ junction: derivation of $I$ - $V$ equation

Definitions:

$p_{p0}$ : equilibrium hole density in the neutral  $p$ -region

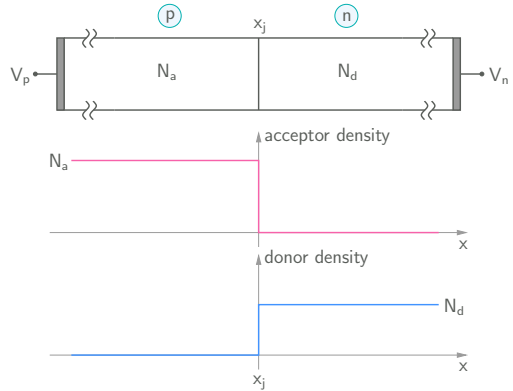


## $pn$ junction: derivation of $I$ - $V$ equation

Definitions:

$p_{p0}$ : equilibrium hole density in the neutral  $p$ -region

$p_{n0}$ : equilibrium hole density in the neutral  $n$ -region



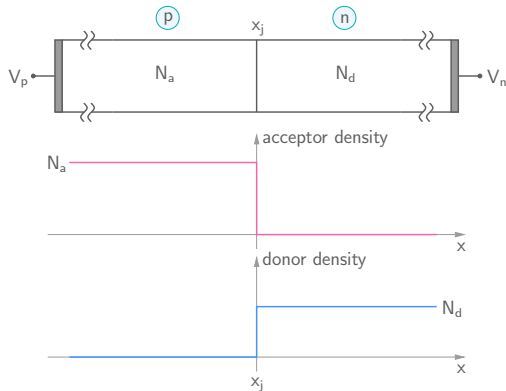
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## $pn$ junction: derivation of $I$ - $V$ equation

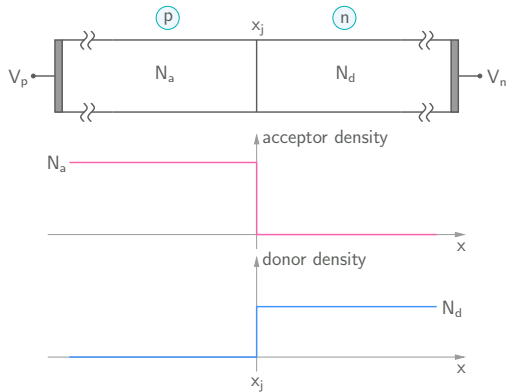
Definitions:

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## $pn$ junction: derivation of $I$ - $V$ equation

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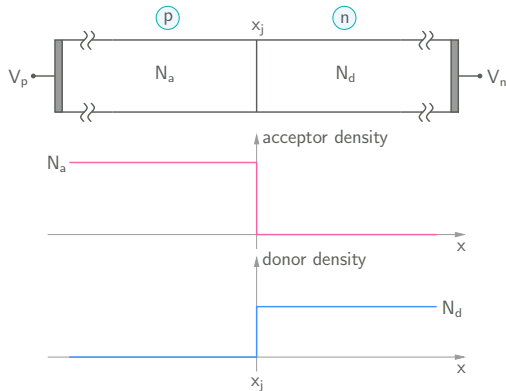
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$p_{p0}$  and  $n_{n0}$  are majority carrier densities.



## $pn$ junction: derivation of $I$ - $V$ equation

### Definitions:

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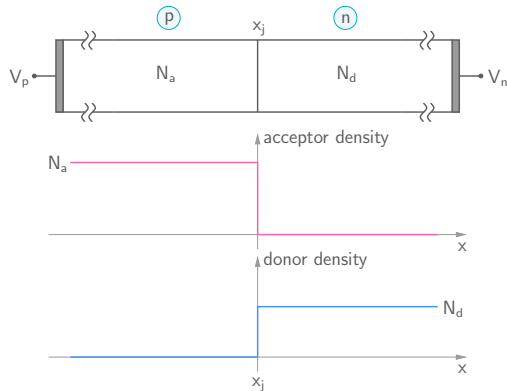
$p_{n0}$ : equilibrium hole density in the neutral  $n$ -region

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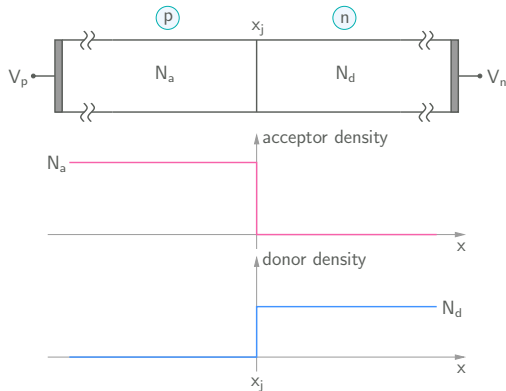
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Example:  $N_a = 5 \times 10^{16} \text{ cm}^{-3}$ ,  $N_d = 10^{18} \text{ cm}^{-3}$  ( $T = 300 \text{ K}$ ).



## $pn$ junction: derivation of $I$ - $V$ equation

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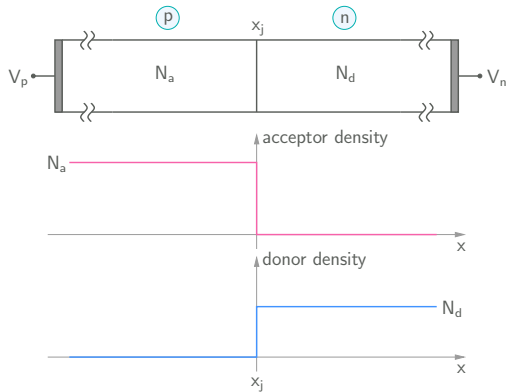
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$\rightarrow p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3}$ ,



## *pn* junction: derivation of *I*-*V* equation

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$p_{n0}$ : equilibrium hole density in the neutral *n*-region

$n_{p0}$ : equilibrium electron density in the neutral *p*-region

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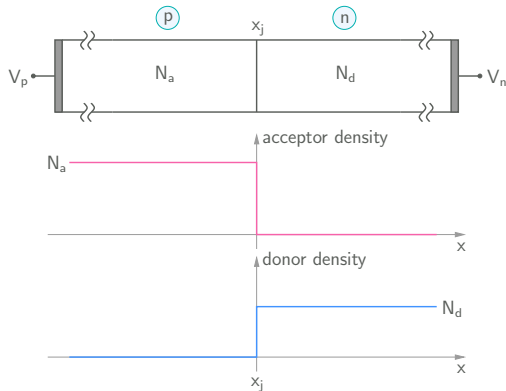
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$\rightarrow p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3}$ ,

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## *pn* junction: derivation of *I*-*V* equation

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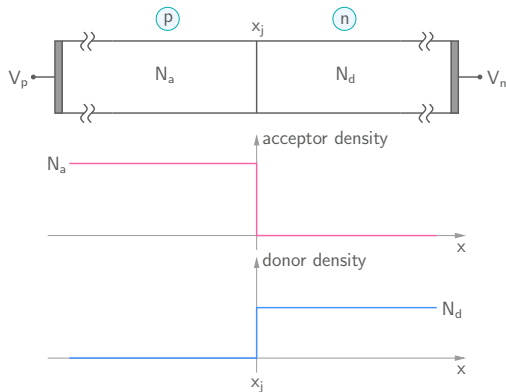
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Example:  $N_a = 5 \times 10^{16} \text{ cm}^{-3}$ ,  $N_d = 10^{18} \text{ cm}^{-3}$  ( $T = 300 \text{ K}$ ).

$$\rightarrow p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3},$$

$$n_{n0} \approx N_d = 10^{18} \text{ cm}^{-3},$$

$$n_{p0} = \frac{n_i^2}{p_{p0}} = \frac{(1.5 \times 10^{10})^2}{5 \times 10^{16}} = 4.5 \times 10^3 \text{ cm}^{-3},$$



