

# 31352-Power Electronics I

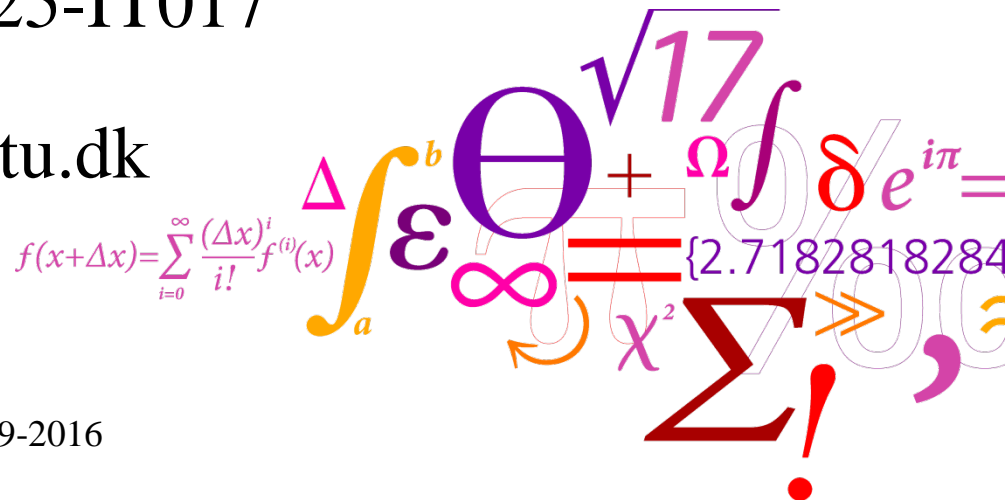
## — Lecture III: Isolated DC-DC Converters

Lecturer : Ziwei Ouyang

07-Sept.-2016, kl: 9:00-12:00

Location: B325-IT017

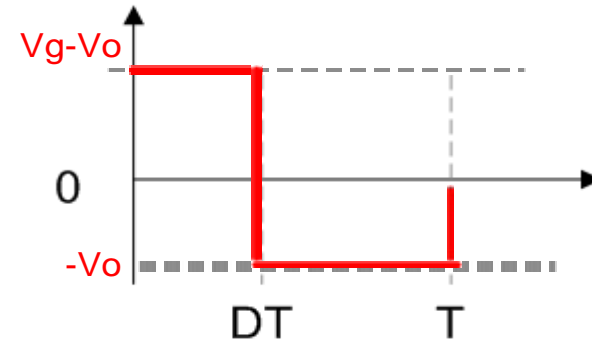
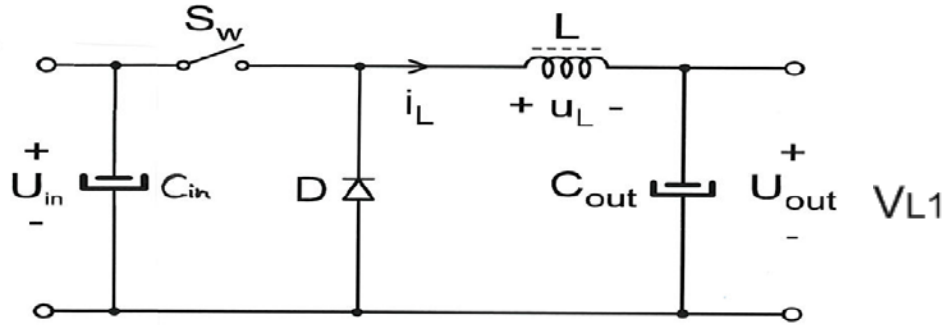
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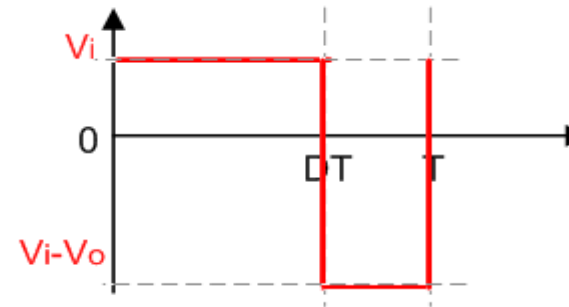
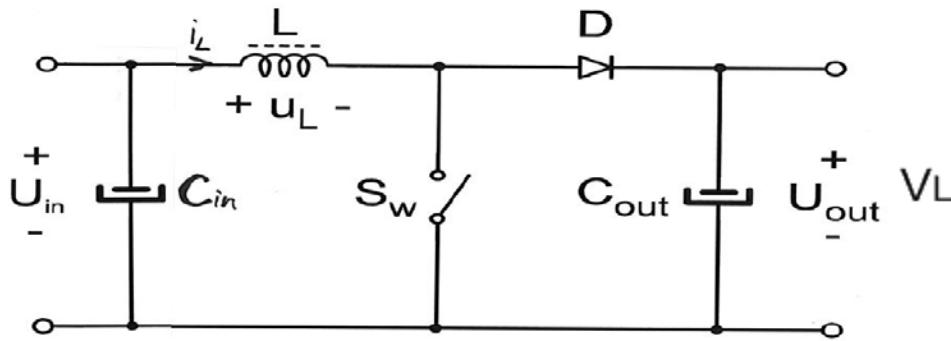
# Agenda

- Review nonisolated DC-DC converters
- Forward converter
- Flyback converter
- Push-pull converter
- Half-bridge & Full bridge

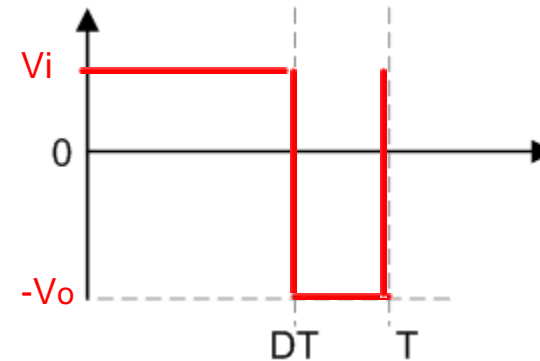
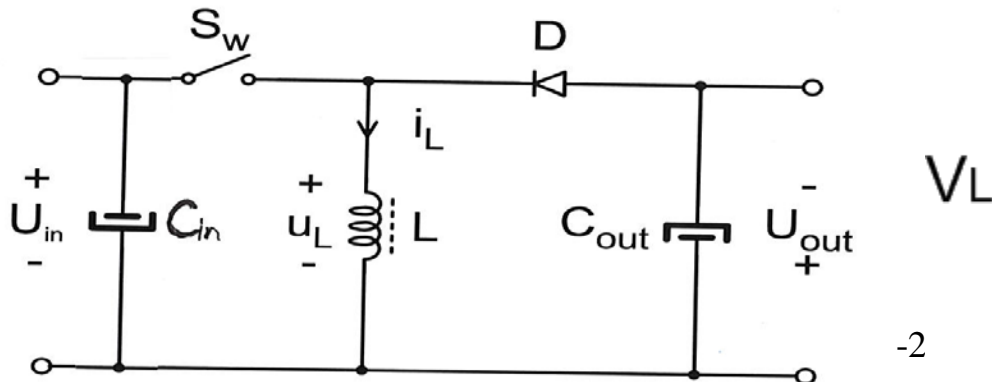
# Review nonisolated DC-DC Converters



$$\frac{V_o}{V_{in}} = D$$



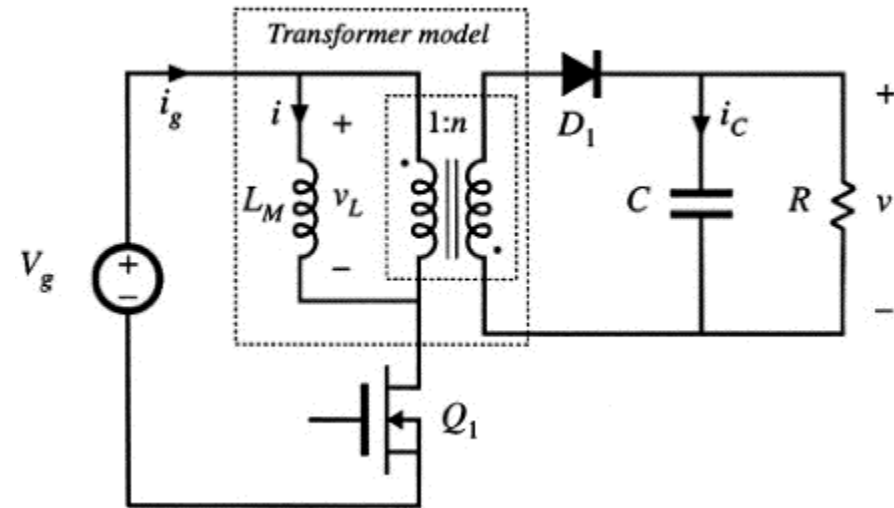
$$\frac{V_o}{V_{in}} = \frac{1}{1 - D}$$



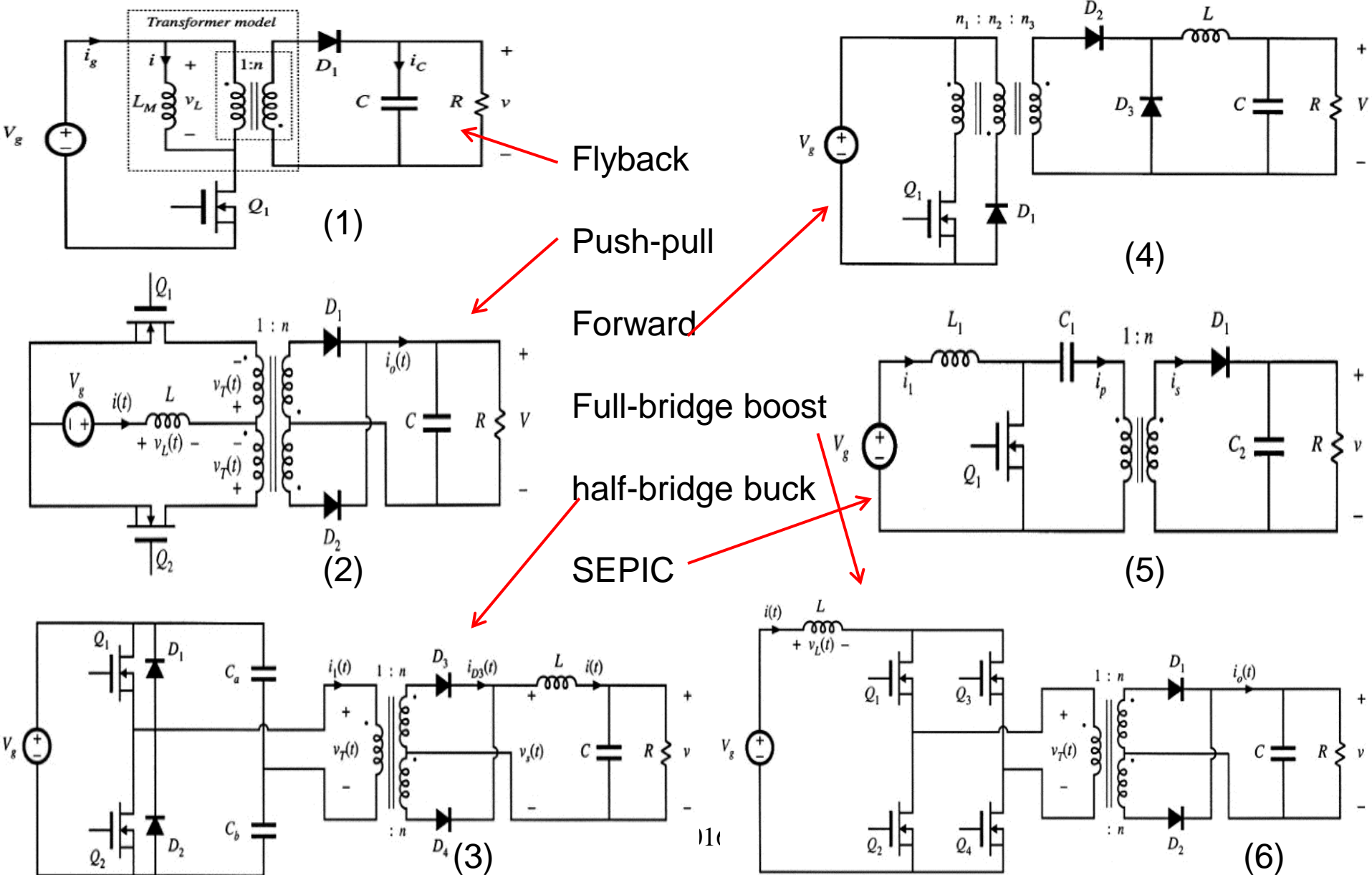
$$\frac{V_o}{V_{in}} = \frac{D}{1 - D}$$

