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Engineering Academy
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**Result Oriented Coaching For
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GATE 2016

**Detailed Solutions For
Electronics & Communication Engg**

**Date: 30-01-2016
Forenoon Session**

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Q.1 – Q.5 Carry One Mark Each

01. Which of the following is CORRECT with respect to grammar and usage?

Mount Everest is _____

- (A) the highest peak in the world
(B) highest peak in the world
(C) one of highest peak in the world
(D) one of the highest peak in the world

01. Ans: (A)

Sol: Before superlative article 'the' has to be used. 'one of' the expression should take plural noun and so option 'C' and 'D' can't be the answer.

02. The policeman asked the victim of a theft, "What did you _____?"

- (A) loose (B) lose (C) loss (D) louse

02. Ans: (B)

Sol: 'lose' is verb.

03. Despite the new medicine's _____ in treating diabetes, it is not _____ widely

- (A) effectiveness --- prescribed (B) availability ---- used
(B) prescription ---- available (D) acceptable--- proscribed

03. Ans: (A)

Sol: 'effectiveness' is noun and 'prescribed' is verb. These words are apt and befitting with the word 'medicine.'

04. In a huge pile of apples and oranges, both ripe and unripe mixed together, 15% are unripe fruits, Of the unripe fruits, 45% are apples, Of the ripe ones, 66% are oranges. If the pile contains a total of 5692000 fruits, how many of them are apples?

- (A) 2029198 (B) 2467482 (C) 2789080 (D) 3577422

04. Ans: (A)

Sol: Total no. of fruits = 5692000

Unripe type of apples = 45% of 15% of 5692000

$$= \frac{45}{100} \times \frac{15}{100} \times 5692000$$
$$= 384210$$

$$\text{Ripe type of apples} = \frac{34}{100} \times \frac{85}{100} \times 5692000 = 1644988$$

$$\therefore \text{Total no. of apples} = 384210 + 1644988 = 2029198$$

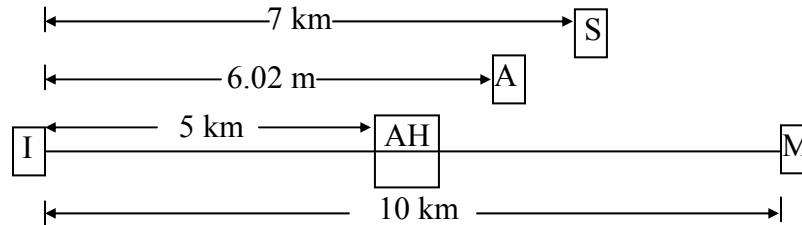
05. Michael lives 10 km away from where I live. Ahmed lives 5km away and Susan lives 7 km away from where I live. Arun is farther away than Ahmed but closer than Susan from where I live. From the information provided here, what is one possible distance (in km) at which I live from Arun's place?

- (A) 3.00 (B) 4.99 (C) 6.02 (D) 7.01



05. Ans: (c)

Sol: From given data, the following diagram is possible



I = I live

AH = Ahmed lives; M = Michael lives

S = Susan lives; A = Arun lives

→ Arun lives farther away than Ahmed means more than 5 km but closer than Susan means less than 7 km, from given alternatives, option 'C' only possible.

Q.6 – Q.10 Carry two marks each

06. A person moving through a tuberculosis prone zone has a 50% probability of becoming infected. However, only 30% of infected people develop the disease. What percentage of people moving through a tuberculosis prone zone remains infected but does not show symptoms of disease?

(A) 15

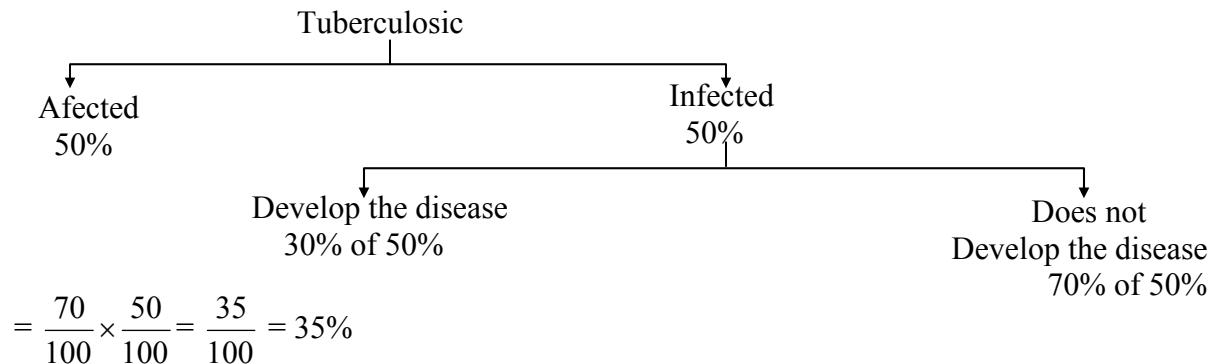
(B) 33

(C) 35

(D) 37

06. Ans: (C)

Sol:



07. In a world filled with uncertainty, he was glad to have many good friends. He had always assisted them in times of need and was confident that they would reciprocate. However, the events of the last week proved him wrong.

Which of the following inference(s) is/are logically valid and can be inferred from the above passage?

(i) His friends were always asking him to help them

(ii) He felt that when in need of help, his friends would let him down.

(iii) He was sure that his friends would help him when in need.

(iv) His friends did not help him last week.

(A) (i) and (ii)

(B) (iii) and (iv)

(C) (iii) only

(D) (iv) only

07. Ans: (B)

Sol: The words 'was confident that they would reciprocate' and 'last week proved him wrong' lead to statements iii and iv as logically valid inferences.

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08. Leela is older than her cousin Pavithra, pavithra's brother Shiva is older than Leela. When Pavithra and shiva are visiting Leela, all three like to play chess. Pavithra wins more often than Leela does. Which one of the following statements must be TRUE based on the above?
- (A) When Shiva plays chess with Leela and Pavithra. he often loses.
 (B) Leela is the oldest of the three.
 (C) Shiva is a better chess palyer than Pavithra
 (D) Pavithra is the young of the three

08. Ans: (D)

Sol: From given data, the following arrangement is possible

Shiva

Leela

Pavithra

Among four alternatives, option D is TRUE.

09. If $q^{-a} = \frac{1}{r}$ and $r^{-b} = \frac{1}{s}$ and $S^{-c} = \frac{1}{q}$, the value of abc is ____.

(A) $(rqs)^{-1}$

(B) 0

(C) 1

(D) $r+q+s$

09. Ans: (C)

Sol: $q^{-a} = \frac{1}{r} \Rightarrow \frac{1}{q^a} = \frac{1}{r} \Rightarrow q^a = r$

$$r^{-b} = \frac{1}{s} \Rightarrow \frac{1}{r^b} = \frac{1}{s} \Rightarrow s = r^b$$

$$s^{-c} = \frac{1}{q} \Rightarrow \frac{1}{s^c} = \frac{1}{q} \Rightarrow s^c = q$$

$$q^a = r \Rightarrow (s^c)^a = r \Rightarrow s^{ac} = r$$

$$(s^{ac})^b = s$$

$$s^{abc} = s'$$

$$\therefore abc = 1$$

10. P,Q,R and S are working on a project. Q can finish the taks in 25 days, working alone for 12 hours a day. R can finish the task in 50 days, working alone for 12 hours per day. Q worked 12 hours a day but took sick leave in the beginning for two days. R worked 18 hours a day on all days. What is the ratio of work done by Q and R after 7 days from the start of the project?

(A) 10:11

(B) 11:10

(C) 20:21

(D) 21:20

10. Ans: (C)

Sol: Q can finish the task = 25 days, 12 hrs/day

$$= 300 \text{ hrs, } 1 \text{ hr} = \frac{1}{300} \text{ th}$$

R can finish the task = 50 days, 12 hrs/day



$$= 50 \times 12$$

$$= 600 \text{ hrs, } 1 \text{ hr} = \frac{1}{600} \text{ th}$$

Q working hours $\Rightarrow (7 - 2) \times 12 = 60 \text{ hrs}$

R working hours $\Rightarrow 7 \times 18 = 126 \text{ hrs}$

After 7 days, the ratio of work done by Q and R

$$\begin{array}{lcl} Q & : & R \\ \hline 60 & : & 126 \\ 300 & : & 600 \\ 20 & : & 21 \end{array}$$

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Q.1 – Q.25 Carry one mark each.

01. Let $M^4 = I$ (where I denotes the identity matrix) and $M \neq I$, $M^2 \neq I$ and $M^3 \neq I$. Then, for any natural number k , M^{-1} equals:

(A) M^{4k+1}

(B) M^{4k+2}

(C) M^{4k+3}

(D) M^{4k}

01. Ans: (C)

Sol: $M^4 = I$

$$\Rightarrow M^8 = M^4 = I \Rightarrow M^7 = M^{-1}$$

$$\Rightarrow M^{12} = M^8 = I \Rightarrow M^{11} = M^{-1}$$

$$\Rightarrow M^{16} = M^{12} = I \Rightarrow M^{15} = M^{-1}$$

.....

$$\therefore M^{-1} = M^{4K+3}, K \text{ is a natural number}$$

02. The second moment of a Poisson-distributed random variables is 2. The mean of the random variable is ____.

02. Ans: $\lambda = 1$

Sol: $E(x^2) = 2$

$$V(X) = E(X^2) - (E(X))^2$$

Let mean of the poisson random variable be x

$$x = 2 - x^2$$

$$x^2 + x - 2 = 0$$

$$x = 1, -2$$

$$\therefore \text{Mean is } \lambda = 1$$

*mean & var
of poisson RV are
same*

03. Given the following statements about a function $f: \mathbb{R} \rightarrow \mathbb{R}$, select the right option:

P : If $f(x)$ is continuous at $x = x_0$, then it is also differentiable at $x = x_0$

Q : If $f(x)$ is continuous at $x = x_0$, then it may not be differentiable at $x = x_0$

R : If $f(x)$ is continuous at $x = x_0$, then it is also different at $x = x_0$

(A) P is true, Q is false, R is false

(B) P is false, Q is true, R is true

(C) P is false, Q is true, R is false

(D) P is true, Q is false, R is true

03. Ans: (B)

Sol: Since continuous function may not be differentiable. But differentiable function is always continuous.