## $pn$ junction: derivation of $I$ - $V$ equation

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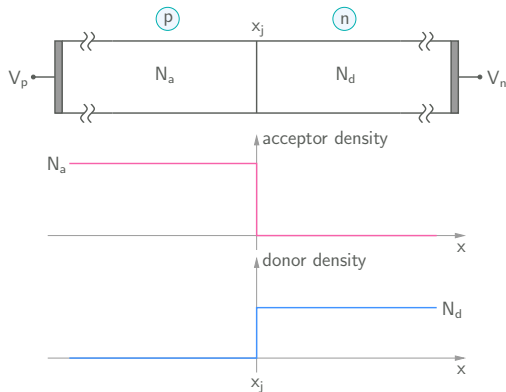
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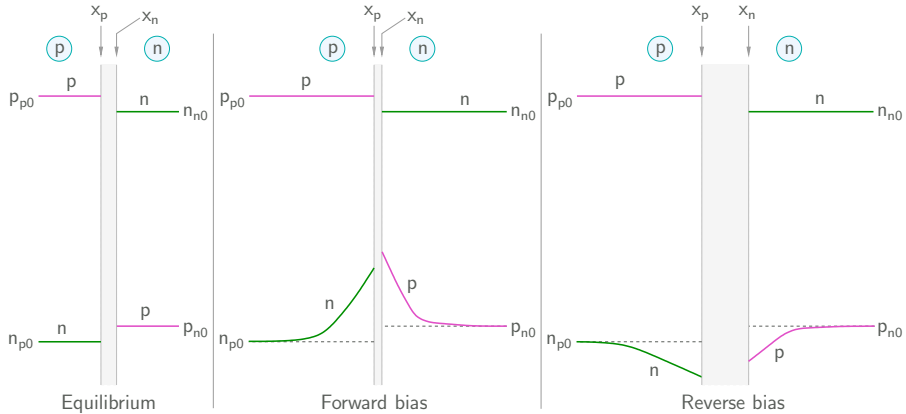
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$$p_{n0} = \frac{n_i^2}{n_{n0}} = \frac{(1.5 \times 10^{10})^2}{10^{18}} = 2.25 \times 10^2 \text{ cm}^{-3}.$$

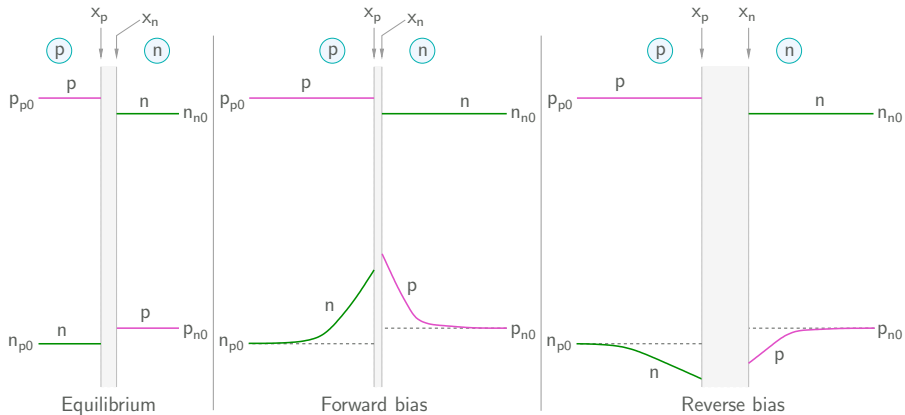


## $pn$ junction: derivation of $I$ - $V$ equation



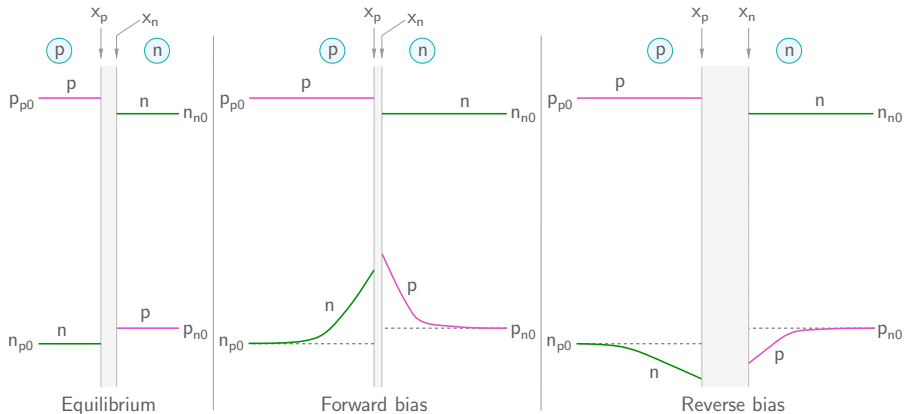


## $pn$ junction: derivation of $I$ - $V$ equation



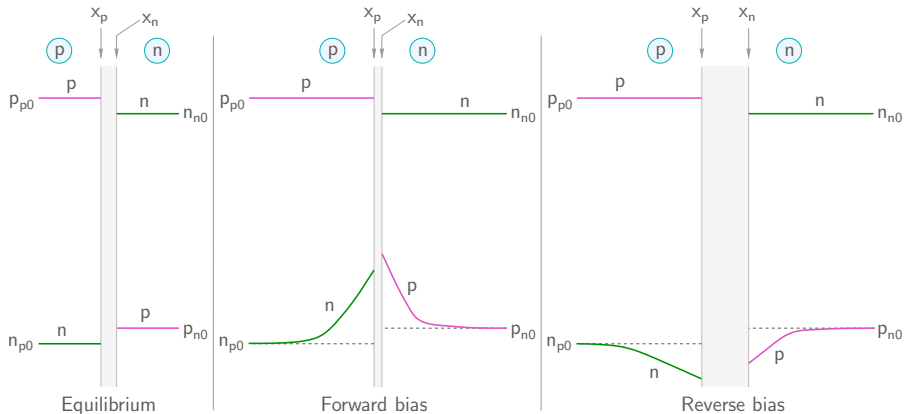
\* Since  $V_j = V_{bi} - V_a$  and  $W \propto \sqrt{V_j}$ , the depletion region is narrower under forward bias, wider under reverse bias.

## $pn$ junction: derivation of $I$ - $V$ equation



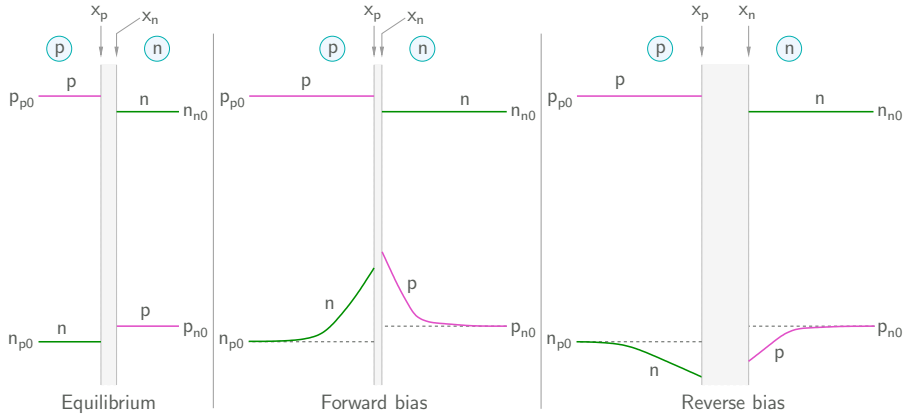
- \* Since  $V_j = V_{bi} - V_a$  and  $W \propto \sqrt{V_j}$ , the depletion region is narrower under forward bias, wider under reverse bias.
- \* Equilibrium concentrations (note: log scale for  $n$  and  $p$ ):  
 $p = p_{p0}$ ,  $n = n_{p0}$  in the neutral  $p$ -region.