# Galvanic Isolation

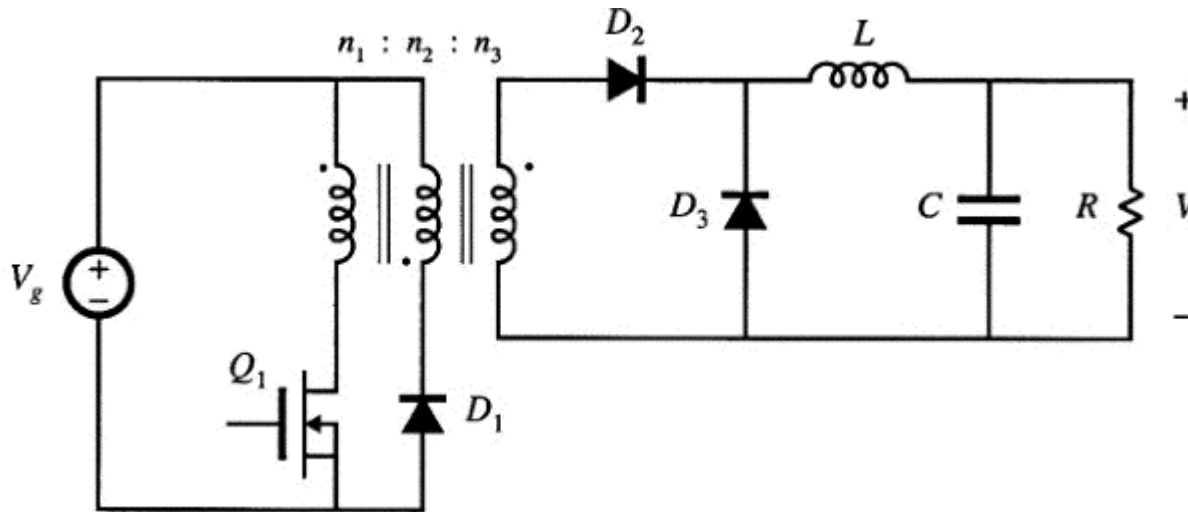
- ✓ for safety, preventing accidental current from reaching ground through a person's body
- ✓ to break ground loops, to suppress electrical noise in sensitive devices
- ✓ to transfer power between two circuits
- ✓ to change the transfer ratio



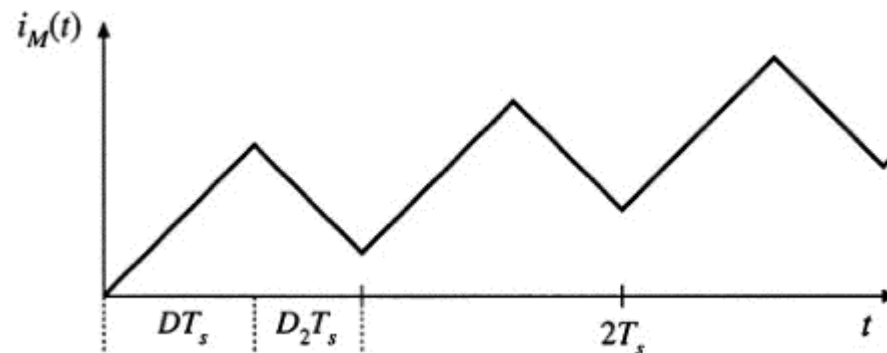
# Topologies



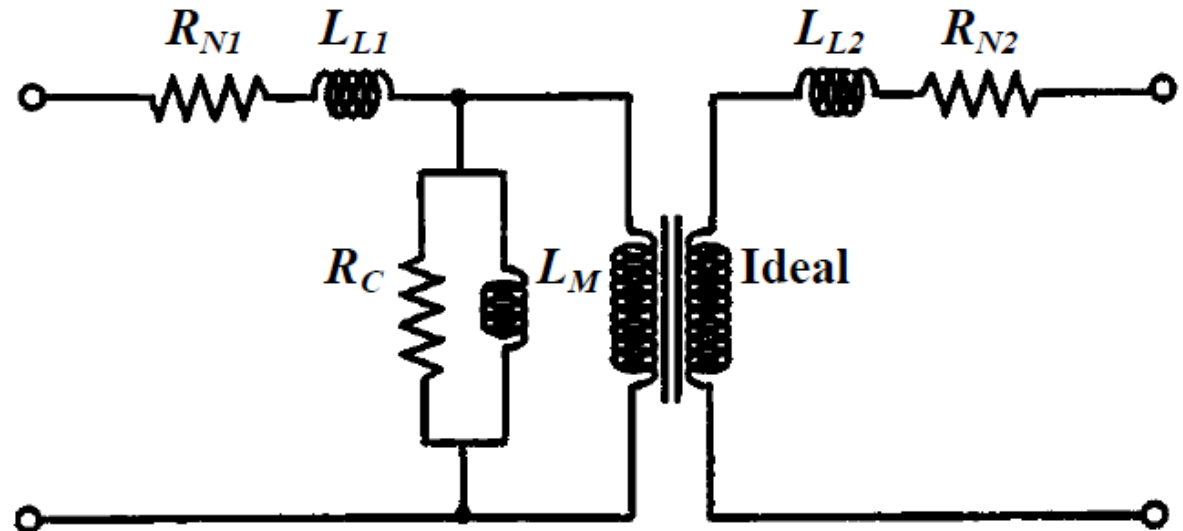
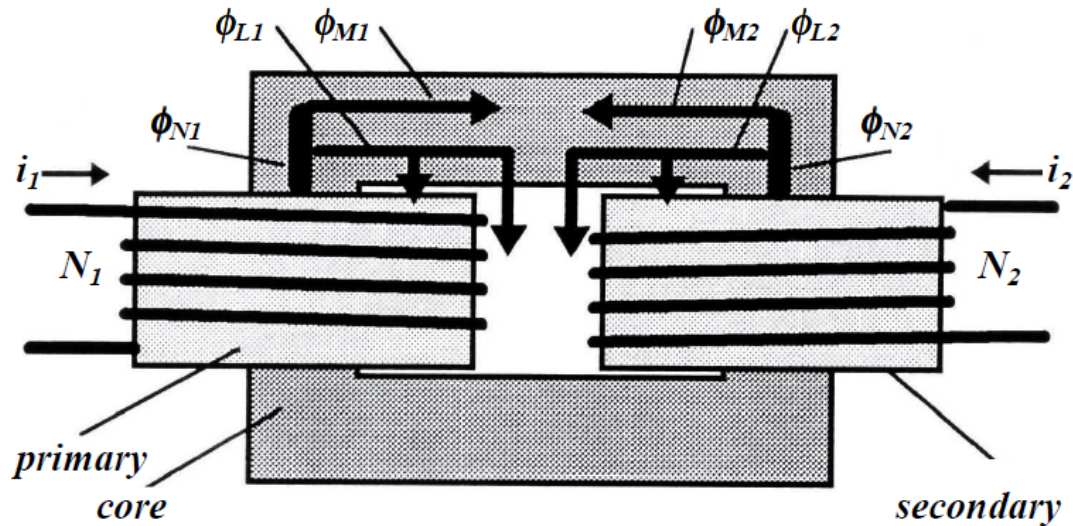
# Forward Converter



