# APPENDIX A ERROR BOUND FOR REDUCTION

Given a vector  $\mathbf{x} \in \mathbb{R}^{1 \times n}$ , let  $\mathbf{x}' \in \mathbb{R}^{1 \times n}$  be its perturbed version. Define  $\mathbf{y} = \mathbf{x}\mathbf{A}$  and  $\mathbf{y}' = \mathbf{x}'\mathbf{A}$ . The condition number  $\kappa(\mathbf{A})$  provides an upper bound on the amplification induced by linear transformation:

$$\frac{\|\mathbf{y}' - \mathbf{y}\|}{\|\mathbf{y}\|} \le \kappa(\mathbf{A}) \cdot \frac{\|\mathbf{x}' - \mathbf{x}\|}{\|\mathbf{x}\|}.$$
 (14)

In the reduction, the destination tends to decode the summation of the messages of the transmitters, i.e.,  $\sum_{t \in \mathcal{T}} \mathbf{x}_t$ . Analyzing the bound on the distortion rate of this summation is challenging, as  $\|\sum_{t \in \mathcal{T}} \mathbf{x}_t\|$  may be close to zero. To simplify the analysis, we consider the distortion rate of decoding the message  $\mathbf{x}_t$  by assuming that the symbols transmitted by other sink nodes are zero. This distortion rate provides an upper bound on the distortion rate of  $\|\sum_{t \in \mathcal{T}} \mathbf{x}_t\|$ .

Recall that the transmitted message of node  $t \in \mathcal{T}$  on its outgoing links is given by

$$\mathbf{s}_t = \mathbf{x}_t((\overline{\mathbf{F}}_t^{-1})^T + \mathbf{E}_t^{(1)}),$$

where

$$\mathbf{E}_t^{(1)} = (\mathbf{E}_t^{(rep)})^T + \mathbf{E}_t^{(mul)}.$$

After transmitted by t,  $\mathbf{s}_t$  is recoded by the intermediate nodes and finally decoded at  $\rho$ .

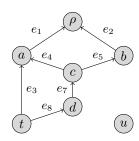


Fig. 5: Three paths from t to  $\rho$ 

Fig. 5 illustrates an example of transmitting the message of t to the sink node  $\rho$  in a butterfly network. There are three paths:  $t \to e_3 \to e_1 \to \rho$ ,  $t \to e_8 \to e_7 \to e_4 \to e_1 \to \rho$ , and  $t \to e_8 \to e_7 \to e_5 \to e_2 \to \rho$ , which are not mutually disjoint. Ideally, we can use vector-matrix multiplication to represent the process of recoding and decoding:

$$\begin{bmatrix} s_{t,1} & s_{t,2} \end{bmatrix} \begin{bmatrix} \beta_{e_3,e_1}^* & 0 \\ \beta_{e_8,e_7}^* \beta_{e_7,e_4}^* \beta_{e_4,e_1}^* & \beta_{e_8,e_7}^* \beta_{e_7,e_5}^* \beta_{e_5,e_2}^* \end{bmatrix} \begin{bmatrix} a_{11} & a_{21} \\ a_{12} & a_{22} \end{bmatrix}$$

$$= \begin{bmatrix} s_{t,1} & s_{t,2} \end{bmatrix} \overline{\mathbf{F}}_t^T = \mathbf{s}_t \overline{\mathbf{F}}_t^T.$$

Note that  $\beta^*_{e_i,e_j}=\beta_{e_j,e_i}$  and  $\begin{bmatrix} a_{11}&a_{21}\\a_{12}&a_{22}\end{bmatrix}$  is the transpose of the generator matrix  ${\bf A}$ .

Ideally, the decoded message is denoted as

$$\mathbf{x}_t^{(0)} = \mathbf{s}_t \overline{\mathbf{F}}_t^T.$$

The inequality (14) and (7) suggest that the error between  $\mathbf{x}_t^{(0)}$  and the original message  $\mathbf{x}_t = [\mathbf{x}_t(\overline{\mathbf{F}}_t^{-1})^T](\overline{\mathbf{F}}_t^T)$  is lower bounded by:

$$\frac{\|\mathbf{x}_{t}^{(0)} - \mathbf{x}_{t}\|}{\|\mathbf{x}_{t}\|} \leq \kappa(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{s}_{t} - \mathbf{x}_{t}(\overline{\mathbf{F}}_{t}^{-1})^{T}\|}{\|\mathbf{x}_{t}(\overline{\mathbf{F}}_{t}^{-1})^{T}\|}$$

$$\leq \kappa^{2}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}}.$$
(15)

