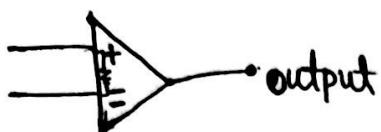


Operational Amplifiers & Application

জেনেরেল Transistor দ্বাৰা cascade

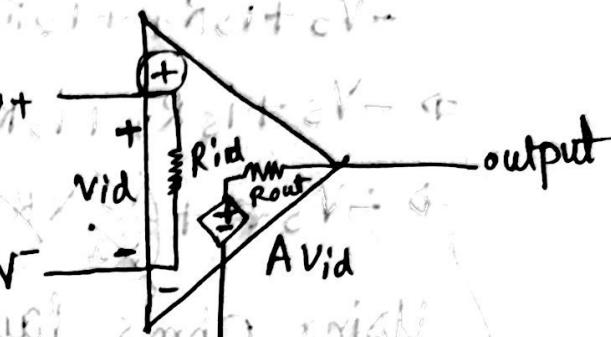


Amplifier

V^+ → Non inverting input

V^- → Inverting input

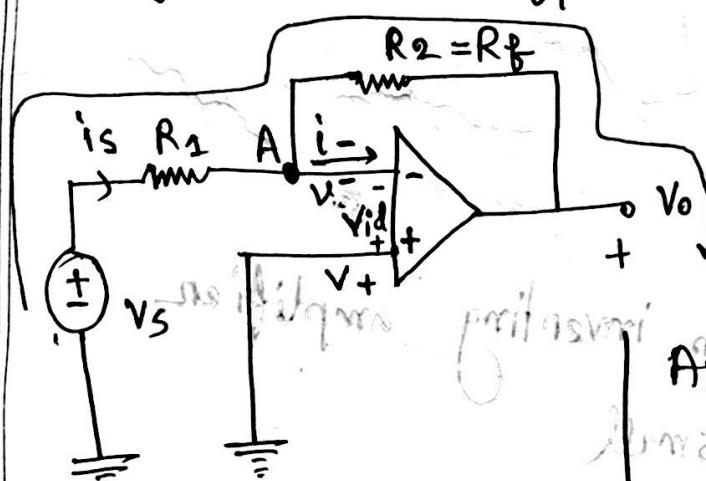
A → open circuit voltage gain



$$R_{out} = 0 \\ R_{id} = \text{high}$$

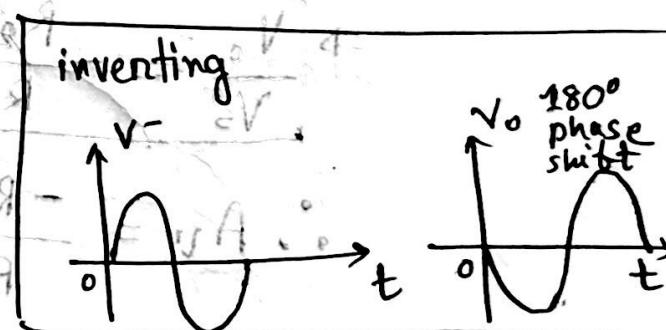
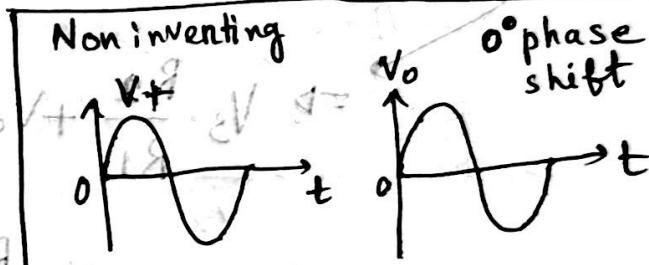
Inverting Amplifier

$$\text{Voltage gain} \approx A_V = \frac{V_o}{V_i}$$



$$\therefore V_{id} = V^+ - V^-$$

$$\therefore R_{id} = \text{infinity}, i_- = 0$$



Apply KCL at node A,

$$i_s = i_2 + i_-$$

$$\therefore i_s = i_2 + 0$$

$$\therefore i_s = i_2$$

$$A_v = \frac{\text{Output}}{\text{Input}}$$

negative & positive input

Applying KVL,

$$-V_s + i_s R_1 + i_2 R_2 + V_o = 0$$

$$\Rightarrow -V_s + i_s R_1 + i_s R_2 + V_o = 0$$

$$\Rightarrow -V_s + \left(\frac{V_s}{R_1} \times R_1 \right) + \left(\frac{V_s}{R_1} \times R_2 \right) + V_o = 0$$