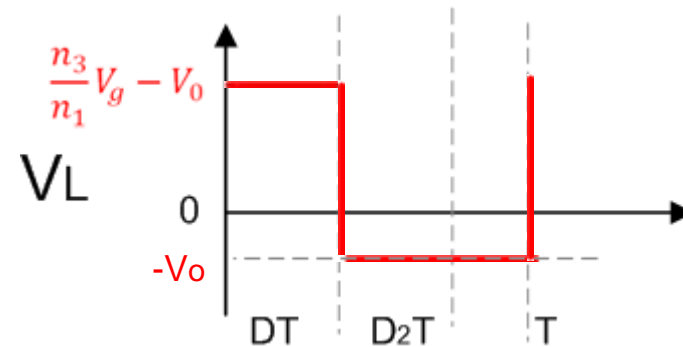
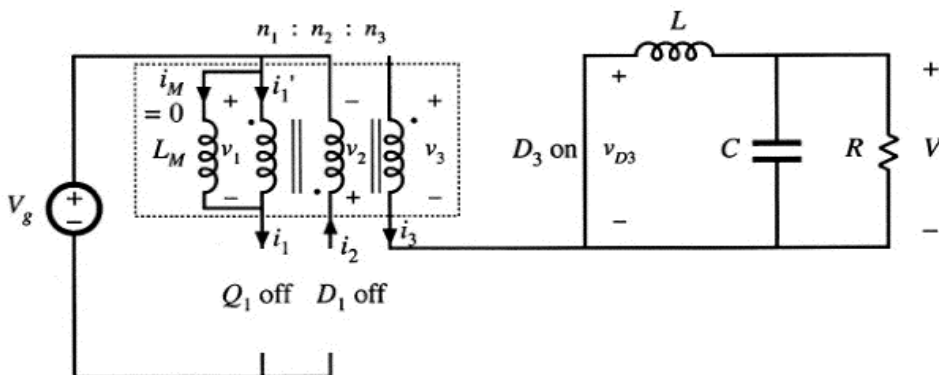
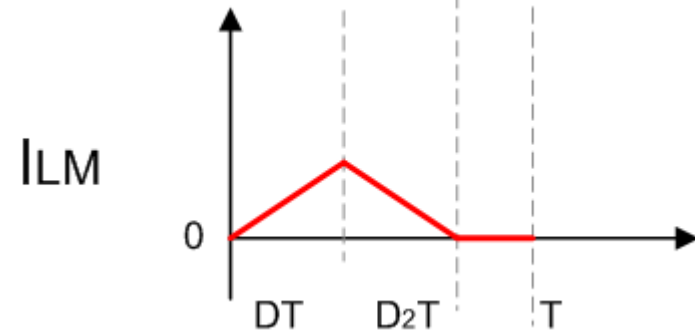
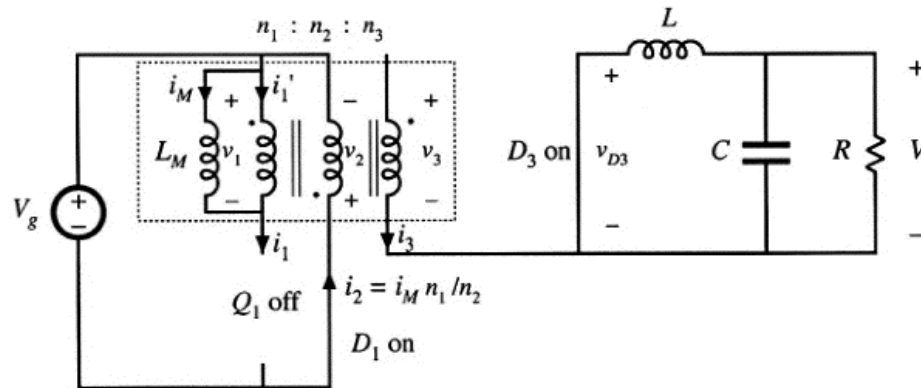
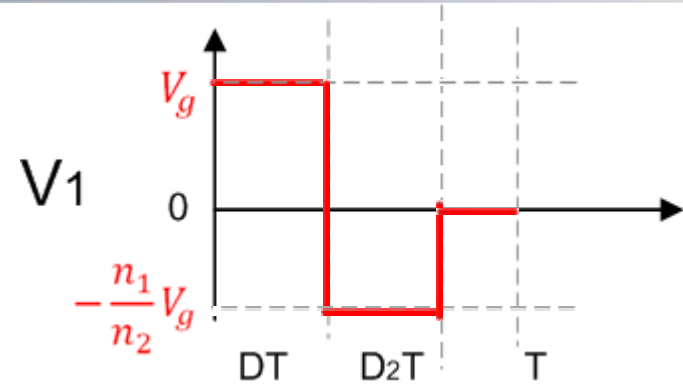
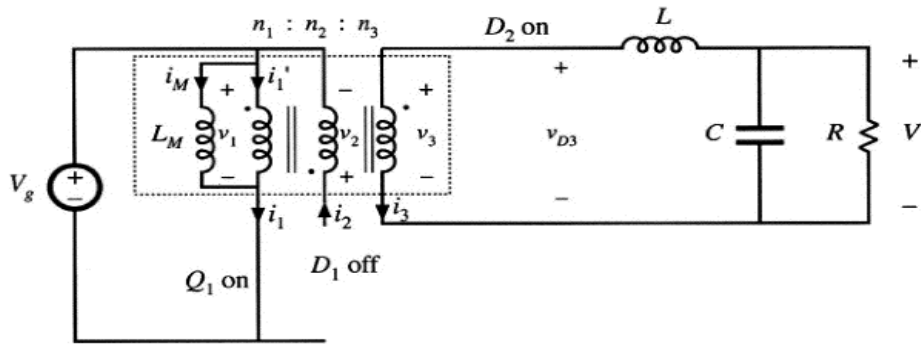
Why is there three windings transformer?



# Magnetizing Inductance



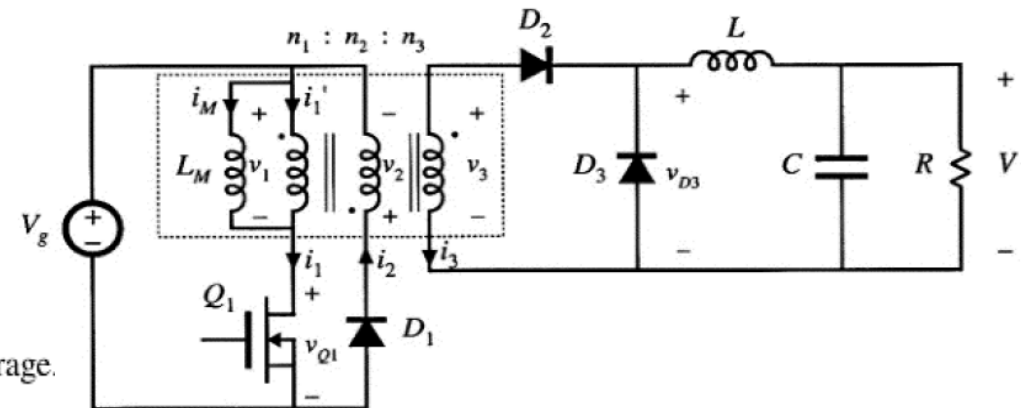
# Forward Converter (CCM)





# Forward Converter

## Duty cycle limitaion:



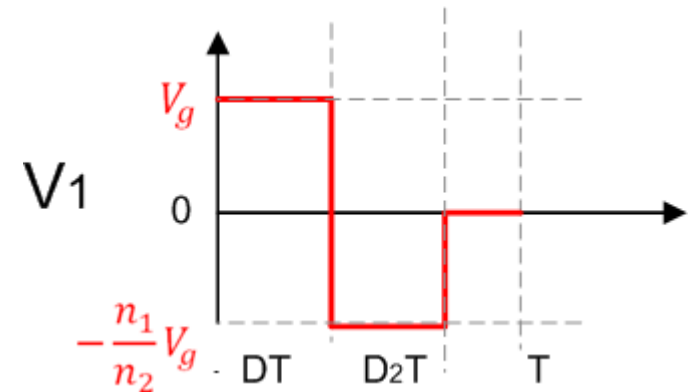
the primary winding voltage  $v_1(t)$  must have zero average.

$$\langle v_1 \rangle = D(V_g) + D_2(-V_g n_1/n_2) + D_3(0) = 0$$

$$D_2 = \frac{n_2}{n_1} D$$

the duty cycle  $D_3$  cannot be negative

$$D \leq \frac{1}{1 + \frac{n_2}{n_1}}$$



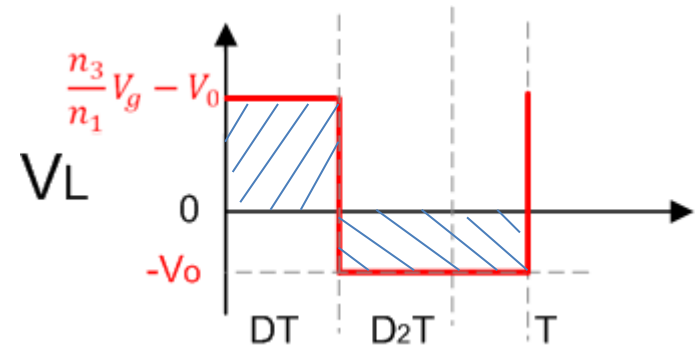
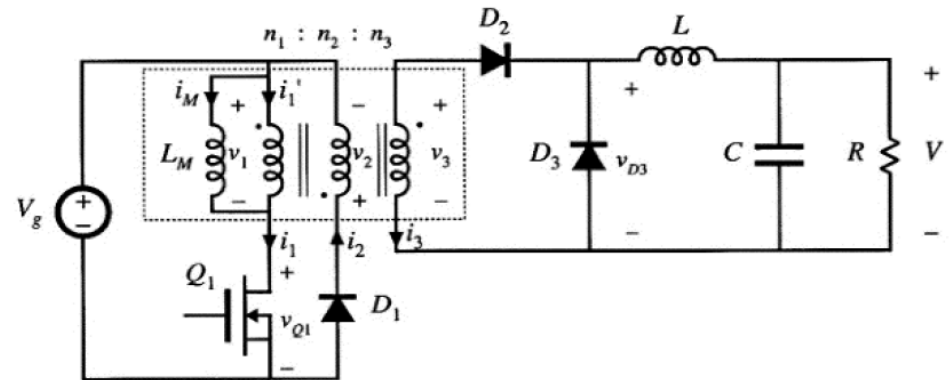
# Forward Converter

Transfer function:

$$\frac{n_3}{n_1} V_g D = V_o (1 - D)$$



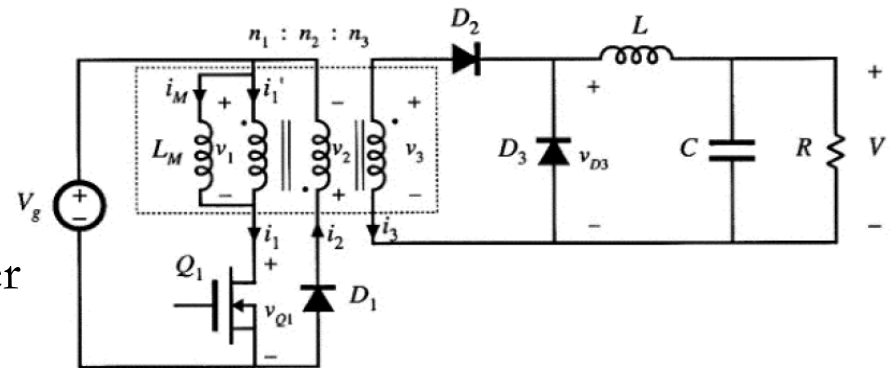
$$\frac{V_o}{V_g} = \frac{n_3}{n_1} D$$



# Group Discussion

## Question 1:

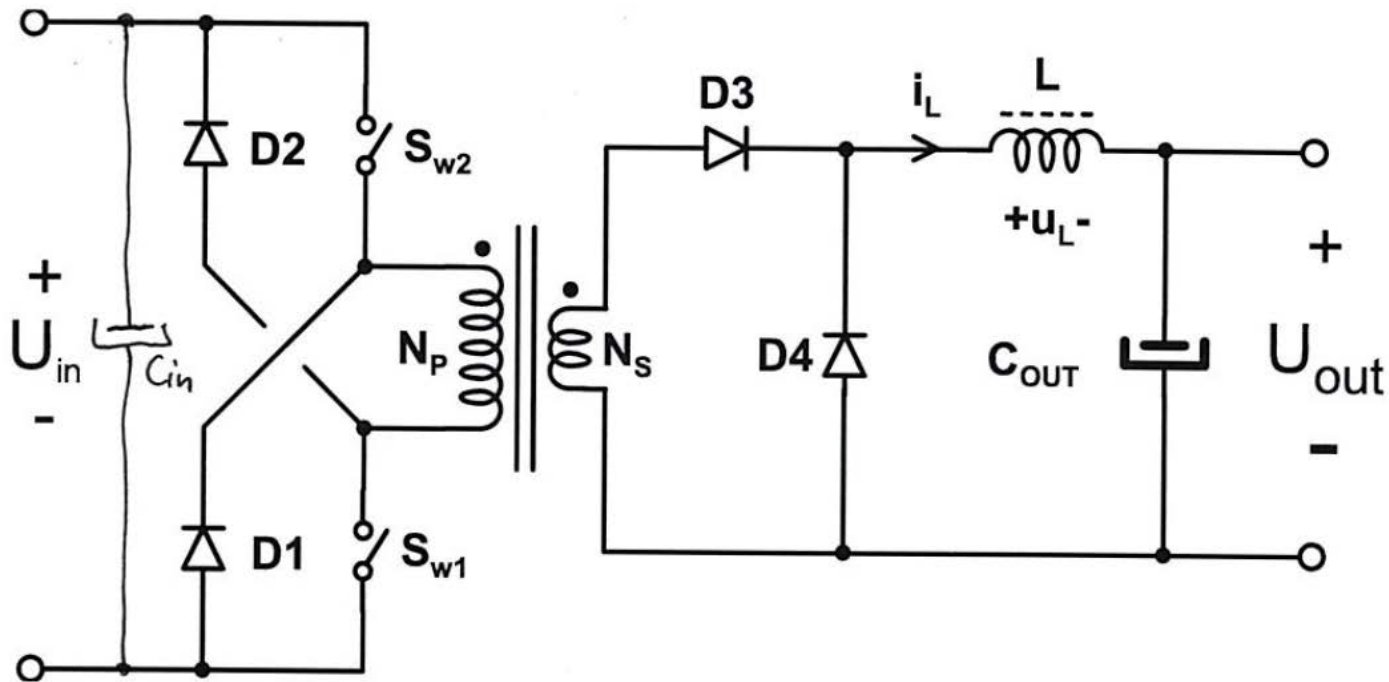
You are going to design a 120W forward converter. The input voltage is 48 V and the output voltage is 12 V. The transformer turns ratio is 1:1:1, and the converter is operated at 100k Hz and in CCM mode. The magnetizing inductance is 12uH. The output inductance is assumed to be infinite. What is the max drain-source voltage of MOSFET,  $V_{ds}$ , and what is the peak current of the MOSFET,  $I_{ds}$ ?