## $pn$ junction: derivation of $I$ - $V$ equation



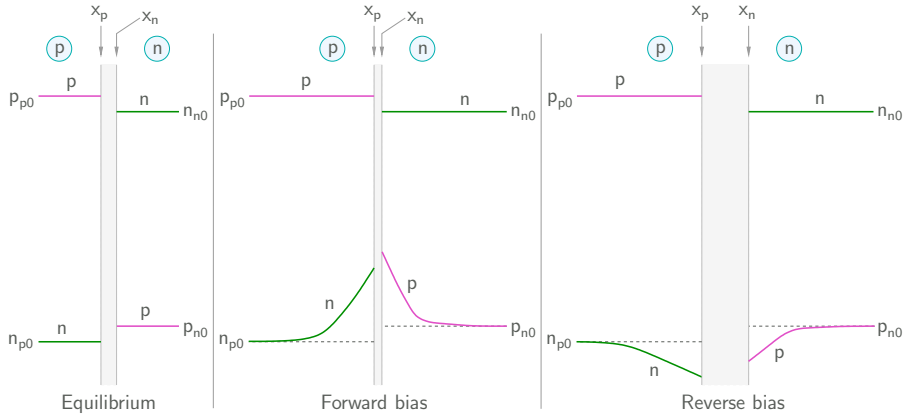
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## $pn$ junction: derivation of $I$ - $V$ equation



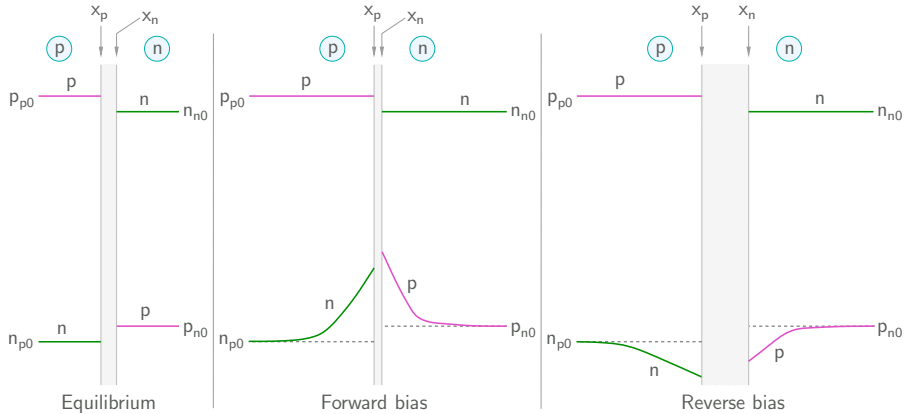
$$J_p^{\text{diff}} \approx -J_p^{\text{drift}}$$

## $pn$ junction: derivation of $I$ - $V$ equation



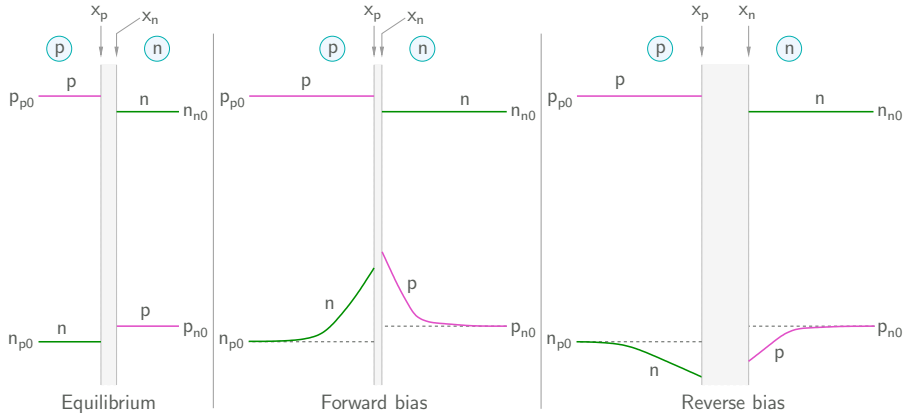
$$J_p^{\text{diff}} \approx -J_p^{\text{drift}} \rightarrow q \mu_p p \mathcal{E} = q D_p \frac{dp}{dx},$$

## $pn$ junction: derivation of $I$ - $V$ equation



$$J_p^{\text{diff}} \approx -J_p^{\text{drift}} \rightarrow q \mu_p p \mathcal{E} = q D_p \frac{dp}{dx}, \text{ i.e., } \mathcal{E} = -\frac{d\psi}{dx} = \frac{D_p}{\mu_p} \frac{1}{p} \frac{dp}{dx}.$$

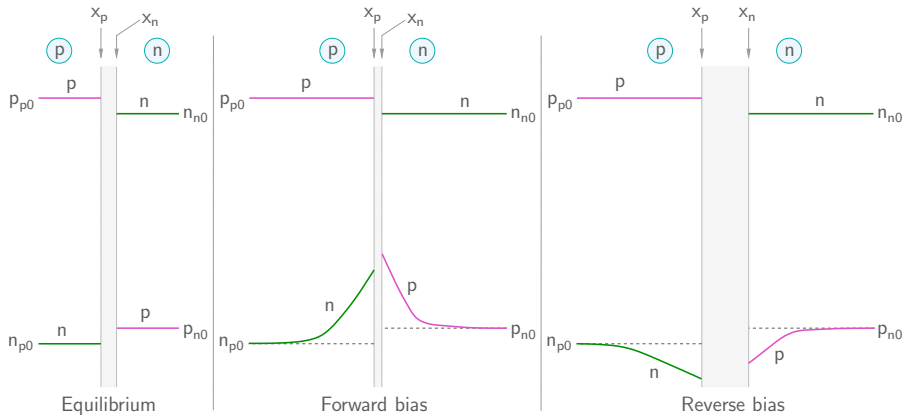
## $pn$ junction: derivation of $I$ - $V$ equation



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$$\frac{D}{\mu} = \frac{kT}{q} \rightarrow \int d\psi = -V_T \int \frac{1}{p} dp$$

## $pn$ junction: derivation of $I$ - $V$ equation

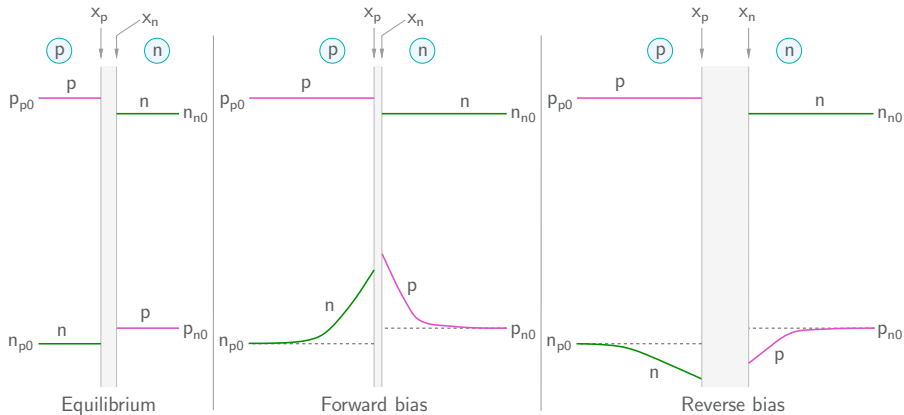


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$$\frac{D}{\mu} = \frac{kT}{q} \rightarrow \int d\psi = -V_T \int \frac{1}{p} dp \rightarrow \psi \Big|_{x_1}^{x_2} = -V_T \log \frac{p(x_2)}{p(x_1)}$$