Using Ohm's law,

$$i_s = \frac{V_s}{R_1}$$

$$\Rightarrow V_s \frac{R_2}{R_1} + V_o = 0$$

$$\Rightarrow V_o = -V_s \frac{R_2}{R_1}$$

$$\Rightarrow \boxed{\frac{V_o}{V_s} = -\frac{R_2}{R_1}}$$

$$\therefore A_v = -\frac{R_2}{R_1}$$

Voltage gain for inverting amplifier

$R_1 > R_2$ = gain small

$R_2 > R_1$ = gain বড় হবে
output resistance ক্ষুণ্ণু এবং voltage divider.

20 dB

$$\Rightarrow 10 \times 2$$

$$\Rightarrow 10 \log(100)$$

$\infty = \text{open circuit}$

To find output resistance, Thvenin's theorem

$$A_v = 20 \text{ dB} = (-10^{\frac{20}{10}}) = -100 \quad [\text{Inverting amplifier}]$$

$$R_1 = 20 \text{ k}\Omega$$

$$\therefore A_v = -\frac{R_2}{R_1}$$

$\frac{V_o}{V_s} = -\frac{R_2}{R_1}$

$$\Rightarrow -100 = -\frac{R_2}{20 \text{ k}}$$

$$\Rightarrow R_2 = 2 \text{ M}\Omega$$

(Ans)

Determine R_2 for

inverting amplifier circuit, where $A_v = 20 \text{ dB}$ and $R_1 = 20 \text{ k}\Omega$.