04. Which one of the following is a property of the solutions to the Laplace equation : $\nabla^2 f = 0$?

(A) The solution have neither maxima nor minima anywhere except at the boundaries

(B) The solution are not separable in the coordinates

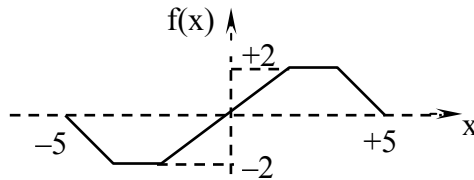
(C) The solution are not continuous

(D) The solution are not dependent on the boundary conditions

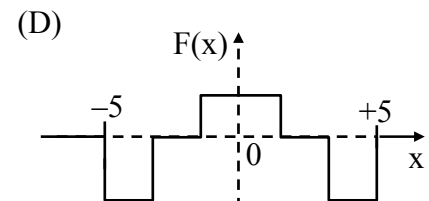
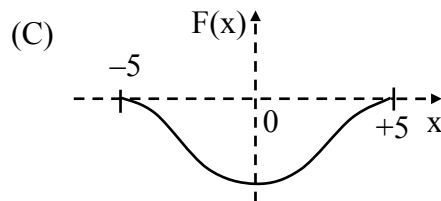
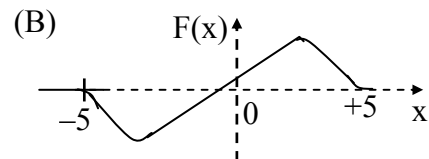
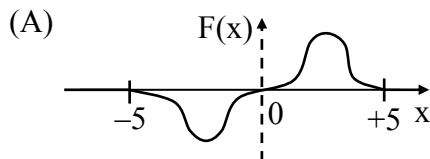
04. Ans: (A)



05. Consider the plot of $f(x)$ versus x as shown below

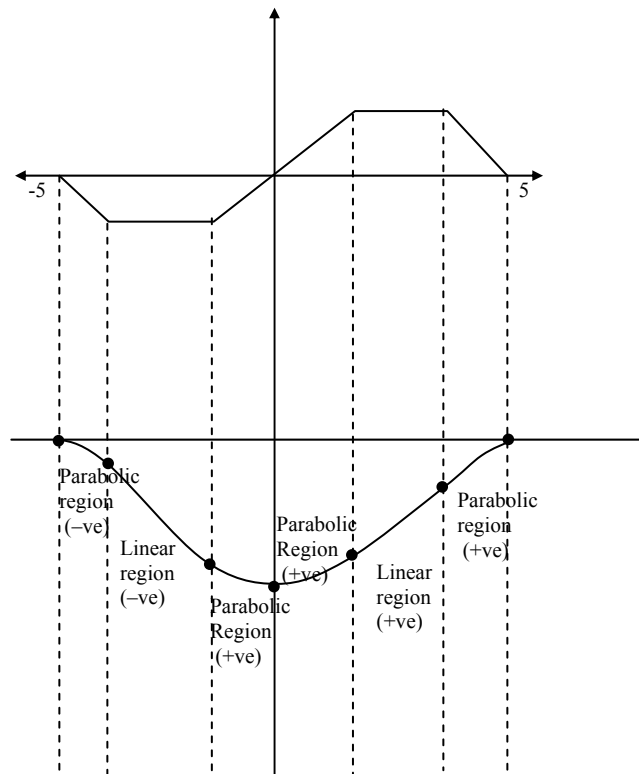


Suppose $F(x) = \int_{-5}^x f(y) dy$. which one of the following is a graph of $F(x)$?



05. Ans: (C)

Sol: integration of ramp is parabolic, integration of step is ramp.





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HEARTY CONGRATULATIONS TO OUR IES - 2015 TOPPERS

Total no. of selections in IES 2015 - EC : 52 EE : 36 CE : 24 ME : 28

01 EC LIAZ M YOUSUF	01 ME PRATAP SINGH	02 EE PARITHA SARATHI TRIPATHY	02 EC SAURABH PRATAP SINGH	02 CE PIYUSH PATHAK	03 EE NIKKI BANSAL	03 EC BIDHARTH SABHARWAL	04 EC PIYUSH VIJAY
04 EE KAJA NACA SAI HEMANTH	04 CE AMIT SHARMA	05 EE NAGENDRA TIWARI	05 CE DHIRAJ AGARWAL	05 EC MANAS PANDA	06 EE ANAS FERDZ	06 EC SIMON SAMUEL	07 EC PIYUSH PRASHANKAR KUMBHAR
07 EE AMAL SEBASTIAN	08 ME BANDI SREENIHAR	08 EE DHARMINI SACHIN	09 ME K. KRISHNA CHAITANYA	09 EC SHRUTI KUSHWAHA	09 EE SUDHAKAR KUMAR	10 EE VISHAL RATHI	10 CE AISHWARYA ALOK

EC (52 selections): [Grid of 52 portraits]

ME (28 selections): [Grid of 28 portraits]

EE (36 selections): [Grid of 36 portraits]

CE (24 selections): [Grid of 24 portraits]

24 SELECTIONS IN TOP 10

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06. Which one of the following is an eigen function of the class of all continuous-time, linear, time-invariant systems ($u(t)$ denotes the unit-step function) ?

- (A) $e^{j\omega_0 t} u(t)$ (B) $\cos(\omega_0 t)$ (C) $e^{j\omega_0 t}$ (D) $\sin(\omega_0 t)$

06. Ans: (C)

Sol: If the input to a system is its eigen signal, the response has the same form as the eigen signal



07. A continuous –time function $x(t)$ is periodic with period T . The function is sampled uniformly with a sampling period T_s . In which one of the following cases is the sampled signal periodic ?

- (A) $T = \sqrt{2} T_s$ (B) $T = 1.2T_s$ (C) Always (D) Never

07. Ans: (B)

Sol: A discrete time signal $x(n) = \cos(\omega_0 n)$ is said to be periodic if $\frac{\omega_0}{2\pi}$ is a rational number.

08. Consider the sequence $x[n] = a^n u[n] + b^n u[n]$, where $u[n]$ denotes the unit-step sequence and $0 < |a| < |b| < 1$. The region of convergence (ROC) of the z-transform of $x[n]$ is

- (A) $|z| > |a|$ (B) $|z| > |b|$ (C) $|z| < |a|$ (D) $|a| < |z| < |b|$

08. Ans: (B)

Sol: $x(n) = (a)^n x(n) + (b)^n x(n)$, given $0 < |a| < |b| < 1$

$$\text{Roc} = (|z| > |a|) \cap (|z| > |b|) = |z| > |b|$$

09. Consider a two-port network with the transmission matrix : $T = \begin{pmatrix} A & B \\ C & D \end{pmatrix}$. If the network is reciprocal, then

- (A) $T^{-1} = T$ (B) $T^2 = T$ (C) Determinant (T) = 0 (D) Determinant(T) = 1

09. Ans: (D)

Sol: A two port network is reciprocal in transmission parameters if $AD - BC = 1$
i.e Determinant(T) = 1

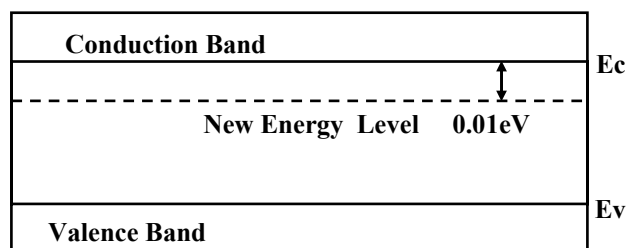
10. A continuous-time sinusoid of frequency 33 Hz is multiplied with a periodic Dirac impulse train of frequency 46 Hz. The resulting signal is passed through an ideal analog low-pass filter with a cutoff frequency of 23Hz. The functional frequency (in Hz) of the output is ____.

10. Ans: 13

Sol: $f_m = 33\text{Hz}$, $f_s = 46\text{Hz}$

The frequency in sampled signal are $= \pm 33, 13, 79, 59, 125 \dots$. The above frequencies are passed to a LPF of cutoff frequency 23Hz. The output frequency are $= 13\text{Hz}$.

11. A small percentage of impurity is added to intrinsic semiconductor at 300 K. Which one of the following statements is true for the energy band diagram shown in the following figure?





- (A) Intrinsic semiconductor doped with pentavalent atoms to form n-types semiconductor
- (B) Intrinsic semiconductor doped with trivalent atoms to form n-types semiconductor
- (C) Intrinsic semiconductor doped with pentavalent atoms to form p-types semiconductor
- (D) Intrinsic semiconductor doped with trivalent atoms to form p-type semiconductor

11. Ans: (A)

Sol: Donor energy level close to conduction band.

12. Consider the following statements for a metal oxide semiconductor field effect transistor (MOSFET):

P : As channel length reduces, OFF-state current increases

Q : As channel length reduces, output resistance increases

R : As channel length reduces, threshold voltage remains constant

S: As channel reduces, ON current increases.

Which of the above statements are INCORRECT?

- (A) P and Q (B) P and S (C) Q and R (D) R and S

12. Ans: (C)

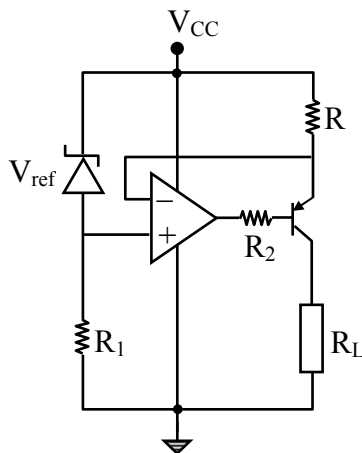
Sol: P: TRUE

Q: FALSE, As channel length reduces, output resistance reduces

R: FALSE: As channel length reduces, threshold voltage reduces

S: TRUE

13. Consider the constant current source shown in the figure below. Let β represent the current gain of the transistor



The load current I_0 through R_L is

(A) $I_0 = \left(\frac{\beta + 1}{\beta} \right) \frac{V_{ref}}{R}$

(B) $I_0 = \left(\frac{\beta}{\beta + 1} \right) \frac{V_{ref}}{R}$

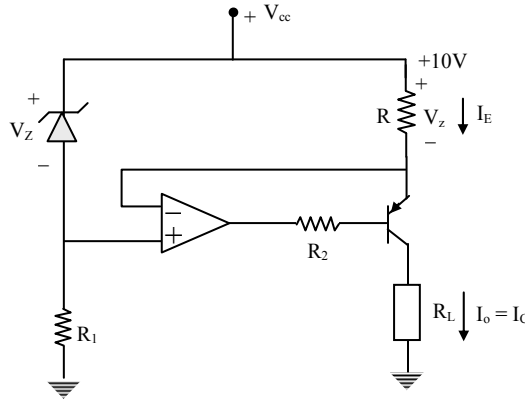
(C) $I_0 = \left(\frac{\beta + 1}{\beta} \right) \frac{V_{ref}}{2R}$

(D) $I_0 = \left(\frac{\beta}{\beta + 1} \right) \frac{V_{ref}}{2R}$



13. Ans: (B)

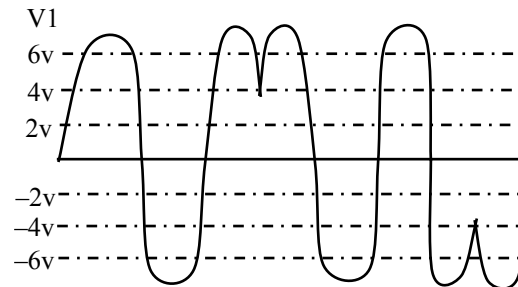
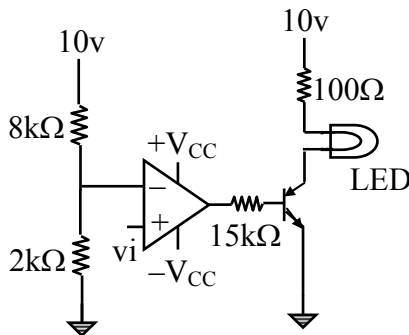
Sol:



$$V_P = V_N \text{ [Virtual short]}$$

$$I_0 = I_C = \left(\frac{\beta}{\beta + 1} \right) I_E = \left(\frac{\beta}{\beta + 1} \right) \frac{V_Z}{R}$$

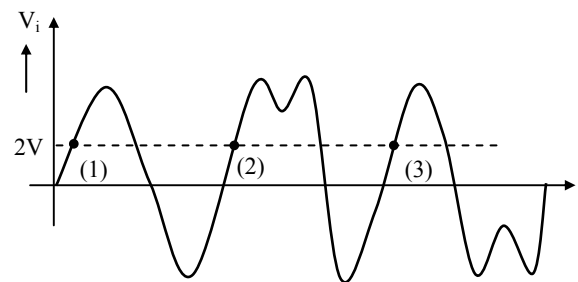
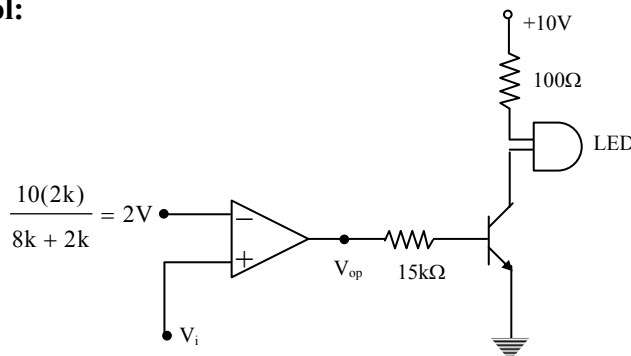
14. The following signal V_i of peak voltage 8V is applied to the non-inverting terminal of an ideal Opamp. The transistor has $V_{BE} = 0.7 \text{ V}$, $\beta = 100$; $V_{LED} = 1.5 \text{ V}$, $V_{CC} = 10 \text{ V}$ and $-V_{CC} = -10 \text{ V}$.



