

# Random Processes

- Stationary RP  $\begin{cases} \text{WSS} \\ \text{SSS} \end{cases}$

- Ergodicity

- Stationary (WSS) RP

↳ Autocorrelation function

$$R_x(\tau) = E[X_t X_{t+\tau}]$$

1.  $R_x(0) = E[X_t^2]$   $\rightarrow$  Energy  $\rightarrow \infty$   
(Not an energy signal)

$\int_{-\infty}^{\infty} x^2(t) P_{\text{noise}} dt$   $\rightarrow$  Average Power

$\int_{-T/2}^{T/2} E[X_t^2] dt \rightarrow T \times \underbrace{E[X_t^2]}_{\text{does not depend on 't' } (\because \text{WSS})}$

$R_x(0)$   $\rightarrow$  Average Power (Time duration)

2.  $G_x(\omega) := F\{R_x\}$  ?

a.  $R_x(0) \geq 0$       b.  $R_x(\tau) = R_x(-\tau), \forall \tau.$   
 $\Rightarrow$  even function

c. If  $\{x_t\}_{t \in I}$  is real valued, then  
 $R_x(\tau) \in \mathbb{R}, \forall \tau.$

$$G_X(\omega) = \int_{-\infty}^{\infty} R_X(\tau) e^{-j\omega\tau} d\tau$$

$$G_X(\omega) = \int_{-\infty}^{\infty} R_X(\tau) e^{j\omega\tau} d\tau$$

$$s = -\tau, \quad ds = -d\tau$$

$$\tau = -\infty, \quad s = \infty$$

$$\tau = \infty, \quad s = -\infty$$

$$= - \int_{\infty}^{-\infty} R_X(-s) e^{-j\omega s} ds$$

$$= \int_{-\infty}^{\infty} R_X(s) e^{-j\omega s} ds = G_X(\omega)$$

$$\Rightarrow G_X(\omega) \in \mathbb{R}$$

$$2. G_X(\omega) = G_X(-\omega), \quad \forall \omega \in \mathbb{R}$$

$$3. G_X(\omega) \geq 0, \quad \forall \omega \in \mathbb{R} \text{ [Pending]}$$

$G_X(\omega) \rightarrow$  even  
 $\rightarrow \mathbb{R}$   
 $\rightarrow$  non-negative  
 $\rightarrow$  func of  $\omega$  (frequency)

$G_X(\omega)$  is the Power Spectral Density of  $X_t$

$G_X(\omega) \rightarrow$  Quantify the amount of noise present in a communication channel.

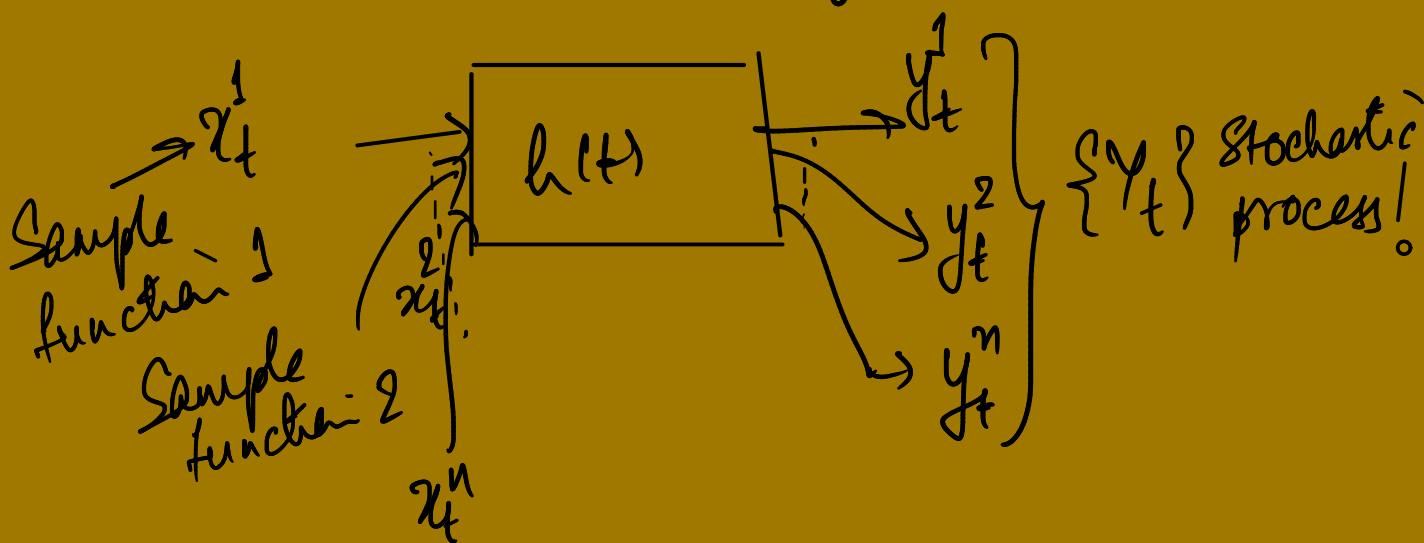
# Stochastic Processes and LTI systems



$$y(t) = \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau = \int_{-\infty}^{\infty} h(t-\tau) x(\tau) d\tau$$

$$= (h * x)(t) = (x * h)(\tau).$$

What happens if the i/p to the LTI system is a Stochastic Process  $\{X_t\}$ ?