B Non-Inverting :-

Using VDR,

$$V_1 = \frac{V_o R_1}{R_1 + R_2}$$

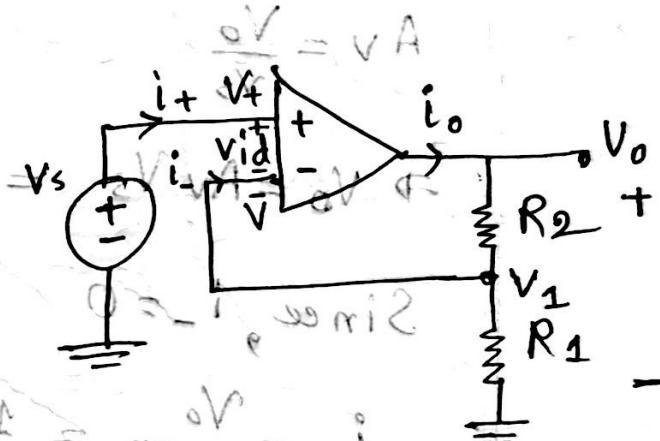
$$V_{id} = V_+ - V_- = V_s - V_1$$

As $i_- = 0$, $V_{id} = 0$

$$\therefore V_{id} = V_s - V_1$$

$$\Rightarrow 0 = V_s - V_1$$

$$\Rightarrow V_s = V_1$$



$$V_s = \frac{V_o R_1}{R_1 + R_2}$$

$$\Rightarrow \frac{V_s}{V_o} = \frac{R_1}{R_1 + R_2}$$

$$\Rightarrow \frac{V_o}{V_s} = \frac{R_1 + R_2}{R_1}$$

$$\frac{V_o}{V_s} = 1 + \frac{R_2}{R_1}$$

cannot be satisfied

$$\therefore A_V = 1 + \frac{R_2}{R_1}$$

Non-inventing Example

$$R_1 \approx 3k\Omega$$

$$R_2 = 43 \text{ k}\Omega$$

$$V_s = 0.1V$$

$$A_V = \frac{V_o}{V_i}$$

$$\Rightarrow V_0 = A V_s$$

Since, $i_- = 0$

$$i_o = \frac{V_o}{R_2 + R_1} = \frac{1.53}{43k\Omega + 3k\Omega} = 33.3 \mu A$$

$$\frac{19}{2} \cdot V = 2V$$

18 24

$$\frac{9+19}{19} = \frac{0.8}{2.8}$$

$$EV - cV = bIV$$

$$-EV - eV = 0$$

42

~~Final~~ $\beta = \text{feedback factor}$

$$V_o = AV_{id} = A(V_+ - V_-)$$

$$= A(V_s - V_i)$$

$$= A(V_s - \beta V_o)$$

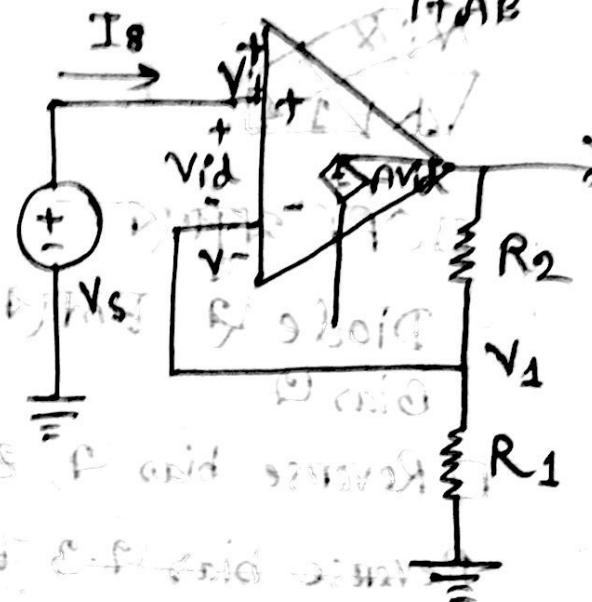
Voltage gain, $A_v = \frac{V_o}{V_s}$

$$V_s = \frac{R_1 V_o}{R_1 + R_2}$$

$$\beta = \frac{R}{R_1 + R_2} \times V_o$$

$$= \beta V_o$$

$$\begin{aligned} V_o + A\beta V_o &= AV_s \\ \Rightarrow V_o(1 + AB) &= AV_s \\ \Rightarrow \frac{V_o}{V_s} &= \frac{A}{1 + AB} \\ \Rightarrow A_v &= \frac{A}{1 + AB} \end{aligned}$$



$$A_v = 1 + \frac{R_2}{R_1} = \frac{1}{\beta}$$

Ideal gain

$$A_v = \frac{A}{1 + AB}$$

Actual gain

Gain error

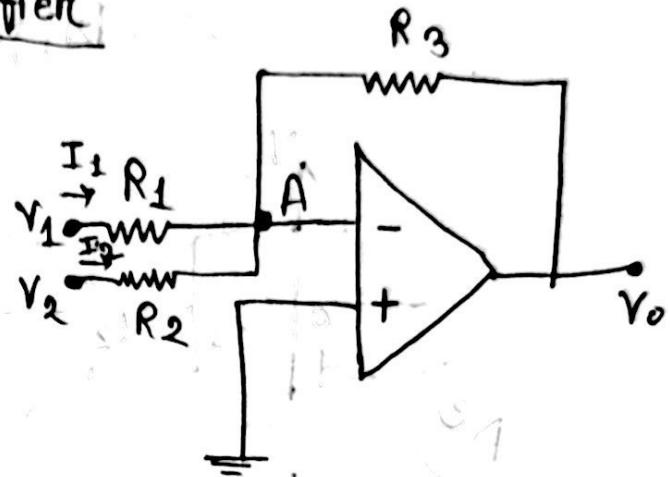
Ideal gain - Actual gain = gain error.

$$GE = \frac{1}{\beta(1 + AB)}$$

Summing Amplifier

$$I_1 = \frac{V_1}{R_1}$$

$$I_2 = \frac{V_2}{R_2}$$



Applying KCL at node A,

$$I_1 + I_2 = I_3$$

$$\Rightarrow \frac{V_1}{R_1} + \frac{V_2}{R_2} = -\frac{V_0}{R_3}$$

(1)

$$\text{if, } V_1 = V_2 = V_3$$

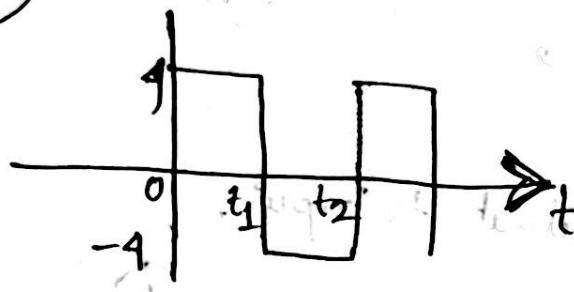
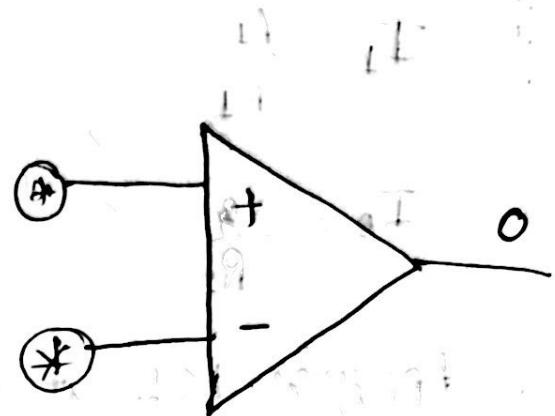
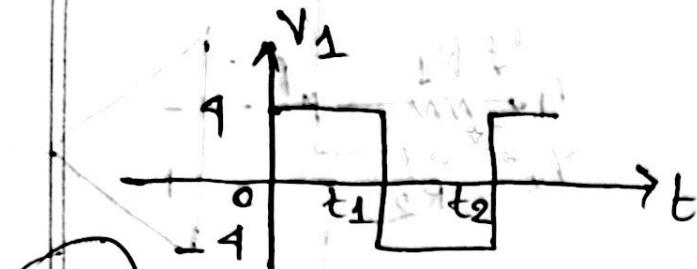
$$\text{From eqn (1), } \frac{V_1}{R_1} + \frac{V_2}{R_2} = -\frac{V_0}{R_3}$$

