

LAB 10 : NOISE SIMULATION

ITI_ANALOG_IC_DESIGN

Lab 10 : NOISE SIMULATION



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Analog IC Design Lab 10 Noise Simulation

PART 1: LPF AC Noise Analysis

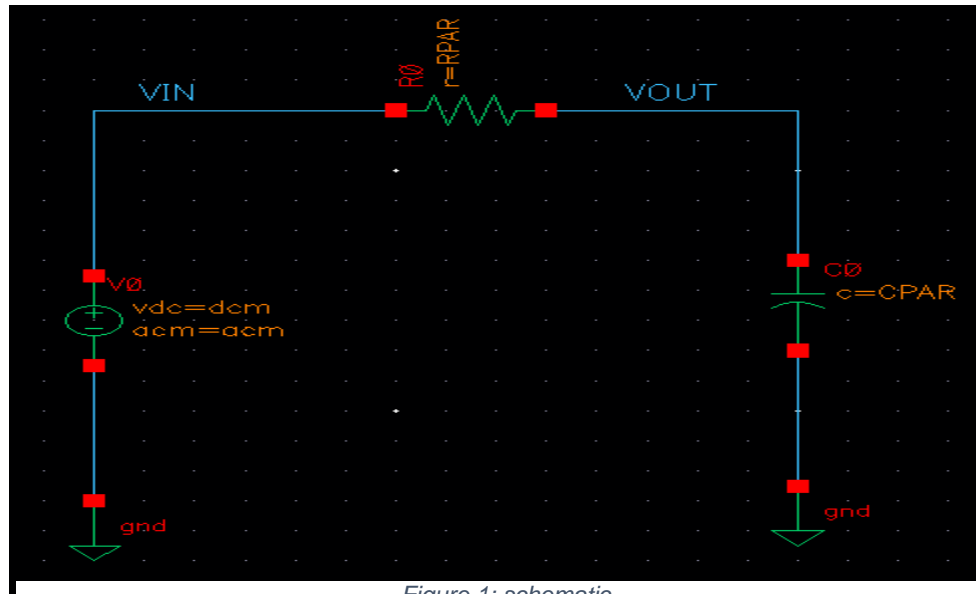
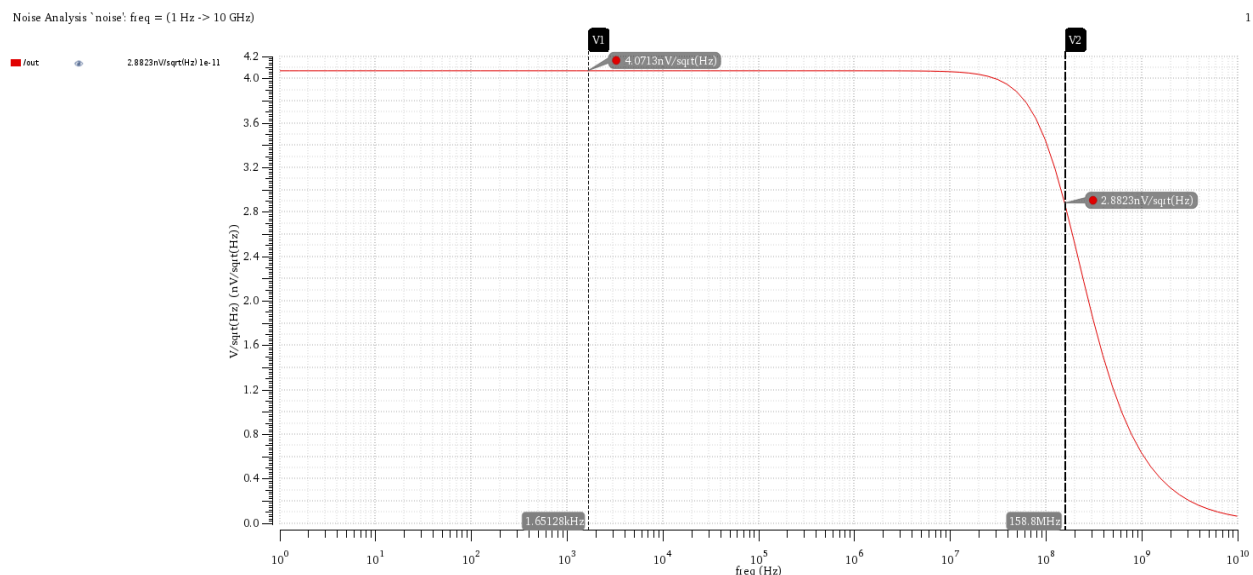


Figure 1: schematic

6) Report output noise vs frequency. Annotate voltage noise density and bandwidth in the plot.



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7) Calculate rms output noise using rms noise function in the calculator.