- A. 12V
- B. 48V
- C. 60V
- D. 96V

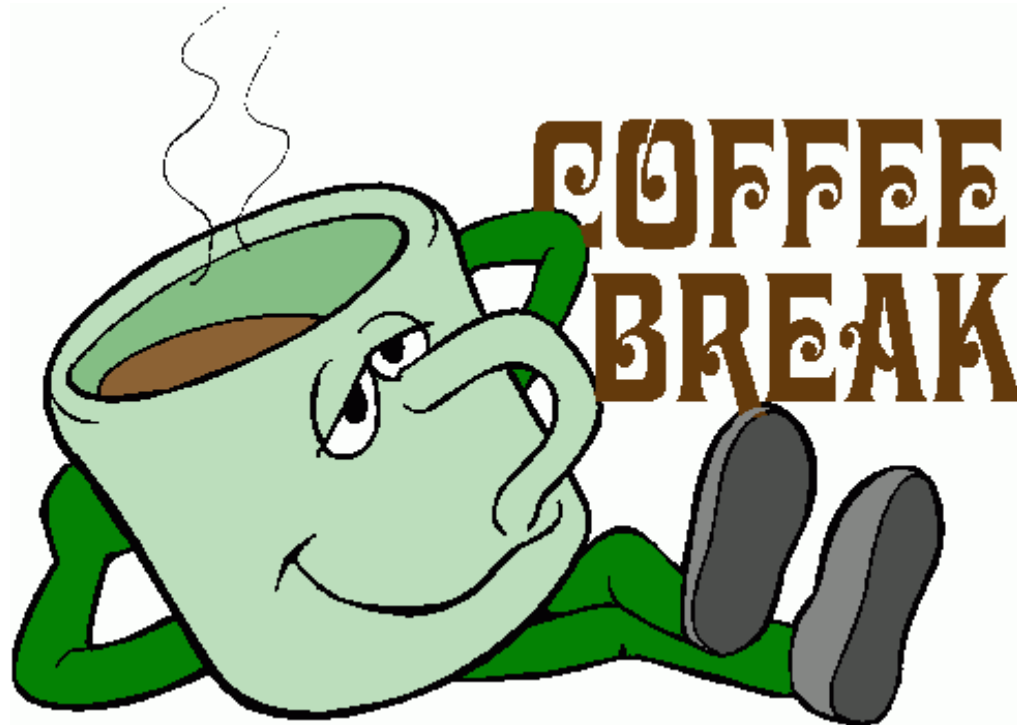
- a. 2.5A
- b. 5A
- c. 10A
- d. 20A

# Two-switch Forward Converter

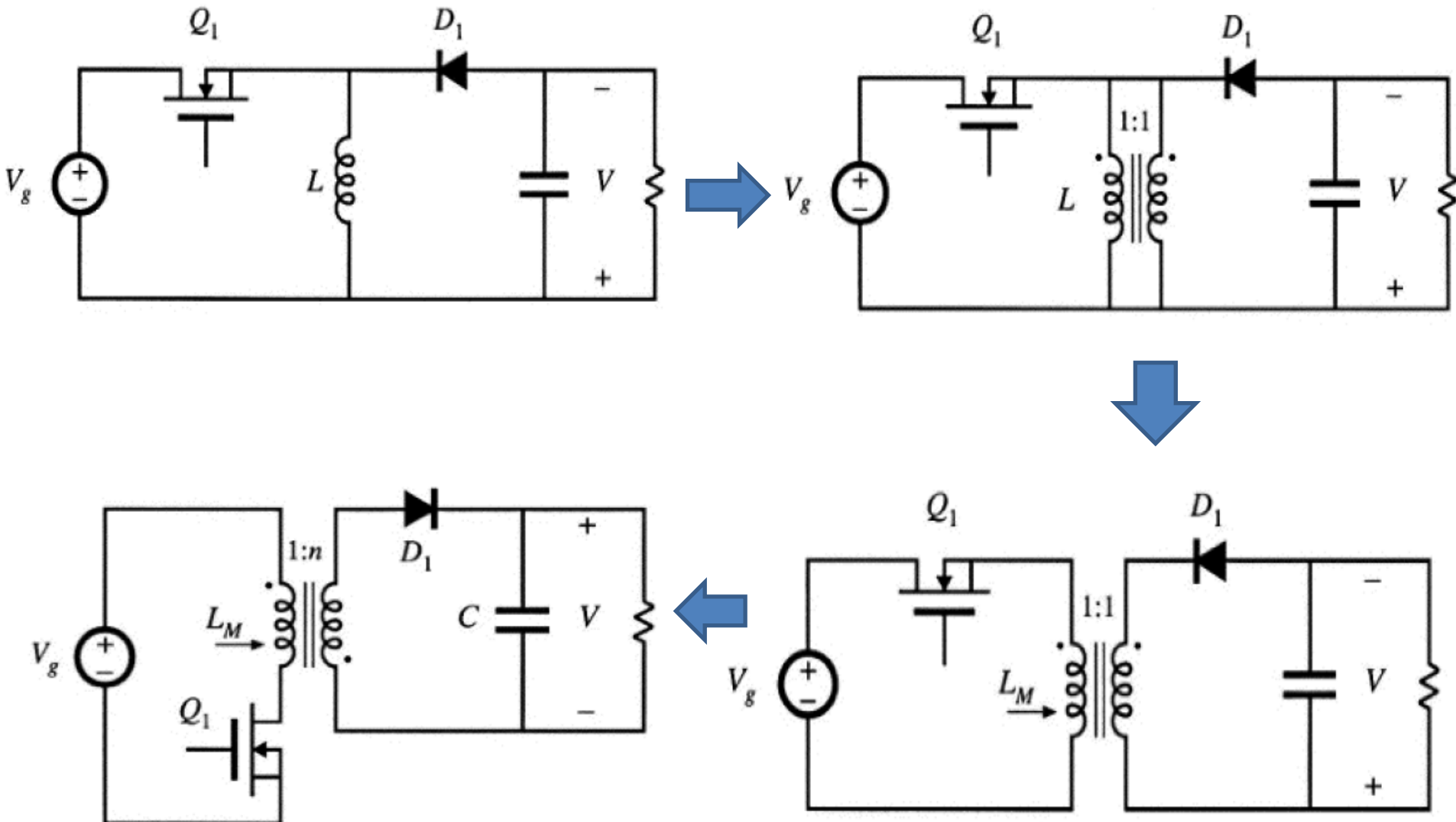


- + The maximum switch blocking voltage is equal to  $U_{IN}$  (clamped!)
- + Snubber is regenerative (for the magnetizing current)
- + Very robust (Inherently robust)

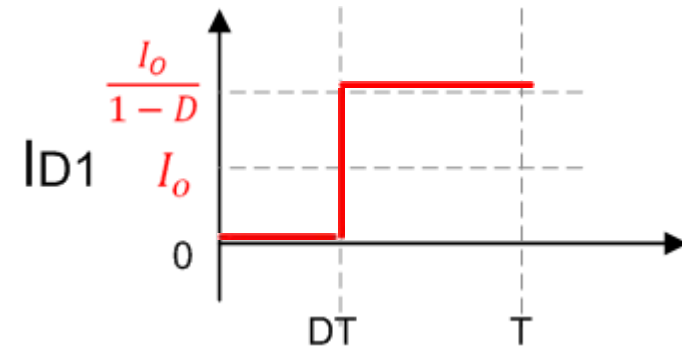
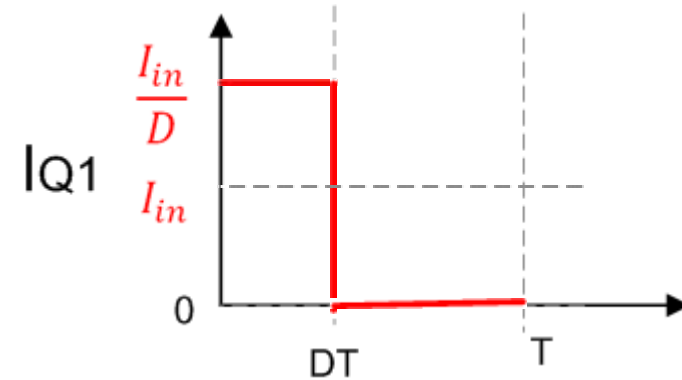
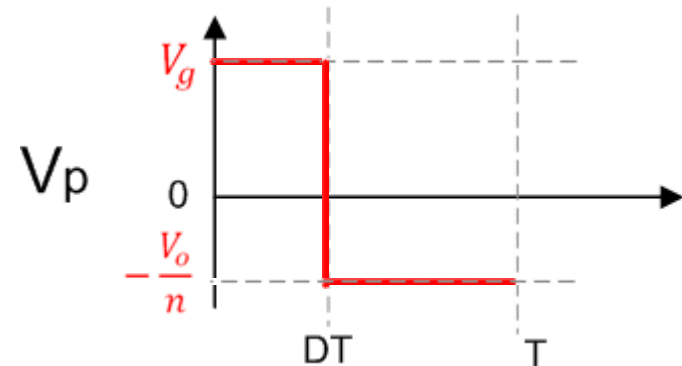
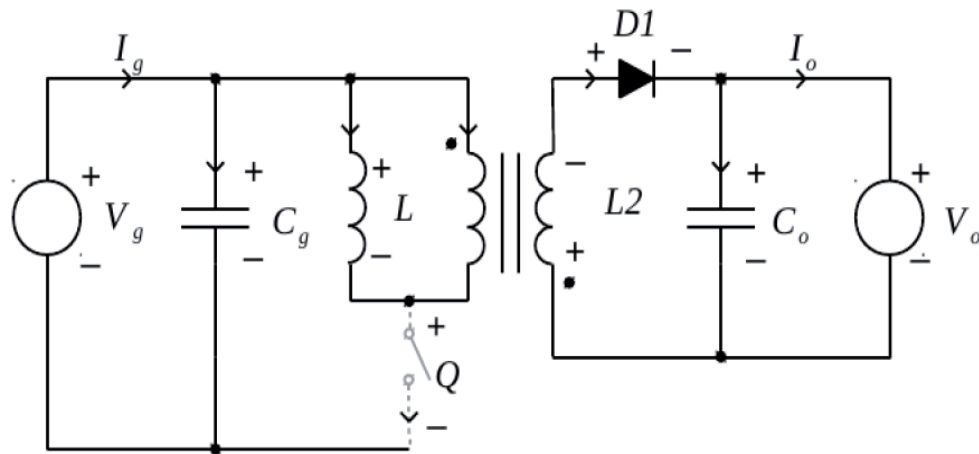
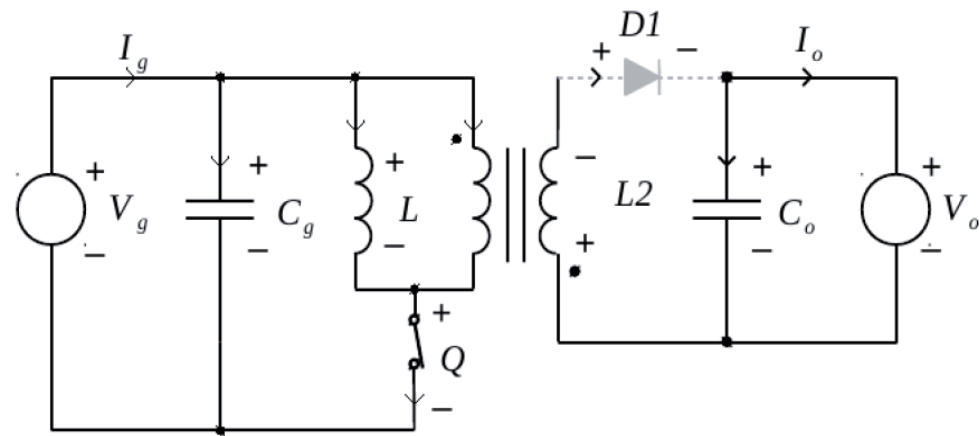
- Requires a high-side driver for switch-2
- Requires two switch elements
- The duty-cycle can never be more than 50%
- Excites only the core in one flux-direction