$$\Rightarrow V_s \left(\frac{1}{R_1} + \frac{1}{R_2} \right) = -\frac{V_0}{R_3}$$

$$\Rightarrow R_3 \left(\frac{1}{R_1} + \frac{1}{R_2} \right) = -\frac{V_0}{V_s}$$

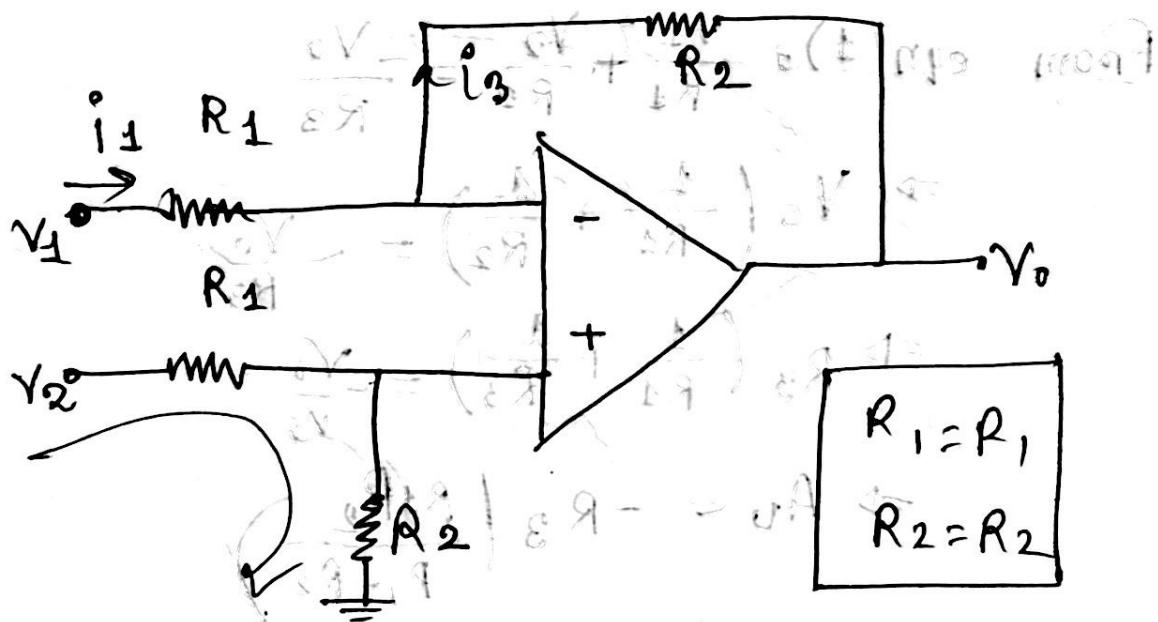
$$\Rightarrow A_v = -R_3 \left(\frac{R_1 R_2}{R_1 + R_2} \right)$$