The number of times the LED glows is ———

14. Ans: 3

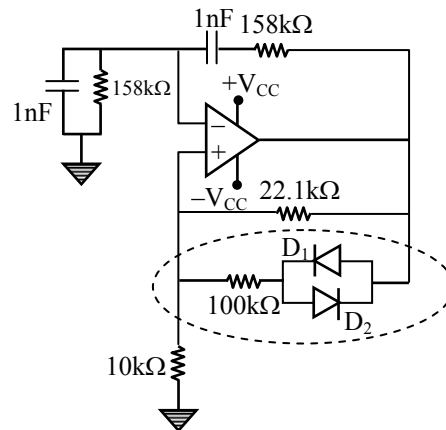
Sol:



V_i crosses 2V, 3times
Therefore the LED glows 3 times



15. Consider the oscillator circuit shown in the figure. The function of the network (shown in dotted lines) consisting of the $100\text{ k}\Omega$ resistor in series with the two diodes connected back-to-back is to:

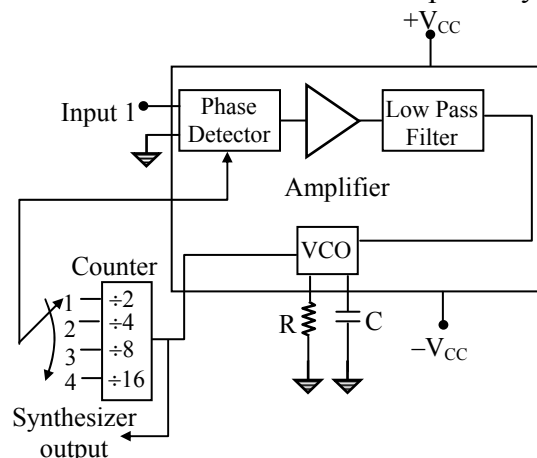


- (A) introduce amplitude stabilization by preventing the op-amp from saturating and thus producing sinusoidal oscillations of fixed amplitude
- (B) introduce amplitude stabilization by forcing the op-amp to swing between positive and negative saturation and thus producing square wave oscillations of fixed amplitude
- (C) introduce frequency stabilization by forcing the circuit to oscillate at a single frequency
- (D) enable the loop gain to take on a value that produces square wave oscillations

15. Ans: (A)

Sol: The circuit shown is a wein bridge oscillator. The amplitude of oscillations can be determined and stabilized by using a nonlinear control network. As the oscillations grow, the diodes start to conduct causing the effective resistance in the feedback to decrease. Equilibrium will be reached at the output amplitude that causes the loop gain to be exactly unity.

16. The block diagram of a frequency synthesizer consisting of Phase Locked Loop (PLL) and a divide-by-N counter (comprising $\div 2$, $\div 4$, $\div 8$, $\div 16$ outputs) is sketched below. The synthesizer is excited with a 5 kHz signal (Input 1). The free-running frequency of the PLL is set to 20 kHz . Assume that the commutator switch makes contacts repeatedly in the order 1-2-3-4.



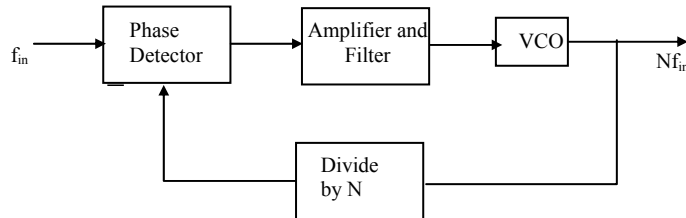
The corresponding frequencies synthesized are:



- (A) 10kHz, 20kHz, 40kHz, 80 kHz
(B) 20kHz, 40kHz, 80kHz, 160 kHz
(C) 80kHz, 40kHz, 20kHz, 10kHz
(D) 160kHz, 80kHz, 40kHz, 20kHz

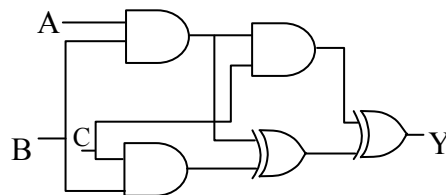
16. Ans: (A)

Sol:



f_{in}	Divide by N	VCO output(Nf_{in})
5kHz	2	10kHz
5kHz	4	20kHz
5kHz	8	40kHz
5kHz	16	80kHz

17. The output of the combinational circuit given below is



- (A) $A+B+C$ (B) $A(B+C)$ (C) $B(C+A)$ (D) $C(A+B)$

17. Ans: (C)

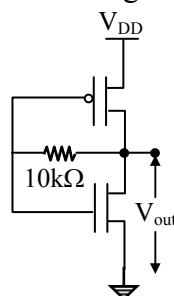
Sol: $Y = ABC \oplus AB \oplus BC = AB(C \oplus 1) \oplus BC$

$$= AB\bar{C} \oplus BC = B(A\bar{C} \oplus C) = B[\overline{A\bar{C}}C + A\bar{C} \cdot \bar{C}]$$

$$= B[(\bar{A} + C)C + A\bar{C}] = B[\bar{A}C + C + A\bar{C}]$$

$$= B[C + A\bar{C}] = B[C + A] \rightarrow Y = B(A + C)$$

18. What is the voltage V_{out} in the following circuit?



- (A) 0V (B) $(|V_T \text{ of PMOS}| + V_T \text{ of NMOS})/2$
(C) Switching threshold of inverter (D) V_{DD}

18. Ans: (D)



19. Match the inferences X, Y, and Z, about a system, to the corresponding properties of the elements of first column in Routh's Table of the system characteristic equation.

X: The system is stable

P: ... When all elements are positive

Y: The system is unstable ...

Q: ... When any one element is zero

Z: The test breaks down

R: ... When there is a change in sign of coefficients

(A) $X \rightarrow P, Y \rightarrow Q, Z \rightarrow R$

(B) $X \rightarrow Q, Y \rightarrow P, Z \rightarrow R$

(C) $X \rightarrow R, Y \rightarrow Q, Z \rightarrow P$

(D) $X \rightarrow P, Y \rightarrow R, Z \rightarrow Q$

19. Ans: (D)

Sol:

P	+	Stable
	+	
	+	
Q	+	Routh array breaks
	0	
	+	
R	+	Un stable
	-	
	+	

X - P, Y - R, Z - Q

20. A closed-loop control system is stable if the Nyquist plot of the corresponding open-loop transfer function

(A) encircles the s-plane point $(-1 + j0)$ in the counterclockwise direction as many times as the number of right-half s-plane poles.

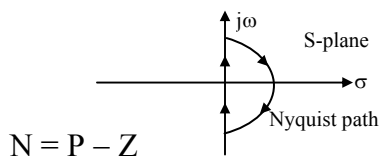
(B) encircles the s-plane point $(0 - j1)$ in the clockwise direction as many times as the number of right-half s-plane poles.

(C) encircles the s-plane point $(-1 + j0)$ in the counterclockwise direction as many times as the number of left-half s-plane poles.

(D) encircles the s-plane point $(-1 + j0)$ in the counterclockwise direction as many times as the number of right-half s-plane zeros.

20. Ans: (A)

Sol:



$$N = P - Z$$

For closed loop stability $Z = 0, N = P$

$\therefore (-1, j0)$ should be encircled in Counter clock wise direction equaling P poles in RHP.

21. Consider binary data transmission at a rate of 56 kbps using baseband binary pulse amplitude modulation (PAM) that is designed to have a raised-cosine spectrum. The transmission bandwidth (in kHz) required for a roll-off factor of 0.25 is _____.

21. Ans: 35

Sol: $R_b = 56$ kbps, $\alpha = 0.25$

$$BW = \frac{R_b}{2} [1 + \alpha] = \frac{56}{2} [1 + 0.25] \text{ kHz} = 35 \text{ kHz}$$



22. A superheterodyne receiver operates in the frequency range of 58 MHz-68MHz. The intermediate frequency f_{IF} and local oscillator frequency f_{LO} are chosen such that $f_{IF} \leq f_{LO}$. It is required that the image frequencies fall outside the 59 MHz – 68 MHz band. The minimum required f_{IF} (in MHz) is _____.

22. Ans: 5

Sol: $f_s = 58 \text{ MHz} - 68 \text{ MHz}$

When $f_s = 58 \text{ MHz}$

$$f_{si} = f_s + 2f_{IF} > 68 \text{ MHz}$$

$$2f_{IF} > 10 \text{ MHz}$$

$$f_{IF} \geq 5 \text{ MHz}$$

23. The amplitude of a sinusoidal carrier is modulated by a single sinusoid to obtain the amplitude modulated signal $s(t) = 5 \cos 1600\pi t + 20 \cos 1800\pi t + 5 \cos 2000\pi t$. The value of the modulation index is _____.

23. Ans: $\mu = 0.5$

Sol: $S(f) = 5 \cos 1600\pi t + 20 \cos 1800\pi t + 5 \cos 2200\pi t$

$$S(f) = \frac{A_c \mu}{2} \cos 2\pi(f_c - f_m)t + A_c \cos 2\pi f_c t + \frac{A_c \mu}{2} \cos 2\pi(f_c + f_m)t$$

$$A_c = 20$$

$$A_c \mu = 10$$

$$\frac{A_c \mu}{2} = 5$$

$$\mu = \frac{10}{20} = 0.5$$

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24. Concentric spherical shells of radii 2m, 4m, and 8m carry uniform surface charge densities of 20 nC/m^2 , -4 nC/m^2 and ρ_s , respectively. The value of ρ_s (nC/m^2) required to ensure that the electric flux density $\vec{D} = \vec{0}$ at radius 10 m is ———.

