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Homework 2

ADSI Problem 1.4: Filter decomposition

1) Decompose a FIR filter with one zero outside the unit circle

Let the system function for a FIR filter be given by

$$H(z) = 1 - 3z^{-1} + \frac{5}{2}z^{-2} - z^{-3}$$

1. Decompose the system function into a product of a minimum-phase filter and an all-pass filter, $H(z) = H_{min}(z)H_{ap}(z)$.

Any system function can be decomposed into a product of a minimum-phase filter and an all-pass filter using the following formula:

$$H(z) = H_{\rm min}(z) H_{\rm ap}(z)$$

The minimum phase filter can be computed as follows:

$$H_{\min}(z) = \prod_{i} -\frac{1}{a_{i}^{*}} H_{i}(z) (1 - a_{i}z^{-1})$$

where $H_i(z)$ corresponds to the part of the transfer function where the i'th zero is inside the unit circle and $a_i = \frac{1}{z_i^*}$ and z_i is the zero outside the unit circle.

The allpass filter can be calculated:

$$H_{\rm ap}(z) = \prod_{i} \frac{z^{-1} - a_i^*}{1 - a_i z^{-1}}$$

The filter decomposition algorithm has following steps:

 convert transfer function into pole-zero representation in order to find the zeros that are outside the unit circle

- 2. compute a and its conjugate a^*
- 3. find $H_1(z)$ which corresponds to the part of the transfer function where zeros are inside the unit circle
- 4. plugin the numbers for the formula for the minimum-phase filter
- 5. plugin the numbers for the formula for the allpass filter
- 6. put everything together:

Step 1: convert transfer function into pole-zero representation in order to find the zeros that are outside the unit circle

```
syms z;
rts = roots([1, -3, 5/2, -1])
rts = 3 \times 1 complex
   2.0000 + 0.0000i
   0.5000 + 0.5000i
   0.5000 - 0.5000i
H_outside = 1; % Represents part of H where zeros are outside the unit circle
H_inside = 1; % Represents part of H where zeros are inside the unit circle
zeros outside = [];
for i = 1:numel(rts)
    root = rts(i);
    if abs(root) > 1
        H_{outside} = H_{outside} * (1 - root*z^-1);
         zeros outside = [zeros outside; root];
    else
         H inside = H inside * (1 - root*z^-1);
    end
end
z0 = 0;
if numel(zeros outside) == 1
    z0 = zeros outside(1);
else
    disp('Something is wrong! The transfer function has more than one zero outside the unit cit
end
H outside = expand(H outside);
H_inside = expand(H_inside);
% Sanity check
H = expand(H_inside * H_outside)
H =
   \frac{5}{2z^2} - \frac{3}{z} - \frac{1}{z^3} + 1
```

The zero-pole representation of the transfer function is:

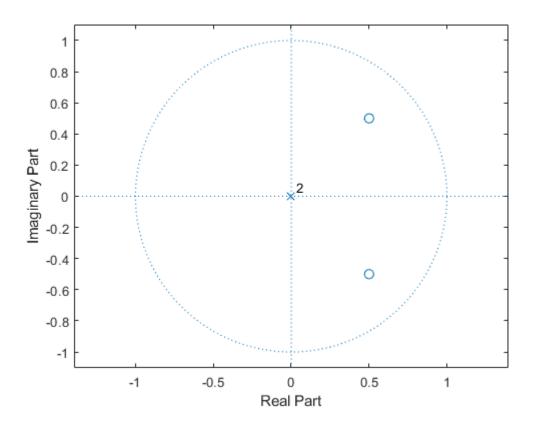
H_outside

$$H_{\text{outside}} = 1 - \frac{2}{z}$$

H_inside

$$H_{inside} = \frac{1}{2z^2} - \frac{1}{z} + 1$$

zplane([1, -1, 1/2]) % Should all have zeros inside the unit circle



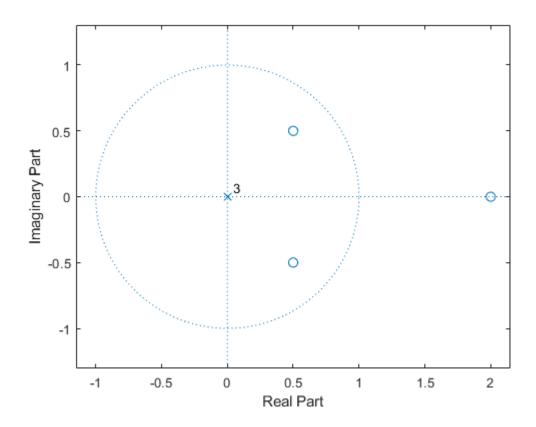
$$H(z) = (1-2z^{-1}) \bigg(1-z^{-1} + \frac{1}{2}z^{-2} \bigg)$$

