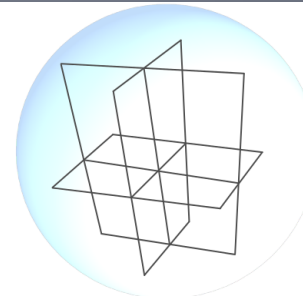


Compressive Sensing

CS 754
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Outline of the Lectures

- Theorem 3: Sketch of Proof and comments on the theorem
- Design of compressed sensing matrices
- Compressive classification

Theorem 3

- Suppose the matrix $\mathbf{A} = \Phi\Psi$ of size m by n (where sensing matrix Φ has size m by n , and basis matrix Ψ has size n by n) has RIP property of order $2S$ where $\delta_{2S} < 0.41$. Let the solution of the following be denoted as θ^* , (for signal $\mathbf{f} = \Psi\theta$, measurement vector $\mathbf{y} = \Phi\Psi\theta$):
$$\min \|\theta\|_1 \text{ such that } \|\mathbf{y} - \Phi\Psi\theta\|_2 \leq \varepsilon$$

Then we have:

θ_S is created by retaining the S largest magnitude elements of θ , and setting the rest to 0.

$$\|\theta^* - \theta\|_2 \leq \frac{C_0}{\sqrt{S}} \|\theta - \theta_S\|_1 + C_1 \varepsilon$$

Comments on Theorem 3

- Theorem 3 is a direct extension of Theorem 2 for the case of noisy measurements.
- It states that the solution of the convex program (see previous slides) gives a reconstruction error which is the sum of two terms: (1) the error of an oracle solution where the oracle told us the S largest coefficients of the signal \mathbf{f} at the correct indices, and (2) a term proportional to the noise variance.
- The constants C_0 and C_1 are very small (less than or equal to 5.5 and 6 respectively for $\delta_{2S} = 0.25$), they are increasing functions of just δ_{2S} .

Sketch of the proof

- The proof can be found in a paper by Candes [“The restricted isometry property and its implications for compressed sensing”](#), published in 2008.
- The proof as such is just between 1 and 2 pages long.

Sketch of the proof

- The proof uses various properties of vectors in Euclidean space.
- The Cauchy Schwartz inequality: $|\mathbf{v} \cdot \mathbf{w}| \leq \|\mathbf{v}\|_2 \|\mathbf{w}\|_2$
- The triangle inequality: $\|\mathbf{v} + \mathbf{w}\|_2 \leq \|\mathbf{v}\|_2 + \|\mathbf{w}\|_2$
- Reverse triangle inequality: $|\|\mathbf{v}\|_2 - \|\mathbf{w}\|_2| \leq \|\mathbf{v} - \mathbf{w}\|_2$

Sketch of the proof

- Relationship between various norms:

$$\|\mathbf{v}\|_2 \leq \|\mathbf{v}\|_1 = |\mathbf{v}| \cdot \mathbf{1} \leq \sqrt{n} \|\mathbf{v}\|_2$$

$$\|\mathbf{v}\|_1 \leq \sqrt{k} \|\mathbf{v}\|_2 \text{ if } \mathbf{v} \text{ is a } k\text{-sparse vector}$$

- Refer to Theorem 3. For the sake of simplicity alone, we shall assume Ψ to be the identity matrix.
- Hence $\mathbf{x} = \boldsymbol{\theta}$. Even if Ψ were not identity, the proof as such does not change.

Sketch of the proof: Step 1

This result is called the **Tube constraint**.

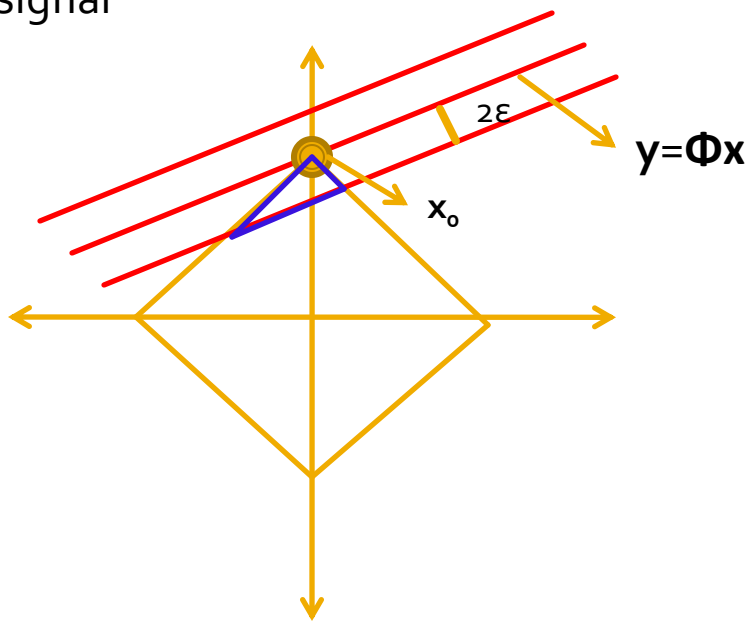
- We have:

$$\|\Phi(\mathbf{x}^* - \mathbf{x}_0)\|_2 \leq \|\Phi\mathbf{x}^* - \mathbf{y}\|_2 + \|\Phi\mathbf{x}_0 - \mathbf{y}\|_2 \leq 2\varepsilon$$

In the following,
 \mathbf{x}_0 = true signal

Triangle inequality

Given constraint + feasibility
of solution \mathbf{x}^*



Sketch of the proof: Preparing for steps 2 and 3

- Define vector $\mathbf{h} = \mathbf{x}^* - \mathbf{x}_0$.
- Decompose \mathbf{h} into vectors $\mathbf{h}_{T_0}, \mathbf{h}_{T_1}, \mathbf{h}_{T_2}, \dots$ which are all at the most s -sparse.
- T_0 is the set of indices corresponding to the s largest absolute value elements of \mathbf{x}_0 ,
- T_1 is the set of indices corresponding to the s largest absolute value elements of $\mathbf{h}_{(T_0-c)} = \mathbf{h} - \mathbf{h}_{T_0}$,
- T_2 is the set of indices corresponding to the s largest absolute value elements of $\mathbf{h}_{(T_0 \cup T_1)-c} = \mathbf{h} - \mathbf{h}_{T_0} - \mathbf{h}_{T_1}$,
- and so on.