Test	Output	Nominal	Spec	Weight	Pass/Fail
lab_10_noise:part1:1	RMS	64.32u			
lab_10_noise:part1:1	BW	158.8M			

Hand analysis:

$$V_n = \sqrt{4KTR} = 4.07 \text{ nV} / \sqrt{\text{HZ}}$$

$$\text{BW} = \frac{1}{2\pi \times RC} = 159.1549 \text{ MHZ}$$

$$\text{RMS} = \sqrt{\frac{KT}{C}} = 64.36 \text{ uV}$$

	V_n	BW	RMS
Simulation	4.0713 nV/ $\sqrt{\text{HZ}}$	158.8 MHZ	64.32 uV
Hand analysis	4.07 nV/ $\sqrt{\text{HZ}}$	159.1549 MHZ	64.36 uV

Plot output noise overlaid on the same plot. Using log-scale for y-axis. Comment on the results.

Noise Response

1

output noise: V / sqrt(Hz)
 output noise: V / sqrt(Hz) 4.0713nV/sqrt(Hz) 1000
 output noise: V / sqrt(Hz) 12.8747nV/sqrt(Hz) 10000
 output noise: V / sqrt(Hz) 40.7133nV/sqrt(Hz) 100000
 output noise: V / sqrt(Hz) 128.747nV/sqrt(Hz) 1000000

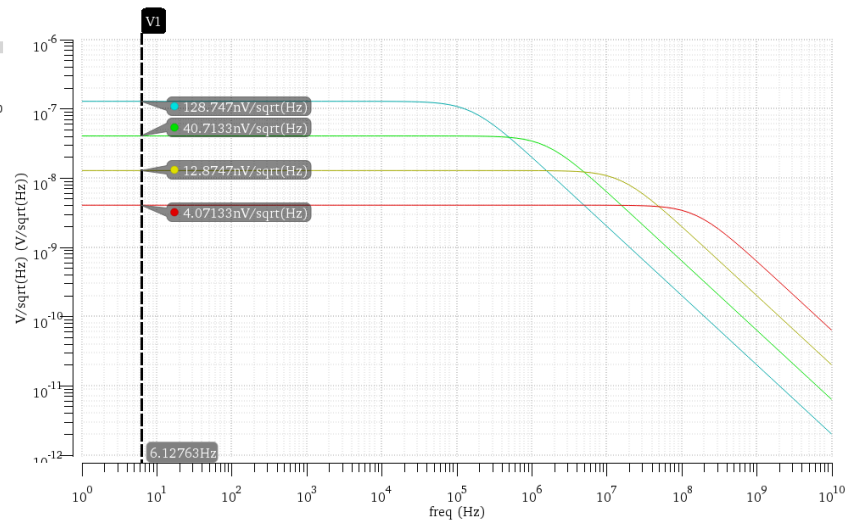


Figure 3: V_n VS 4 VALUES OF R

COMMENT

When R increases, the noise density also increases. This is expected because the thermal noise spectral density is given by $V_n = \sqrt{4KTR}$, so larger R directly raises V_n .

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11) Calculate the rms noise using the calculator. Comment on the results.

Parameters: RPAR=1k					
1	lab_10_noise:part1:1	RMS	64.32u		
1	lab_10_noise:part1:1	BW	158.8M		
Parameters: RPAR=10k					
2	lab_10_noise:part1:1	RMS	64.62u		
2	lab_10_noise:part1:1	BW	15.88M		
Parameters: RPAR=100k					
3	lab_10_noise:part1:1	RMS	64.65u		
3	lab_10_noise:part1:1	BW	1.588M		
Parameters: RPAR=1M					
4	lab_10_noise:part1:1	RMS	64.66u		
4	lab_10_noise:part1:1	BW	158.8k		

Figure 4: rms and BW from simulation

COMMENT

The calculated RMS output noise voltage remains approximately constant as expected despite the change in the resistor value R because there is no dependence for rms on R where $\text{RMS} = \sqrt{\frac{KT}{C}}$, but BW decrease because it depends on R values where $\text{BW} = \frac{1}{2\pi \times RC}$.



PART 2: LPF Transient Noise Analysis

3) Report the noise output waveform. Annotate the min and max values.