24. Ans: - 0.25

Sol: Given:

$$\rho_{s1} = 20 \text{ nC/m}^2$$

$$\rho_{s1} = -4 \text{ nC/m}^2$$

$$\rho_{s3} = ? \text{ (unknown)}$$

Electric flux density at $r = 10 \text{ m}$

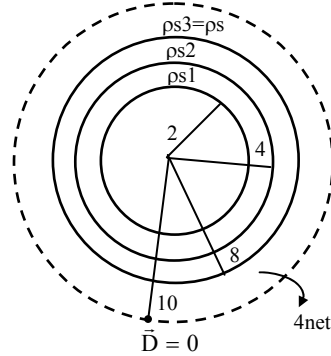
is given by

$$\vec{D} = \left(\frac{\psi_{\text{net leaving the sphere of radius } r = 10\text{m}}}{\text{Area of sphere of radius } r = 10\text{m}} \right) \hat{a}_r$$

but $\vec{D} = 0$

$$\psi_{\text{net}} \text{ at } r = 10 = 0 \quad 20 \times 10^{-9} \times 4\pi(2)^2 + (-4 \times 10^{-9}) \times 4\pi(4)^2 + \rho_{s3} 4\pi(8)^2 = 0$$

$$\therefore \rho_{s3} = \rho_s = -0.25 \text{ nC/m}^2$$



25. The propagation constant of a lossy transmission line is $(2 + j5) \text{ m}^{-1}$ and its characteristic impedance is $(50 + j0) \Omega$ at $\omega = 10^6 \text{ rad S}^{-1}$. The values of the line constants L, C, R, G are, respectively,

(A) $L = 200 \mu\text{H/m}$, $C = 0.1 \mu\text{F/m}$, $R = 50 \Omega/\text{m}$, $G = 0.02 \text{ S/m}$

(B) $L = 250 \mu\text{H/m}$, $C = 0.1 \mu\text{F/m}$, $R = 100 \Omega/\text{m}$, $G = 0.04 \text{ S/m}$

(C) $L = 200 \mu\text{H/m}$, $C = 0.2 \mu\text{F/m}$, $R = 100 \Omega/\text{m}$, $G = 0.02 \text{ S/m}$

(D) $L = 250 \mu\text{H/m}$, $C = 0.2 \mu\text{F/m}$, $R = 50 \Omega/\text{m}$, $G = 0.04 \text{ S/m}$

25. Ans: (B)

Sol: Given

Propagation constant, $P = (2 + j5) \text{ m}^{-1}$,

characteristic impedance $z_0 = 50 \Omega$,

angular frequency $\omega = 10^6 \text{ rad/sec}$,

$$P = \sqrt{(R + j\omega L)(G + j\omega C)}$$

$$z_0 = \sqrt{\frac{(R + j\omega L)}{(G + j\omega C)}}$$

$$Pz_0 = R + j\omega L$$

$$\Rightarrow R + j\omega L = (100 + j250)$$

$$\therefore R = 100 \Omega/\text{m}$$

$$L = \frac{250}{10^6} = 250 \mu\text{H/m}$$

$$\frac{P}{z_0} = G + j\omega C$$

$$G + j\omega C = \left(\frac{2}{50} + j\frac{5}{50} \right)$$



$$\therefore G = 0.04 \text{ s/m}$$

$$C = \frac{5}{50 \times 10^6} = 0.1 \mu\text{F/m}$$

Therefore line constants L, C, R & G are respectively L = 250 $\mu\text{H/m}$, C = 0.1 $\mu\text{F/m}$, R = 100 Ω/m , G = 0.04 s/m

Q.26 – Q.55 carry two marks each.

26. The integral $\frac{1}{2\pi} \iint_D (x + y + 10) dx dy$, where D denotes the disc: $x^2 + y^2 \leq 4$, evaluates to ____.

26. Ans: 20

Sol: Converting to polar coordinates, we get

$$\begin{aligned} \frac{1}{2\pi} \iint_D (x + y + 10) dx dy &= \frac{1}{2\pi} \int_{r=0}^2 \int_{\theta=0}^{2\pi} (r \cos \theta + r \sin \theta + 10) r dr d\theta \\ &= \frac{1}{2\pi} \int_{r=0}^2 \int_{\theta=0}^{2\pi} (r^2 \cos \theta + r^2 \sin \theta + 10r) dr d\theta \\ &= \frac{1}{2\pi} \int_{\theta=0}^{2\pi} \left\{ \frac{r^3}{3} \cos \theta + \frac{r^3}{3} \sin \theta + 5r^2 \right\}_0^2 d\theta \\ &= \frac{1}{2\pi} \int_0^{2\pi} \left\{ \frac{8}{3} \cos \theta + \frac{8}{3} \sin \theta + 20 \right\} d\theta \\ &= \frac{1}{2\pi} \left\{ \frac{8}{3} \sin \theta - \frac{8}{3} \cos \theta + 20\theta \right\}_0^{2\pi} \\ &= \frac{1}{2\pi} \left\{ \left(-\frac{8}{3} + 40\pi \right) - \left(-\frac{8}{3} \right) \right\} \\ &= 20 \end{aligned}$$

27. A sequence $x[n]$ is specified as

$$\begin{bmatrix} x[n] \\ x[n-1] \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}^n \begin{bmatrix} 1 \\ 0 \end{bmatrix}, \text{ for } n \geq 2.$$

The initial conditions are $x[0] = 1$, $x[1] = 1$ and $x[n] = 0$ for $n < 0$. The value of $x[12]$ is ____

27. Ans: 233

Sol:
$$\begin{bmatrix} x(n) \\ x(n-1) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}^n \begin{bmatrix} 1 \\ 0 \end{bmatrix}, n \geq 2$$

$$n = 2$$

$$\begin{bmatrix} x(2) \\ x(1) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}^2 \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 2 \\ 1 \end{bmatrix}$$



$$x(2) = 2, x(1) = 1$$

$$n = 3$$

$$\begin{bmatrix} x(3) \\ x(2) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}^3 \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 3 & 2 \\ 2 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 3 \\ 2 \end{bmatrix}$$

$$x(3) = 3, x(2) = 2$$

From the above values we can write the recursive relation as

$$x(n) = x(n-1) + x(n-2)$$

$$x(2) = x(1) + x(0) = 1 + 1 = 2$$

$$x(3) = x(2) + x(1) = 2 + 1 = 3$$

$$x(4) = x(3) + x(2) = 3 + 2 = 5$$

$$x(5) = x(4) + x(3) = 5 + 3 = 8$$

$$x(6) = x(5) + x(4) = 8 + 5 = 13$$

$$x(7) = x(6) + x(5) = 13 + 8 = 21$$

$$x(8) = x(7) + x(6) = 21 + 13 = 34$$

$$x(9) = x(8) + x(7) = 34 + 21 = 55$$

$$x(10) = x(9) + x(8) = 55 + 34 = 89$$

$$x(11) = 89 + 55 = 144$$

$$x(12) = 144 + 89 = 233$$

28. In the following integral, the contour C encloses the points $2\pi j$ and $-2\pi j$. The value of the integral

$$-\frac{1}{2\pi} \oint_C \frac{\sin z}{(z - 2\pi j)^3} dz \text{ is } \text{-----}.$$

28. Ans: - 133.87

Sol: $-\frac{1}{2\pi} \oint_C \frac{\sin z}{(z - 2\pi j)^3} dz = -\frac{1}{2\pi} \times 2\pi j \frac{f''(2\pi j)}{2!}$

$$f(z) = \sin z$$

$$f''(z) = -\sin z$$

$$\therefore f''(z_0) = -\sin 2\pi j$$

$$\frac{1}{2\pi} \oint_C \frac{\sin z}{(z - 2\pi j)^3} dz = -\frac{1}{2\pi} \times 2\pi j \left(\frac{-\sin(2\pi j)}{2} \right)$$

$$= j \times j \frac{\sinh 2\pi}{2}$$

$$= -\frac{1}{2} (\sinh 2\pi)$$

$$= -133.87$$



29. The region specified by $\{(\rho, \phi, Z) : 3 \leq \rho \leq 5, \frac{\pi}{8} \leq \phi \leq \frac{\pi}{4}, 3 \leq z \leq 4.5\}$ in cylindrical coordinates has volume of _____.

29. Ans: 4.714

Sol: Given region of cylinder

$$3 \leq \rho \leq 5,$$

$$\frac{\pi}{8} \leq \phi \leq \frac{\pi}{4},$$

$$3 \leq z \leq 4.5$$

The differential volume of cylinder is given by

$$dv = \rho d\rho d\phi dz$$

$$\text{Volume, } v = \int_{\rho=3}^5 \int_{\phi=\frac{\pi}{8}}^{\frac{\pi}{4}} \int_{z=3}^{4.5} \rho d\rho d\phi dz$$

$$= \frac{\rho^2}{2} \Big|_3^5 \times \phi \Big|_{\frac{\pi}{8}}^{\frac{\pi}{4}} \times z \Big|_3^{4.5} = \frac{1}{2} (25 - 9) \times \left(\frac{\pi}{4} - \frac{\pi}{8} \right) \times (4.5 - 3)$$

$$\therefore v = 4.71 \text{ m}^3$$

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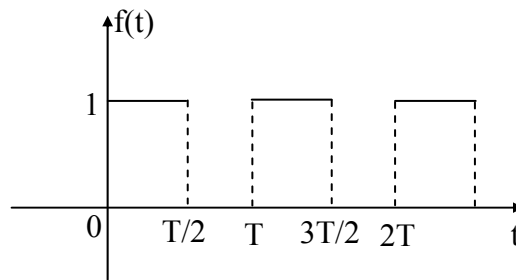
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30. The Laplace transform of the causal periodic square wave of period T shown in the figure below is



(A) $F(s) = \frac{1}{1 + e^{-sT/2}}$

(B) $F(s) = \frac{1}{s \left(1 + e^{-\frac{sT}{2}} \right)}$

(C) $F(s) = \frac{1}{s(1 - e^{-sT})}$

(D) $F(s) = \frac{1}{1 - e^{-sT}}$

30. Ans: (B)

Sol: One period of signal $x_1(t) = u(t) - u(t-T/2)$

$$X_1(s) = \frac{1}{s} - \frac{e^{-sT/2}}{s} = \frac{1 - e^{-sT/2}}{s}$$

$$X(s) = \frac{1}{1 - e^{-sT}} X_1(s) = \frac{1 - e^{-sT/2}}{s(1 - e^{-sT})} = \frac{1}{s(1 + e^{-sT/2})}$$

31. A network consisting of a finite number of linear resistor (R), inductor (L), and capacitor (C) elements, connected all in series or all in parallel, is excited with a source of the form

$\sum_{k=1}^3 a_k \cos(k\omega_0 t)$, where $a_k \neq 0$, $\omega_0 \neq 0$. The source has nonzero impedance. Which one of the following is a possible form of the output measured across a resistor in the network?