Difference Amplifier



এখন দুটি output হবে,

$$V_+ = V_1 = 1 \text{ V}$$



$$V_0 = V_+ - i_2 R_2$$

$$= V_+ - i_1 R_2$$

$$i_1 = \frac{V_1 - V_-}{R_1}$$

$$= V_+ - \frac{V_+ - V_-}{R_1} R_2$$

$$\boxed{\begin{aligned} R_1 &= R_1 \\ R_2 &= R_2 \end{aligned}}$$

$$= V_+ - \frac{R_2}{R_1} (V_+ - V_-)$$

$$= V_- - \frac{R_2}{R_1} V_1 + \frac{R_2}{R_1} V_- \text{ (assuming } iB = 0)$$

$$= \left(1 + \frac{R_2}{R_1} \right) V_- - \frac{R_2}{R_1} V_1$$

Applying VDR,

$$V_+ = \frac{V_2 R_2}{R_1 + R_2}$$

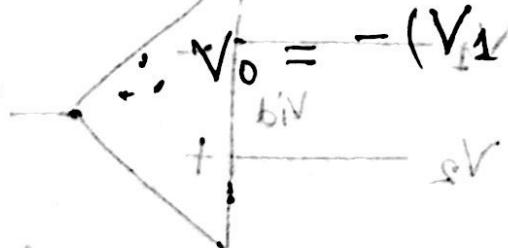
using option V short circuiting V_+ \rightarrow mV

$$V_- = V_+ = \frac{V_2 R_2}{R_1 + R_2}$$

$$\Rightarrow \left(1 + \frac{R_2}{R_1} \right) V_- - \frac{R_2}{R_1} V_1 = \frac{V_2 R_2}{R_1 + R_2}$$

$$\therefore V_o = \frac{-R_2}{R_1} (V_1 - V_2)$$

if $R_1 = R_2$ \rightarrow unit gain amplifier



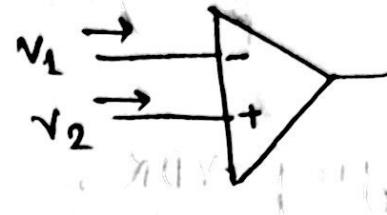
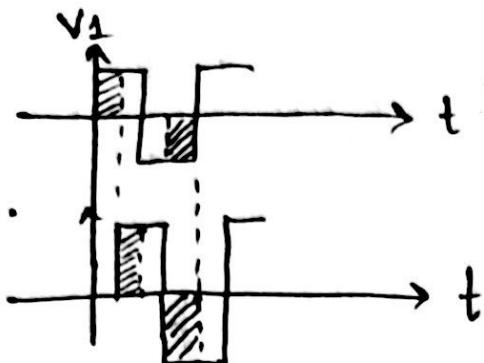
$$\frac{V_o + V}{2} = 0 \text{ V}$$

This is due to A = 0
using option V

\rightarrow spillover factor

$$\text{For moment } A = 0 \text{ V} \quad \left(\frac{V_o + V}{2} \right)_{\text{mV}} + (\&V - \&V) \text{ mV} = 0 \text{ V}$$

~~C22~~ C22/28
 CMRR → Common Mode Rejection Ratio



$A_{dm} \rightarrow$ Differential Mode Voltage gain

For difference Amplifier, $+V = -V$

$$A_{dm} = -\frac{R_2}{R_1}$$

$$\therefore V_o = -\frac{R_2}{R_1} (V_1 - V_2)$$

$$= A_{dm} (V_1 - V_2)$$

$$V_{id} = V^+ - V^- \propto V_1 - V_2 \approx V_2 - V_1$$

V_{ic} = Common mode Voltage

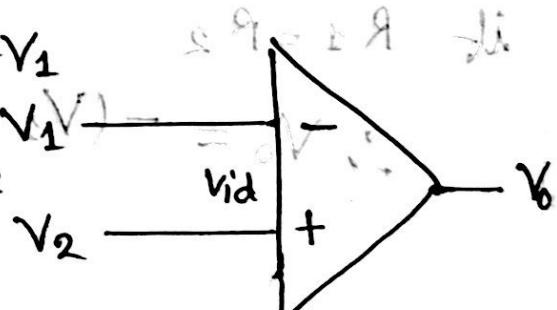
$$V_{ic} = \frac{V_1 + V_2}{2}$$

Output Voltage :-

$$V_o = A_{dm} (V_1 - V_2) + A_{cm} \frac{V_1 + V_2}{2}$$

$A \Rightarrow$ open circuit
Voltage gain

$A_{cm} \Rightarrow$ Common mode
Voltage gain.



$$\therefore V_o = A_{dm}(V_{id}) + A_{cm}(V_{ic})$$

$$\therefore V_1 = V_{ic} + \frac{V_{id}}{2}$$

$$V_2 = V_{ic} - \frac{V_{id}}{2}$$

$$\therefore V_o = A_{dm}(V_{id}) + A_{cm}(V_{ic})$$

$$\Rightarrow V_o = A_{dm}\left(V_{id} + \frac{A_{cm}(V_{ic})}{A_{dm}}\right)$$

$$\therefore V_o = A_{dm}\left(V_{id} + \frac{V_{ic}}{\frac{A_{cm}}{A_{dm}}}\right)$$

$$\therefore V_o = A_{dm}\left(V_{id} + \frac{V_{ic}}{CRMM}\right) = 20 \log_{10}(CMRR)$$