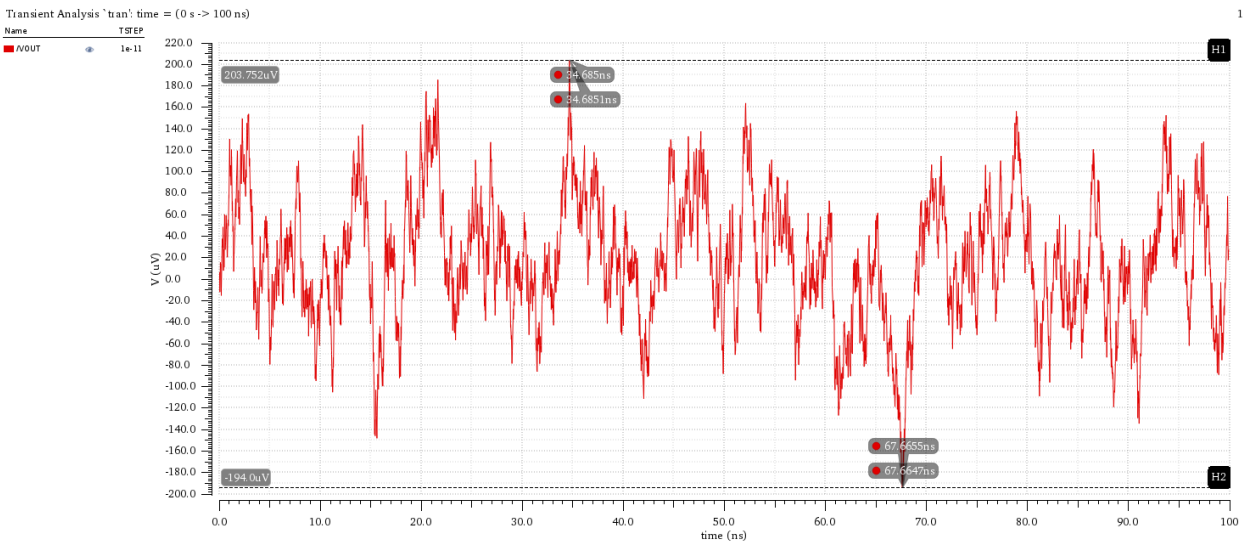


Figure 6: Noise output waveform

Test	Output	Nominal
lab_10_noise:part1:1	Y_MAX	203.8u
lab_10_noise:part1:1	Y_MIN	-194.2u

Figure 5: Y_MAX and Y_MIN from simulation

4) Use the rms function in the calculator to calculate the rms noise Compare it to the value calculated in Part 1.

Expression	Value
rms(v("/VOUT"...	60.88E-6

	RMS
PART 1	64.32 uV
PART 2	60.88 uV

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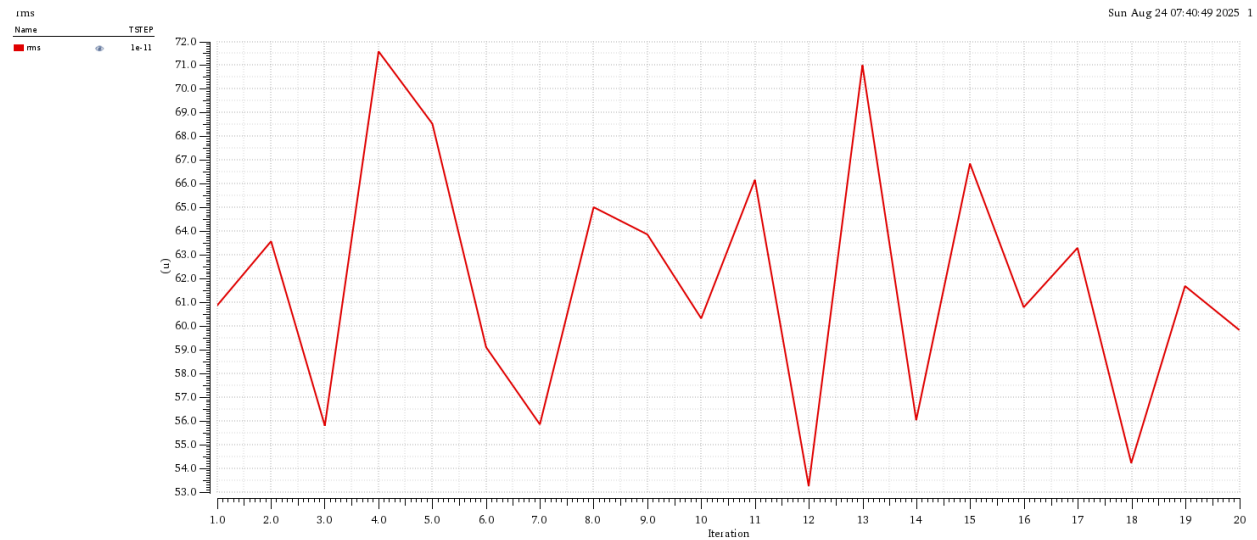


5) Repeat the simulation with $TSTEP = TAU/10$. Does the calculated rms noise increase or decrease? Why?

lab_10_noise:part1:1	rms	61.59u			
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- The simulated RMS noise result shows an increase, which is not the expected trend. Normally, increasing T_STEP should lower the maximum noise frequency (F_{max}) and therefore reduce the RMS noise value. The observed increase is likely due to a limitation or artifact of the simulation setup/machine

7) Report the rms noise vs iteration.