ans =
$$1 \times 4$$

1.0000 -3.0000 2.5000 -1.0000

To find the zero z_0 that is outside the unit circle, we can plot zplane:

zplane([1, -3, 5/2, -1]);



In this exercise, the zero outside the unit circle is $z_0 = 2$.

z0

z0 = 2

Step 2: compute a and its conjugate a^*

$$a = 1/z0$$

a = 0.5000

 $a_{conj} = 0.5000$

$$a = \frac{1}{z_0} = \frac{1}{2}$$

$$a^* = \left(\frac{1}{z_0}\right)^* = \left(\frac{1}{2}\right)^* = \frac{1}{2}$$

Step 3: find $H_1(z)$ which corresponds to the part of the transfer function where zeros are inside the unit circle

H_inside

H_inside =
$$\frac{1}{2z^2} - \frac{1}{z} + 1$$

The zero-pole representation of the transfer function is:

$$H(z) = (1 - 2z^{-1})\left(1 - z^{-1} + \frac{1}{2}z^{-2}\right)$$

therefore $H_1(z)$ is

$$H_1(z) = 1 - z^{-1} + \frac{1}{2}z^{-2}$$

Step 4: plugin the numbers for the formula for minimum-phase filter

$$H_{\min}(z) = -\frac{1}{a^*}H_1(z)(1 - az^{-1})$$

$$H_{\min}(z) = -\frac{1}{\left(\frac{1}{2}\right)} \left(1 - z^{-1} + \frac{1}{2}z^{-2}\right) (1 - 0.5z^{-1})$$

$$H_{\min}(z) = -2\Big(1-z^{-1}+\frac{1}{2}z^{-2}\Big)(1-0.5z^{-1})$$

ans =
$$1 \times 4$$

-2.0000 3.0000 -2.0000 0.5000

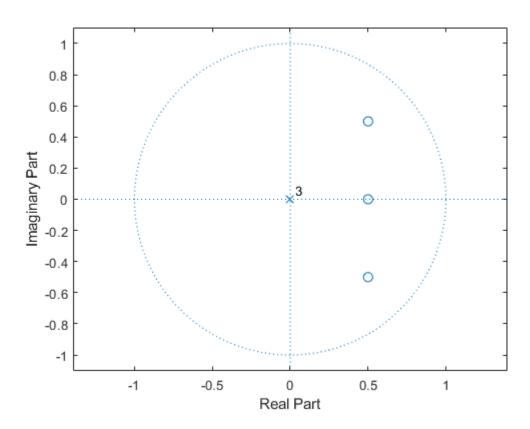
$$H_{\min}(z) = -2 + 3z^{-1} - 2z^{-2} + \frac{1}{2}z^{-3}$$

$$H_{min} = expand((-1/a_{conj}) * H_{inside} * (1 - a*z^-1))$$

H_min =
$$\frac{3}{z} - \frac{2}{z^2} + \frac{1}{2z^3} - 2$$

isminphase(H_min_b)

zplane(H_min_b)



Step 5: compute the allpass filter:

$$H_{\rm ap}(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$

$$H_{\rm ap}(z) = \frac{z^{-1} - 0.5}{1 - 0.5z^{-1}}$$

$$H_{ap} = (z^{-1} - a_{conj}) / (1 - a*z^{-1})$$

H_ap =
$$-\frac{\frac{1}{z} - \frac{1}{2}}{\frac{1}{2z} - 1}$$

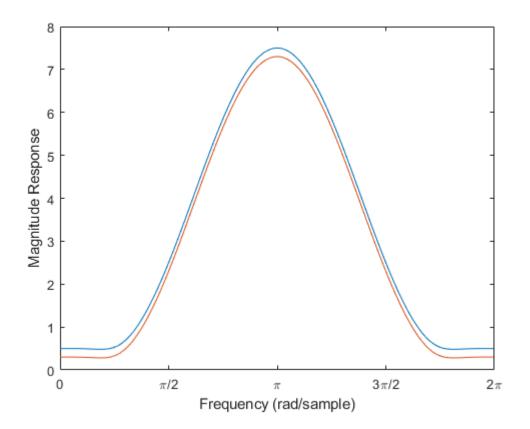
Step 6: put everything together:

$$\begin{split} H_{\min}(z) &= -2 + 3z^{-1} - 2z^{-2} + \frac{1}{2}z^{-3} \\ H_{\mathrm{ap}}(z) &= \frac{z^{-1} - 0.5}{1 - 0.5z^{-1}} \\ H(z) &= H_{\min}(z)H_{\mathrm{ap}}(z) \\ H(z) &= \left(-2 + 3z^{-1} - 2z^{-2} + \frac{1}{2}z^{-3}\right) \left(\frac{z^{-1} - 0.5}{1 - 0.5z^{-1}}\right) \end{split}$$

- 2) Show that H(z) and $H_{\min}(z)$ have the same magnitude response
- 2. Demonstrate that H(z) and $H_{min}(z)$ have the same amplitude response.

To show that H(z) and $H_{\min}(z)$ have the same magnitude response, we can plot them. If the two graphs are on top of each other then they have the same magnitude response.

```
% Coefficients for H(z)
b = [1, -3, 5/2, -1];
a = 1;
[H, w] = freqz(b,a,'whole');
% Coefficients for H_min(z)
H_{min_b} = [-2, 3, -2, 1/2];
H_{min}a = 1;
[H_min_z, H_min_w] = freqz(H_min_b, H_min_a, 'whole');
% plot(H_min_w, log10(abs(H_min_z)));
% The offset ensures that the two graphs are not on top of each other
offset = -0.2;
plot(w, abs(H), H_min_w, abs(H_min_z) + offset);
set(gca,'XTick',0:pi/2:2*pi)
set(gca,'XTickLabel',{'0','\pi/2','\pi','3\pi/2','2\pi'})
xlabel('Frequency (rad/sample)')
ylabel('Magnitude Response')
xlim([0, 2*pi]);
```

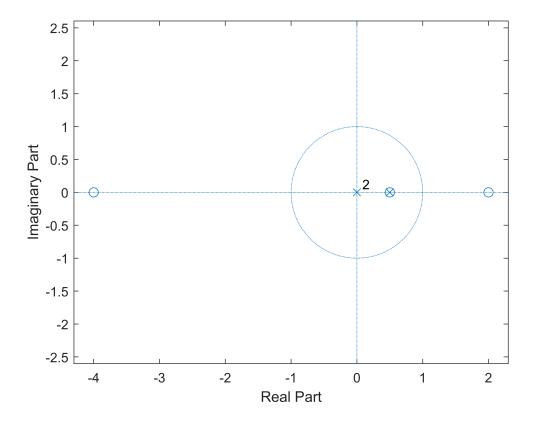


3) Decompose a filter that has two zeros outside the unit circle

3. Repeat for the following filter

$$H(z) = \frac{1 + \frac{3}{2}z^{-1} - 9z^{-2} + 4z^{-3}}{1 - \frac{1}{2}z^{-1}}$$

Plot the zeros and the poles:



From the plot, we can see that there is a zero and a pole at 0.5. They will cancel each other.

rts = 3×1 -4.0000 2.0000 0.5000

Now, the transfer function can be rewritten as:

$$H(z) = \frac{(1-(-4)z^{-1})(1-2z^{-1})\left(1-\frac{1}{2}z^{-1}\right)}{1-\frac{1}{2}z^{-1}} = (1-(-4)z^{-1})(1-2z^{-1}) = (1-(-4)z^{-1})(1-2z^{-1})$$