# Flyback Converter



# Flyback Converter



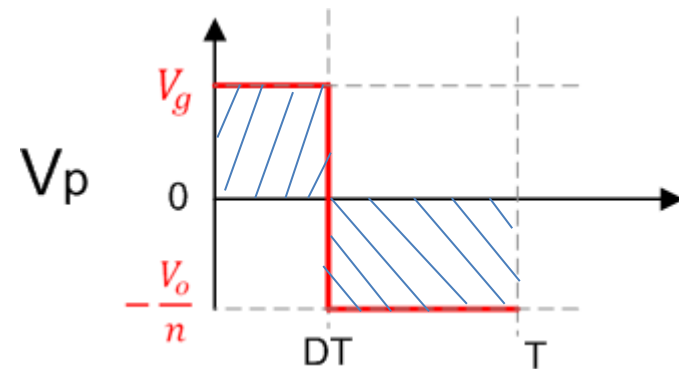
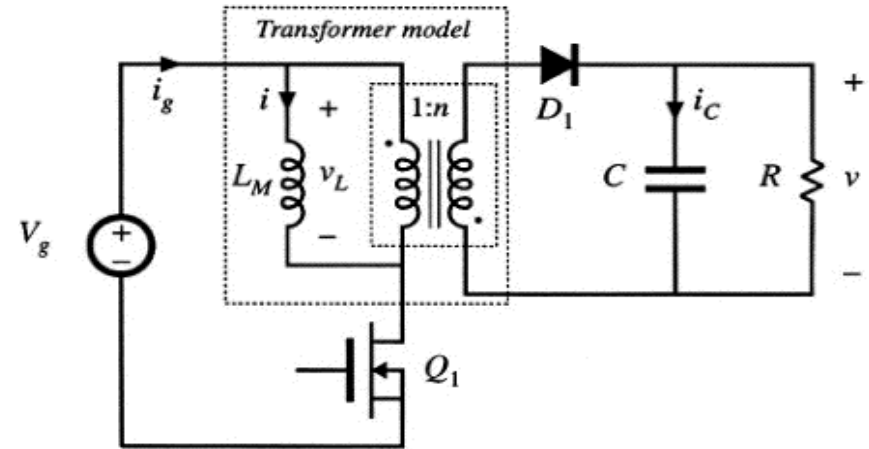
# Flyback Converter

Transfer function:

$$V_g DT = \frac{V_o}{n} (1 - D) T$$



$$\frac{V_o}{V_g} = \frac{nD}{1 - D}$$

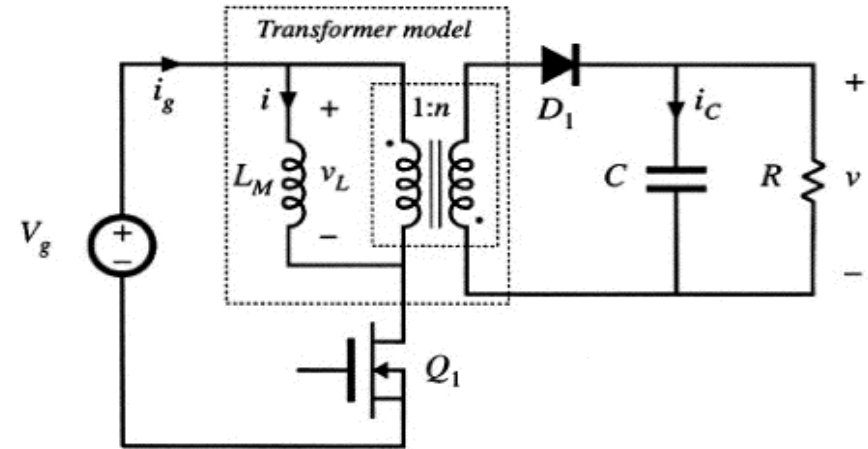




# Group Discussion

## Question 2:

This is Flyback converter. The input voltage is 48 V and the output voltage is 12 V. The transformer turns ratio is 4:1, and the converter is operated at 100k Hz and in CCM mode. The magnetizing inductance is 100uH. What is the max drain-source voltage stress of MOSFET,  $V_{ds}$ , and what is the max voltage stress of the diode  $D_1$ ?



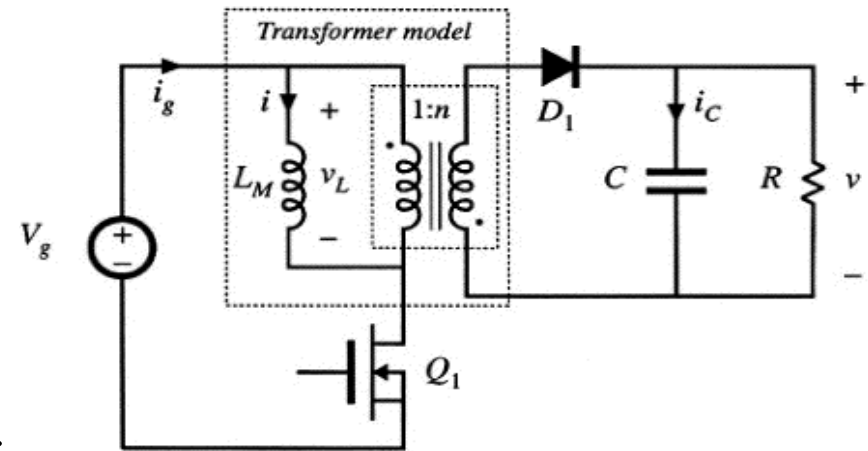
- A. 12V
- B. 24V
- C. 48V
- D. 96V

- a. 12V
- b. 24V
- c. 48V
- d. 96V

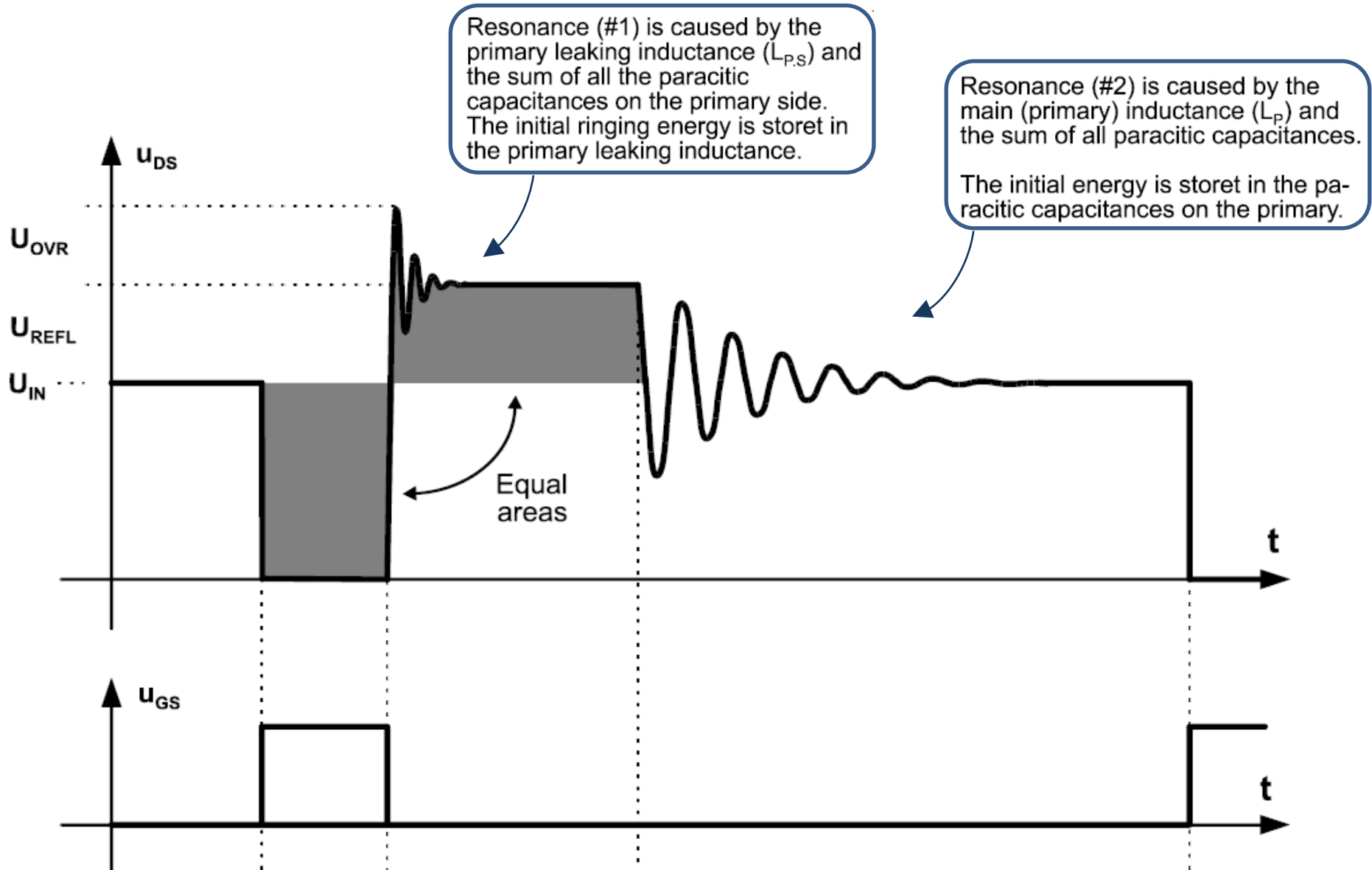
# Flyback Converter

## Disadvantages:

- ❑ higher component stress
- ❑ spike and ringing due to leakage inductance
- ❑ half of core material B-H loop is utilized