Additionally, the following inequality holds:

$$\frac{\|\mathbf{x}_{t}^{(0)}\|}{\|\mathbf{x}_{t}\|} \leq \frac{\|\mathbf{x}_{t}^{(0)} - \mathbf{x}_{t}\| + \|\mathbf{x}_{t}\|}{\|\mathbf{x}_{t}\|} 
\leq 1 + \kappa^{2}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}}.$$
(16)

In practice, recoding and decoding involve floating-point arithmetic, and round-off errors are unavoidable. We model these errors by introducing an error term  $\mathbf{E}_t^{(2)}$ . The decoded message, denoted by  $\mathbf{x}_t^{(1)}$ , is then modeled as:

$$\mathbf{x}_t^{(1)} = \mathbf{s}_t(\overline{\mathbf{F}}_t^T + \mathbf{E}_t^{(2)}).$$

It can be derived that:

$$\frac{\|\mathbf{x}_{t}^{(1)} - \mathbf{x}_{t}^{(0)}\|}{\|\mathbf{x}_{t}^{(0)}\|} \le \kappa(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(2)}\|_{op}}{\|\overline{\mathbf{F}}_{t}\|_{op}}.$$
(17)

By combining inequalities (17) and (16), we derive the upper bound for the distortion rate  $\frac{\|\mathbf{x}_t^{(1)} - \mathbf{x}_t\|}{\|\mathbf{x}_t\|}$  as follows:

$$\frac{\|\mathbf{x}_{t}^{(1)} - \mathbf{x}_{t}\|}{\|\mathbf{x}_{t}\|}$$

$$\leq \frac{\|\mathbf{x}_{t}^{(1)} - \mathbf{x}_{t}^{(0)}\| + \|\mathbf{x}_{t}^{(0)} - \mathbf{x}_{t}\|}{\|\mathbf{x}_{t}\|}$$

$$= \frac{\|\mathbf{x}_{t}^{(1)} - \mathbf{x}_{t}^{(0)}\|}{\|\mathbf{x}_{t}^{(0)}\|} \cdot \frac{\|\mathbf{x}_{t}^{(0)}\|}{\|\mathbf{x}_{t}\|} + \frac{\|\mathbf{x}_{t}^{(0)} - \mathbf{x}_{t}\|}{\|\mathbf{x}_{t}\|}$$

$$\leq \kappa(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(2)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{(2)}\|_{op}} \cdot (1 + \kappa^{2}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}})$$

$$+ \kappa^{2}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}}$$

$$\leq \kappa^{3}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op} \cdot \|\mathbf{E}_{t}^{(2)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}}$$

$$+ \kappa^{2}(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(1)}\|_{op}}{\|\overline{\mathbf{F}}_{t}^{-1}\|_{op}} + \kappa(\overline{\mathbf{F}}_{t}) \cdot \frac{\|\mathbf{E}_{t}^{(2)}\|_{op}}{\|\overline{\mathbf{F}}_{t}\|_{op}}.$$
(18)

### APPENDIX B

### **EXAMPLE OF NETWORK CODES CONSTRUCTION**

Consider multicast in a combination network, where the source node is  $\rho$  and the sink nodes are  $\{d,e,f\}$ , as illustrated in Fig. 6. We use this example to illustrate how to construct a coding scheme using float32 by applying **Algorithm I** and **Algorithm II**, along with the optimization technique.

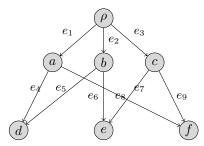


Fig. 6: Combination network

For this multicast, we know that  $\min_{t \in \mathcal{T}} \operatorname{MaxFlow}(\rho, t) = r = 2$ . We first construct a graph  $G^0$ , which consists of the source node  $\rho$  and its adjacent nodes a, b, c, and its outgoing links, as illustrated by Fig. 7.