(A) $\sum_{k=1}^3 b_k \cos(k\omega_0 t + \phi_k)$, where $b_k \neq a_k, \forall k$

(B) $\sum_{k=1}^4 b_k \cos(k\omega_0 t + \phi_k)$, where $b_k \neq 0, \forall k$

(C) $\sum_{k=1}^3 a_k \cos(k\omega_0 t + \phi_k)$

(D) $\sum_{k=1}^3 a_k \cos(k\omega_0 t + \phi_k)$

31. Ans: (A)

Sol: Consider a series RLC-Circuit with voltage source

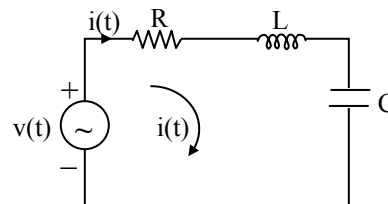
Here

$$V(t) = a_1 \cos \omega_0 t + a_2 \cos 2\omega_0 t + a_3 \cos 3\omega_0 t$$

$$i(t) = b_1 \cos(\omega_0 t + \phi_1) + b_2 \cos(2\omega_0 t + \phi_2) + b_3 \cos(3\omega_0 t + \phi_3)$$

$$i(t) = \sum_{k=1}^3 b_k \cos(k\omega_0 t + \phi_k)$$

Where $b_k \neq a_k$ for all k





32. A first-order low-pass filter of time constant T is excited with different input signals (with zero initial conditions up to $t = 0$). Match the excitation signals X, Y, Z with the corresponding time responses for $t \geq 0$:

X: Impulse

P: $1 - e^{-t/T}$

Y: Unit step

Q: $t - T(1 - e^{-t/T})$

Z: Ramp

R: $e^{-t/T}$

(A) $X \rightarrow R, Y \rightarrow Q, Z \rightarrow P$

(B) $X \rightarrow Q, Y \rightarrow P, Z \rightarrow R$

(C) $X \rightarrow R, Y \rightarrow P, Z \rightarrow Q$

(D) $X \rightarrow P, Y \rightarrow R, Z \rightarrow Q$

32. Ans: (C)

Sol: $H(s) = \frac{1}{1 + s\tau}$

$V_0(s) = H(s) \cdot V_i(s)$

(I) if $v_i(t) = \delta(t)$

$V_i(s) = 1$

$V_0(s) = H(s) \cdot V_i(s)$

$= \frac{1}{1 + s\tau}$

$v_0(t) = \frac{1}{\tau} e^{-\frac{t}{\tau}}$

(II) if $v_i(t) = u(t)$

$V_i(s) = 1/s$

$V_0(s) = \frac{1}{s(1 + s\tau)} = \frac{1}{s} - \frac{1}{s + \frac{1}{\tau}}$

$v_0(t) = (1 - e^{-t/\tau})$

(III) $v_i(t) = r(t) \Rightarrow V_i(s) = \frac{1}{s^2}$

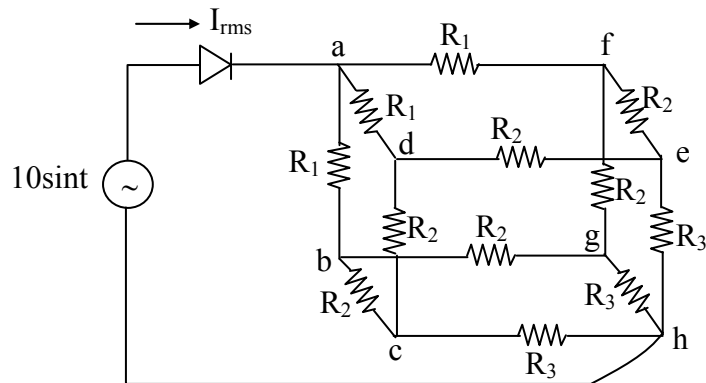
$V_0(s) = H(s) \cdot V_i(s) = \frac{1}{s^2(1 + s\tau)}$

$= \frac{1}{s^2} - \frac{\tau}{s} + \frac{\tau}{s + \frac{1}{\tau}}$

$V_0(t) = t - \tau(1 - e^{-t/\tau})$



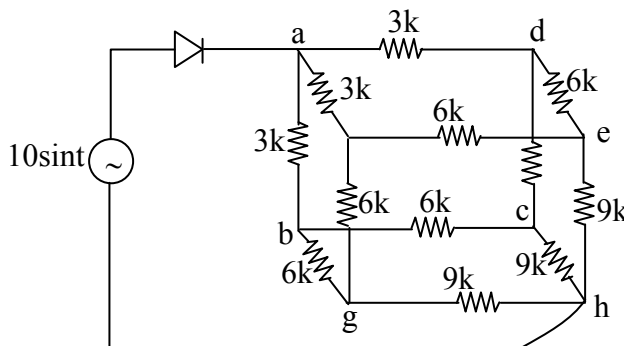
33. An AC voltage source $V = 10 \sin(t)$ volts is applied to the following network. Assume that $R_1 = 3 \text{ k}\Omega$, $R_2 = 6 \text{ k}\Omega$ and $R_3 = 9 \text{ k}\Omega$, and that the diode is ideal.



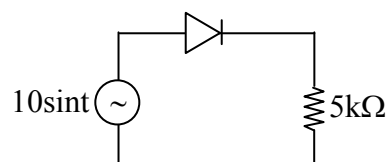
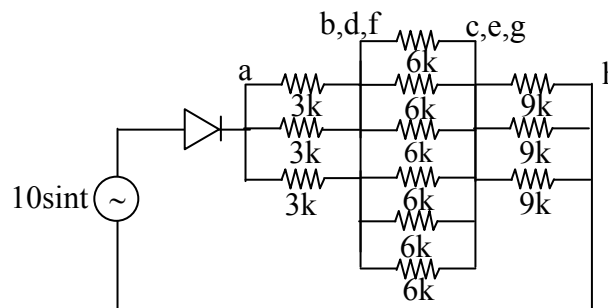
Rms current I_{rms} (in mA) through the diode is _____

33. Ans: 1

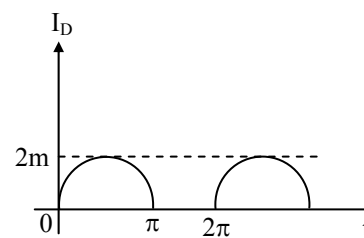
Sol:



We can join nodes that are at same potential so network becomes

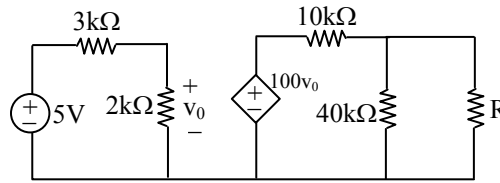


$$I_{D(\text{RMS})} = \frac{2 \text{ m}}{2} = 1 \text{ mA}$$





34. In the circuit shown in the figure, the maximum power (in watt) delivered to the resistor R is ____.



34. Ans: 0.8

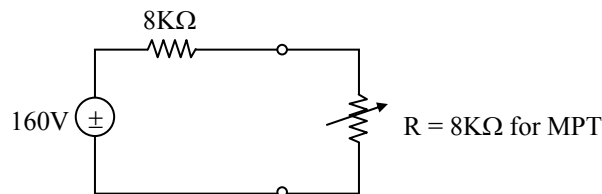
Sol: $V_0 = \left(\frac{5}{5k}\right) \cdot 2k = 2v$

$$V_{oc} = V_{th} = V_{40k} = i_{40k} \cdot 40k = \left(\frac{200}{50k}\right) \cdot 40k = 160V$$

$$\Rightarrow I_{sc} = \frac{200}{10k} = 20mA$$

$$\text{So, } R_{th} = \frac{V_{th}}{I_{sc}} = \frac{160v}{20mA} = 8k\Omega$$

$$P_{max} = \frac{V_s^2}{4R_L} = \frac{(160)^2}{4 \times 8k} = 0.8W$$



35. Consider the signal $x[n] = 6\delta[n+2] + 3\delta[n+1] + 8\delta[n] + 7\delta[n-1] + 4\delta[n-2]$.
If $X(e^{j\omega})$ is the discrete-time Fourier transform of $x[n]$,

then $\frac{1}{\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \sin^2(2\omega) d\omega$ is equal to ____.

35. Ans: 8

Sol: Plancherel's relation is

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y(e^{j\omega}) d\omega = \sum_{n=-\infty}^{\infty} x(n)y(n)$$

$$Y(e^{j\omega}) = \sin^2(2\omega) = \frac{1 - \cos(4\omega)}{2}$$

$$= \frac{1}{2} - \frac{1}{4} e^{j4\omega} - \frac{1}{4} e^{-j4\omega}$$

$$y(n) = \frac{1}{2} \delta(n) - \frac{1}{4} \delta(n+4) - \frac{1}{4} \delta(n-4)$$

$$y(n) = \left\{ -\frac{1}{4}, 0, 0, 0, \frac{1}{2}, 0, 0, 0, -\frac{1}{4} \right\}$$

$$x(n) = \{6, 3, 8, 7, 4\}$$

$$\frac{1}{\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y(e^{j\omega}) d\omega = 2 \sum_{n=-\infty}^{\infty} x(n)y(n)$$

$$2 \sum_{n=-\infty}^{\infty} x(n)y(n) = 2 \times 8 \times \frac{1}{2} = 8$$



36. Consider a silicon p-n junction with a uniform acceptor doping concentration of 10^{17} cm^{-3} on the p-side and a uniform donor doping concentration of 10^{16} cm^{-3} on the n-side. No external voltage is applied to the diode. Given: $kT/q = 26 \text{ mV}$, $n_i = 1.5 \times 10^{10} \text{ cm}^{-3}$, $\epsilon_{\text{si}} = 12 \epsilon_0$, $\epsilon_0 = 8.85 \times 10^{-14} \text{ F/m}$, and $q = 1.6 \times 10^{-19} \text{ C}$.

The charge per unit junction area (nC cm^{-2}) in the depletion region on the p-side is ———.

36. Ans: 4.836

Sol: $\epsilon = 12\epsilon_0$

$$= 12 \times 8.85 \times 10^{-14} \text{ F/m}$$

$$N_D = 10^{16} \text{ cm}^{-3}$$

$$= 10^{22} \text{ m}^{-3}$$

$$N_A = 10^{17} \text{ cm}^{-3}$$

$$= 10^{23} \text{ m}^{-3}$$

$$V_0 = \frac{kT}{q} \ln \left[\frac{N_A N_D}{n_i^2} \right] = 0.026 \ln \left[\frac{10^{23} \times 10^{22}}{(1.5 \times 10^{10})^2} \right] = 0.757 \text{ V}$$

$$W = \sqrt{\frac{2\epsilon}{q} V_0 \left(\frac{1}{N_A} + \frac{1}{N_D} \right)} = \sqrt{\frac{2 \times 12 \times 8.85 \times 10^{-14}}{1.6 \times 10^{-19}} \times 0.757 \left(\frac{1}{10^{23}} + \frac{1}{10^{22}} \right)}$$

$$= 3.325 \times 10^{-8} \text{ m}$$

$$= 3.325 \times 10^{-6} \text{ cm}$$

$$W_p = \frac{N_D}{N_A + N_D} \omega = \frac{10^{22}}{10^{22} + 10^{23}} \times 3.325 \times 10^{-8} = 3.023 \times 10^{-9} \text{ m}$$

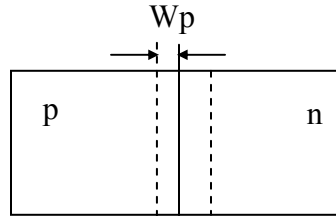
$$= 3.023 \times 10^{-7} \text{ cm}$$

$$Q = W_p N_A e A$$

$$\Rightarrow \frac{Q}{A} = W_p N_A e = 3.023 \times 10^{-7} \times 10^{17} \times 1.6 \times 10^{-19}$$

$$= 4.836 \times 10^{-9} \text{ cm}^{-2}$$

$$= 4.836 \text{ nc-cm}^{-2}$$



37. Consider an n-channel metal oxide semiconductor field effect transistor (MOSFET) with a gate-to-source voltage of 1.8 V . Assume that $\frac{W}{L} = 4$, $\mu_n C_{\text{ox}} = 70 \times 10^{-6} \text{ A V}^{-2}$, the threshold voltage is 0.3 V , and the channel length modulation parameter is 0.09 V^{-1} . In the saturation region, the drain conductance (in micro siemens) is ———.