Let us find the coefficients of the rewritten transfer function:

$$H_{\text{new}} = \exp((1+4*z^{-1}) * (1-2*z^{-1}))$$

$$H_{\text{new}} = \frac{2}{z} - \frac{8}{z^2} + 1$$

% Automatically extract the coefficients
b_new = coeffs(expand(H_new * z^2), 'all')

```
b new = (1 \ 2 \ -8)
% Find the zeros of H new
rts = roots(b_new)
rts =
syms z;
H_outside = 1; % Represents part of H where zeros are outside the unit circle
H_inside = 1; % Represents part of H where zeros are inside the unit circle
zeros_outside = [];
for i = 1:numel(rts)
    root = rts(i);
    if abs(root) > 1
        H_outside = H_outside * (1 - root*z^-1);
        zeros_outside = [zeros_outside; root];
    else
        H_inside = H_inside * (1 - root*z^-1);
    end
end
% Sanity check
H = expand(H_inside * H_outside)
H =
  \frac{2}{z} - \frac{8}{z^2} + 1
% Finally compute the H_min and H_ap
H_{min} = 1;
H_ap = 1;
for i = 1:numel(zeros outside)
    a = 1/zeros_outside(i);
    a_conj = conj(a);
    H_{min} = H_{min} * (-1/a_{conj}) * H_{inside} * (1 - a*z^-1);
```

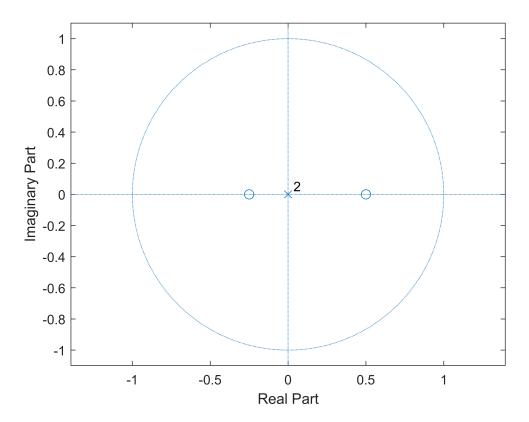
 $H_ap = H_ap * ((z^-1 - a_{conj}) / (1 - a*z^-1));$

% Check that H min is in fact minimum-phase

% Check that all zeros are within the unit circle!
H_min_b = coeffs(expand(H_min * z^10), 'all');

end

zplane(H min b)



% Display H_min
expand(H_min)

ans =
$$\frac{2}{z} + \frac{1}{z^2} - 8$$

We can write $H_{\rm min}(z) = -8 + 2z^{-1} + z^{-2}$

% Display H_ap in order to rewrite it H_ap

$$\begin{array}{l} {\rm H_ap} \ = \\ \\ - \frac{{{{\left({\frac{1}{z} - \frac{1}{2}} \right)} \ \left({\frac{1}{z} + \frac{1}{4}} \right)}}}{{{{\left({\frac{1}{2}z - 1} \right)\ \left({\frac{1}{4}z + 1} \right)}}} \end{array}$$

Expanding the numerator, we get:

$$H_ap_num = expand(-1*(1/z - 1/2)*(1/z + 1/4))$$

H_ap_num =
$$\frac{1}{4z} - \frac{1}{z^2} + \frac{1}{8}$$

$$H_ap_den= expand((1/(2*z) - 1)*(1/(4*z) + 1))$$

H_ap_den =
$$\frac{1}{4z} + \frac{1}{8z^2} - 1$$

We can multiply both the numerator and denominator with 8, to get nice numbers:

ans =
$$\frac{\frac{2}{z} - \frac{8}{z^2} + 1}{\frac{2}{z} + \frac{1}{z^2} - 8}$$

Now, we can write the transfer function for the allpass filter:

$$H_{\rm ap}(z) = \frac{1 + 2z^{-1} - 8z^{-2}}{-8 + 2z^{-1} + z^{-2}}$$

So we have:

$$H(z) = H_{\min}(z)H_{\mathrm{ap}}(z)$$

where
$$H_{\min}(z) = -8 + 2z^{-1} + z^{-2}$$
 and $H_{ap}(z) = \frac{1 + 2z^{-1} - 8z^{-2}}{-8 + 2z^{-1} + z^{-2}}$