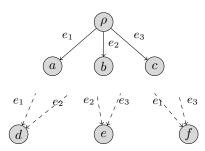


Fig. 7: Graph  $G^0$ 

We construct the generator matrix **A** using the approach outlined in Appendix F, as given below

$$\mathbf{A} = \begin{pmatrix} 1.2987896 & 1.1710904 & -0.12769923 \\ -0.60240215 & 0.8235837 & 1.4259859 \end{pmatrix}$$

The transfer matrices are then given by

$$\begin{split} \widetilde{\mathbf{F}}_d^0 &= \begin{pmatrix} 1.2987896 & 1.1710904 \\ -0.60240215 & 0.8235837 \end{pmatrix}, \\ \widetilde{\mathbf{F}}_e^0 &= \begin{pmatrix} 1.1710904 & -0.12769923 \\ 0.8235837 & 1.4259859 \end{pmatrix}, \\ \widetilde{\mathbf{F}}_f^0 &= \begin{pmatrix} 1.2987896 & -0.12769923 \\ -0.60240215 & 1.4259859 \end{pmatrix}. \end{split}$$

It can be verified that all the transfer matrix are full rank.

In the next step, edge  $e_4$  is added to the graph. The graph  $G^1$  is illustrated in Fig. 8.

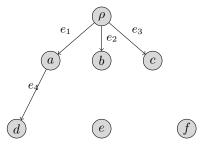


Fig. 8: Graph  $G^1$ 

Since  $In(a) = \{e_1\}$ , we have  $\mathbf{c}_{e_4} = \beta_{e_1,e_4} \mathbf{c}_{e_1}$ . We pick

$$\widetilde{\mathbf{F}}_d^1 = \begin{pmatrix} 1.2987896 & 1.1710904 \\ -0.60240215 & 0.8235837 \end{pmatrix}, \beta_{e_1,e_4} = 1.$$

Next, edge  $e_5$  is added to the graph. The graph  $G^2$  is illustrated in Fig. 9.

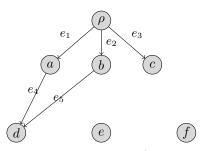


Fig. 9: Graph  $G^2$ 

We pick

$$\widetilde{\mathbf{F}}_d^2 = \begin{pmatrix} 1.2987896 & 1.1710904 \\ -0.60240215 & 0.8235837 \end{pmatrix}, \beta_{e_2, e_5} = 1.$$

The remaining steps are omitted, as they are similar. We finally obtain the following coding scheme.

• Generator Matrix:

$$\mathbf{A} = \begin{pmatrix} 1.2987896 & 1.1710904 & -0.12769923 \\ -0.60240215 & 0.8235837 & 1.4259859 \end{pmatrix}$$

• Transfer Matrix and Condition Number:

$$\begin{aligned} \overline{\mathbf{F}}_d &= \begin{pmatrix} 1.2987896 & 1.1710904 \\ -0.60240215 & 0.8235837 \end{pmatrix}, \ \kappa(\overline{\mathbf{F}}_d) = 1.73. \\ \overline{\mathbf{F}}_e &= \begin{pmatrix} 1.1710904 & -0.12769923 \\ 0.8235837 & 1.4259859 \end{pmatrix}, \ \kappa(\overline{\mathbf{F}}_e) = 1.73. \\ \overline{\mathbf{F}}_f &= \begin{pmatrix} 1.2987896 & -0.12769923 \\ -0.60240215 & 1.4259859 \end{pmatrix}, \ \kappa(\overline{\mathbf{F}}_f) = 1.73. \end{aligned}$$

• Coefficient:

$$\beta_{e_1,e_4} = 1, \ \beta_{e_1,e_8} = 1, \ \beta_{e_2,e_5} = 1,$$
  
 $\beta_{e_2,e_6} = 1, \ \beta_{e_3,e_7} = 1, \ \beta_{e_3,e_9} = 1.$ 

Below is the dual coding scheme for implementing the reduction from nodes  $t \in \mathcal{T}$  to node  $\rho$ .