Sketch of the proof: Step 2

- We will assume \mathbf{x}_0 is s -sparse (later we will relax this to assume it is compressible).
- We now establish the so-called **cone constraint**.

$$\|x^*\|_1 = \|x_0 + h\|_1 \leq \|x_0\|_1$$

The vector h has its origin at \mathbf{x}_0 and it lies in the intersection of the L_1 ball and the tube.

$$\therefore \sum_{i \in T_0} |x_{0i} + h_i| + \sum_{i \in T_{0-c}} |x_{0i} + h_i| \leq \|x_0\|_1$$

o-valued

$$\therefore \|x_0\|_1 - \|h_{T_0}\|_1 + \|h_{T_{0-c}}\|_1 \leq \|x_0\|_1$$

$$\therefore \|h_{T_{0-c}}\|_1 \leq \|h_{T_0}\|_1$$

The vector h must also necessarily obey this constraint – the cone constraint.

Sketch of the proof: Preparing for steps 3 and onwards

- We will now prove that such a vector \mathbf{h} is orthogonal to the null-space of Φ .
- In fact, we will prove that $\|\Phi\mathbf{h}\|_2 \approx \|\mathbf{h}\|_2$.
- In other words, we will prove that the magnitude of \mathbf{h} is not much greater than 2ε , which means that the solution \mathbf{x}^* of the optimization problem is “close enough” to \mathbf{x}_0 .

Sketch of the proof: Step 3

- In step 3, we use a bunch of algebraic manipulations to prove that the magnitude of \mathbf{h} outside of $T_0 \cup T_1$ is upper bounded by the magnitude of \mathbf{h} on $T_0 \cup T_1$.

- In other words, we prove that:

$$\|\mathbf{h}_{(T_0 \cup T_1)^c}\|_2 \leq \|\mathbf{h}_{T_0}\|_2 \leq \|\mathbf{h}_{T_0 \cup T_1}\|_2$$

- The algebra involves various inequalities mentioned earlier.

Sketch of the proof: Steps 4 and 5

- We now prove that the magnitude of \mathbf{h} on $T_0 \cup T_1$ is upper bounded by a reasonable quantity.
- For this, we show using the RIP of Φ of order $2s$ and a series of manipulations that:

$$(1 - \delta_{2s}) \|h_{T_0 \cup T_1}\|_2^2 \leq \|\Phi h_{T_0 \cup T_1}\|_2^2 \leq \|h_{T_0 \cup T_1}\|_2 (2\varepsilon \sqrt{1 + \delta_{2s}} + \sqrt{2} \delta_{2s} \|h_{T_0}\|_2)$$

- This implies that

$$\|h_{T_0 \cup T_1}\|_2 \leq \left(\frac{2\varepsilon \sqrt{1 + \delta_{2s}}}{1 - \delta_{2s}} + \frac{\sqrt{2} \delta_{2s}}{1 - \delta_{2s}} \|h_{T_0 \cup T_1}\|_2 \right) \rightarrow \|h_{T_0 \cup T_1}\|_2 \leq \frac{2\sqrt{1 + \delta_{2s}} \varepsilon}{1 - \delta_{2s}(\sqrt{2} + 1)} \rightarrow \|h\|_2 \leq \frac{4\varepsilon \sqrt{1 + \delta_{2s}}}{1 - \delta_{2s}(\sqrt{2} + 1)}$$

The proof for compressible signals

- The steps change a bit. The cone constraint changes to:

$$\|x^*\|_1 = \|x_0 + h\|_1 \leq \|x_0\|_1$$

$$\therefore \sum_{i \in T_0} |x_{0i} + h_i| + \sum_{i \in T_{0-c}} |x_{0i} + h_i| \leq \|x_0\|_1$$

$$\therefore \|x_{0,T_0}\|_1 - \|h_{T_0}\|_1 + \|h_{T_{0-c}}\|_1 - \|x_{0,T_{0-c}}\|_1 \leq \|x_0\|_1$$

$$\therefore \|h_{T_{0-c}}\|_1 \leq \|h_{T_0}\|_1 + 2\|x_{0,T_{0-c}}\|_1$$

The proof for compressible signals

- All the other steps remain as is, except the last one which produces the following bound:

$$\|h\|_2 \leq \frac{2\sqrt{1+\delta_{2s}}}{1-\delta_{2s}(1+\sqrt{2})}\varepsilon + \frac{1+\delta_{2s}(\sqrt{2}-1)}{1-\delta_{2s}(\sqrt{2}+1)} \frac{\|x_0 - x_{0,T0}\|_1}{\sqrt{s}}$$

Sketch of the proof: Step 3

- Step 3 of the proof uses the following corollary of the RIP for two s -sparse unit vectors with disjoint support: $|\Phi \mathbf{x}_1 \bullet \Phi \mathbf{x}_2| \leq \delta_{2s}$