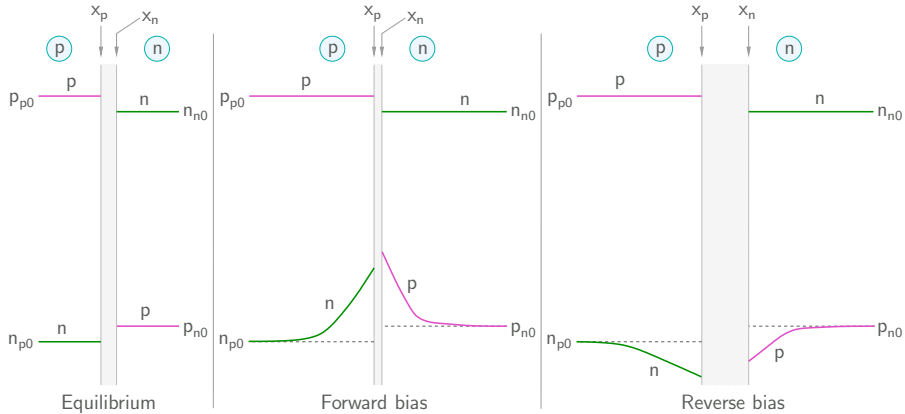
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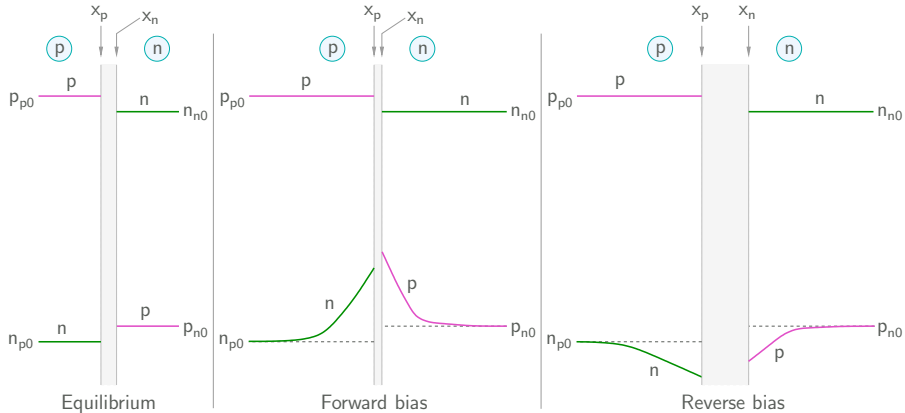
$$\frac{D}{\mu} = \frac{kT}{q} \rightarrow \int d\psi = -V_T \int \frac{1}{p} dp \rightarrow \psi \Big|_{x_1}^{x_2} = -V_T \log \frac{p(x_2)}{p(x_1)} \rightarrow \frac{p(x_n)}{p(x_p)} = \exp \left( \frac{\psi(x_p) - \psi(x_n)}{V_T} \right).$$

## $pn$ junction: derivation of $I$ - $V$ equation



$$J_p^{\text{diff}} \approx -J_p^{\text{drift}} \rightarrow \frac{p(x_n)}{p(x_p)} = \exp\left(\frac{\psi(x_p) - \psi(x_n)}{V_T}\right).$$

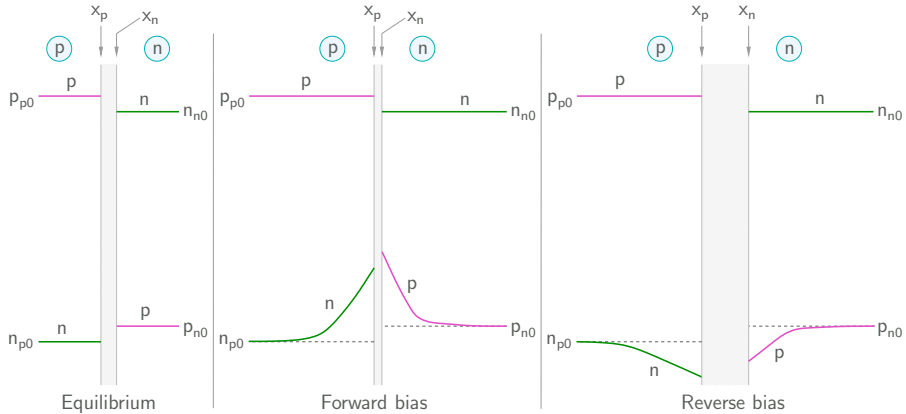
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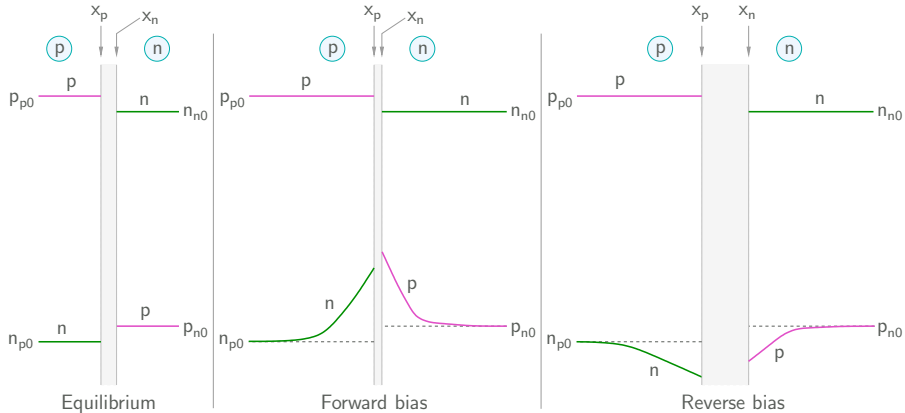
$$J_p^{\text{diff}} \approx -J_p^{\text{drift}} \rightarrow \frac{p(x_n)}{p(x_p)} = \exp\left(\frac{\psi(x_p) - \psi(x_n)}{V_T}\right).$$

$$J_n^{\text{diff}} \approx -J_n^{\text{drift}} \rightarrow \frac{n(x_n)}{n(x_p)} = \exp\left(\frac{\psi(x_n) - \psi(x_p)}{V_T}\right).$$

## $pn$ junction: derivation of $I$ - $V$ equation

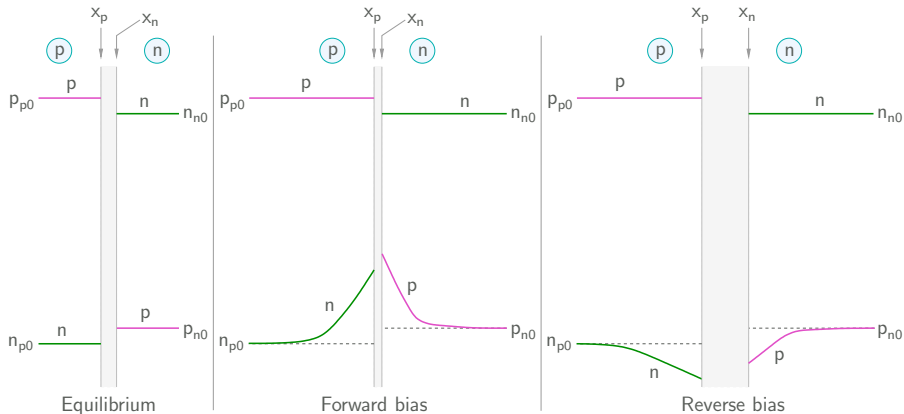


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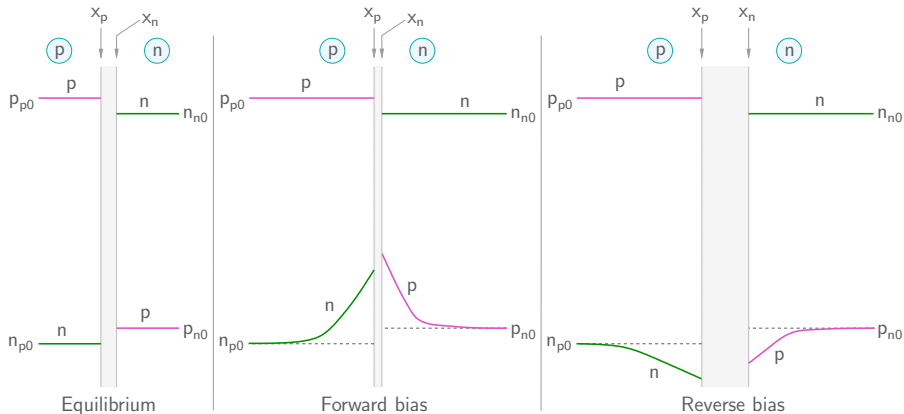
\* When a bias is applied, the minority carrier concentrations in the neutral regions can change substantially.