· Decoding Matrix:

$$\mathbf{A}^T = \begin{pmatrix} 1.2987896 & -0.60240215 \\ 1.1710904 & 0.8235837 \\ -0.12769923 & 1.4259859 \end{pmatrix}$$

• Generator Matrix of each Source:

$$\mathbf{A}_{d} = \begin{pmatrix} 0.463957 & 0.33935675 \\ -0.65972114 & 0.7316592 \end{pmatrix}$$

$$\mathbf{A}_{e} = \begin{pmatrix} 0.8033138 & -0.463957 \\ 0.07193799 & 0.65972114 \end{pmatrix}.$$

$$\mathbf{A}_{f} = \begin{pmatrix} 0.8033138 & 0.33935675 \\ 0.07193798 & 0.7316591 \end{pmatrix}.$$

#### • Coefficient:

$$\beta_{e_4,e_1} = 1, \ \beta_{e_8,e_1} = 1, \ \beta_{e_5,e_2} = 1,$$
  
 $\beta_{e_6,e_2} = 1, \ \beta_{e_7,e_3} = 1, \ \beta_{e_9,e_3} = 1.$ 

## PROOF OF LEMMA 3

Note that  $\phi(\alpha) = \prod_{i=1}^m (\alpha^T \alpha + \gamma_i^2)^{r/2}$  and  $f(\alpha) =$  $\prod_{i=1}^{m} |b_i^T \alpha|$ . Assume that  $r \geq 2$ . If r = 1, then a direct transmission strategy is sufficient, and the network coding strategy provides no additional benefit.

We denote the domain of  $\frac{\phi(\alpha)}{f(\alpha)}$  as  $\mathcal{D} = \{\alpha : f(\alpha) \neq 0\}$ . Fix an arbitrary point  $\alpha' \in \mathcal{D}$ . Considering  $0 < f(\alpha) \leq$  $\prod_{i=1}^m (\|\mathbf{b}_i\| \cdot \|\boldsymbol{\alpha}\|)$  and  $\phi(\boldsymbol{\alpha}) \geq \prod_{i=1}^m \gamma_i^r > 0$ , it follows that  $\frac{\phi(\alpha)}{f(\alpha)} \ge \prod_{i=1}^m \frac{\gamma_i^r}{\|\mathbf{b}_i\| \cdot \|\alpha\|}$ . Then there exists l > 0 such that for any  $\alpha$  satisfying  $\|\alpha\| \le l$ , we have  $\frac{\phi(\alpha)}{f(\alpha)} \ge \prod_{i=1}^m \frac{\gamma_i^r}{\|\mathbf{b}_i\| \cdot l} >$  $\frac{\phi(\alpha')}{f(\alpha')}$ . In addition, there exists  $\epsilon>0$  such that for any  $\alpha$ satisfying  $f(\alpha) \leq \epsilon$ , we have  $\frac{\phi(\alpha)}{f(\alpha)} \geq \frac{\prod_{i=1}^{m} \gamma_i^r}{\epsilon} > \frac{\phi(\alpha')}{f(\alpha')}$ . On the other hand, as  $\alpha \to \infty$ ,  $\frac{\phi(\alpha)}{f(\alpha)}$  becomes unbounded. Since  $\phi(\alpha) > \prod_{i=1}^{m} (\alpha^T \alpha)^{r/2} = \prod_{i=1}^{m} \|\alpha\|^2$ , it follows that  $\frac{\phi(\alpha)}{f(\alpha)} > \frac{1}{n} \frac{1}{n}$  $\prod_{i=1}^{m} \frac{\|\mathbf{\alpha}\|}{\|\mathbf{b}_i\|}$ . As a result, there exists u > 0 such that for any  $\alpha$  satisfying  $\|\alpha\| \geq u$ , we have  $\frac{\phi(\alpha)}{f(\alpha)} > \prod_{i=1}^m \frac{u}{\|\mathbf{b}_i\|} > \frac{\phi(\alpha')}{f(\alpha')}$ . Define  $\mathcal{D}_{\alpha'} = \{\alpha: l \leq \|\alpha\| \leq u, f(\alpha) \geq \epsilon\}$ . It follows that the global minimum of  $\frac{\phi(\alpha)}{f(\alpha)}$  over  $\mathcal{D}$  is either  $\alpha'$  or lies within  $\mathcal{D}$ . If the global minimum is  $\alpha'$  then  $\alpha'$ done. Otherwise, by the Extreme Value Theorem, since  $\frac{\phi(\alpha)}{f(\alpha)}$ 

within  $\mathcal{D}_{\alpha'}$ . If the global minimum is  $\alpha'$ , then the proof is is continuous and  $\mathcal{D}_{\alpha'}$  is closed and bounded, there exists  $\alpha'$  such that  $\alpha^* = \min_{\alpha \in \mathcal{D}_{\alpha'}} \frac{\phi(\alpha)}{f(\alpha)} = \min_{\alpha \in \mathcal{D}} \frac{\phi(\alpha)}{f(\alpha)}$ .