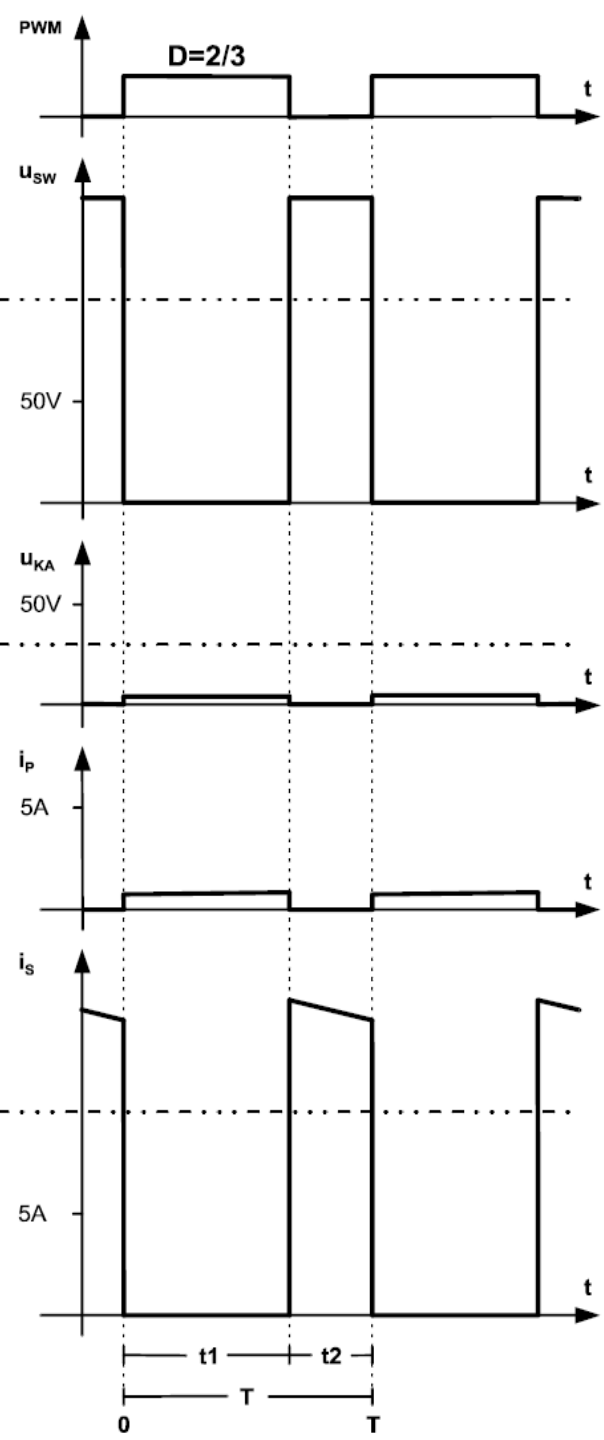
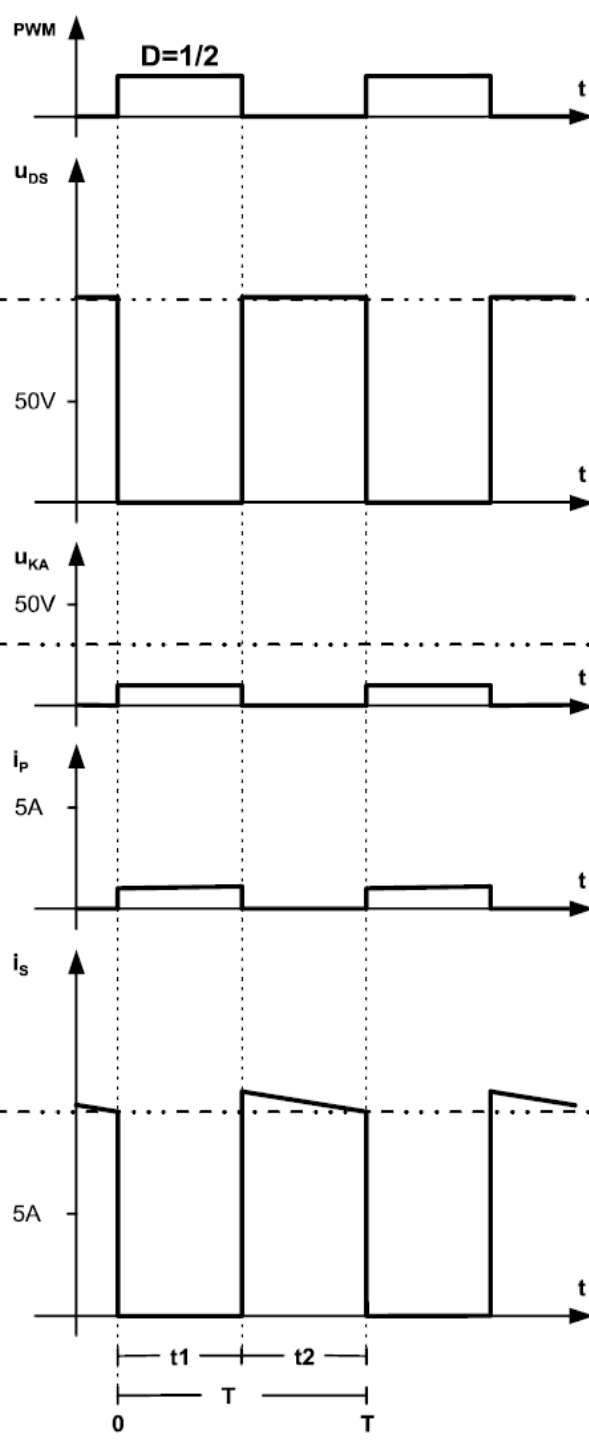
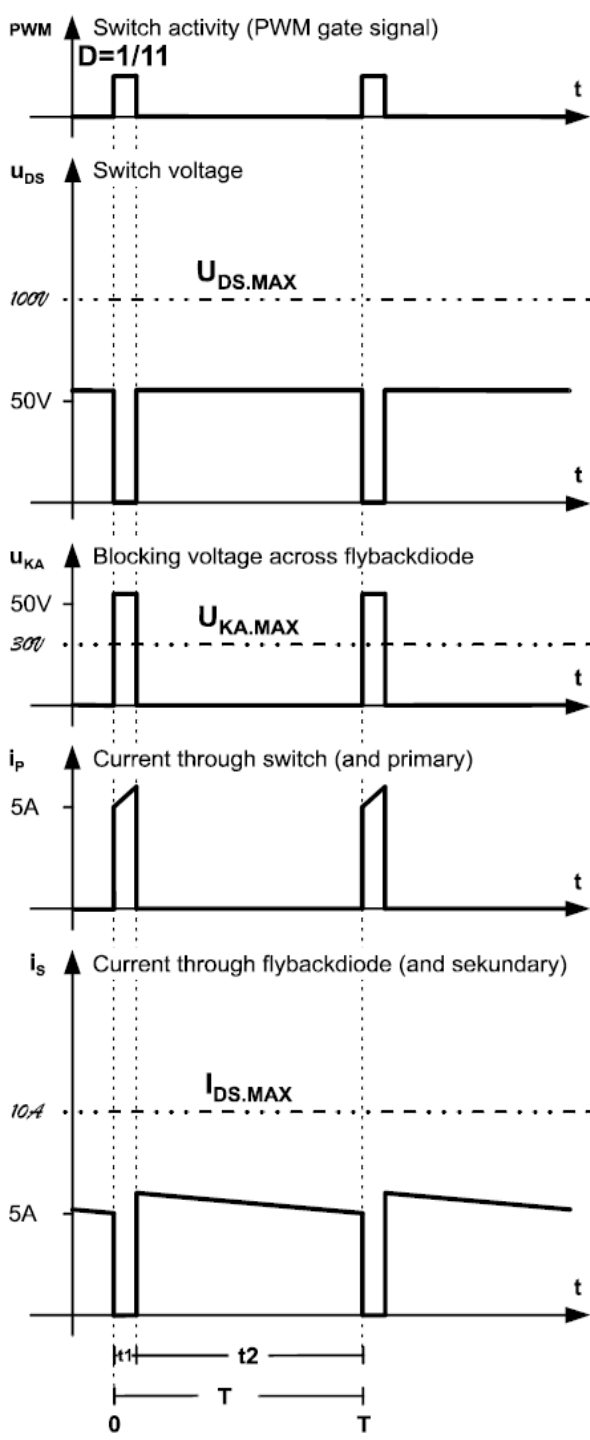