## $pn$ junction: derivation of $I$ - $V$ equation

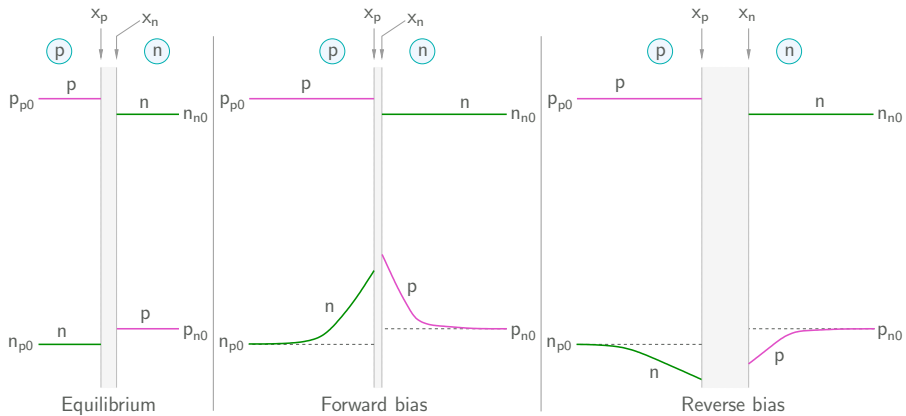


- \* When a bias is applied, the minority carrier concentrations in the neutral regions can change substantially.
- \* There is a corresponding change in the majority carrier concentrations as well, and it serves to keep these regions charge-neutral.

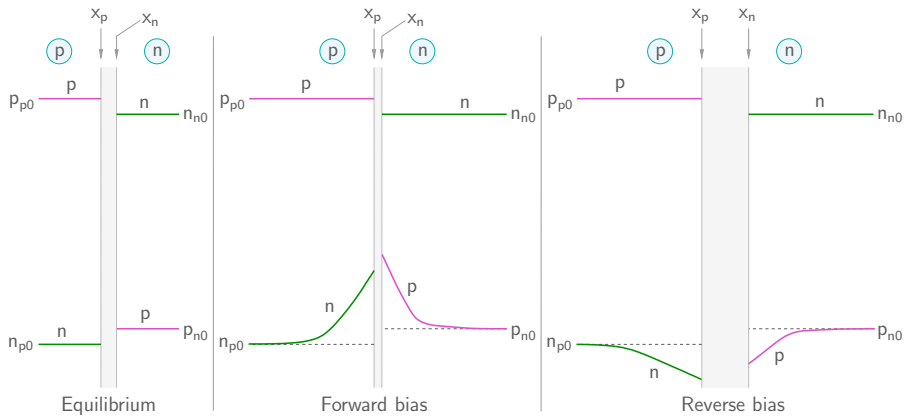
## $pn$ junction: derivation of $I$ - $V$ equation



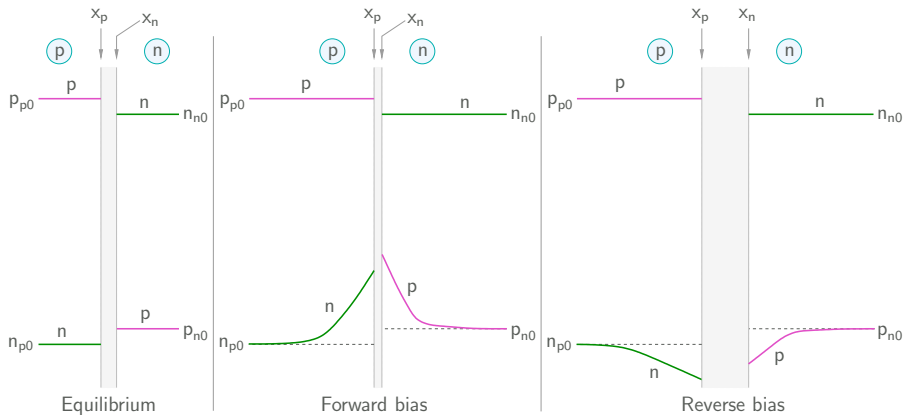
- \* When a bias is applied, the minority carrier concentrations in the neutral regions can change substantially.
- \* There is a corresponding change in the majority carrier concentrations as well, and it serves to keep these regions charge-neutral.
- \* Low-level injection:  $\Delta n \approx \Delta p \ll p_{p0}$  in the neutral  $p$ -region  $\rightarrow p(x) \approx p_{p0}$  for  $x \leq x_p$   
 $\Delta p \approx \Delta n \ll n_{n0}$  in the neutral  $n$ -region  $\rightarrow n(x) \approx n_{n0}$  for  $x \geq x_n$





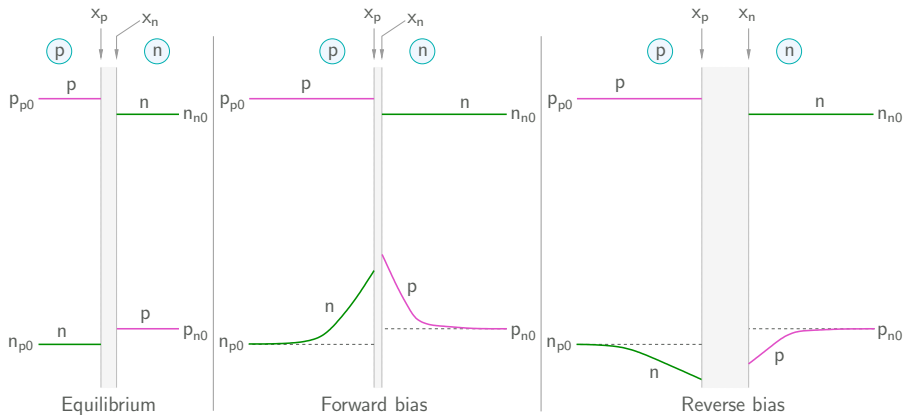


$$* \frac{p(x_n)}{p(x_p)} = \exp\left(\frac{\psi(x_p) - \psi(x_n)}{V_T}\right), \quad \frac{n(x_n)}{n(x_p)} = \exp\left(\frac{\psi(x_n) - \psi(x_p)}{V_T}\right). \text{ Also, } \psi(x_n) - \psi(x_p) = V_j.$$



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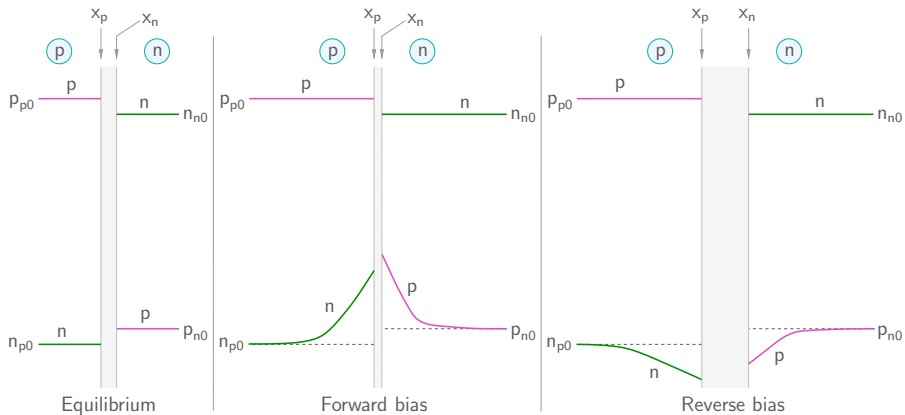
\* Low-level injection:  $p(x) \approx p_{p0}$  for  $x \leq x_p$ , and  $n(x) \approx n_{n0}$  for  $x \geq x_n$ .