### APPENDIX D PROOF OF THEOREM 1

In this section, we show that the coding scheme created by **Algorithm I and II** ensures that  $\overline{\mathbf{F}}_t$  for  $t \in \mathcal{T}$  has rank r.

We first show that  $\mathbf{F}_t^i$  always maintains rank r during the updating, provided that the desired coding coefficients specified in the algorithm can be found.

Initially, the generator matrix  $\mathbf{A} \in \mathbb{F}^{r \times |\operatorname{Out}(\rho)|}$  can be constructed such that  $\mathbf{F}_t^0$  for  $t \in \mathcal{T}$  has rank r. The condition  $r \leq |\operatorname{Out}(\rho)|$  is sufficient for the existence of such **A**.

Assume that at the (i-1)-th step,  $\widetilde{\mathbf{F}}_t^{i-1}$  has rank r. In the *i*-th step, edge e with tail v is added to the graph. We show that the update in the i-th step of the algorithm ensures that  $\mathbf{F}_t^i$  also has rank r. Let  $\mathcal{T}_e$  denote the set of sink nodes with e on one of its paths from the source node. Let  $U = \langle \mathbf{c}_{e'} : e' \in$  $\operatorname{In}(v)$  denote the space span by vectors  $\{\mathbf{c}_{e'}: e' \in \operatorname{In}(v)\}$ . Suppose e is part of the j-th path of node t. Let  $\tilde{\mathbf{f}}_{t,j}^{i-1}$  be the column of  $\widetilde{\mathbf{F}}_t^{i-1}$  corresponding to the j-th path, and  $\widetilde{\mathbf{F}}_{t,-j}^{i-1}$  be the matrix obtained from  $\widetilde{\mathbf{F}}_t^{i-1}$  by removing  $\widetilde{\mathbf{f}}_{t,j}^{i-1}$ . Let  $V_t =$  $\langle \text{columns of } \widetilde{\mathbf{F}}_{t,-j}^{i-1} \rangle$ . We just need to show that there exists  $\mathbf{c}_e \in U \setminus \cup_{t \in \mathcal{T}_e} (U \cap V_t)$ .

Let's view this problem from the perspective of linear subspaces. For any  $t \in \mathcal{T}_e$ , since  $\widetilde{\mathbf{f}}_{t,j}^{i-1} \notin V_t$  and  $\widetilde{\mathbf{f}}_{t,j}^{i-1} \in U$ , we have  $\dim(U \cap V_t) \leq \dim(U) - 1 < \dim(U)$ . It follows that  $\mathbf{c}_e \in U \setminus \bigcup_{t \in \mathcal{T}_e} (U \cap V_t)$  exists. In fact, since  $|\mathcal{T}_e|$  is finite, nearly all of the points in linear space U lie in the set  $U \setminus \bigcup_{t \in \mathcal{T}_e} (U \cap V_t)$ . This implies that, when using floatingpoint numbers, randomly choosing  $\beta_{e',e}$  for each  $e' \in \text{In}(v)$ is sufficient to obtain a  $\mathbf{c}_e \in U \setminus \cup_{t \in \mathcal{T}_e} (U \cap V_t)$ . We construct  $\widetilde{\mathbf{F}}_t^i$  from  $\widetilde{\mathbf{F}}_t^{i-1}$  by replacing  $\widetilde{\mathbf{f}}_{t,j}^{i-1}$  with  $\mathbf{c}_e$ . This ensures that  $\widetilde{\mathbf{F}}_t^i$ is full rank. Similar randomized approach can be utilized to construct the matrix A.

In each step i of the algorithm,  $\tilde{\mathbf{f}}_{t,i}^{i}$  always takes the value of  $\mathbf{c}_e$  where e is the most downstream edge in the truncation in  $G^i$  along the j-th edge-disjoint path from the source node  $\rho$  to the sink node t. Therefore, when the algorithm stops, we have  $\bar{\mathbf{F}}_t = \mathbf{F}_t$ , which has full rank.