37. Ans: 28.35

Sol: Drain conductance in saturation region is, $g_d = \frac{1}{r_d} = \lambda I_D$

$$I_D = \frac{1}{2} \mu_n C_{\text{ox}} \frac{W}{L} [V_{\text{gs}} - V_T]^2 = \frac{1}{2} \times 70 \times 10^{-6} \times 4 [1.8 \text{ V} - 0.3 \text{ V}]^2$$

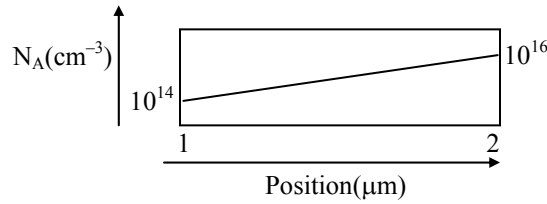
$$= 0.315 \text{ mA}$$

$$g_d = 0.09 \times 0.315 \text{ mA/V}$$

$$g_d = 28.35 \times 10^{-6} \text{ A/V} = 28.35 \mu \text{ Seimens}$$



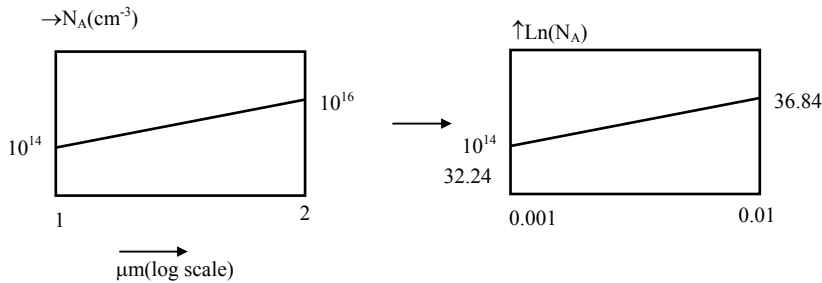
38. The figure below shows the doping distribution in a P-type semiconductor in **log scale**.



The magnitude of the electric field (in kV/cm) in the semiconductor due to non uniform doping is _____.

38. Ans: 0.0133

Sol:



$$qD_p \frac{dp}{dx} = q\mu_p p \varepsilon$$

$$\mu_p V_T \frac{dp}{dx} = \mu_p p \varepsilon$$

$$\varepsilon = \frac{V_T}{p} \frac{dp}{dx} \quad p \cong N_A$$

$$\varepsilon = \frac{V_T}{N_A} \frac{dN_A}{dx} \Rightarrow \varepsilon = V_T \frac{d}{dx} \ln[N_A(x)]$$

$$\log_{10} x_1 = 1 \mu\text{m} \Rightarrow x_1 = 10^1 \mu\text{m} = 0.001 \text{ cm}$$

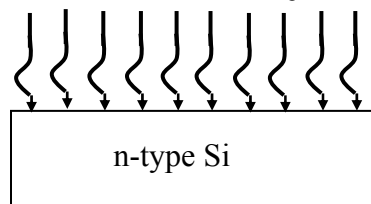
$$\log_{10} x_2 = 2 \mu\text{m} \Rightarrow x_2 = 10^2 \mu\text{m} = 0.01 \text{ cm}$$

$$\ln(10^{14}) = 32.23$$

$$\ln(10^{16}) = 36.84$$

$$\varepsilon = 0.026 \left[\frac{36.84 - 32.23}{0.01 - 0.001} \right] = 0.0133 \text{ /cm}$$

39. Consider a silicon sample at $T = 300 \text{ K}$, with a uniform donor density $N_d = 5 \times 10^{16} \text{ cm}^{-3}$ illuminated uniformly such that the optical generation rate is $G_{\text{opt}} = 1.5 \times 10^{20} \text{ cm}^{-3} \text{ s}^{-1}$ through out the sample. The incident radiation is turned off at $t = 0$. Assume low-level injection to be valid and ignore surface effects. The carrier lifetimes are $\tau_{\text{po}} = 0.1$ and $\tau_{\text{no}} = 0.5 \mu\text{s}$.





The hole concentration at $t = 0$ and the hole concentration at $t = 0.3 \mu\text{s}$, respectively, are

- (A) $1.5 \times 10^{13} \text{ cm}^{-3}$ and $7.47 \times 10^{11} \text{ cm}^{-3}$
 (B) $1.5 \times 10^{13} \text{ cm}^{-3}$ and $8.23 \times 10^{11} \text{ cm}^{-3}$
 (C) $7.5 \times 10^{13} \text{ cm}^{-3}$ and $3.73 \times 10^{11} \text{ cm}^{-3}$
 (D) $7.5 \times 10^{13} \text{ cm}^{-3}$ and $4.12 \times 10^{11} \text{ cm}^{-3}$

39. Ans: (C)

Sol: $P_n(t) = P_{n_0} + P_n(0)e^{-t/\tau_p}$

at low level injection $\Rightarrow P_{n_0}$ neglective

$$GR = \frac{P_n(0)}{\tau_{n_0}}$$

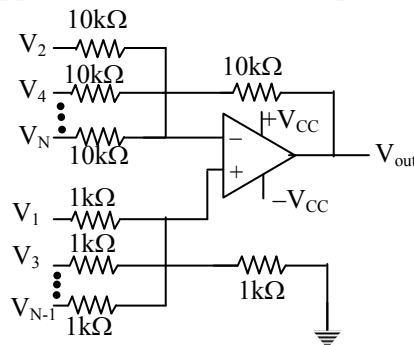
$$\Rightarrow P_{n_0}(0) = GR \times \tau_{n_0} = 1.5 \times 10^{20} \times 0.5 \times 10^{-6} = 7.5 \times 10^{13} / \text{cm}^3$$

$$\text{At } t = 0 \Rightarrow P(t) = P_n(0). e^0 = 7.5 \times 10^{13} / \text{cm}^3$$

$$\text{At } t = 0.3 \mu\text{s} \Rightarrow P(t) = P_n(0)e^{-\frac{0.3}{0.1}} = 3.73 \times 10^{11} / \text{cm}^3$$

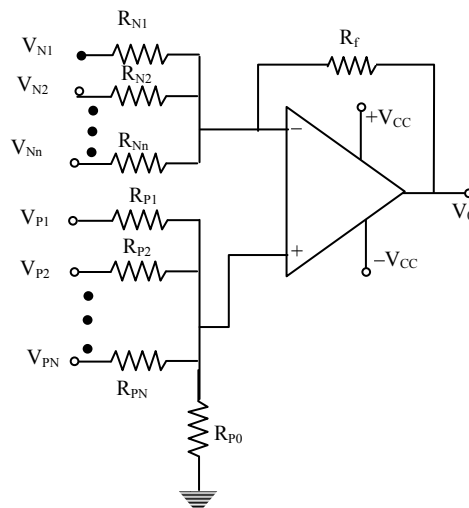
incorrect

40. An ideal opamp has voltage source $V_1, V_2, V_3, V_5, \dots, V_{N-1}$ connected to the non-inverting input and $V_2, V_4, V_6 \dots V_N$ connected to the inverting input as shown in the figure below ($+V_{CC} = 15$ volt, $-V_{CC} = -15$ volt). The voltage $V_1, V_2, V_3, V_4, V_5, V_6 \dots$ are 1, $-1/2, 1/3, -1/4, 1/5, -1/6, \dots$ volt, respectively. As N approaches infinity, the output voltage (in volt) is _____



40. Ans: $V_0 = 15$

Sol:





Using superposition it can shown that the output

$$V_0 = \left[1 + \frac{R_f}{R_N} \right] \left[\frac{R_p}{R_{p1}} V_{p1} + \frac{R_p}{R_{p2}} V_{p2} + \dots \frac{R_p}{R_{pN}} V_{pn} \right] - \left[\frac{R_f}{R_{N1}} V_{N1} + \frac{R_f}{R_{N2}} V_{N2} + \dots \frac{R_f}{R_{Nn}} V_{Nn} \right]$$

Where $R_N = R_{N1} || R_{N2} || \dots || R_{Nn}$ and $R_p = R_{p1} || R_{p2} \dots R_{pN} || R_{pO}$

In the problem given

$$R_f = R_{N1} = R_{N2} = \dots = R_{Nn} = 10k\Omega$$

$$R_{p1} = R_{p2} = R_{p3} = \dots = R_{pN} = R_{pO} = 1k\Omega$$

$$\therefore V_0 = \left[1 + \frac{10k}{\left(\frac{10k}{n} \right)} \right] \left[\frac{1k}{(1+n)} V_{p1} + \frac{\left(\frac{1k}{1+n} \right)}{1k} V_{p2} + \dots \right] - \left[\frac{10k}{10k} V_{N1} + \frac{10k}{10k} V_{N2} + \dots \right]$$

$$\therefore V_0 = (V_{p1} + V_{p2} + \dots V_{pn}) - (V_{N1} + V_{N2} + \dots V_{Nn})$$

If the series approaches ∞ then

$$V_0 = \left(1 + \frac{1}{3} + \frac{1}{5} + \frac{1}{7} \dots \right) - \left(\frac{-1}{2} - \frac{1}{4} - \frac{1}{6} \dots \right)$$

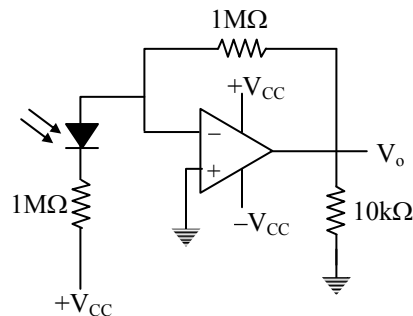
$$= 1 + \frac{1}{2} + \frac{1}{3} + \frac{1}{4} + \frac{1}{5} + \dots$$

$$= \infty$$

This series is called harmonic series which is a divergent infinite series

$$\therefore V_0 = +\infty = +V_{sat} = +V_{CC} = +15V$$

41. A p-i-n photo diode of responsivity $0.8A/W$ is connected to the inverting input of an ideal opamp as shown in the figure, $+V_{CC} = 15V$, $-V_{CC} = -15V$, Load resistor $R_L = 10k\Omega$. If $10\mu W$ of power is incident on the photodiode, then the value of the photocurrent (in μA) through the load is _____.