- Proof of corollary:

$$(1 - \delta_{2s}) \|\mathbf{x}_1 - \mathbf{x}_2\|^2 \leq \|\Phi(\mathbf{x}_1 - \mathbf{x}_2)\|^2 \leq (1 + \delta_{2s}) \|\mathbf{x}_1 - \mathbf{x}_2\|^2, \text{ by RIP}$$

$$|\Phi \mathbf{x}_1 \bullet \Phi \mathbf{x}_2| = \frac{1}{4} \left| \|\Phi \mathbf{x}_1 + \Phi \mathbf{x}_2\|^2 - \|\Phi \mathbf{x}_1 - \Phi \mathbf{x}_2\|^2 \right|$$

$$\leq \frac{1}{4} \left((1 + \delta_{2s}) \|\mathbf{x}_1 + \mathbf{x}_2\|^2 - (1 - \delta_{2s}) \|\mathbf{x}_1 - \mathbf{x}_2\|^2 \right)$$

$$\leq \frac{1}{4} \left((1 + \delta_{2s}) (\|\mathbf{x}_1\|^2 + \|\mathbf{x}_2\|^2 + 2\mathbf{x}_1 \bullet \mathbf{x}_2) - (1 - \delta_{2s}) (\|\mathbf{x}_1\|^2 + \|\mathbf{x}_2\|^2 - 2\mathbf{x}_1 \bullet \mathbf{x}_2) \right)$$

$$\leq \delta_{2s} \because \|\mathbf{x}_1\|^2 = \|\mathbf{x}_2\|^2 = 1, \mathbf{x}_1 \bullet \mathbf{x}_2 = 0$$

Sketch of the proof: Step 3

- This step also uses the following corollary of the RIP for two s -sparse unit vectors with disjoint support: $|\Phi \mathbf{x}_1 \bullet \Phi \mathbf{x}_2| \leq \delta_{2s}$
- What if the original vectors \mathbf{x}_1 and \mathbf{x}_2 were not unit-vectors, but both were s -sparse?

$$\frac{|\Phi \mathbf{x}_1 \bullet \Phi \mathbf{x}_2|}{\|\mathbf{x}_1\|_2 \|\mathbf{x}_2\|_2} \leq \delta_{2s} \rightarrow |\Phi \mathbf{x}_1 \bullet \Phi \mathbf{x}_2| \leq \delta_{2s} \|\mathbf{x}_1\|_2 \|\mathbf{x}_2\|_2$$

Comments on the theorem

- The bound is

$$\|\mathbf{h}\|_2 \leq \frac{4\sqrt{1+\delta_{2s}}}{1-(1+\sqrt{2})\delta_{2s}} \varepsilon + \frac{1}{\sqrt{s}} \frac{1+\delta_{2s}(\sqrt{2}-1)}{1-(1+\sqrt{2})\delta_{2s}} \|\mathbf{x}_0 - \mathbf{x}_{0,\text{TO}}\|_1$$

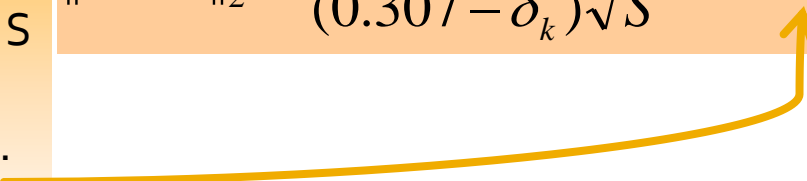
- Note the requirement that δ_{2s} should be less than $2^{0.5}-1$.
- You can prove that the two constant factors – one before ε and the other before $\|\mathbf{x}_0 - \mathbf{x}_{0,\text{TO}}\|_1$, are both increasing functions of δ_{2s} in the domain $[0,1]$.
- So sensing matrices with smaller values of δ_{2s} are always nicer!

Theorem 6 (Cai et al): Improvement over Theorem 3

- Suppose the matrix $\mathbf{A}=\Phi\Psi$ of size m by n (where sensing matrix Φ has size m by n , and basis matrix Ψ has size n by n) has RIP property of order S where $\delta_S < 0.307$. Let the solution of the following be denoted as θ^* , (for signal $\mathbf{f} = \Psi\theta$, measurement vector $\mathbf{y}=\Phi\Psi\theta$):
$$\min \|\theta\|_1 \text{ such that } \|\mathbf{y} - \Phi\Psi\theta\|_2^2 \leq \varepsilon$$

Then we have:

θ_S is created by retaining the S largest magnitude elements of θ , and setting the rest to 0.

$$\|\theta^* - \theta\|_2 \leq \frac{1}{(0.307 - \delta_k)\sqrt{S}} \|\theta - \theta_S\|_1 + \frac{1}{0.307 - \delta_k} \varepsilon$$


Ψ Need not be orthonormal

- Theorems 3,5,6 refer to orthonormal bases for the signal to have sparse or compressible representations.
- However that is not a necessary condition.
- There exist the so-called “**over-complete bases**” in which the number of columns exceeds the number of rows ($n \times K, K > n$).
- Such matrices afford even sparser signal representations.

Ψ Need not be orthonormal

- Why? We explain with an example.
- A cosine wave (with grid-aligned frequency) will have a sparse representation in the DCT basis \mathbf{V}_1 .
- An impulse signal has sparse representation in the identity basis \mathbf{V}_2 .
- Now consider a signal which is the superposition of a small number of cosines and impulses.
- The combined signal has sparse representation in neither the DCT basis nor the identity basis.
- But the combined signal will have a sparse representation in the combined dictionary $[\mathbf{V}_1 \mathbf{V}_2]$.