8) Use the calculator to calculate the average rms noise. Compare the calculated value with the rms noise previously obtained in Part 1 and Part 2.

lab_10_noise:part1:1	AVERAGE	61.97u			
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	RMS
PART 1	64.32 uV
PART 2	60.88 uV
AVERAGE	61.97 uV



PART 3: 5T OTA AC Noise Analysis

- 1) Create a new testbench. Connect the 5T OTA you designed in Lab 07 in unity-gain buffer configuration. Similar to Part 1, run ac noise analysis.

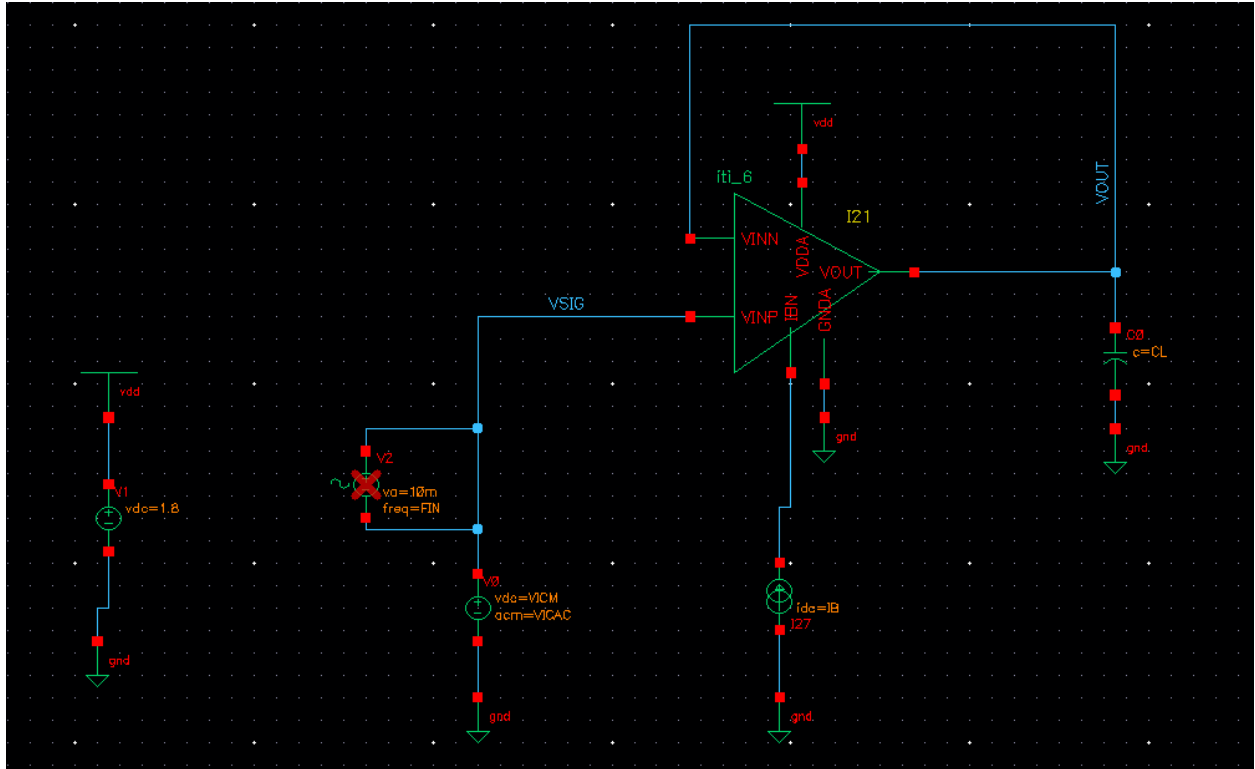


Figure 8: SCHEMATIC

- 2) Report output thermal noise vs frequency. Annotate noise density and bandwidth in the plot. Compare the simulation results with hand analysis.