41. Ans: – 800 μA

Sol: The photo diode with Responsivity $0.8A/W$

$$\therefore \text{Diode current} = 0.8A/W [10\mu W]$$

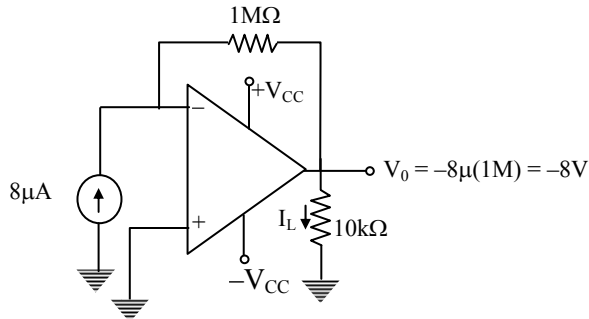
$$= 8 \times 10^{-6} A$$

$$V_0 = -8\mu (1M) = -8V$$



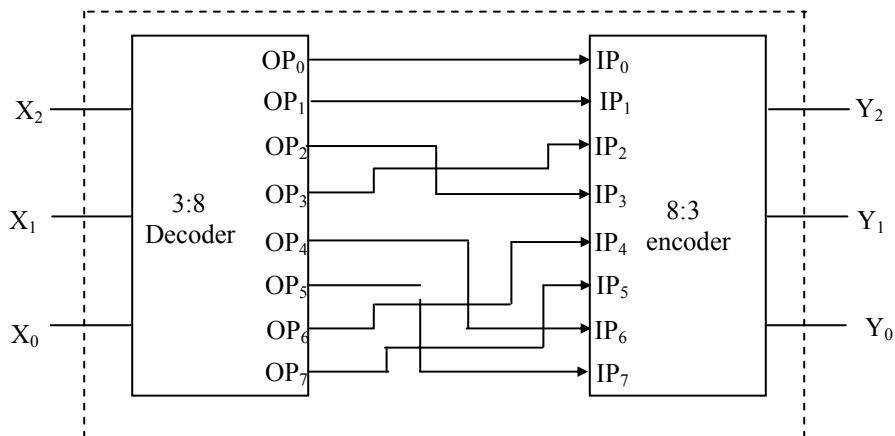
$$I_L = \frac{-8}{10k} = -8 \times 10^{-4} \text{ A} = -800 \times 10^{-6} \text{ A}$$

$$= -800 \mu\text{A}$$



Therefore the value of photo current throughput the load is $-800 \mu\text{A}$

42. Identify the circuit below.



(A) Binary to Gray code converter

(B) Binary to XS3 converter

(C) Gray to Binary converter

(D) XS3 Binary converter

42. Ans: (A) (No Answer) if considering $OP_6 \rightarrow IP_5$, $OP_7 \rightarrow IP_4$

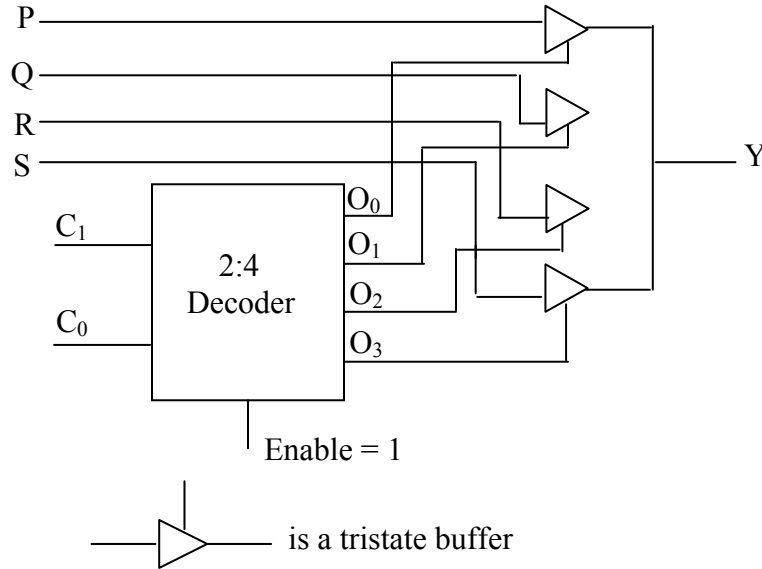
Sol:

X_2	X_1	X_0	OP_0	OP_1	OP_2	OP_3	OP_4	OP_5	OP_6	OP_7	IP_0	IP_1	IP_2	IP_3	IP_4	IP_5	IP_6	IP_7	Y_2	Y_1	Y_0
0	0	0	1	0	0	0	0	0	0	0	1	0	0	0	0	0	0	0	0	0	0
0	0	1	0	1	0	0	0	0	0	0	0	1	0	0	0	0	0	0	0	0	1
0	1	0	0	0	1	0	0	0	0	0	0	0	1	0	0	0	0	0	1	1	0
0	1	1	0	0	0	1	0	0	0	0	0	0	0	1	0	0	0	0	1	0	0
					
					
1	1	0	0	0	0	0	0	0	1	0	0	0	0	0	1	0	0	0	1	0	1
1	1	1	0	0	0	0	0	0	0	1	0	0	0	1	0	0	0	0	1	0	0

Thus it is a Binary to Gray code converter



43. The functionally implemented by the circuit below is



- (A) 2-to-1 multiplexer
(C) 7-to-1 multiplexer

- (B) 4-to-1 multiplexer
(D) 6-to-1 multiplexer

43. **Ans: (B)**

Sol:

C ₁	C ₀	Y
0	0	P
0	1	Q
1	0	R
1	1	S

Hence it is a “4 to 1 multiplexer”

44. In a 8085 system, a PUSH operation requires more clock cycles than a POP operation, which one of the following options is the correct reason for this?

- (A) For POP, the data transceivers remain in the same direction as for instruction fetch (memory to processor), whereas for PUSH their direction has to be reversed
(B) Memory write operations are slower than memory read operations in an 8085 bases system.
(C) The stack pointer needs to be pre-determined before writing registers in a PUSH, whereas a POP operation uses the address already in the stack pointer.
(D) Order of register has to be interchanged for a PUSH operation, whereas POP uses their natural order.

44. **Ans: (C)**

Sol: Push takes 12T states due to pre decrement and pop takes 10T states.

45. The open-loop transfer function of a unity-feedback control system is

$$G(S) = \frac{K}{s^2 + 5s + 5}$$

The value of K at the breakaway point of the feedback control system's root-locus plot is ____



45. Ans: 1.25

Sol: Break away point $\frac{dk}{ds} = 0$

$$\frac{d}{ds} \left(\frac{1}{s^2 + 5s + 5} \right) = 0$$

$$0 - (2s + 5) = 0$$

$s = -2.5$ is a break away point

K Value is Obtain From Magnitude Condition

$$\left| \frac{K}{s^2 + 5s + 5} \right|_{s=-2.5} = 1$$

$$\Rightarrow \left| \frac{K}{6.25 - 12.5 + 5} \right| = 1$$

$$\Rightarrow K = 1.25$$

46. The open-loop transfer function of a unity-feedback control system is given by

$$G(S) = \frac{K}{s(s+2)}$$

For the peak overshoot of the closed-loop system to a unit step input to be 10% , the value of K is

46. Ans: 2.87

Sol: Given % $M_p = 10\%$

$$M_p = 0.1$$

$$\Rightarrow M_p = e^{-\pi\xi / \sqrt{1-\xi^2}}$$

$$0.1 = e^{-\pi\xi / \sqrt{1-\xi^2}}$$

$$\Rightarrow \ln(0.1) = \frac{-\pi\xi}{\sqrt{1-\xi^2}}$$

$$\Rightarrow 2.3 = \frac{\pi\xi}{\sqrt{1-\xi^2}}$$

$$\xi = 0.59$$

$$\text{Given } G(s) = \frac{K}{s(s+2)}$$

CE:-

$$1 + G(s) = 0 \Rightarrow s^2 + 2s + K = 0$$

$$2\zeta\omega_n = 2$$

$$2 \times 0.59 \times \omega_n = 2$$

$$\omega_n = 1.69 \text{ r/sec}$$

$$K = \omega_n^2 = 2.87$$



47. The transfer function of a linear time invariant systems is given by $H(s) = 2s^4 - 5s^3 + 5s - 2$
The number of zeros in the right half of the s-plane is _____.

47. Ans: 3

Sol: TF $H(s) \Rightarrow 2s^4 - 5s^3 + 5s - 2$
RH – Criteria

①	$+S^4$	2	0	-2
②	$-S^3$	-5	+5	
	$+S^2$	2	-2	
③	$+S^1$	0(2)		
	$-S^0$	-2		

3 Sign Changes

3 Roots (Zeros) in the RH -S-Plane.

48. Consider a discrete memoryless source with alphabet $S = \{s_0, s_1, s_2, s_3, s_4, \dots\}$ and respective probabilities of occurrence $P = \left\{ \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \frac{1}{32}, \dots \right\}$. The entropy of the source (in bits) is _____.

48. Ans: 2

Sol: $H = \frac{1}{2} \log_2 + \frac{1}{4} \log_2 + \frac{1}{8} \log_2 + \frac{1}{16} \log_2 + \dots$

$$H = \frac{1}{2} \times 1 + \frac{1}{4} \times 2 + \frac{1}{8} \times 3 + \frac{1}{16} \times 4$$

$$= \sum_{n=0}^{\infty} n \left(\frac{1}{2} \right)^n = \frac{\frac{1}{2}}{\left(1 - \frac{1}{2} \right)^2} = 2$$

49. A digital communication system uses a repetition code for channel encoding/decoding. During transmission, each bit is repeated three times (0 is transmitted as 000, and 1 is transmitted as 111). It is assumed that the source puts out symbols independently and with equal probability. The decoder operates as follows: In a block of three received bits, if the number of zeros exceeds the number of ones, the decoder decides in favor of a 0, and if the number of ones exceeds the number of zeros, the decoder decides in favor of a 1. Assuming a binary symmetric channel with crossover probability $p = 0.1$, the average probability of error is _____.

49. Ans: 0.028

Sol: $P_e = P^3 + 3P^2(1-P)$

$$P = 0.1$$

$$P_e = (0.1)^3 + 3 \times (0.1)^2 (1-0.1) = 0.001 + 3 \times 0.01 \times 0.9 = 0.001 + 0.027 = 0.028$$



50. An analog pulse $s(t)$ is transmitted over an additive white Gaussian (AWGN) channel. The received signal is $r(t) = s(t) + n(t)$, where $n(t)$ is additive white Gaussian noise with power spectral density $\frac{N_0}{2}$. The received signal is passed through a filter with impulse response $h(t)$. Let E_s and E_n denote the energies of the pulse $s(t)$ and the filter $h(t)$, respectively. When the signal-to-noise ratio (SNR) is maximized at the output of the filter (SNR_{\max}), which of the following holds?