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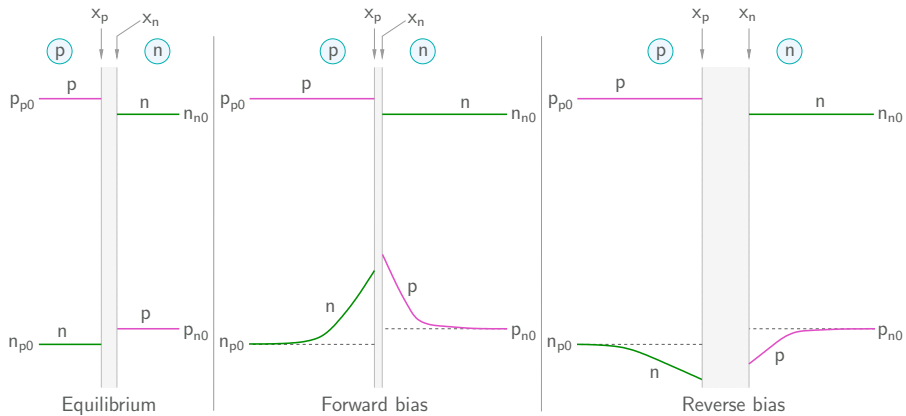
$$\rightarrow \frac{p(x_n)}{p_{p0}} = e^{-V_j/V_T} \rightarrow p(x_n) = p_{p0} e^{-V_j/V_T}$$



$$* \frac{p(x_n)}{p(x_p)} = \exp\left(\frac{\psi(x_p) - \psi(x_n)}{V_T}\right), \quad \frac{n(x_n)}{n(x_p)} = \exp\left(\frac{\psi(x_n) - \psi(x_p)}{V_T}\right). \text{ Also, } \psi(x_n) - \psi(x_p) = V_j.$$

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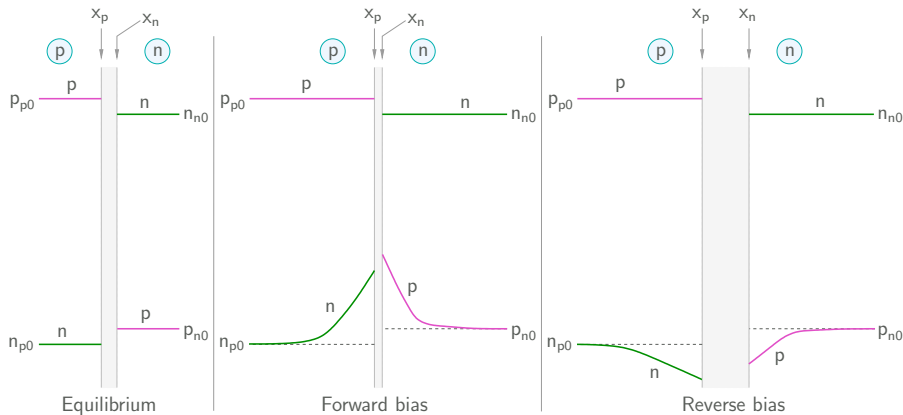
$$\begin{aligned} \rightarrow \frac{p(x_n)}{p_{p0}} &= e^{-V_j/V_T} \rightarrow p(x_n) = p_{p0} e^{-V_j/V_T} \\ \frac{n_{n0}}{n(x_p)} &= e^{V_j/V_T} \rightarrow n(x_p) = n_{n0} e^{-V_j/V_T}. \end{aligned}$$



$$p(x_n) = p_{p0} e^{-V_j/V_T}$$

$$n(x_p) = n_{n0} e^{-V_j/V_T}$$

$$V_j = V_{bi} - V_a$$

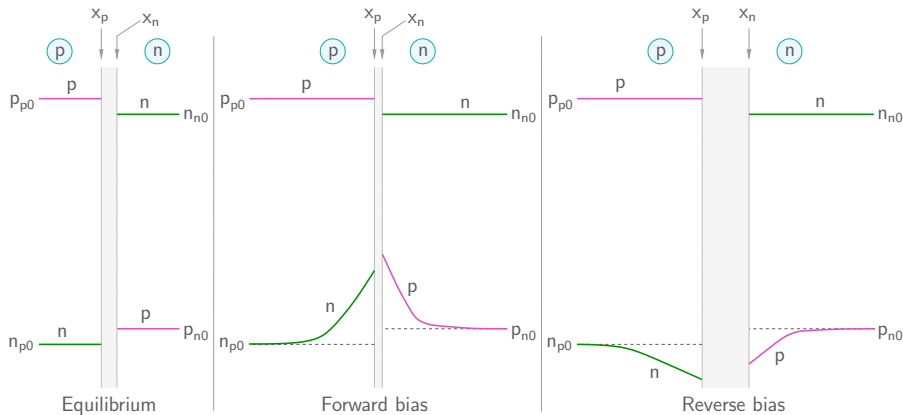


$$p(x_n) = p_{p0} e^{-V_j/V_T}$$

$$n(x_p) = n_{n0} e^{-V_j/V_T}$$

$$V_j = V_{bi} - V_a$$

$$\text{Equilibrium: } p(x_n) = p_{n0} = p_{p0} \exp\left(\frac{-V_{bi}}{V_T}\right), \quad n(x_p) = n_{p0} = n_{n0} \exp\left(\frac{-V_{bi}}{V_T}\right).$$



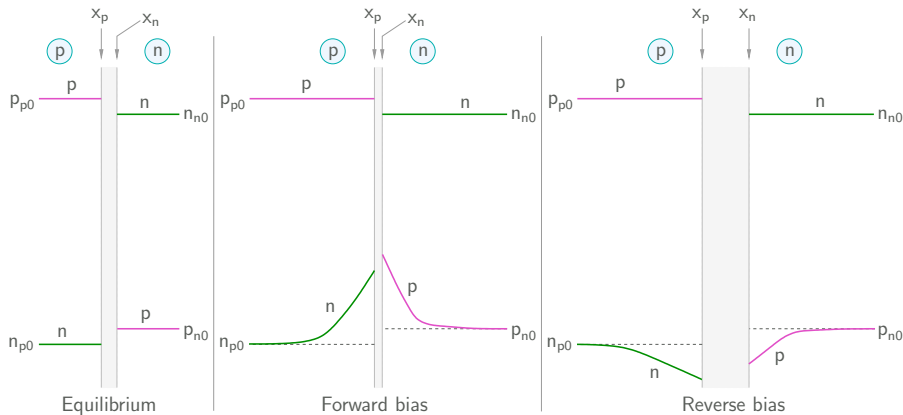
$$p(x_n) = p_{p0} e^{-V_j/V_T}$$

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With bias:  $p(x_n) = p_{p0} \exp\left(\frac{-V_{bi} + V_a}{V_T}\right), \quad n(x_p) = n_{n0} \exp\left(\frac{-V_{bi} + V_a}{V_T}\right).$



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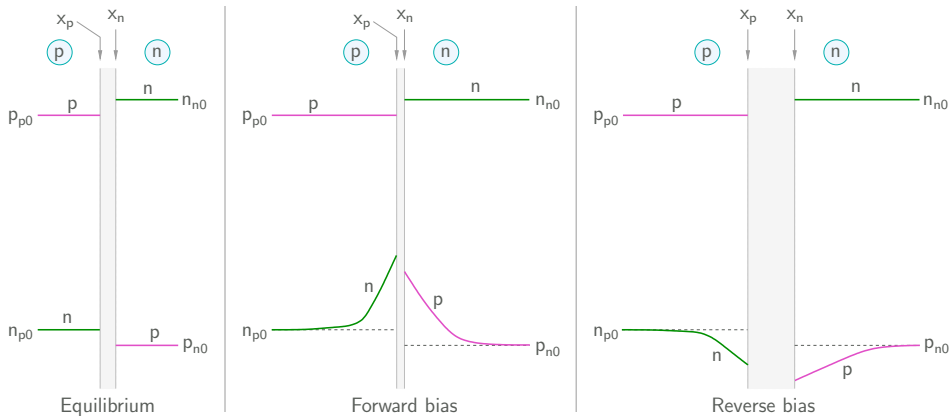
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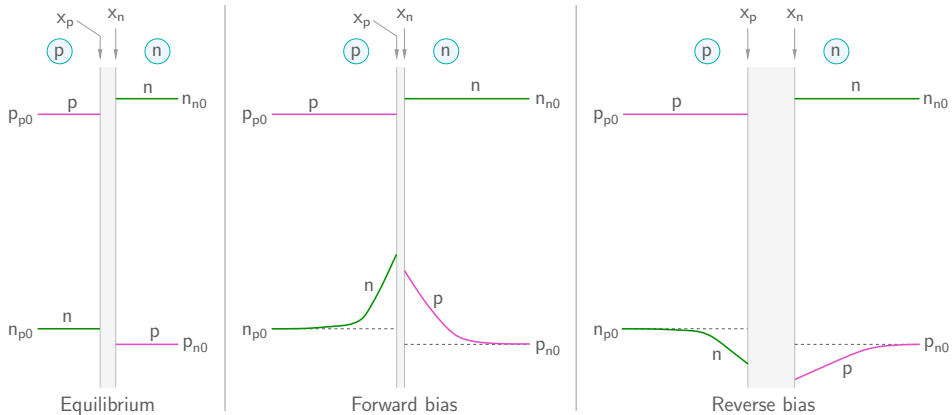
$$p(x_n) = p_{n0} \exp\left(\frac{V_a}{V_T}\right), \quad n(x_p) = n_{p0} \exp\left(\frac{V_a}{V_T}\right).$$