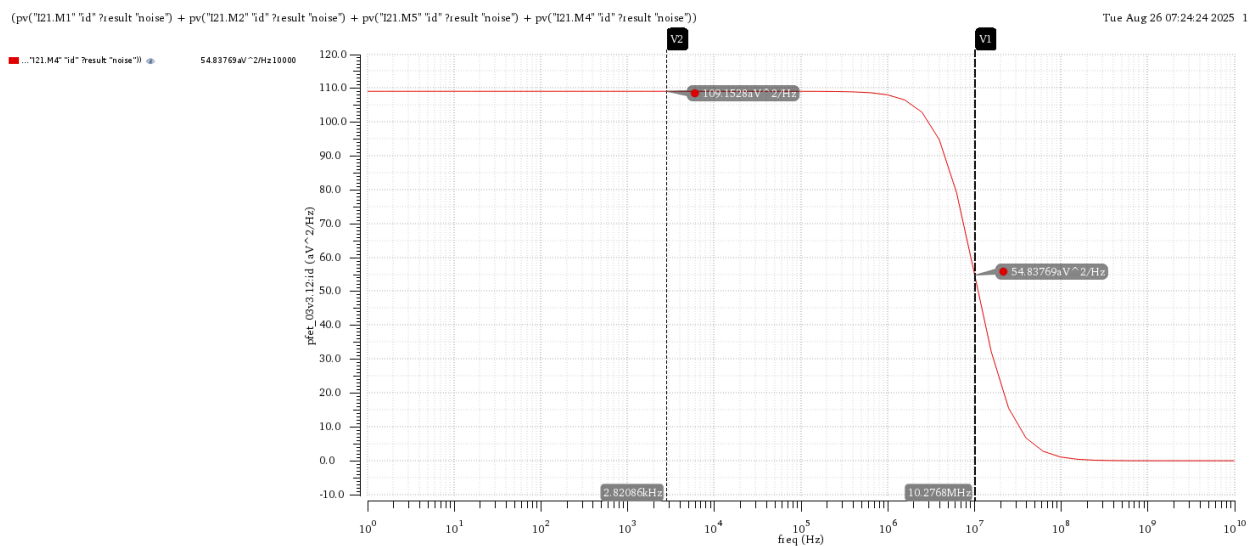


Figure 9: output thermal noise VS freq

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Hand analysis:

Assume $\gamma = 2/3$

From lab 7: $gm_n = 320\mu S$, $gm_p = 200\mu S$, $CL = 5p$

$$\overline{V_{nout}^2} = \frac{8KT\gamma}{gm_n} \left(1 + \frac{gm_p}{gm_n}\right) = 108.517 \text{ aV}^2/\text{HZ}$$

$$BW = GBW = \frac{gm_n}{2\pi \times CL} = 10.1859 \text{ MHZ}$$

	$\overline{V_{nout}^2}$	BW
Simulation	109.1528 aV ² / HZ	10.2768 MHZ
Hand analysis	108.517 aV ² / HZ	10.1859 MHZ

3) Report total output noise (thermal + flicker) vs frequency. Estimate the Flicker noise corner.

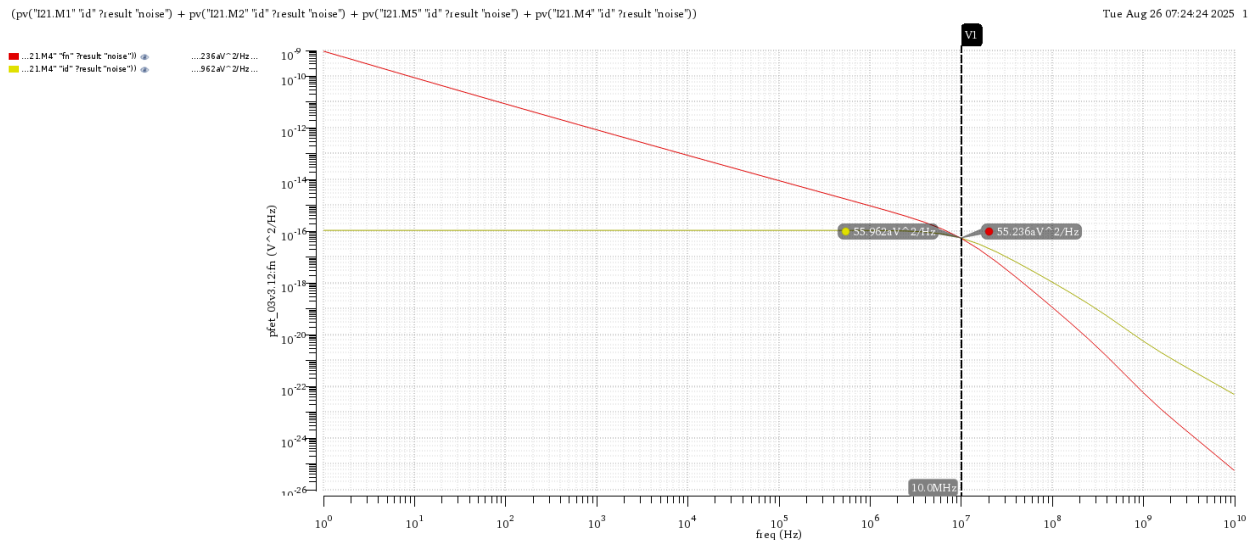


Figure 10: thermal and flicker noise VS freq

The flicker noise corner frequency (f_c): is the frequency at which the flicker ($1/f$) noise spectral density equals the thermal (white) noise spectral density.

Flicker corner $\approx 10 \text{ MHZ}$

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4) Calculate rms output noise (calculate the rms noise due to thermal noise only using Noise Summary).