# Signals in a “real” flyback converter

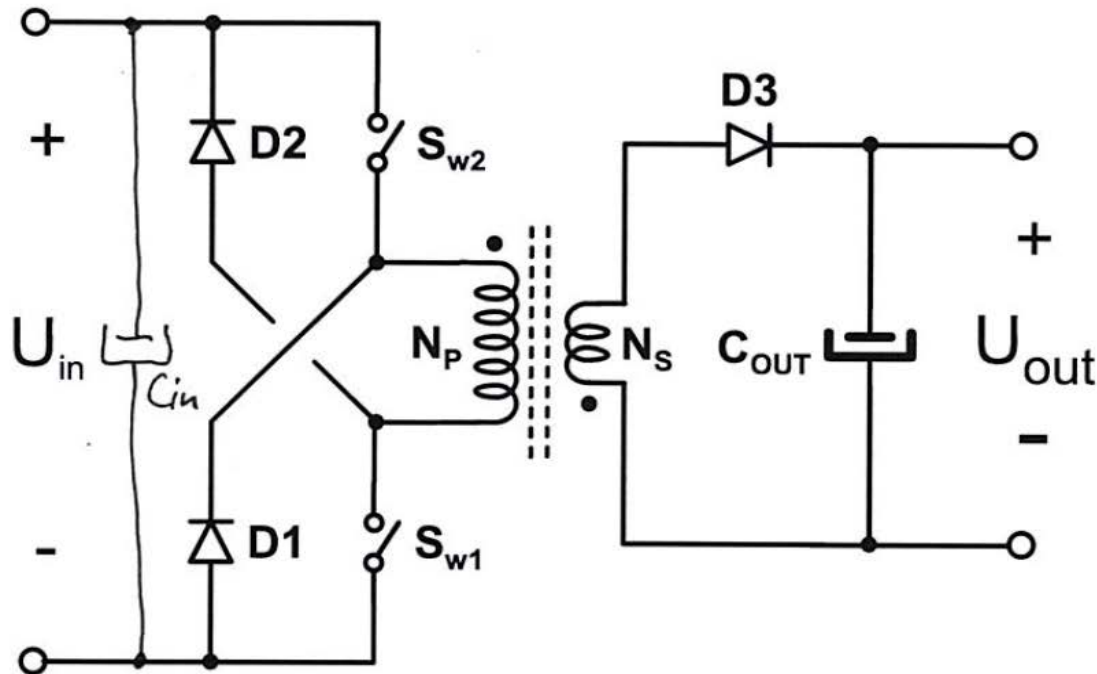


# Design Considerations

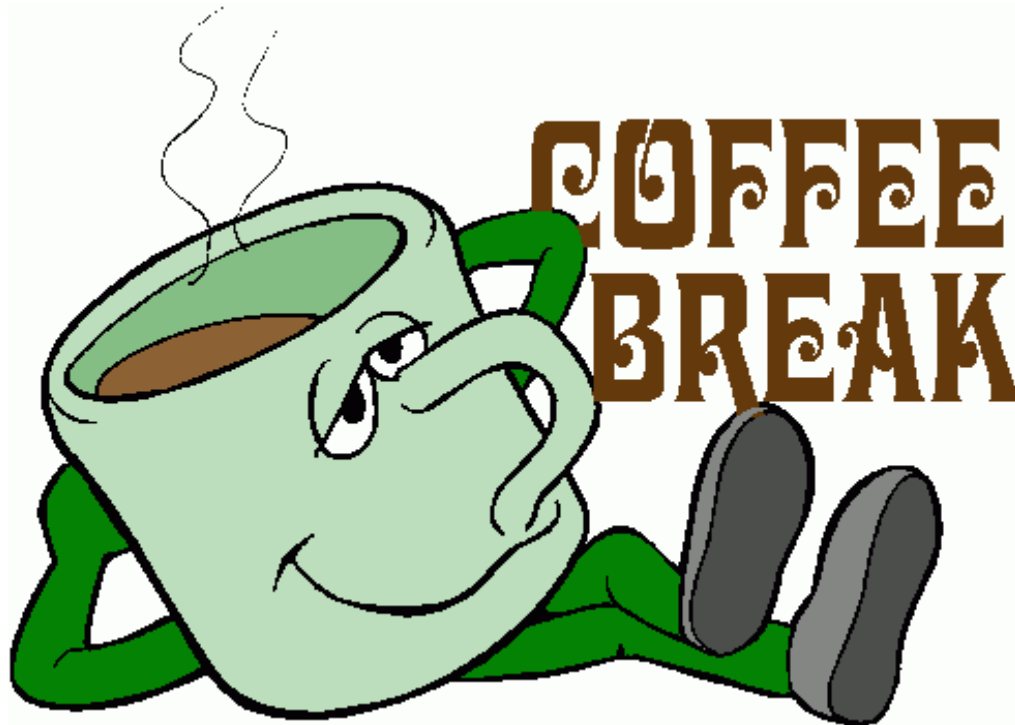
	$U_{IN}=50VDC, U_{OUT}=5VDC,$ $I_{OUT}=5A, \Delta I_S=1A, f_{sw}=50kHz$	$n=N_s/N_p=1,$ $D=1/11$	$n=N_s/N_p=0.1,$ $D=1/2$	$N_s/N_p=0.05$ $D=2/3$
Blocking voltage across switch	$U_{DS} = U_{OUT} \cdot \frac{1}{n} + U_{IN}$	55V	100V	150V
Blocking voltage across flyback diode	$U_{KA} = U_{IN} \cdot n + U_{OUT}$	55V	10V	7.5V
Peak current on primary	$I_{PP} = n \cdot I_{SP}$	6A	1.05A	0.775A
Peak current on secondary	$I_{SP} = \frac{I_o - \frac{1}{2} \cdot \Delta I_S (D-1)}{(1-D)}$	6A	10.5A	15.5A
Primary inductance	$L_P = n^2 \cdot L_S$	90.91uH	5mH	13.33mH
Comments:	$\left( L_S = \frac{U_{OUT} \cdot (1-D)T}{\Delta I_S} \right)$	+ smallest output capacitor -High blocking voltage diode --High loss in primary winding	+small input capacitor +low blocking voltage diode - High output capacitor	+ smallest input capacitor +lowest blocking voltage diode -highest output capacitor -high blocking voltage MOSFET



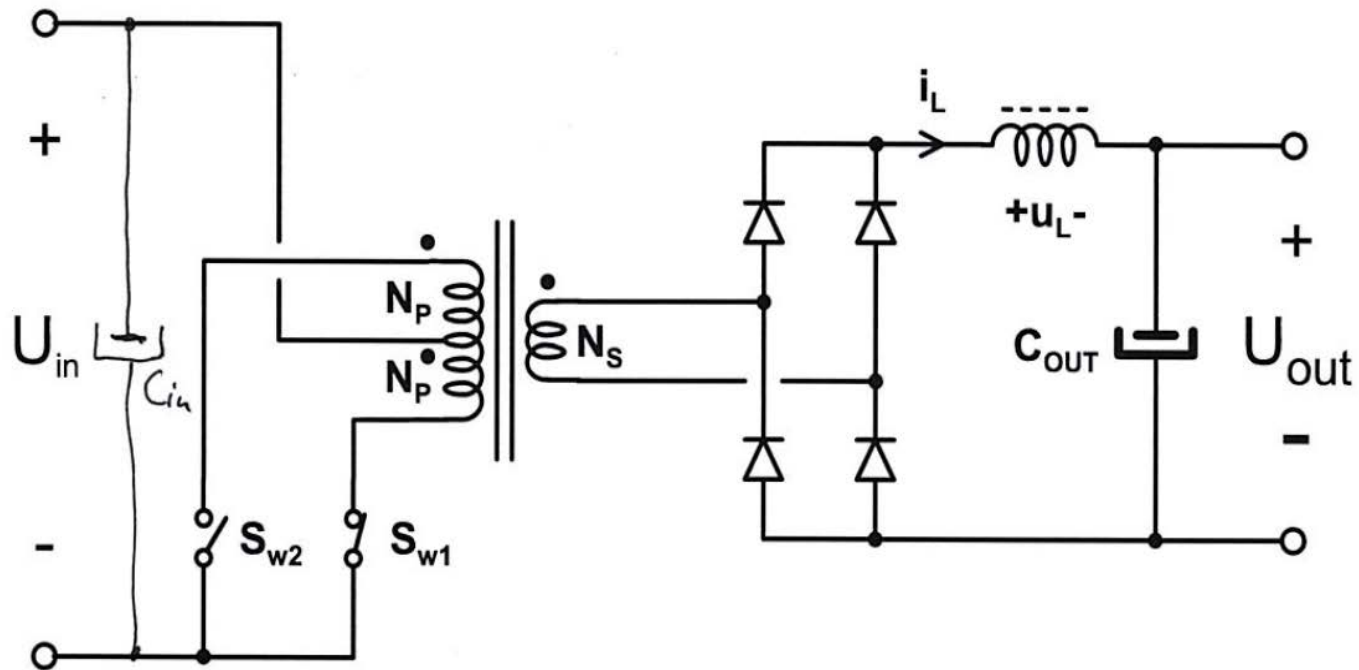
# Two-Switch Flyback Converter



- + The maximum switch blocking voltage is equal to  $U_{IN}$  (clamped!)
- + Snubber is regenerative (For the primary leaking inductance)
- + Very robust (Inherently robust)
- Requires a high-side driver for switch-2
- Requires two switch elements
- The duty-cycle can never be more than 50%
- Excites only the core in one flux-direction



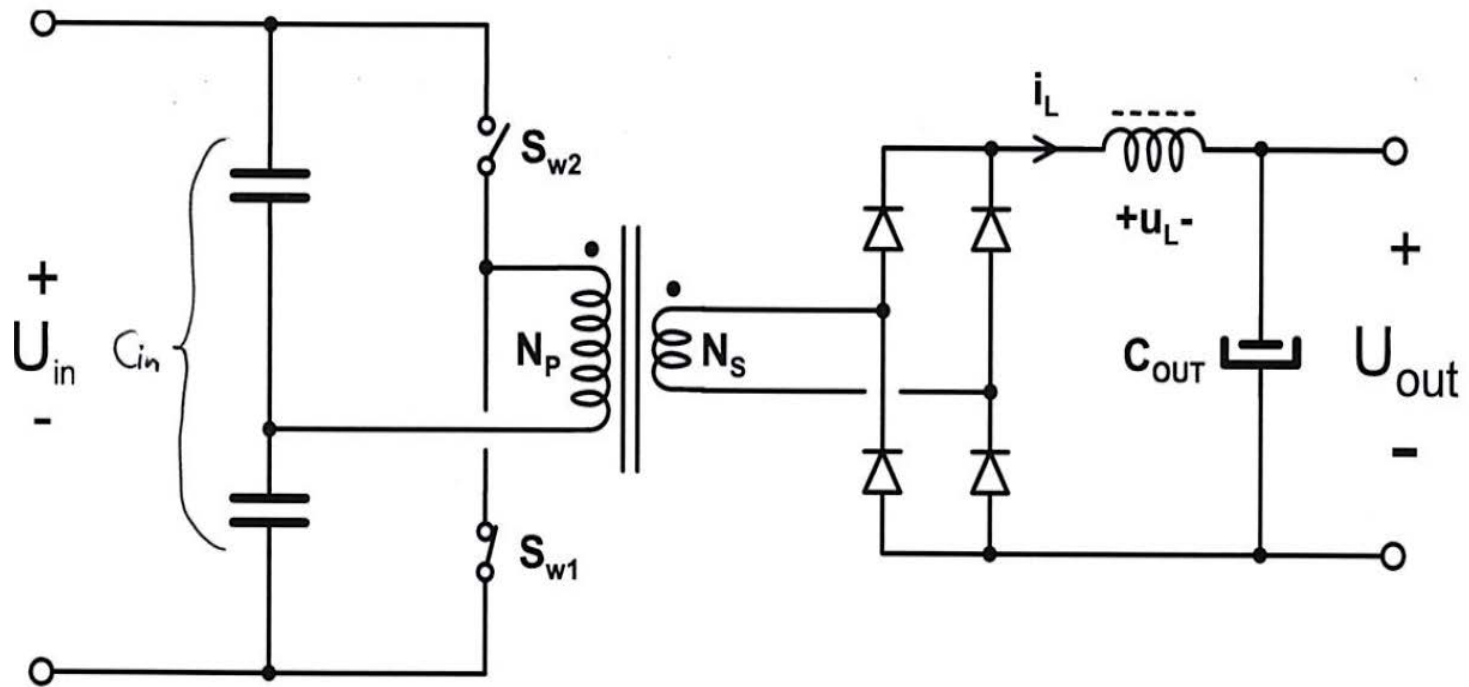
# Push-Pull Converter



- + Does not require any high-side driver for switch-2
- + Theoretically the duty-cycle can be 100%
- + Excites the core in both flux-directions
- The switch blocking voltage is at least equal to  $2 \times U_{IN}$ , (not well controlled!)
- Requires a split primary (or two equal primary windings)
- Requires two switch elements
- Diode conduction losses is comparatively high at low output voltages