Designing Sensing Matrices in CS

Designing Compressed Sensing Matrices

- We know that certain classes of random matrices satisfy the RIP with very high probability.
- However, we also know that small RICs are desirable.
- This gives rise to the question: Can we design matrices with **smaller RIC** than a randomly generated matrix?

Designing Compressed Sensing Matrices

- Unfortunately, there is no known efficient algorithm for even computing the RIC given a fixed matrix!
- But we know that the mutual coherence of $\Phi\Psi$ is an upper bound to the RIC: $\delta_s \leq \mu(s-1)$
- So we can design a CS matrix by starting with a random one, and then performing **a gradient descent on the mutual coherence** to reach a matrix with a smaller mutual coherence!

Designing Compressed Sensing Matrices

- The procedure is summarized below:

Randomly pick a m by n matrix Φ .

Repeat until convergence

{

$$\Phi \leftarrow \Phi - \alpha \frac{\partial}{\partial \Phi} (\mu(\Phi \Psi))$$

}

Pick the step-size adaptively so that you actually descend on the mutual coherence.

$$\mu(\Phi \Psi) = \max_{i \neq j} \left| \frac{(\Phi \Psi)_i}{\|(\Phi \Psi)_i\|_2} \cdot \frac{(\Phi \Psi)_j}{\|(\Phi \Psi)_j\|_2} \right|$$

Designing Compressed Sensing Matrices

- The aforementioned is one example of a procedure to “design” a CS matrix – as opposed to picking one randomly.
- Note that mutual coherence has one more advantage over RIC – the former is **not tied to any particular sparsity level!**
- But one must bear in mind that the mutual coherence is an upper bound to the RIC!

Designing CS matrices: Method 1

- The main problem is how to find a derivative of the “max” function which is non-differentiable!
- Use the softmax function which is differentiable:

$$\lim_{\beta \rightarrow \infty} \frac{1}{\beta} \log \left(\sum_{i=1}^n \exp(\beta x_i) \right) = \max \{x_i\}_{i=1}^n$$

Designing CS matrices: Method 1

- This method can now be used to design CS matrices.
- As the mutual coherence function is expected to be very non-convex, one must run a multi-start strategy in practice.
- For each “start”, you begin with a random Φ , do a gradient descent on μ till convergence.
- Repeat this procedure many times, each time beginning from a different randomly chosen initial condition.
- Choose the value of μ that is the smallest – among all these starts.

Designing CS matrices: Method 1

- In many cases, one needs additional constraints on the matrix Φ .
- For example, in a Hitomi video camera architecture, Φ is a concatenation of non-negative and diagonal matrices.
- The non-negativity can be imposed by means of **projected** gradient descent.
- See the next slide for the modified algorithm to maintain non-negativity.

Designing CS matrices: Method 1

Randomly pick a m by n matrix Φ .

Repeat until convergence

{

$$\Phi \leftarrow \Phi - \alpha \frac{\partial}{\partial \Phi} (\mu(\Phi \Psi))$$

Set negative entries in Φ to 0.

}

Pick the step-size adaptively so that you actually descend on the mutual coherence **after** setting the negative entries to 0.



Figure 9: Demosaicing close-ups, examples $\{1, 2, 3\}$. Clockwise: inputs, reconstructions with $\{\text{random}, \{\text{circularly}, \text{non-circularly}\} \text{designed}\}$ matrices

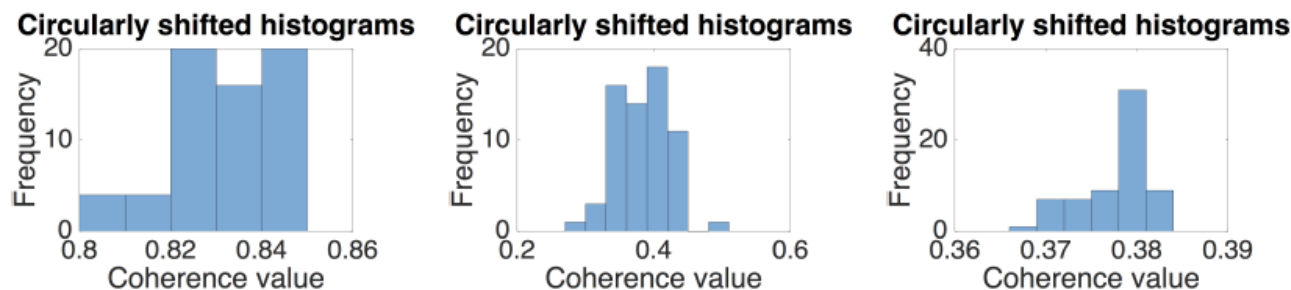


Figure 15: Left to right: Circularly-shifted coherence histograms for $\{\text{random}, \text{non-circularly optimized}, \text{circularly optimized}\}$ matrices

Designing CS Matrices: Method 2

- This method does not directly target μ but instead considers the Gram matrix $\mathbf{D}^T \mathbf{D}$ where $\mathbf{D} = \Phi \Psi$ with all columns unit - normalized.
- The aim is to design Φ in such a way that the Gram matrix resembles the identity matrix as much as possible, in other words we want:

$$\Psi^T \Phi^T \Phi \Psi \approx \mathbf{I}$$

Design CS matrices:

$$E = \| \tilde{\mathbf{D}}^T \tilde{\mathbf{D}} - \mathbf{I} \|_F^2 = \text{Tr}\{(\tilde{\mathbf{D}}^T \tilde{\mathbf{D}} - \mathbf{I})(\tilde{\mathbf{D}}^T \tilde{\mathbf{D}} - \mathbf{I})^T\}$$