Device	Param	Noise Contribution	% Of Total
/I21/M2	fn	6.2516e-05	35.94
/I21/M1	fn	6.20955e-05	35.46
/I21/M5	fn	2.53055e-05	5.89
/I21/M4	fn	2.51117e-05	5.80
/I21/M2	id	2.3342e-05	5.01
/I21/M1	id	2.32691e-05	4.98
/I21/M4	id	1.95629e-05	3.52
/I21/M5	id	1.81707e-05	3.04
/I21/M3	fn	4.58605e-06	0.19
/I21/M3	id	3.26239e-06	0.10
/I21/M0	fn	2.22916e-06	0.05
/I21/M0	id	1.98772e-06	0.04

Integrated Noise Summary (in V) Sorted By Noise Contributors
 Total Summarized Noise = 0.000104283
 Total Input Referred Noise = 0.00263586
 The above noise summary info is for noise data with FIN = 10000.0

I will neglect thermal noise for M0,M3 (tail CM)

$$\text{RMS} = \sqrt{id_1^2 + id_2^2 + id_4^2 + id_5^2} \approx 42.2 \text{ uV}$$

Hand analysis:

$$\text{RMS} = \sqrt{BW \times \overline{V_{nout}^2} \times \frac{\pi}{2}} \approx 41.67 \text{ uV}$$

5) Compare the simulation results (noise density, bandwidth, and rms) with hand analysis.

	$\overline{V_{nout}^2}$	BW	RMS
Simulation	109.1528 aV ² / HZ	10.2768 MHZ	42.2 uV
Hand analysis	108.517 aV ² / HZ	10.1859 MHZ	41.67 uV



PART 4: 5T OTA Transient Noise Analysis

1) Create a new simulation configuration. Keep the 5T OTA connected in unity-gain buffer configuration.

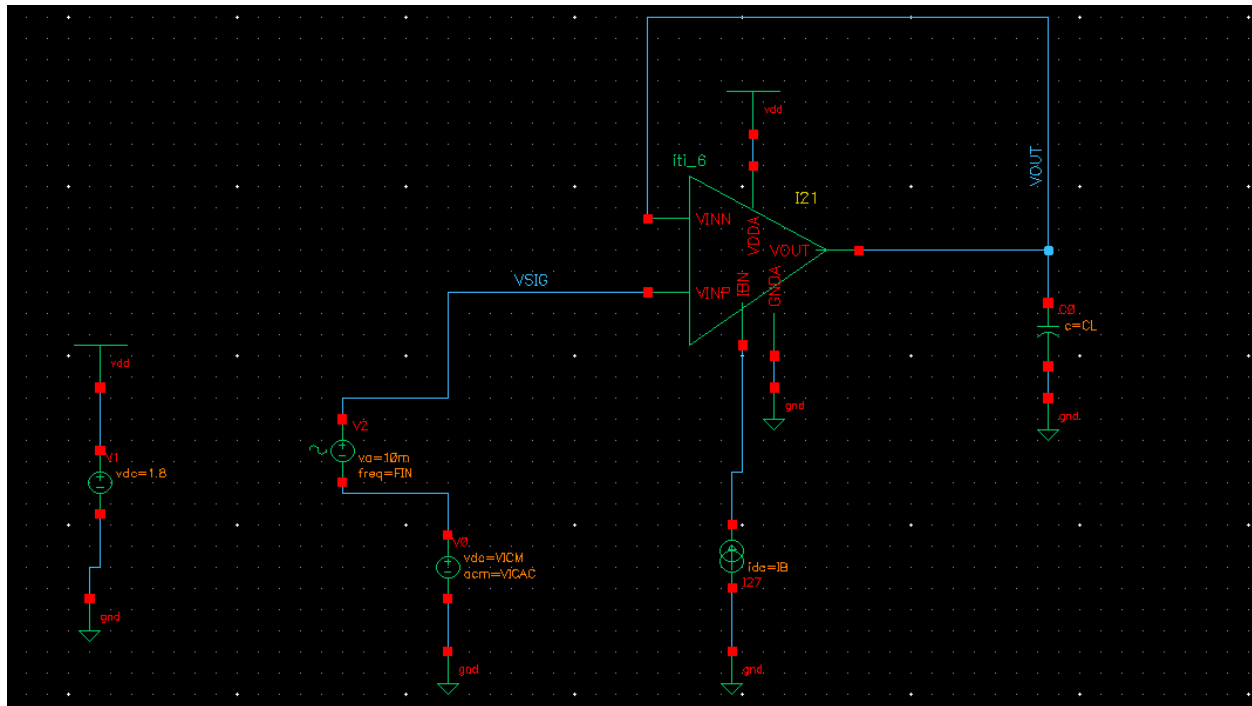
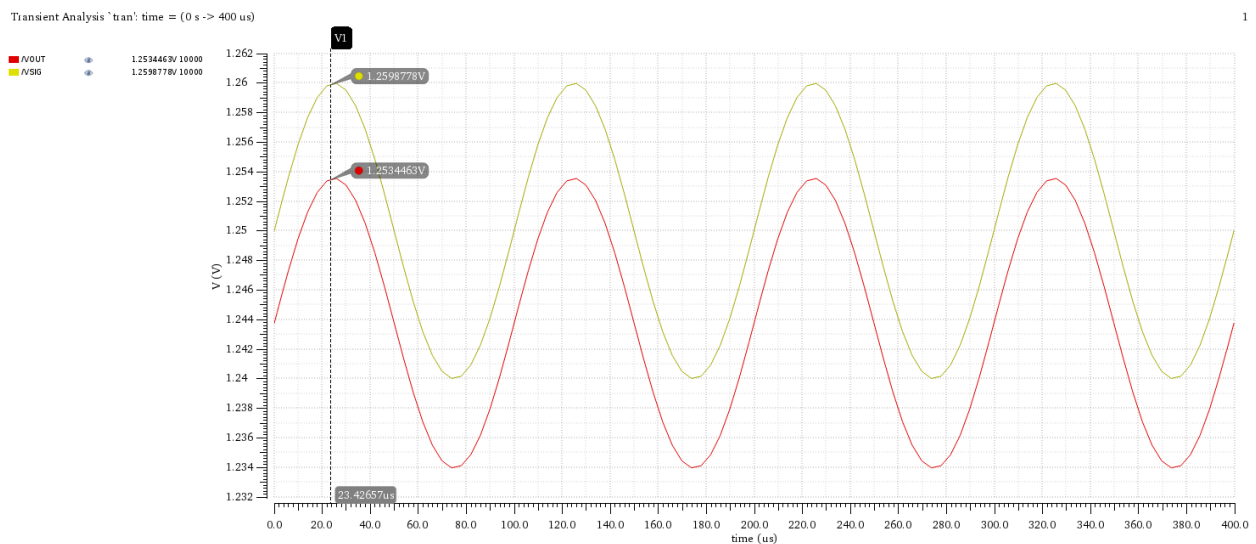