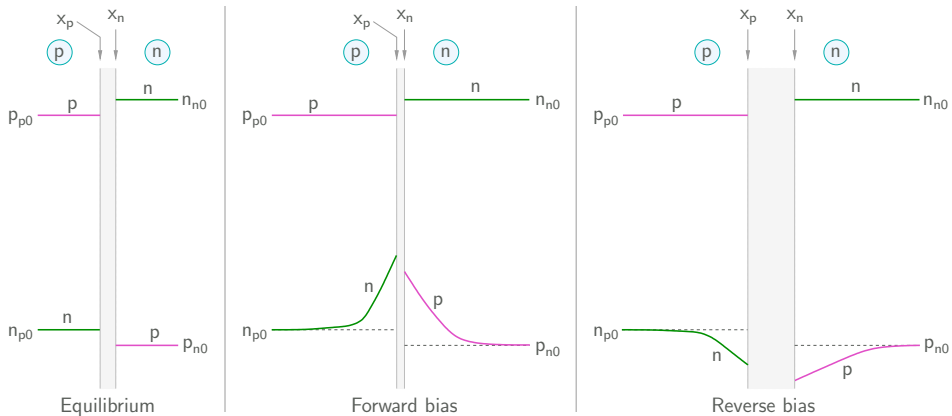


Example: Consider an abrupt, uniformly doped silicon  $pn$  junction at  $T = 300\text{ K}$ , with  $N_a = 5 \times 10^{16}\text{ cm}^{-3}$  and  $N_d = 10^{18}\text{ cm}^{-3}$ . Compute the depletion width and the minority carrier densities at the depletion region edges ( $x_p$  and  $x_n$ ) for an applied bias of  $+0.3\text{ V}$ ,  $+0.6\text{ V}$ ,  $-1\text{ V}$ ,  $-5\text{ V}$ .

( $n_i = 1.5 \times 10^{10}\text{ cm}^{-3}$  for silicon at  $T = 300\text{ K}$ .)

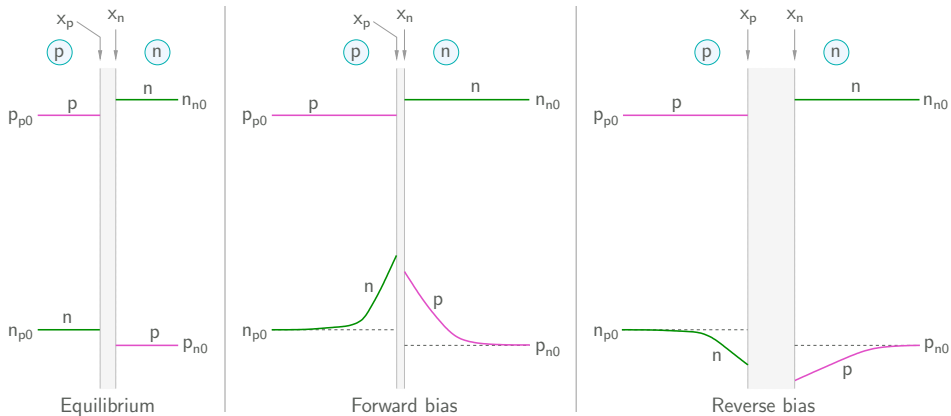


Solution:  $p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3} \rightarrow n_{p0} = \frac{n_i^2}{p_{p0}} = \frac{(1.5 \times 10^{10})^2}{5 \times 10^{16}} = 4.5 \times 10^3 \text{ cm}^{-3}.$



Solution:  $p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3} \rightarrow n_{p0} = \frac{n_i^2}{p_{p0}} = \frac{(1.5 \times 10^{10})^2}{5 \times 10^{16}} = 4.5 \times 10^3 \text{ cm}^{-3}.$

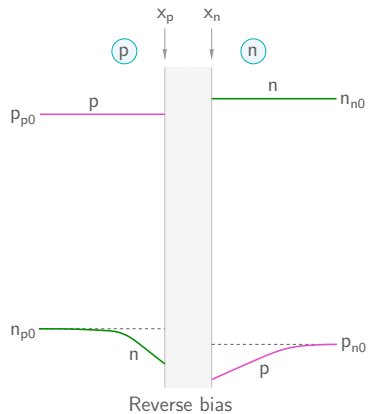
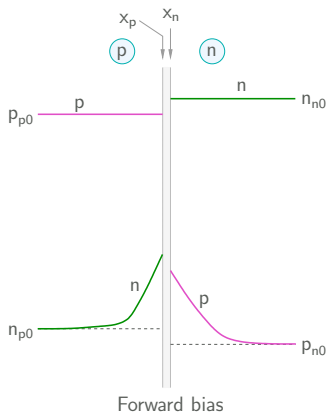
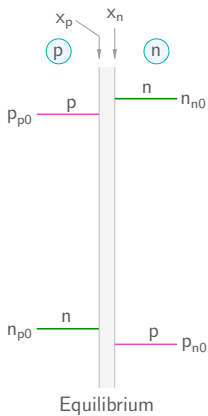
$n_{n0} \approx N_d = 1 \times 10^{18} \text{ cm}^{-3} \rightarrow p_{n0} = \frac{n_i^2}{n_{n0}} = \frac{(1.5 \times 10^{10})^2}{1 \times 10^{18}} = 2.25 \times 10^2 \text{ cm}^{-3}.$

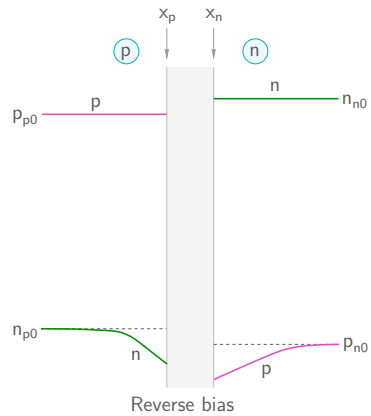
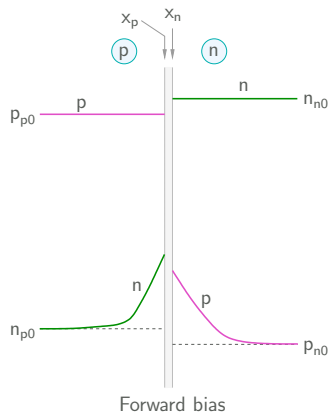
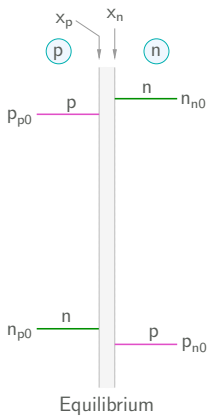


Solution:  $p_{p0} \approx N_a = 5 \times 10^{16} \text{ cm}^{-3} \rightarrow n_{p0} = \frac{n_i^2}{p_{p0}} = \frac{(1.5 \times 10^{10})^2}{5 \times 10^{16}} = 4.5 \times 10^3 \text{ cm}^{-3}.$