$$\nabla E \equiv \frac{\partial E}{\partial \tilde{d}_{ij}} = 4\tilde{\mathbf{D}}(\tilde{\mathbf{D}}^T \tilde{\mathbf{D}} - \mathbf{I})$$

$$\tilde{\mathbf{D}}_{(i+1)} = \tilde{\mathbf{D}}_{(i)} - \hat{\eta} \tilde{\mathbf{D}}_{(i)} (\tilde{\mathbf{D}}_{(i)}^T \tilde{\mathbf{D}}_{(i)} - \mathbf{I})$$

$$\hat{\Phi} = \arg \min_{\tilde{\Phi}} \| \Psi^T \Phi^T \Phi \Psi - \mathbf{I} \|_F^2$$

$$\frac{\partial E}{\partial \phi_{ij}} = 4\Phi\Psi(\Psi^T \Phi^T \Phi \Psi - \mathbf{I})\Psi^T$$

$$\Phi_{(i+1)} = \Phi_{(i)} - \eta \Phi_{(i)} \Psi (\Psi^T \Phi_{(i)}^T \Phi_{(i)} \Psi - \mathbf{I}) \Psi^T.$$

Algorithm 1: Gradient-descent optimization.

Input: Sparse representation basis Ψ_{nn} (if necessary),
Stepsize η , Maximum number of iterations K .

Output: Measurement matrix Φ_{pn} .

begin

 Initialize \mathbf{D} to a random matrix.

for $k=1$ **to** K **do**

for $j=1$ **to** n **do**

$\mathbf{d}^j \leftarrow \mathbf{d}^j / \|\mathbf{d}^j\|_2$

end

$\mathbf{D} \leftarrow \mathbf{D} - \eta \mathbf{D}(\mathbf{D}^T \mathbf{D} - \mathbf{I})$

end

if Ψ_{nn} has been given as input **then**

$\hat{\Phi} \leftarrow \mathbf{D}\Psi^{-1}$

else

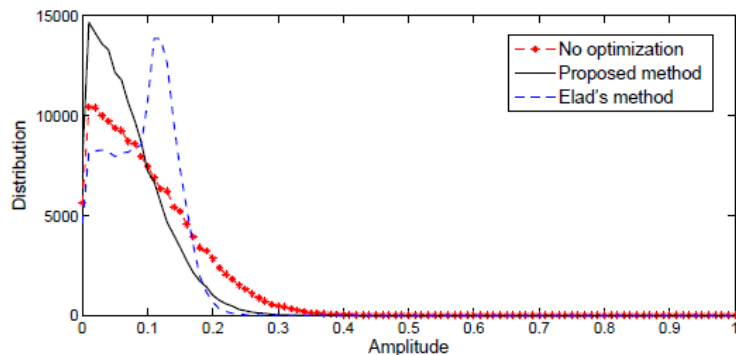
$\hat{\Phi} \leftarrow \mathbf{D}$

end

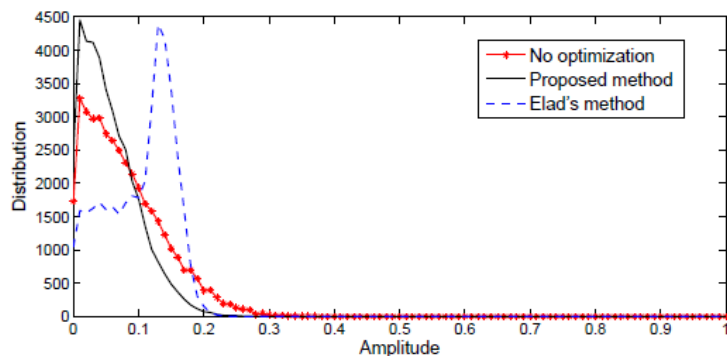
end

Abdoghasemi et al, On optimization of the measurement matrix for compressive sensing, EUSIPCO 2010

Design CS matrices:

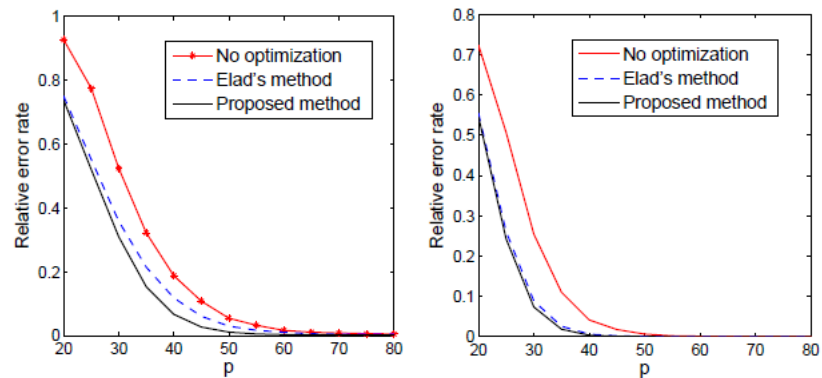


(a) Φ : Random, and Ψ : Random.



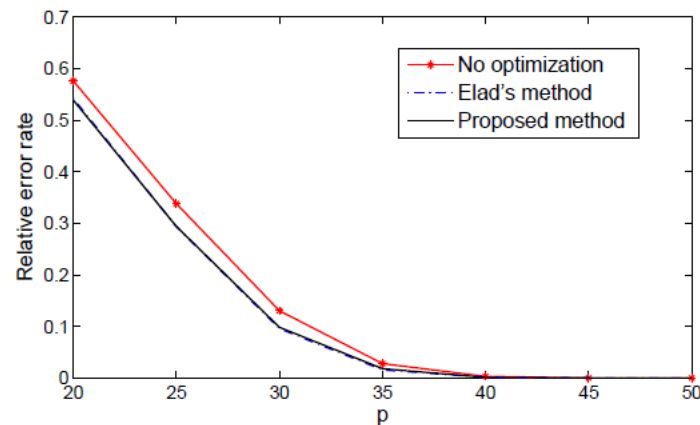
(b) Φ : Random, and Ψ : DCT.