Characterize  $\{Y_t\}$  in terms of  $\{X_t\}$  and  $h(t)$ .

① Assume that  $\{X_t\}$  is WSS.

$$\begin{aligned} m_Y(t) &= E[Y_t] = E\left[\int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau\right] \\ &= \int_{-\infty}^{\infty} h(\tau) \underbrace{E[x(t-\tau)]}_{m_X} d\tau \end{aligned}$$

$$= m_x \int_{-\infty}^{\infty} h(\tau) d\tau$$

$$m_y(t) = m_x h(0) \quad [H \rightarrow \text{freq. response}]$$

$\Rightarrow m_y$  does not depend on 't'.

$$R_y(t_1, t_2) = E[Y_{t_1} Y_{t_2}]$$

$$= E \left[ \int_{-\infty}^{\infty} h(p) x(t_1 - p) dp \int_{-\infty}^{\infty} h(q) x(t_2 - q) dq \right]$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) \underbrace{E[x(t_1 - p) x(t_2 - q)]}_{R_x(t_1 - t_2 - p + q)} dp dq$$

$$R_y(t_1, t_2) = \underbrace{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) R_x(t_1 - t_2 - p + q) dp dq}_{f(t_1 - t_2)}$$

$R_y$  is a function only of  $t_1 - t_2$  rather than the absolute values of  $t_1$  &  $t_2$ .

∴  $\{Y_t\}$  is also a WSS SP!!!

$$R_y(\tau) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) R_x(\tau - p + q) dp dq$$

$$G_y(\omega) \sim G_x(\omega)$$

In case of deterministic signals  $x$  &  $y$ ,  
 $\underline{Y(\omega)} = H(\omega) \underline{X(\omega)}$ ,  $\forall \omega \in \mathbb{R}$

$$G_Y(\omega) = \mathcal{F}\{R_Y\}$$

$$= \int_{-\infty}^{\infty} \left( \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) R_X(\tau - p + q) dp dq \right) e^{-j\omega\tau} d\tau$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) \left( \int_{-\infty}^{\infty} R_X(\tau - p + q) e^{-j\omega\tau} d\tau \right) dp dq$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(p) h(q) G_X(\omega) e^{j(p-q)\omega} dp dq$$

$$= \underbrace{\int_{-\infty}^{\infty} h(p) e^{jp\omega} dp}_{H^*(\omega)} \underbrace{\int_{-\infty}^{\infty} h(q) e^{-jq\omega} dq}_{H(\omega)} \cdot G_X(\omega)$$

$$= H^*(\omega) H(\omega) \cdot G_X(\omega)$$

(if  $h$  is real valued)

$$\boxed{G_Y(\omega) = |H(\omega)|^2 G_X(\omega)}$$

$$0 \leq R_Y(0) = E[Y_t^2] = \int_{-\infty}^{\infty} G_Y(\omega) d\omega = \int_{-\infty}^{\infty} |H(\omega)|^2 G_X(\omega) d\omega$$

$$\int_{-\infty}^{\infty} |H(\omega)|^2 G_X(\omega) d\omega \geq 0 \rightarrow (1)$$

$H(\omega)$  is the Freq. response of any real valued LTI system.

$$H(\omega) = \begin{cases} 1, & \omega_0 \leq \omega \leq \omega_0 + \Delta\omega \\ 0, & \text{else.} \end{cases} \rightarrow (2)$$

Substitute (2) in (1) to get

$$\int_{\omega_0}^{\omega_0 + \Delta\omega}$$

$$G_X(\omega) d\omega \geq 0$$

for any arbitrary  $\omega_0$  and arbitrarily small  $\Delta\omega$ .

$$\Rightarrow G_X(\omega) \geq 0, \forall \omega \in \mathbb{R}.$$

[Completes the justification for interpreting  $G_X(\omega) \triangleq \mathcal{F}\{R_X\}$  as the PSD of  $X_t$ ]

\* Ideal filters:

Low pass filter:  $H(\omega) = \begin{cases} 1 & \omega \leq \omega_0 \\ 0 & \omega > \omega_0 \end{cases}$

Are ideal filters realizable?