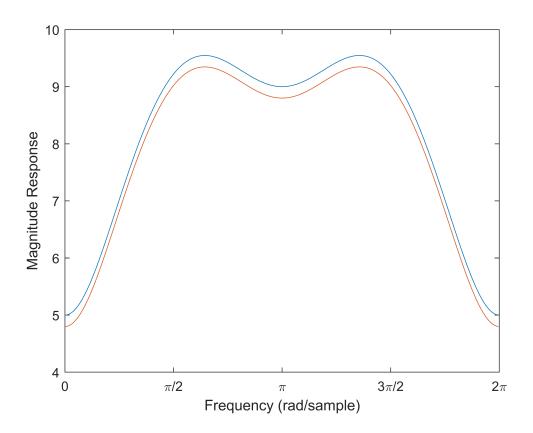
To show that H(z) and $H_{\min}(z)$ have the same magnitude response, we can plot them. If the two graphs are on top of each other then they have the same magnitude response.

$$H(z) = \frac{1 + \frac{3}{2}z^{-1} - 9z^{-2} + 4z^{-3}}{1 - \frac{1}{2}z^{-1}}$$

$$H_{\min}(z) = -8 + 2z^{-1} + z^{-2}$$

```
% Coefficients for H_min(z)
H_min_b = [-8, 2, 1];
H_min_a = 1;
[H_min_z, H_min_w] = freqz(H_min_b, H_min_a, 'whole');
% plot(H_min_w, log10(abs(H_min_z)));

% The offset ensures that the two graphs are not on top of each other offset = -0.2;
plot(w, abs(H), H_min_w, abs(H_min_z) + offset);
set(gca,'XTick',0:pi/2:2*pi)
set(gca,'XTickLabel',{'0','\pi/2','\pi','3\pi/2','2\pi'})
xlabel('Frequency (rad/sample)')
ylabel('Magnitude Response')
xlim([0, 2*pi]);
```



ADSI Problem 1.5 Filter decomposition

1) Show that a FIR filter with a difference equation is not minimum-phase

Consider a FIR filter with the following difference equation

$$y(n) = x(n) + 2x(n-1) + 2x(n-2)$$

1. Show that the FIR filter is not minimum-phase.

To determine whether the filter is a minimum-phase or not, we need get the transfer function of this FIR filter H(z) which is defined as:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2} \cdots + b_M z^{-M}}{1 + a_1 z^{-1} + a_2 z^{-2} \cdots + a_N z^{-N}}$$

This means that we must take the z-transform of difference equation:

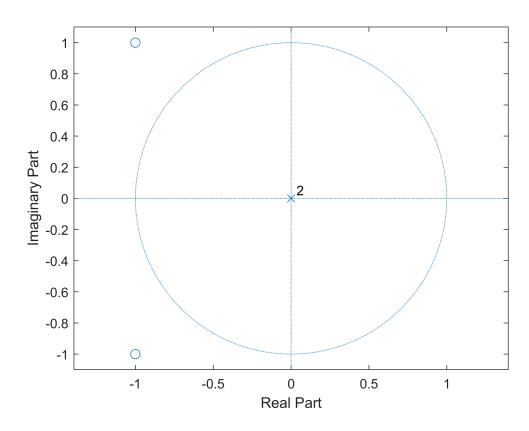
$$Y(z) = X(z) + 2X(z)z^{-1} + 2X(z)z^{-2}$$

$$Y(z) = X(z)(1 + 2z^{-1} + 2z^{-2})$$

$$\frac{Y(z)}{X(z)} = 1 + 2z^{-1} + 2z^{-2}$$

$$H(z) = 1 + 2z^{-1} + 2z^{-2}$$

Once we have the transfer function, we can determine if the filter is a minimum-phase by looking at the zplane plot:



Since all the zeros are outside the unit cirlce, clearly the given FIR filter is not minimum-phase. It is a **maximum-phase**.