Figure 1: Distribution of off-diagonal elements of \mathbf{G} .



(a)

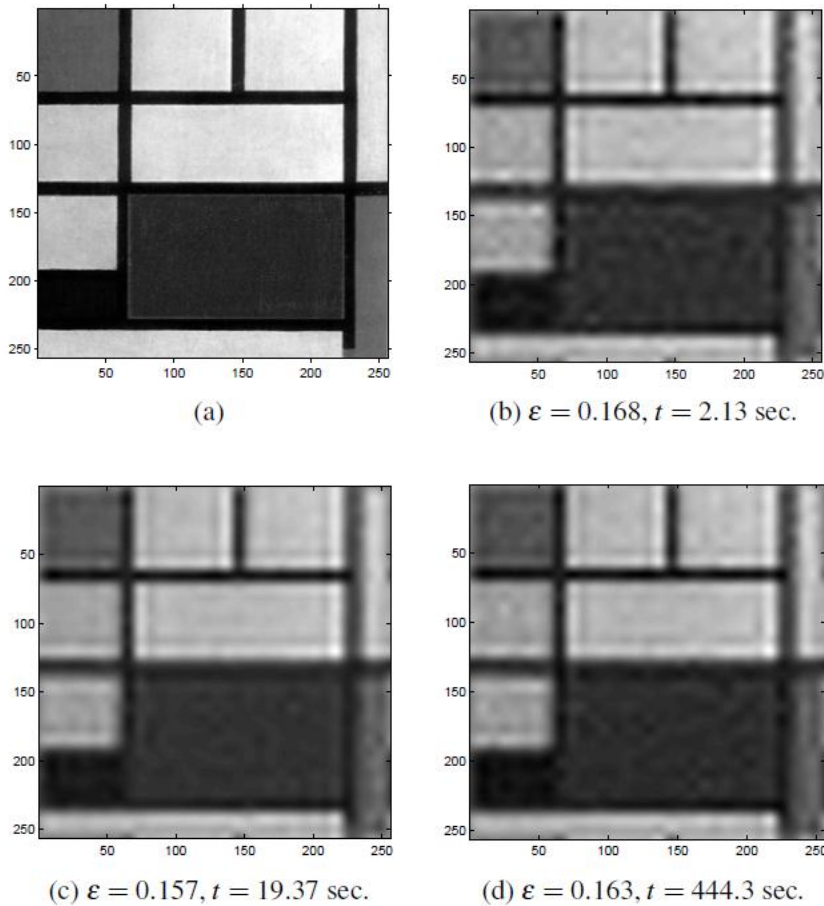
(b)



(c)

Figure 2: Relative error rate vs. the number of measurements p , using (a) IHT, (b) OMP, and (c) BP for reconstruction.

Design CS matrices:



Abdoghasemi et al, On optimization of the measurement matrix for compressive sensing, EUSIPCO 2010

Figure 4: Reconstruction of Mondrian image using HALS-CS method. (a) Original image. Reconstruction with (b) no optimization, (c) proposed optimization, and (d) Elad's optimization of the measurement matrix.

Results

- Leading to smaller values of column-column dot products
- Better reconstruction errors
- For more details refer to Abdoghasemi et al, On optimization of the measurement matrix for compressive sensing, EUSIPCO 2010.

Compressive Classification

Problem definition

- We have extensively examined the issue of reconstruction of signals or images from compressive reconstructions – algorithms, systems as well as theory (theorems).
- Now imagine you had compressive measurements for each of a set of K classes of images.
- The task is to classify the measurements into one of the K categories ***without*** intermediate reconstruction.
- This is called as the problem of **compressive classification**.

Maximum Likelihood Classifier

- Consider a vector \mathbf{y} which is a noisy measurement of vector \mathbf{x} in the following way:
 $\mathbf{y} = \mathbf{x} + \mathbf{n}, \mathbf{n} \sim N(\mathbf{0}, \sigma^2)$.
- Let us suppose that \mathbf{x} belongs to one of P classes, each class containing a single representative vector $\mathbf{s}_i, 1 \leq i \leq P$.
- The likelihood that \mathbf{y} is a noisy sample from the i^{th} class is given as:

$$p(\mathbf{y} | \mathbf{s}_i) = \frac{1}{\sigma \sqrt{2\pi}} \exp\left(-\frac{\|\mathbf{y} - \mathbf{s}_i\|^2}{2\sigma^2}\right)$$

Maximum Likelihood Classifier

- The maximum likelihood classifier assigns to \mathbf{y} the class j such that

$$j = \arg \max_{k \in \{1, 2, \dots, P\}} p(\mathbf{y} | \mathbf{s}_k)$$

$$p(\mathbf{y} | \mathbf{s}_i) = \frac{1}{\sigma \sqrt{2\pi}} \exp\left(-\frac{\|\mathbf{y} - \mathbf{s}_i\|^2}{2\sigma^2}\right)$$

- By using the $-\log$, we see that this reduces to a nearest neighbour classifier with Euclidean distance.

Generalized Maximum Likelihood Classifier (GMLC)

- Consider the earlier problem was an image classification problem where we had P image templates in a database.
- We observed Gaussian noisy versions of one of these templates and wanted to determine which one it was.
- Now in addition, let us suppose that the noisy version of the image were acquired in some different “pose”, i.e. with some translations and/or rotation.
- Let the pose parameters be denoted by a vector θ which belong to a set of values Θ .
- In such a case, this becomes a joint problem – of classification as well as pose estimation.