### APPENDIX E PROOF OF THEOREM 2

Suppose that

$$\alpha_0 = \arg \max_{\alpha: \alpha^T \alpha = 1} f(\alpha)$$

$$\alpha_1 = \arg\min_{\boldsymbol{\alpha}: f(\boldsymbol{\alpha}) \neq 0} \frac{\phi(\boldsymbol{\alpha})}{f(\boldsymbol{\alpha})}.$$

We have

$$\frac{\phi(\|\boldsymbol{\alpha}_1\|\boldsymbol{\alpha}_0)}{f(\|\boldsymbol{\alpha}_1\|\boldsymbol{\alpha}_0)} = \frac{\phi(\boldsymbol{\alpha}_1)}{f(\|\boldsymbol{\alpha}_1\|\boldsymbol{\alpha}_0)} \leq \frac{\phi(\boldsymbol{\alpha}_1)}{f(\boldsymbol{\alpha}_1)}$$

where the first equality holds as  $\alpha_0^T \alpha_0 = 1$ , and the second inequality holds as

$$f(\|\boldsymbol{\alpha}_1\|\boldsymbol{\alpha}_0) = \|\boldsymbol{\alpha}_1\|^m \cdot f(\boldsymbol{\alpha}_0) \ge \|\boldsymbol{\alpha}_1\|^m \cdot f(\frac{\boldsymbol{\alpha}_1}{\|\boldsymbol{\alpha}_1\|}) = f(\boldsymbol{\alpha}_1).$$

On the other hand, since  $\frac{\phi(\|\alpha_1\|\alpha_0)}{f(\|\alpha_1\|\alpha_0)} \geq \frac{\phi(\alpha_1)}{f(\alpha_1)}$ , it follows that  $\frac{\phi(\|\alpha_1\|\alpha_0)}{f(\|\alpha_1\|\alpha_0)} = \frac{\phi(\alpha_1)}{f(\alpha_1)}$ . Therefore, we also have  $f(\alpha_0) = \frac{\phi(\alpha_1)}{f(\alpha_1)}$ .  $f(\frac{\alpha_1}{\|\alpha_1\|})$ . As a result  $\|\alpha_1\|\alpha_0$  minimizes  $\frac{\phi(\alpha)}{f(\alpha)}$  and  $\frac{\alpha_1}{\|\alpha_1\|}$ maximizes  $f(\alpha)$ .

### APPENDIX F CONSTRUCTION OF GENERATOR MATRIX

To optimize the generator matrix A, we propose a heuristic algorithm to minimize  $\prod \kappa(\mathbf{F}_t^0)$ .

First, initialize  $\mathbf{A} \in \mathbb{F}^{r \times |\operatorname{Out}(\rho)|}$  by sampling each entry uniformly from the interval [0, 2]. With high probability  $rank(\mathbf{A}) = r$ . Note that each column in **A** corresponds to a coefficient vector  $\mathbf{c}_e$  for each  $e \in \mathrm{Out}(\rho)$ . The transfer matrix  $\widetilde{\mathbf{F}}_t^0$  for  $t \in \mathcal{T}$  is a submatrix of **A**, consisting of the vectors  $\mathbf{c}_e \in {\mathbf{c}_e : e \in \mathrm{Out}(\rho)}$  if edge e is part of its r paths. With high probability,  $rank(\tilde{\mathbf{F}}_t^0) = r$ . If this condition is not met, we will resample A until it is satisfied. We then replace the columns of **A** that appear in only one transfer matrix  $\mathbf{F}_t^0$  with standard basis vectors from  $\{\hat{\mathbf{e}}_i : i = 1, \dots r\}$ , ensuring that the full-rank property of  $\widetilde{\mathbf{F}}_{t}^{0}$  is preserved.

Next, we optimize the columns of A that are shared by multiple  $\mathbf{F}_t^0$ . Suppose column  $\mathbf{c}_e$  is of these. Let  $\mathcal{T}_e \subset \mathcal{T}$ denote the set of sink nodes where  $c_e$  appears in their transfer matrices. We optimize  $c_e$  by minimizing the upper bound of  $\prod_{t\in\mathcal{T}_e}\kappa(\widetilde{\mathbf{F}}_t^0)$ , as given in (10). Using our optimization framework, this involves determining  $\alpha\in\mathbb{F}^r$  for (11). Subsequently,  $\mathbf{c}_e$  is updated to  $\sum_{i=1}^r\alpha_i\hat{\mathbf{e}}_i$ . This process is iterated for all shared columns, allowing for multiple cycles to improve results, until the decrease in  $\prod_{t\in\mathcal{T}_e}\kappa(\widetilde{\mathbf{F}}_t^i)$  falls below a specified threshold.