2) Find the difference equation for the corresponding minimum-phase FIR filter

We found that the transfer function for the difference equation is:

$$H(z) = 1 + 2z^{-1} + 2z^{-2}$$

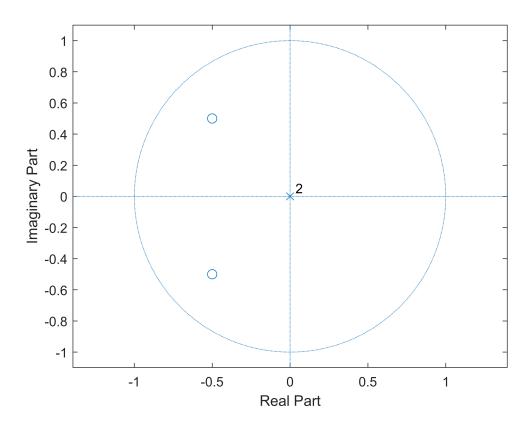
```
% Find the zeros of H
rts = roots([1, 2, 2])
rts = 2 \times 1 complex
  -1.0000 + 1.0000i
  -1.0000 - 1.0000i
syms z;
H_outside = 1; % Represents part of H where zeros are outside the unit circle
H_inside = 1; % Represents part of H where zeros are inside the unit circle
zeros outside = [];
for i = 1:numel(rts)
    root = rts(i);
    if abs(root) > 1
        H_outside = H_outside * (1 - root*z^-1);
        zeros_outside = [zeros_outside; root];
    else
        H_{inside} = H_{inside} * (1 - root*z^{-1});
    end
end
% Sanity check
H = expand(H_inside * H_outside)
```

```
H = \frac{2}{z} + \frac{2}{z^2} + 1
```

```
% Finally compute the H_min and H_ap
H_min = 1;
H_ap = 1;
for i = 1:numel(zeros_outside)
    a = 1/zeros_outside(i);
    a_conj = conj(a);
    H_min = H_min * (-1/a_conj) * H_inside * (1 - a*z^-1);
    H_ap = H_ap * ((z^-1 - a_conj) / (1 - a*z^-1));
end

% Check that H_min is in fact minimum-phase
% Check that all zeros are within the unit circle!
```

H_min_b = coeffs(expand(H_min * z^10), 'all');
zplane(H_min_b)



From the zero-pole plot we can see that we have a minimum-phase filter.

Finally, we can write out the transfer function of the minimum-phase part:

expand(H_min)

ans =
$$\frac{2}{z} + \frac{1}{z^2} + 2$$

$$H_{\min}(z) = 2 + 2z^{-1} + z^{-2}$$

H_ap

$$\begin{aligned} & \text{H_ap =} \\ & \frac{\left(\frac{1}{z} + \frac{1}{2} - \frac{1}{2} \text{ i}\right) \left(\frac{1}{z} + \frac{1}{2} + \frac{1}{2} \text{ i}\right)}{\left(1 + \frac{\frac{1}{2} - \frac{1}{2} \text{ i}}{z}\right) \left(1 + \frac{\frac{1}{2} + \frac{1}{2} \text{ i}}{z}\right)} \end{aligned}$$

Let us simply this expression:

$$H_ap_num = expand((1/z + 1/2 - 1/2i) * (1/z + 1/2 + 1/2i))$$

H_ap_num =
$$\frac{1}{z} + \frac{1}{z^2} + \frac{1}{2}$$

$$H_ap_den = expand((1+ (1/2 -1/2i)*z^-1) * (1 + (1/2+1/2i)*z^-1))$$

H_ap_den =
$$\frac{1}{z} + \frac{1}{2z^2} + 1$$

$$H_{\rm ap}(z) = \frac{\frac{1}{2} + z^{-1} + z^{-2}}{1 + z^{-1} + \frac{1}{2}z^{-2}}$$

Hence:

$$H(z) = H_{\rm min}(z) H_{\rm ap}(z)$$

$$H(z) = (2 + 2z^{-1} + z^{-2}) \left(\frac{\frac{1}{2} + z^{-1} + z^{-2}}{1 + z^{-1} + \frac{1}{2}z^{-2}} \right)$$

We can write $H_{\min}(z) = 2 + 2z^{-1} + z^{-2}$ as difference equation:

$$\begin{split} H_{\min}(z) &= \frac{Y(z)}{X(z)} \\ Y(z) &= X(z)(2 + 2z^{-1} + z^{-2}) \end{split}$$

Now, we take the inverse z-transform:

$$iztrans(2 + 2*z^{-1} + z^{-2})$$

ans =
$$2 \delta_{n-1,0} + \delta_{n-2,0} + 2 \delta_{n,0}$$

We get:

$$y[n] = 2x[n] + 2x[n-1] + x[n-2]$$