Figure 11:SCHEMATIC

4) Plot input and output overlaid and make sure they match well (verify that the circuit behaves as a buffer).



VOUT is approximately equal to VIN, confirming basic buffer operation. The small error exists because the OTA's finite open-loop gain (AOL is not $=\infty$) prevents it from being an ideal buffer.

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6) Report the “noisy” output waveform (zoom-in to highlight the noise). Notice that output signal and noise are superimposed.

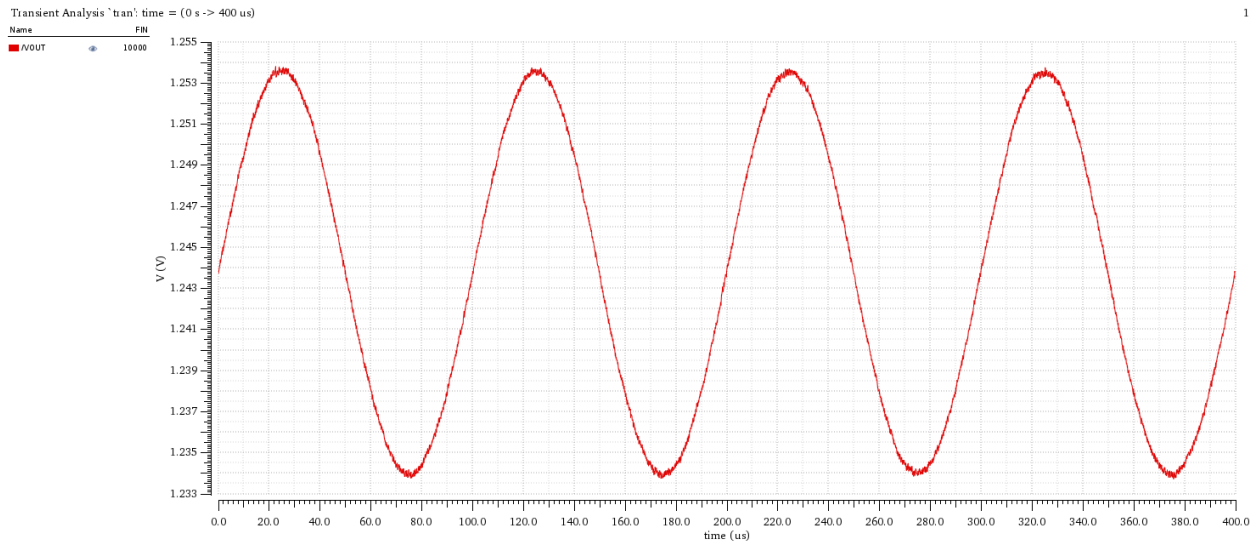


Figure 12: “noisy” output waveform

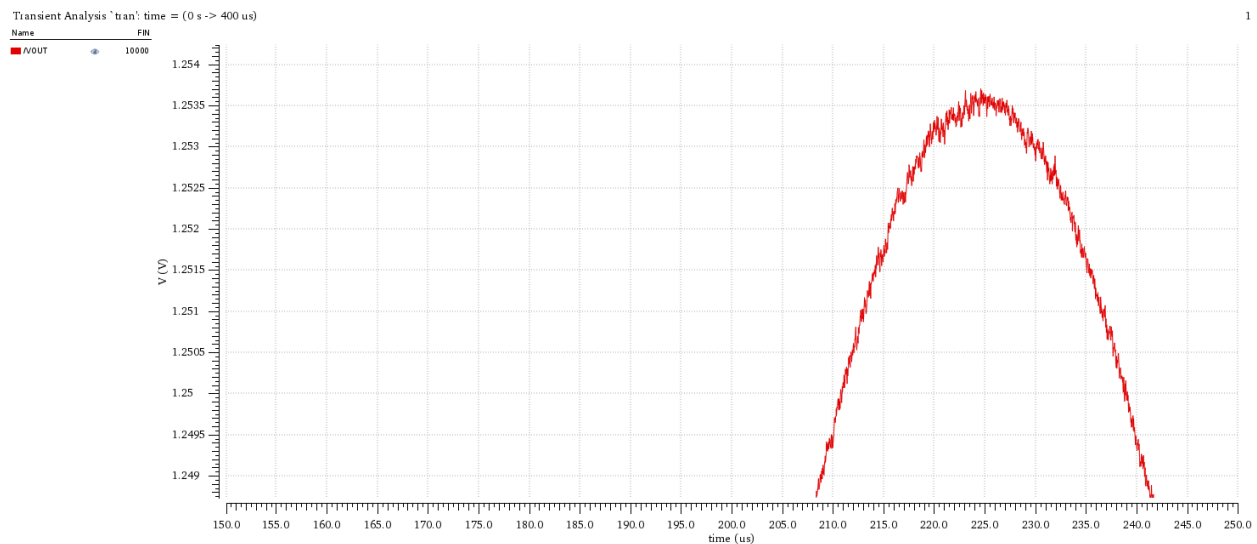