Generalized Maximum Likelihood Classifier (GMLC)

- The problem is solved as follows:

$$j = \arg \max_{k \in \{1, 2, \dots, P\}} p(\mathbf{y} | \mathbf{s}_k, \tilde{\boldsymbol{\theta}}_k)$$
$$p(\mathbf{y} | \mathbf{s}_i, \tilde{\boldsymbol{\theta}}_i) = \frac{1}{\sigma \sqrt{2\pi}} \exp \left(- \frac{\| \mathbf{y} - \mathbf{s}_i(\tilde{\boldsymbol{\theta}}_i) \|^2}{2\sigma^2} \right)$$
$$\tilde{\boldsymbol{\theta}}_i = \arg \max_{\boldsymbol{\theta} \in \Theta} p(\mathbf{y} | \mathbf{s}_i, \boldsymbol{\theta})$$

This denotes image \mathbf{s}_i deformed by pose parameter $\tilde{\boldsymbol{\theta}}_i$. By “deformation” we mean rotation and/or translation in this example. In general, it could mean any other type of transformation including geometric scaling, blurring, etc.

Matched filter: special case of GMLC

- If the parameters θ denoted pure translation, then one way to classify \mathbf{y} is to determine for which i , the following quantity is maximized:

$$\int \mathbf{s}_i(\mathbf{t} - \theta) \mathbf{y}(\mathbf{t}) d\mathbf{t}$$

where $\mathbf{t} = (x, y)$ denotes spatial coordinates.

- This is called a **matched filter** and it is equivalent to GMLC if all the candidate signals \mathbf{s}_i had the same magnitude, and we assume additive white Gaussian noise.

Compressive Classification: MLC

- Consider the following compressive acquisition model:

$$y = \Phi x + \eta, \eta \sim N(0, \sigma^2),$$

$$y \in R^m, x \in R^n, \Phi \in R^{m \times n}, m \ll n$$

- The MLC for this case is now:

$$j = \arg \max_{k \in \{1, 2, \dots, P\}} p(\mathbf{y} \mid \Phi \mathbf{s}_k)$$

$$p(\mathbf{y} \mid \Phi \mathbf{s}_i) = \frac{1}{\sigma \sqrt{2\pi}} \exp\left(-\frac{\|\mathbf{y} - \Phi \mathbf{s}_i\|^2}{2\sigma^2}\right)$$

Compressive Classification: GMLC

- But the compressive measurement \mathbf{y} could be acquired from an image which was acquired in a different pose than any of the images in the database.
- The GMLC is now given as:

$$j = \arg \max_{k \in \{1, 2, \dots, P\}} p(\mathbf{y} | \Phi \mathbf{s}_k, \tilde{\boldsymbol{\theta}}_k)$$
$$p(\mathbf{y} | \Phi \mathbf{s}_i, \tilde{\boldsymbol{\theta}}_i) = \frac{1}{\sigma \sqrt{2\pi}} \exp \left(-\frac{\|\mathbf{y} - \Phi \mathbf{s}_i(\tilde{\boldsymbol{\theta}}_i)\|^2}{2\sigma^2} \right)$$
$$\tilde{\boldsymbol{\theta}}_i = \arg \max_{\boldsymbol{\theta} \in \Theta} p(\mathbf{y} | \Phi \mathbf{s}_i, \boldsymbol{\theta})$$

This has an interesting name – the **smashed filter** (derived from the name “matched filter”), taking into account the compressive nature of the measurements.

RIP and classification

- These are all essentially nearest neighbour classifiers with Euclidean distance.
- What is special about these classifiers?
- Let us assume that the sensing matrix Φ (and hence $\Phi\mathbf{U}$ for any orthonormal \mathbf{U}) obey the restricted isometry property.
- Then for k -sparse signals \mathbf{s}_1 and \mathbf{s}_2 and RIC δ_{2k} , we have:

$$(1 - \delta_{2k}) \|\mathbf{s}_1 - \mathbf{s}_2\|^2 \leq \|\Phi(\mathbf{s}_1 - \mathbf{s}_2)\|^2 \leq (1 + \delta_{2k}) \|\mathbf{s}_1 - \mathbf{s}_2\|^2$$

RIP and classification

- If m is sufficiently large and Φ obeys RIP, we know that the distance between any two sparse vectors in the original n -dimensional space is preserved when you multiply them by Φ .
- In other words, under the conditions of sparsity and RIP, the distance between two compressive measurements is approximately equal to the distance between the original signals.
- The RIP helps classification here – but m needs to be carefully chosen – for an appropriate k .
- Basically this means that the nearest neighbour algorithm from compressive measurements will work almost as well as from the original measurements.

Theorem by Baraniuk and Wakin

- Consider a manifold with K degrees of freedom in \mathbb{R}^n . Consider $0 < \varepsilon < 1$, $0 < \rho < 1$ and a random matrix Φ in $\mathbb{R}^{m \times n}$ with orthonormal rows where m is $O(K \log(n) \log(1/\rho)/\varepsilon^2)$. If $m \leq n$, then with probability at least $1-\rho$, we have the following:

$$(1 - \varepsilon) \sqrt{\frac{m}{n}} \leq \frac{\|\Phi(x_1 - x_2)\|_2}{\|x_1 - x_2\|_2} \leq (1 + \varepsilon) \sqrt{\frac{m}{n}}$$

where x_1 and x_2 are any two points on the manifold.

Theorem by Baraniuk and Wakin

- Clearly for large m , the distances will be preserved better – as per this theorem.
- This theorem makes no assumption on signal sparsity, but just on the dimension of the manifold.