Finite CRMM example

$$= 2500 \left(\left(5.0001 - 4.999 \right) + \left(\frac{\frac{5.0001 + 4.999}{2}}{10^4} \right) \right) \rightarrow 80 \text{ dB}$$

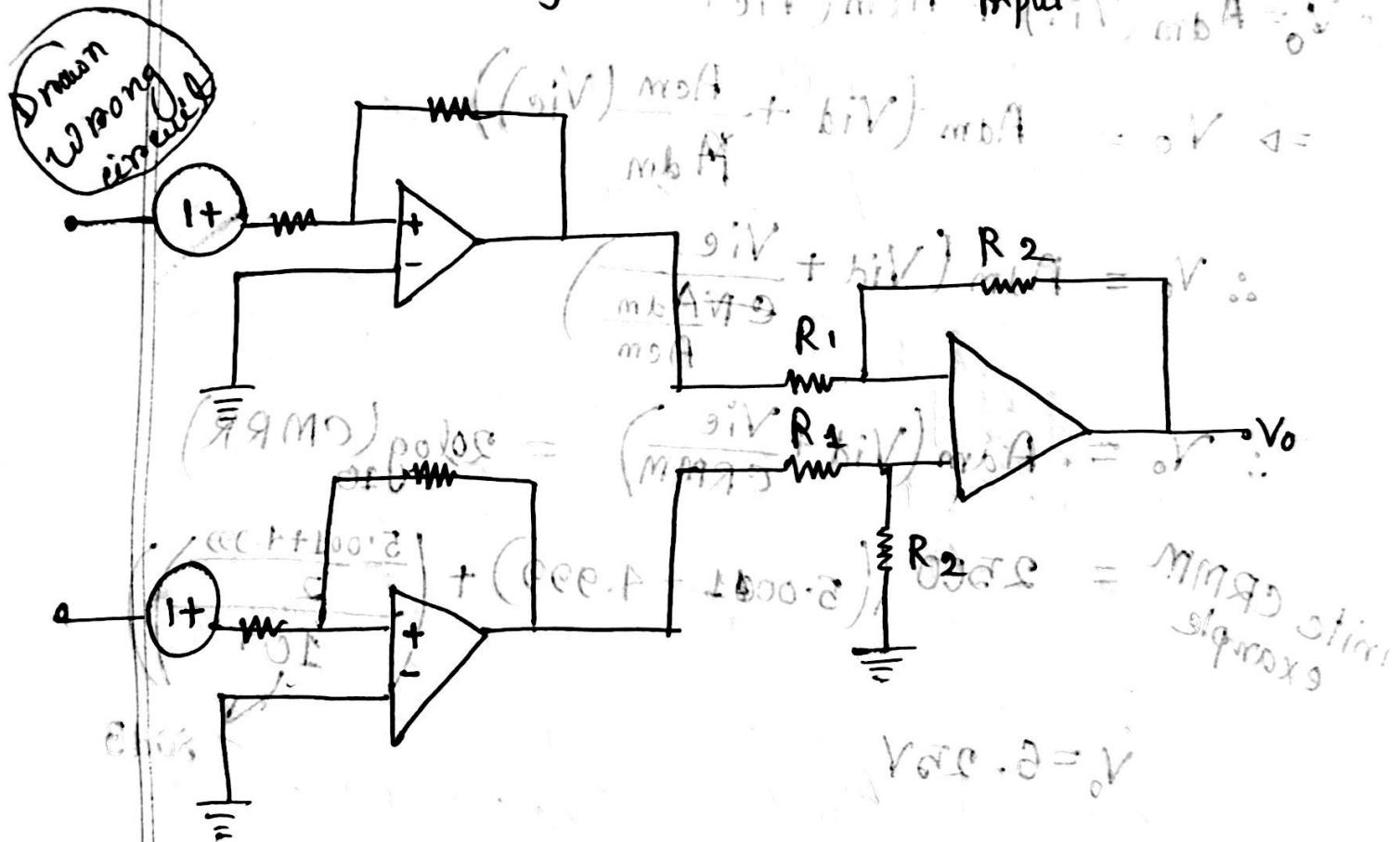
$$V_o = 6.25 \text{ V}$$

$$\begin{aligned} \text{Gain error} &= \frac{6.25 - 5}{5} \times 100\% \\ &= 25\% \end{aligned}$$

Instrumentation amplifier

2 non inverting amplifiers + difference
amplifier Σ cascade.