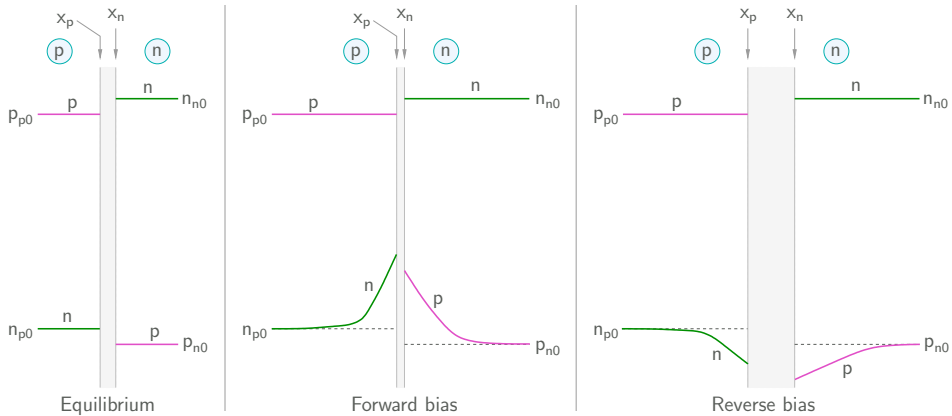
$n_{n0} \approx N_d = 1 \times 10^{18} \text{ cm}^{-3} \rightarrow p_{n0} = \frac{n_i^2}{n_{n0}} = \frac{(1.5 \times 10^{10})^2}{1 \times 10^{18}} = 2.25 \times 10^2 \text{ cm}^{-3}.$

$V_{bi} = V_T \log \left( \frac{N_a N_d}{n_i^2} \right) = (0.0259 \text{ V}) \log \left( \frac{5 \times 10^{16} \times 10^{18}}{(1.5 \times 10^{10})^2} \right) = 0.86 \text{ V}.$



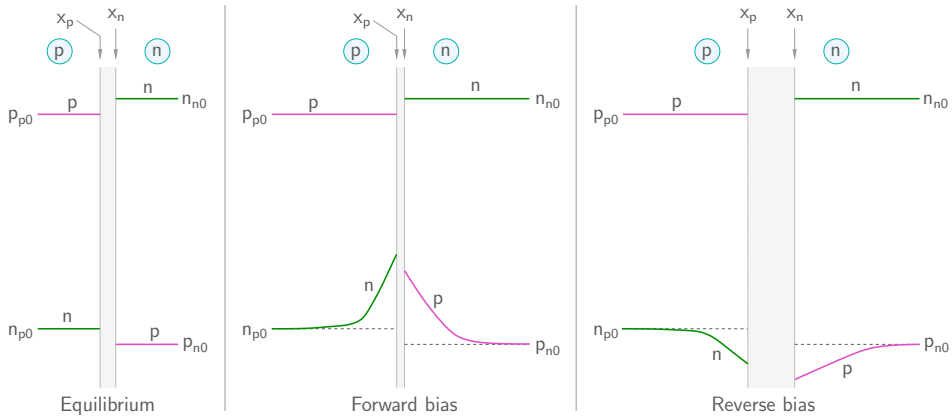


$$V_a = 0.3 \text{ V: } W = \sqrt{\frac{2\epsilon}{q} \frac{N_a + N_d}{N_a N_d} (V_{bi} - V_a)} = 0.12 \mu\text{m}.$$



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$$n(x_p) = n_{p0} \exp\left(\frac{V_a}{V_T}\right) = 4.5 \times 10^3 \times \exp\left(\frac{0.3}{0.0259}\right) = 4.83 \times 10^8 \text{ cm}^{-3}.$$

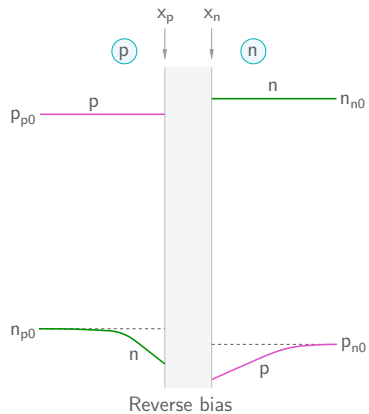
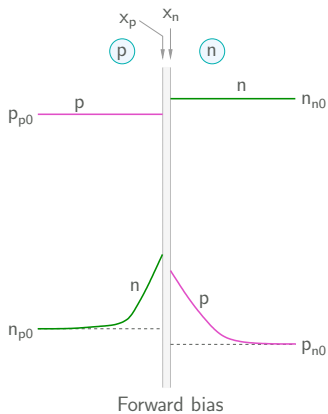
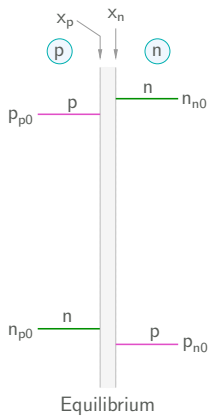


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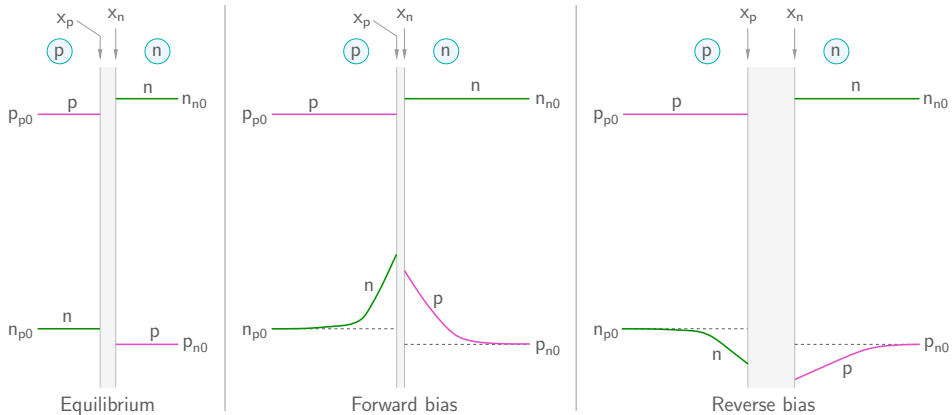
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$$p(x_n) = p_{n0} \exp\left(\frac{V_a}{V_T}\right) = 2.25 \times 10^2 \times \exp\left(\frac{0.3}{0.0259}\right) = 2.41 \times 10^7 \text{ cm}^{-3}.$$



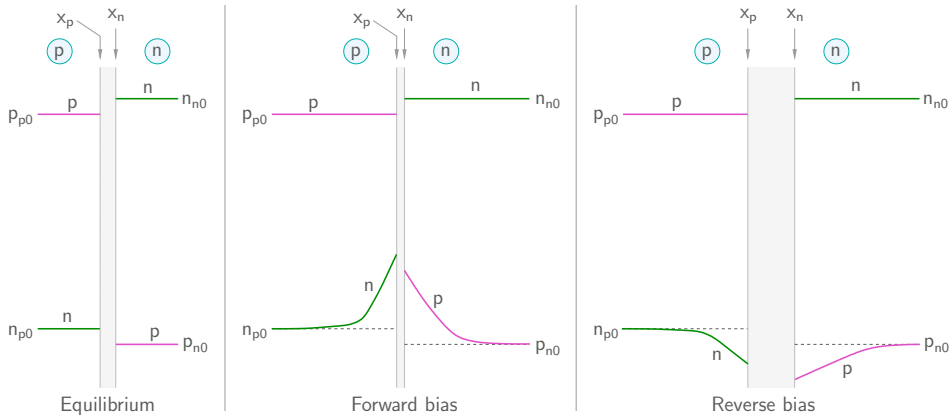


$V_a$ (V)	$W$ ( $\mu\text{m}$ )	$\mathcal{E}_m$ (kV/cm)	$n(x_p)$ ( $\text{cm}^{-3}$ )	$p(x_n)$ ( $\text{cm}^{-3}$ )
0.6	0.08	61.3	$5.18 \times 10^{13}$	$2.59 \times 10^{12}$
0.3	0.12	90.4	$4.83 \times 10^8$	$2.41 \times 10^7$
0.0	0.15	112.2	$4.50 \times 10^3$	$2.25 \times 10^2$
-1.0	0.22	165.3	$7.68 \times 10^{-14} \approx 0$	$3.84 \times 10^{-15} \approx 0$
-5.0	0.40	293.6	$\approx 0$	$\approx 0$



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\* With forward bias, the minority carrier concentrations can increase by several orders of magnitude.



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- \* With forward bias, the minority carrier concentrations can increase by several orders of magnitude.
- \* With reverse bias, the minority carrier concentrations become very small and can be replaced with zero for all practical purposes.