Results



(a) Tank



(b) School Bus



(c) Truck

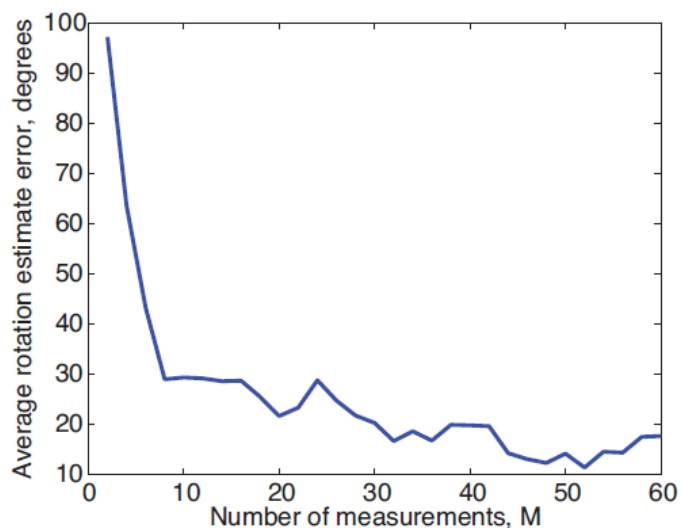
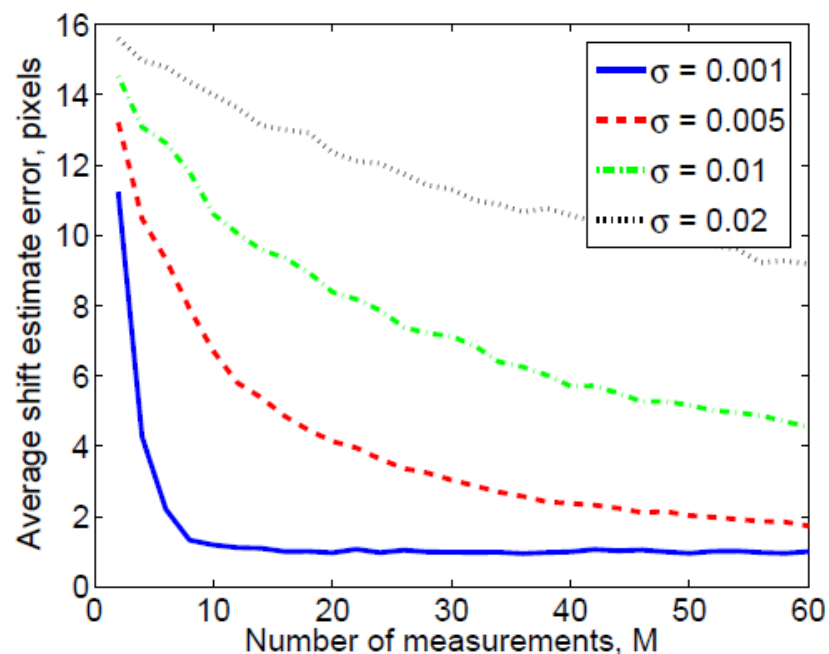
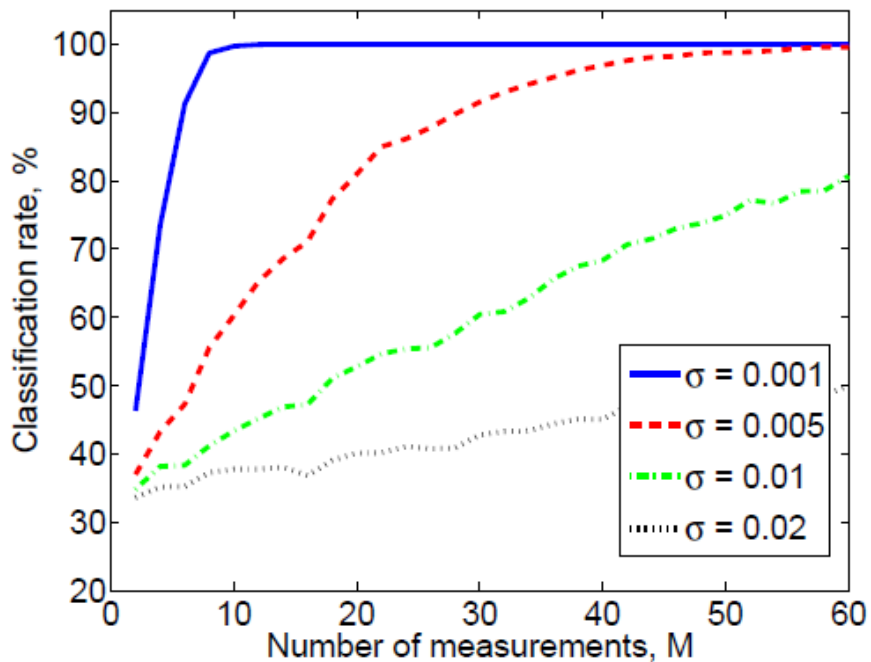
Figure 3. *Models used for classification experiments.*

Image source: [Davenport et al, "The smashed filter for compressive classification and target recognition"](#)

Image size: 128 x 128

Compressive measurements taken by a Rice single pixel camera. Though the sensing matrix of the camera does not obey RIP since it contains values that are 0 or 1, it can be converted into a matrix with entries that are either -1 or +1. This is by taking two measurements of the same scene, where the second measurement is taken by flipping the 0 and 1 values in the first sensing matrix.

Results



[Image source: Davenport et al, "The smashed filter for compressive classification and target recognition"](#)