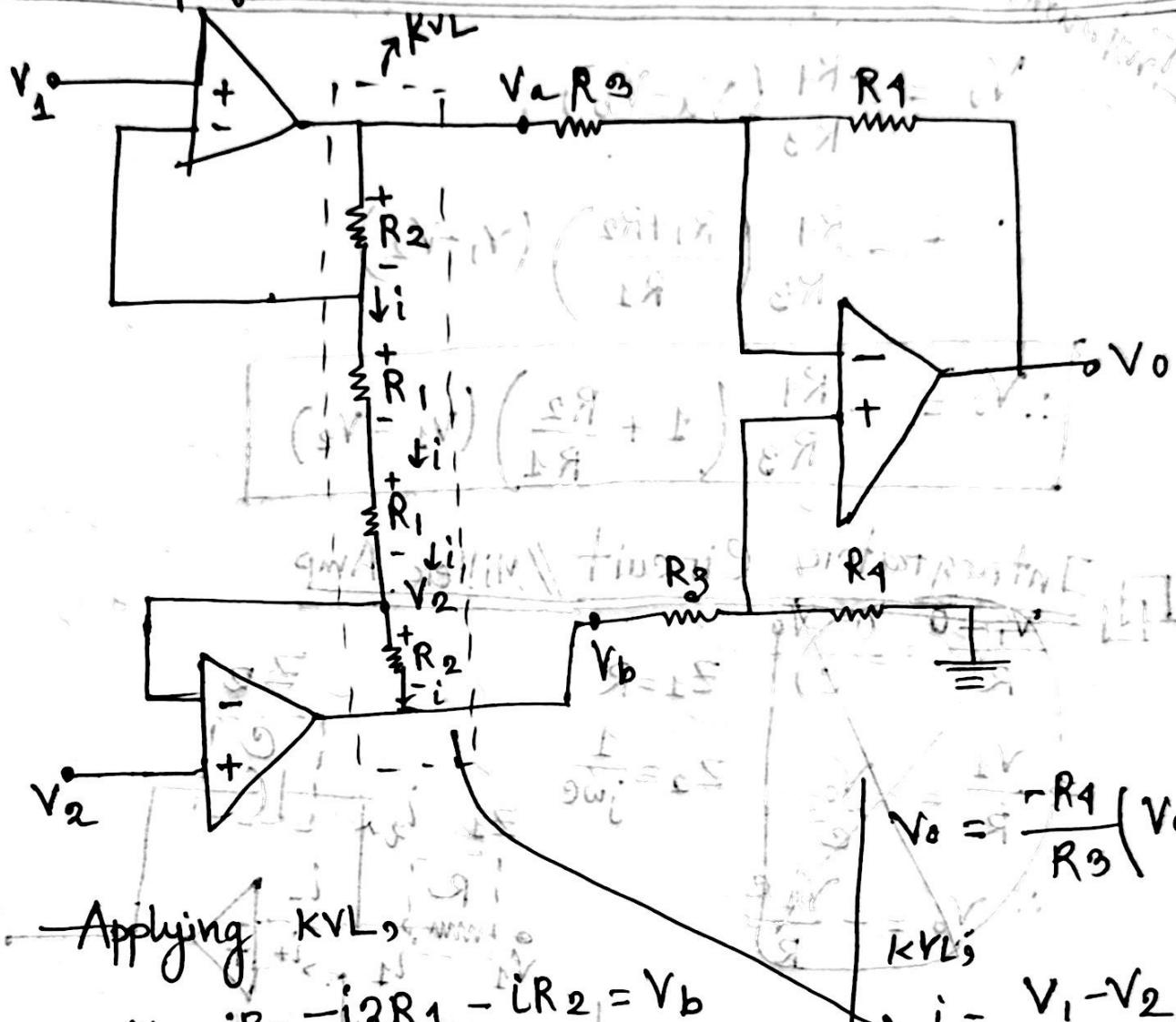
non inverting en output = difference Amplifier