Figure 13: “noisy” output waveform (zoom-in to highlight the noise).

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8) Report the rms noise vs iteration.

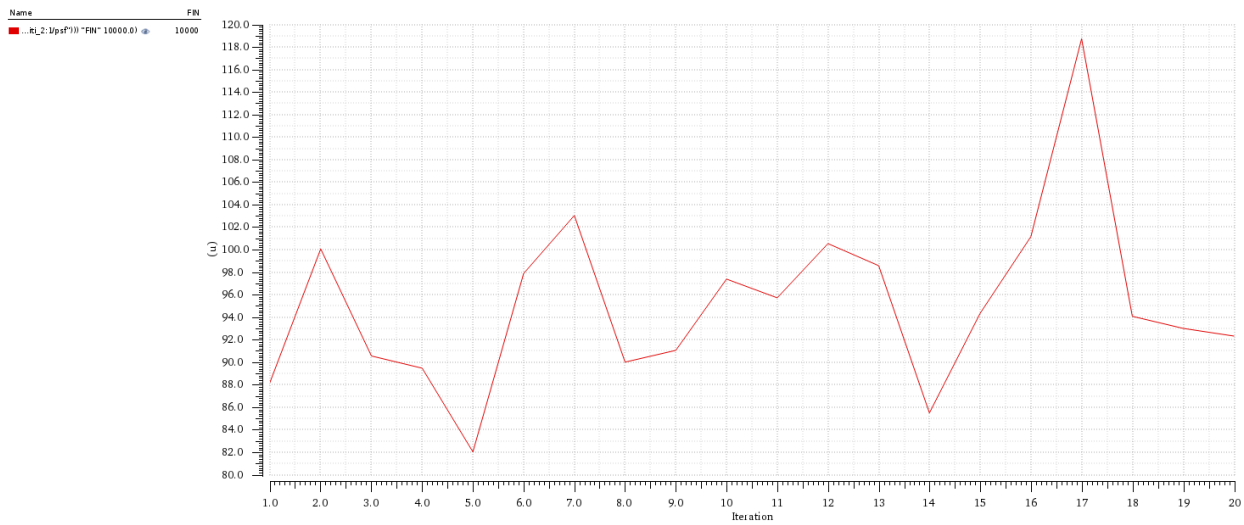


Figure 14: rms noise VS iteration

9) Use the calculator to calculate the average rms noise. Compare the calculated value with the rms noise previously obtained in Part 3.

	FIN	average...psf"))))
1	10.00E3	95.47E-6

	RMS
Part 3	95.47 uV
Part 4	42.2 uV

As expected, the rms noise obtained in Part 4 is larger than in Part 3. In Part 3 we only considered the thermal noise contribution, while in Part 4 both thermal and flicker noise are included, which increases the total noise level.