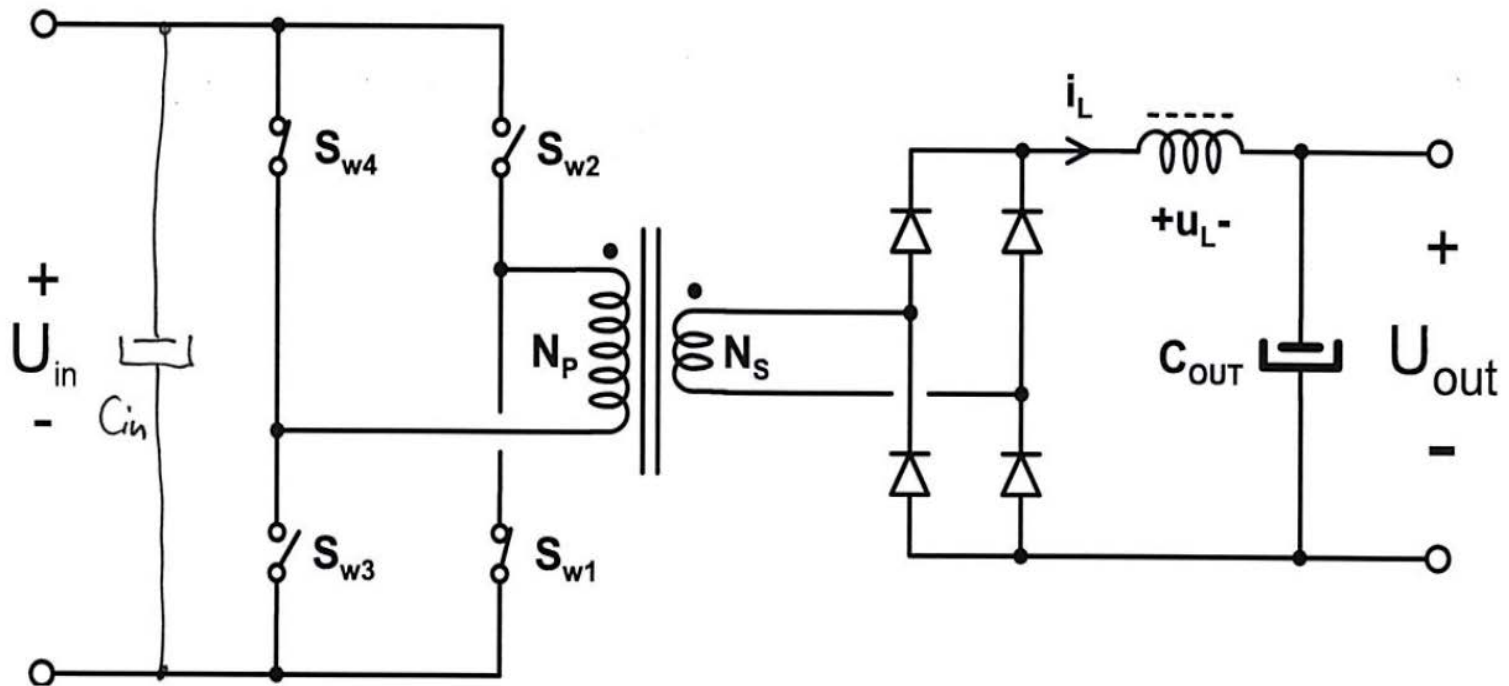


# Half-bridge Converter



- + The maximum switch blocking voltage is equal to  $U_{IN}$  (clamped by FET BD.)
- + Theoretically the duty-cycle can be 100%
- + Excites the core in both flux-directions
- + Does not require a split primary
- Requires a high-side driver for switch-2
- Requires two switch elements
- Diode conduction losses are comparatively high at low output voltages

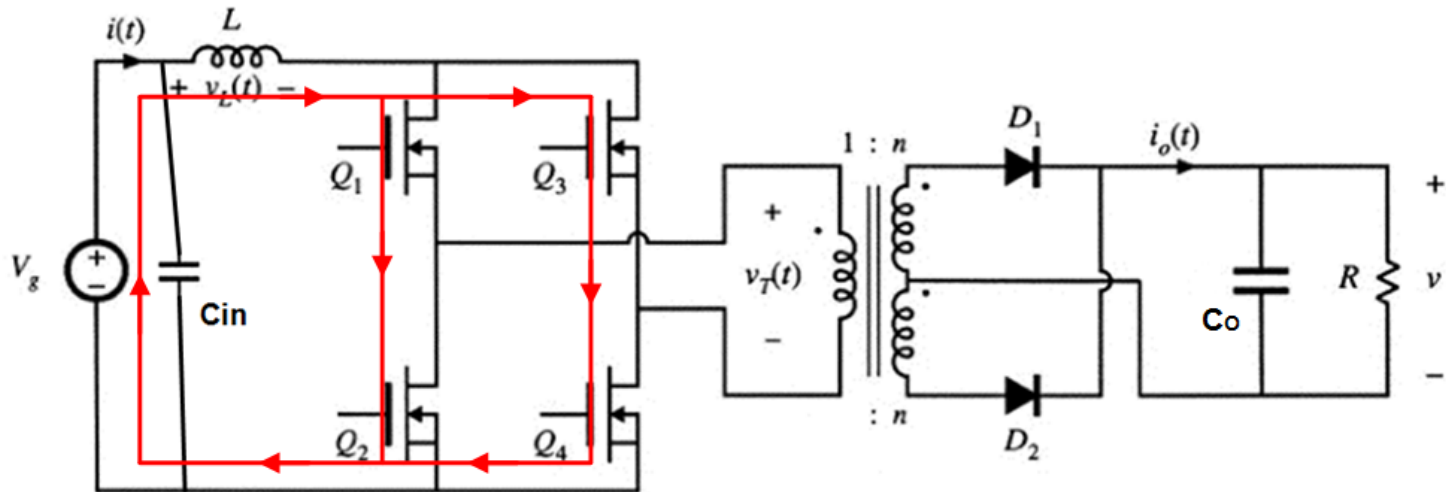
# Full-bridge Buck Converter



- + The maximum switch blocking voltage is equal to  $U_{IN}$  (clamped by FET BD.)
- + Theoretically the duty-cycle can be 100%
- + Excites the core in both flux-directions
- + Does not require a split primary
- Requires a high-side driver for switch-2,4
- Requires four switch elements
- Diode conduction losses is comparatively high at low output voltages

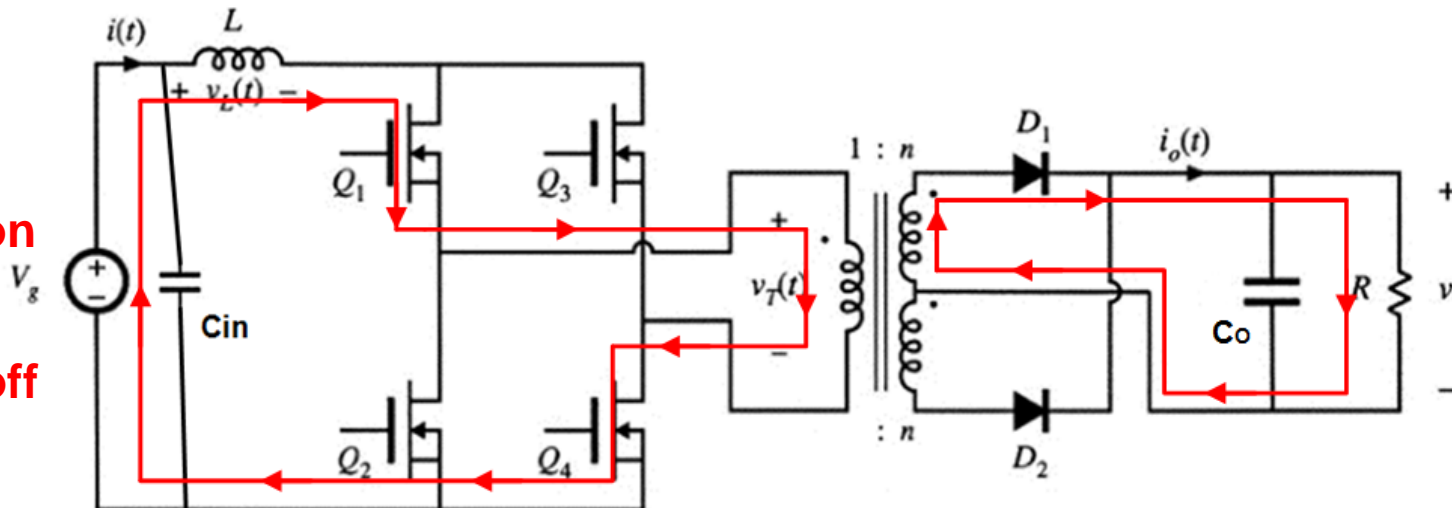
# Full-bridge Boost Converter

All turn on

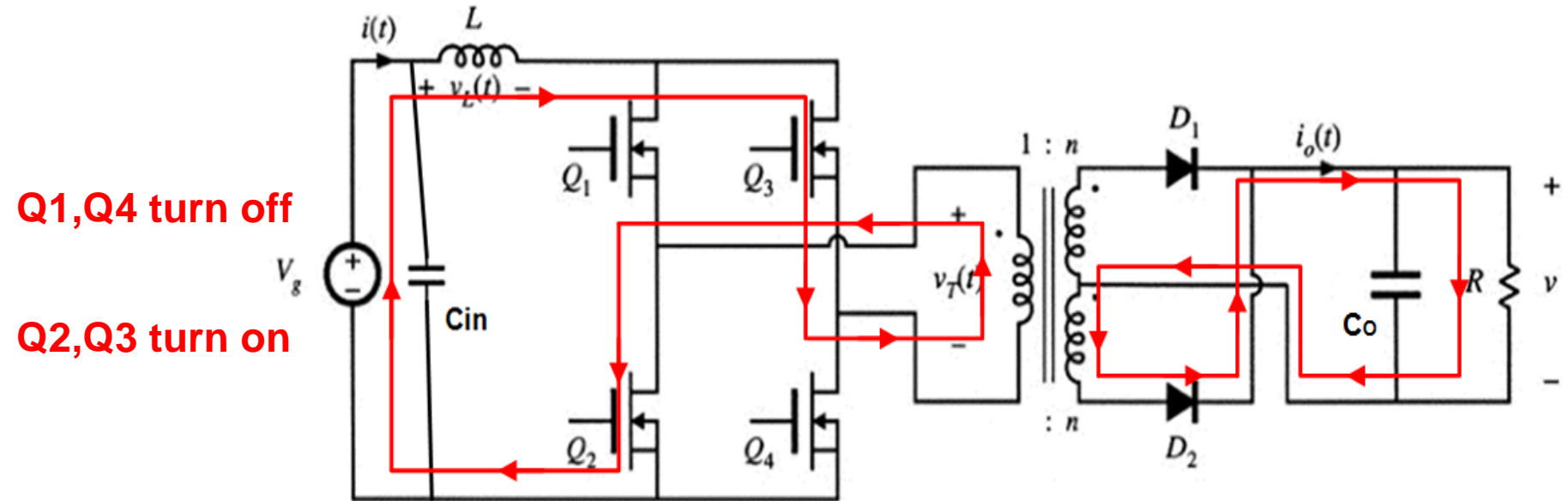


Q1, Q4 turn on

Q2, Q3 turn off



# Full-bridge Boost Converter



Please refer to **Chapter 6** in the text book “fundamentals of power electronics”