$$\text{Gross error} = \frac{\sum_{i=1}^n |y_i - \hat{y}_i|}{n} \times 100\%$$

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Non inverting
Amplifier



Applying KVL,

$$Va - iR_2 - i2R_1 - iR_2 = V_b$$

$$\Rightarrow Va - i2R_2 - i2R_1 = V_b$$

$$\Rightarrow [Va - V_b] = i2(R_1 + R_2)$$

$$= \frac{V_1 - V_2}{2R_1} \times 2 \times (R_1 + R_2)$$

$$\therefore V_a - V_b = \frac{R_1 + R_2}{R_1} (V_1 - V_2)$$

$$f_b(f) \left(\frac{1}{2} \right) \frac{1}{2} - f(f) \frac{1}{2} = 5V$$

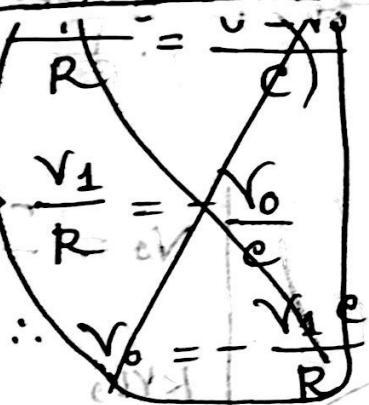
Instrumentation Amplifier

$$V_o = -\frac{R_1}{R_3} (V_a - V_b)$$

$$= -\frac{R_1}{R_3} \left(\frac{R_1 + R_2}{R_1} \right) (V_1 - V_2)$$

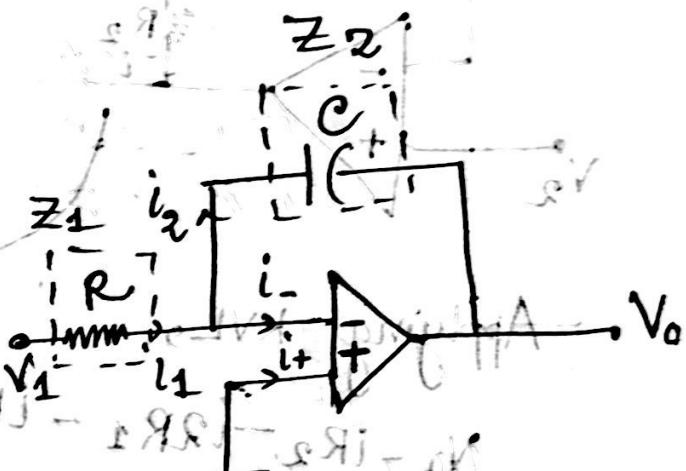
$$\therefore V_o = -\frac{R_1}{R_3} \left(1 + \frac{R_2}{R_1} \right) (V_1 - V_2)$$

Integrator Circuit // Miller Amp



$$Z_1 = R$$

$$Z_2 = \frac{1}{j\omega C}$$



$$A_V = -\frac{Z_2}{Z_1} = -\frac{1}{j\omega C R}$$

$$\therefore A_V = \frac{-1}{j\omega RC} = -\frac{1}{SRC} \quad [S = j\omega]$$

$$V_o = V_c - \frac{1}{C} \int_0^t i_2(t) dt$$

det, $V_c = 0V$

$$V_o(t) = -\frac{1}{C} \int_0^t i_2(t) dt$$

$$V_o = -\frac{1}{C} \int_0^t i_1(t) dt$$

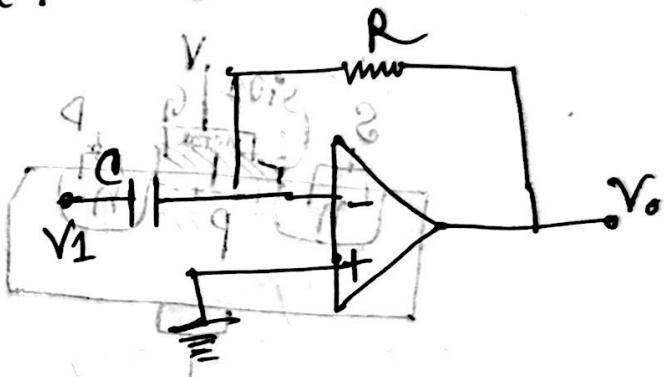
$$= -\frac{1}{RC} \int_0^t v_1(t) dt \quad [i = \frac{v_1}{R}]$$

With V_c ,

$$V_o(t) = -\frac{1}{RC} \int_0^t v_1(t) dt - V_c$$

Differentiate Amp

$$A_v = -\frac{R}{\frac{1}{j\omega C}}$$



$$A_v = -SRC = -j\omega RC$$

$$V_o(t) = -CR \frac{dV_1(t)}{dt}$$

- Non-inverting gain error
- Ideal gain / Actual gain

Quiz 2 → 29/01/2024
Sunday

Summing

buffer
difference

Instrumentation
Integrator
Differential