(A) $E_s = E_h; \text{SNR}_{\max} = \frac{2E_s}{N_0}$

(B) $E_s = E_h; \text{SNR}_{\max} = \frac{E_s}{2N_0}$

(C) $E_s > E_h; \text{SNR}_{\max} > \frac{2E_s}{N_0}$

(D) $E_s < E_h; \text{SNR}_{\max} = \frac{2E_h}{N_0}$

50. Ans: (A)

Sol: The impulse response of the filter is same on the signal so $E_s = E_h$

$$\text{SNR} = \frac{2E_s}{N_0}$$

$$E_s = E_h$$

$$\text{SNR} = \frac{2E_s}{N_0}$$

51. The current density in a medium is given by

$$\vec{J} = \frac{400 \sin \theta}{2\pi(r^2 + 4)} \hat{a}_r \text{ Am}^{-2}$$

The total current and the average current density flowing through the portion of a spherical surface $r = 0.8\text{m}$, $\frac{\pi}{12} \leq \theta \leq \frac{\pi}{4}$, $0 \leq \phi \leq 2\pi$ are given, respectively, by

(A) 15.09 A, 12.86 Am^{-2}

(B) 18.73 A, 13.65 Am^{-2}

(C) 12.86A, 9.23 Am^{-2}

(D) 10.28 A, 7.56 Am^{-2}

51. Ans: correct option is not given

Sol: Current density,

$$\vec{J} = \frac{400 \sin \theta}{2\pi(r^2 + 4)} \bar{a}_r \text{ A/m}^2$$

current passing through the portion of sphere of radius $r = 0.8 \text{ m}$ is given by

$$I = \int_s \vec{J} \cdot d\vec{s} \quad (r = \text{constant})$$

$$d\vec{s} = r^2 \sin \theta d\theta d\phi \hat{a}_r \quad (\because r = 0.8 \text{ m})$$

$$\begin{aligned} I &= \int_{\theta=\frac{\pi}{12}}^{\frac{\pi}{4}} \int_{\phi=0}^{2\pi} \frac{400 \sin \theta}{2\pi(r^2 + 4)} r^2 \sin \theta d\theta d\phi \\ &= \frac{400(0.8)^2}{2\pi(0.8^2 + 4)} \left[\left(\frac{\pi}{4} - \frac{\pi}{12} \right) - \left(\sin\left(\frac{\pi}{2}\right) - \sin\left(\frac{\pi}{6}\right) \right) \right] \times (2\pi) \end{aligned}$$

$$\therefore I = 7.45 \text{ Amp}$$



The average current density through the given sphere surface is

$$J = \frac{I}{\text{Area of } r = 0.8\text{m sphere}}$$

$$= \frac{7.45}{(0.8)^2 \int_{\theta=\pi/2}^{\pi/4} \int_{\phi=0}^{2\pi} \sin \theta d\theta d\phi}$$

$$= \frac{7.45}{1.04}$$

$$\therefore J = 7.15 \text{ A/m}^2$$

52. An antenna pointing in a certain direction has a noise temperature of 50K. The ambient temperature is 290K. The antenna is connected to pre-amplifier that has a noise figure of 2dB and an available gain of 40 dB over an effective bandwidth of 12 MHz. The effective input noise temperature T_e for the amplifier and the noise power P_{ao} at the output of the preamplifier, respectively, are

(A) $T_e = 169.36\text{K}$ and $P_{ao} = 3.73 \times 10^{-10} \text{ W}$

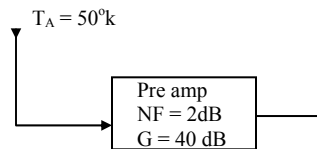
(B) $T_e = 170.8\text{K}$ and $P_{ao} = 4.56 \times 10^{-10} \text{ W}$

(C) $T_e = 182.5\text{K}$ and $P_{ao} = 3.85 \times 10^{-10} \text{ W}$

(D) $T_e = 160.62\text{K}$ and $P_{ao} = 4.6 \times 10^{-10} \text{ W}$

52. Ans: (A)

Sol:



$$10 \log_{10} NF = 2\text{dB}$$

$$\log_{10} NF = 0.2$$

$$NF = 10^{0.2}$$

$$\text{Noise temperature} = (F - 1) T_o$$

$$= (10^{0.2} - 1) 290 = 169.36 \text{ K}$$

$$\text{Noise power i/p} = k T_e B$$

$$= 1.38 \times 10^{-23} \times (169.36 + 50) \times 12 \times 10^6$$

$$\text{Noise power at o/p} = (3.632 \times 10^{-14}) \times 10^4$$

$$= 3.73 \times 10^{-10} \text{ watts}$$

53. Two lossless X-band horn antennas are separated by a distance of 200λ . The amplitude reflection coefficients at the terminals of the transmitting and receiving antennas are 0.15 and 0.18, respectively. The maximum directivities of the transmitting and receiving antennas (over the isotropic antenna) are 18dB and 22dB, respectively. Assuming that the input power in the lossless transmission line connected to the antenna is 2 W, and that the antennas are perfectly aligned and polarization matched, the power (in mW) delivered to the load at the receiver is ____

53. Ans: 2.99

Sol: Given

Lossless horn antennas

$$\eta_T = \eta_R = 1$$



Power Gain = Directivity

Directivity of Txing antenna, $D_T = 18$ dB

$$10 \log D_T = 18$$

$$G_T \text{ (or) } D_T = 63.09$$

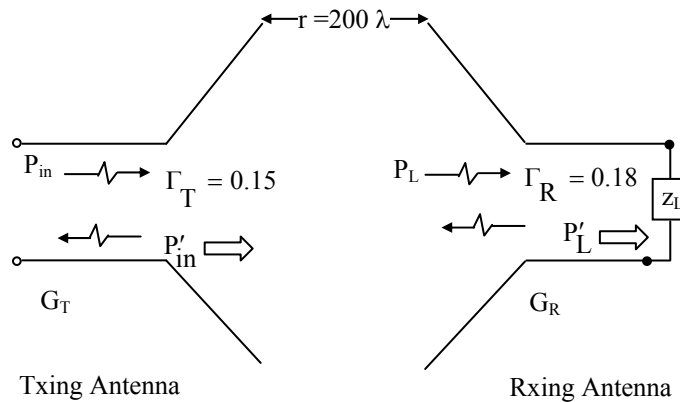
Directivity of Rxing antenna, $D_R = 22$ dB

$$10 \log D_R = 22$$

$$G_R \text{ (or) } D_R = 158.48$$

input power $P_{in} = 2$ W

Spacing, $r = 200 \lambda$



Friis transmission formula is given by

$$P_L = G_T G_R \left[\frac{\lambda}{4\pi r} \right]^2 P'_{in}$$

where:

P'_{in} : Input power (prime indicates power due to reflection)

$$\begin{aligned} P'_{in} &= |1 - \Gamma_T|^2 P_{in} \\ &= |1 - (0.15)^2| \times 2 \end{aligned}$$

$$P'_{in} = 1.955 \text{ W}$$

$$P_L = 63.09 \times 158.48 \left[\frac{\lambda}{4\pi \times 200 \lambda} \right]^2 \times 1.955$$

$$= 3.1 \times 10^{-3}$$

As there is a reflection at the terminals of Rxing antenna power delivered to the load is given by

$$\begin{aligned} P'_L &= \{1 - |\Gamma_R|^2\} \times P_L \\ &= \{1 - (0.18)^2\} \times 3.1 \times 10^{-3} \end{aligned}$$

$$\therefore P'_L = 2.99 \text{ mW}$$



54. The electric field of a uniform plane wave travelling along the negative z direction is given by the following equation:

$$\vec{E}_w = (\hat{a}_x + j\hat{a}_y)E_0 e^{jkz}$$

This wave is incident upon a receiving antenna placed at the origin and whose radiated electric field towards the incident wave is given by the following equation:

$$\vec{E}_a = (\hat{a}_x + 2\hat{a}_y)E_1 \frac{1}{r} e^{-jkr}$$

The polarization of the incident wave, the polarization of the antenna and losses due to the polarization mismatch are, respectively,

- (A) Linear, Circular (clockwise), -5dB
(C) Circular (clockwise), Linear, -3dB

- (B) Circular (clockwise), Linear, -5dB
(D) Circular (anti clockwise), Linear, -3dB

54. Ans: (C)

Sol: Given

Electric field of incident wave is

$$\vec{E}_w = (\hat{a}_x + j\hat{a}_y)E_0 e^{jkz}$$

at $z = 0$;

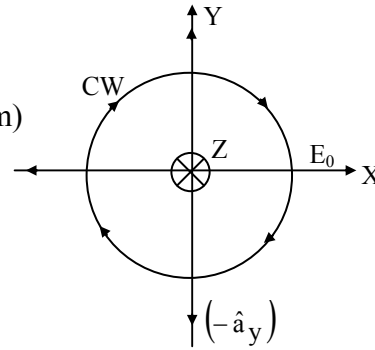
$$\vec{E}_w = E_0 \cos \omega t \hat{a}_x - E_0 \sin \omega t \hat{a}_y \text{ (in time varying form)}$$

at $\omega t = 0$

$$\vec{E}_w = E_0 \hat{a}_x$$

at $\omega t = \frac{\pi}{2}$

$$\vec{E}_w = E_0 (-\hat{a}_y)$$



As a tip of electric field intensity is tracing a circle when time varies, hence the wave is said to be circularly polarized in clockwise direction (or) RHCP. Polarizing vector of incident wave is given by,

$$\hat{P}_i = \frac{\hat{a}_x + j\hat{a}_y}{\sqrt{2}}$$

radiated electric field from the antenna is

$$\vec{E}_a = (\hat{a}_x + 2\hat{a}_y)E_1 \frac{1}{r} e^{-jkr}$$

at $r = 0$

$$\vec{E}_a = E_1 \cos \omega t \hat{a}_x + 2E_1 \cos \omega t \hat{a}_y \text{ (in time varying form)}$$

As both x & y components are in-phase, hence the wave is said to be linear polarized. Polarizing

vector of radiated field is $\hat{P}_a = \frac{(\hat{a}_x + 2\hat{a}_y)}{\sqrt{5}}$ polarizing mismatch; The polarizing mismatch is said to

have, if the polarization of receiving antenna is not same on the polarization of the incident wave. The polarization loss factor (PLF) characterizes the loss of EM power due to polarization mismatch.

$$PLF = |\hat{P}_i \cdot \hat{P}_a|^2$$



in dB; PLF (dB) = 10 log (PLF)

$$\text{PLF} = \left| \left(\frac{\hat{a}_x + j\hat{a}_y}{2} \right) \cdot \left(\frac{\hat{a}_x + 2\hat{a}_y}{\sqrt{5}} \right) \right|^2 = \left| \frac{1+j2}{\sqrt{2}\sqrt{5}} \right|^2 = \frac{1}{2} \text{ (or) } 0.5$$

$$\text{PLF(dB)} = 10 \log 0.5 = -3.0102$$

55. The far-zone power density radiated by a helical antenna is approximated as:

$$\vec{W}_{\text{rad}} = \vec{W}_{\text{average}} \approx \hat{a}_r C_0 \frac{1}{r^2} \cos^4 \theta$$

The radiated power density is symmetrical with respect to ϕ and exists only in the upper hemisphere: $0 \leq \theta \leq \frac{\pi}{2}$; $0 \leq \phi \leq 2\pi$; C_0 is a constant. The power radiated by the antenna (in watts)

and the maximum directivity of the antenna, respectively, are

- (A) $1.5C_0$, 10dB (B) $1.256C_0$, 10dB
(C) $1.256C_0$, 12dB (D) $1.5C_0$, 12dB

55. Ans: (B)

Sol: Given

Power density radiated by the antenna

$$\vec{W}_{\text{rad}} = \frac{C'_0}{r^2} \cos^4 \theta \hat{a}_r \text{ W/m}^2$$

Power radiated (or) average power radiated by the antenna is given by

$$\begin{aligned} P_{\text{rad}} &= \oint_s \vec{W}_{\text{rad}} \cdot d\vec{s} \\ &= \int_{\theta=0}^{\pi/2} \int_{\phi=0}^{2\pi} \frac{C'_0}{r^2} \cos^4 \theta r^2 \sin \theta d\theta d\phi \quad (\because \text{radiated only in the upper hemisphere}) \\ &= C'_0 (2\pi) \frac{1}{5} \end{aligned}$$

$$\therefore P_{\text{rad}} = 1.256 C'_0 \text{ Watt}$$

Maximum directivity of the antenna is given by

$$D = 4\pi \frac{U_{\text{max}}}{P_{\text{rad}}}$$

$$U = r^2 W_{\text{rad}}$$

$$U = r^2 \times \frac{C'_0}{r^2} \cos^4 \theta$$

$$U = C'_0 \cos^4 \theta$$

$$U_{\text{max}} = C'_0$$

$$\therefore D = \frac{4\pi C'_0}{1.256 C_0}$$

$$= 10$$

$$D_{\text{(dB)}} = 10 \log 10$$

$$\therefore D = 10 \text{ dB}$$