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multi level ASK -  
(a) Now suppose our gate is

$\begin{array}{ccccccccc} 0 & 0 & 0 & 1 & 1 & 0 & 0 & 0 & 1 \\ \hline 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 0 \end{array}$   
 $\begin{array}{ccccccccc} 0 & 0 & 0 & 1 & 1 & 0 & 0 & 0 & 1 \\ \hline 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 & 0 \end{array}$



(b) Here BW of the channel shall be  $B_{BW} = \text{band rate}$  as band rate change is dominant factor

(c) Suppose if Band width allocated =  $x$

$$\text{Digital bit rate} = x \times \log_2^04 = x \times 2$$

(d) If  $V = \text{no. of Amplitude levels}$

Then Digital Bit rate

$$= \text{Bandwidth} \times \frac{\log V}{2}$$

where Band rate =  $BH$ .

(e) Now, so long as bandwidth of channel is fixed.

Band rate is fixed.  $\log_2$

So Digital Bit Rate  $\propto$  ~~(No of levels  $(V)$ )~~  
~~(created)~~

(f) (i) Now suppose the Power Rating of the Modulator

$$= P' \text{ (fixed)} \quad \text{Prop. Const}$$

$$(ii) P \propto c^2 \Rightarrow P = k c^2$$

$$c' = \sqrt{\frac{P'}{R}} \rightarrow \text{Amplitude of corner} \\ \text{is fixed.}$$

(iii) Suppose we create  $V$  no of Amplitude levels.

$$\Delta V \text{ (Difference between successive amplitude levels)} = \frac{c'}{V} -$$

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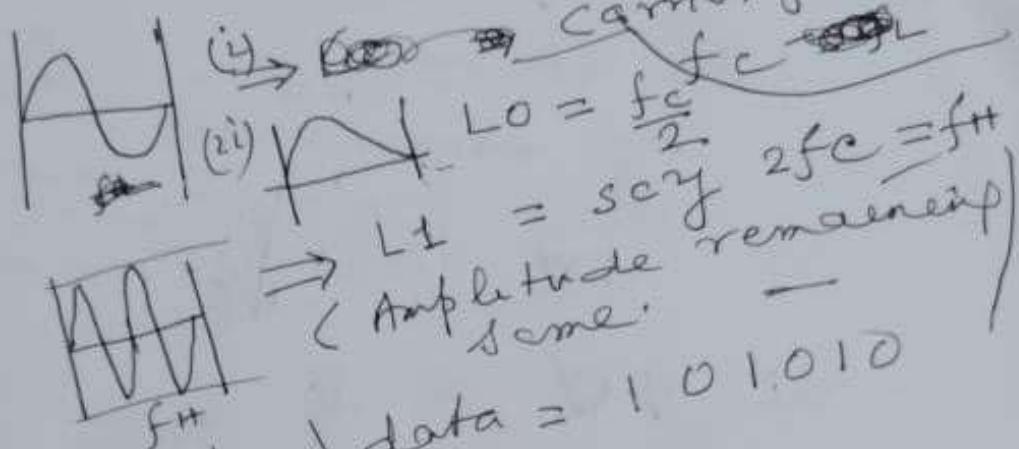
n = 600

- (i) As  $\Delta V_{\text{meas}}$  decreases.
- $\Delta V_{\text{meas}} \propto \Delta V$  decreases.
- (ii) If  $\Delta V$  is too small then it may be corrupted by noise. Then  $\Delta V_{\text{meas}}$  will have a smaller value.  $\Delta V_{\text{meas}} \propto \frac{\Delta V}{N} \times \frac{1}{S} (\text{signal value}) - \text{noise value}$
- (iii)  $\Delta V_{\text{meas}} \propto \frac{\Delta V}{N}$  So we have a maximum value of  $V = \text{no. of levels depending on the noise environment}$  More  $\frac{S}{N}$  value  $\rightarrow$  more  $V$   $\rightarrow$  more  $\Delta V_{\text{meas}}$  But rate can not increase  $V$  beyond limit.

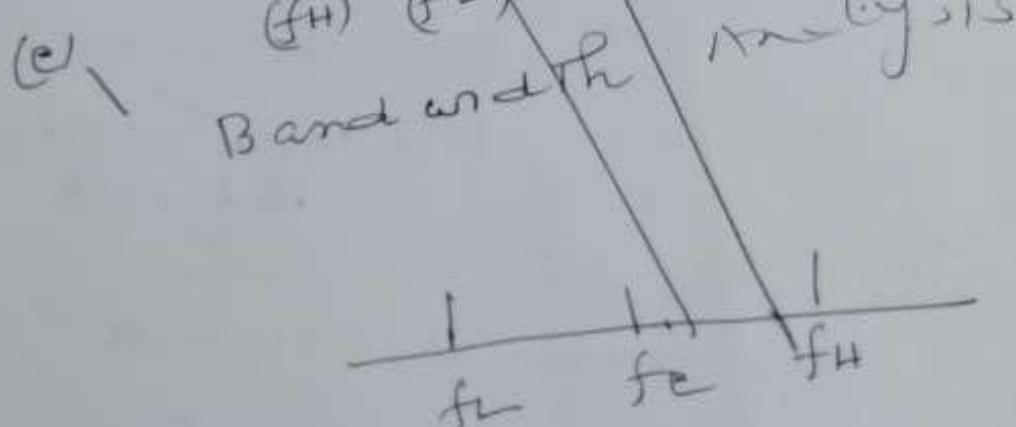
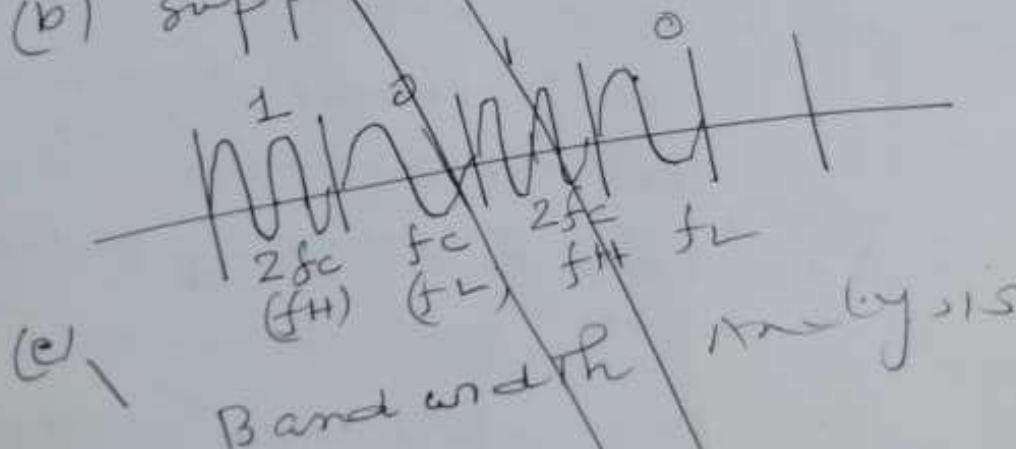
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② Frequency shift keying  
(FSK)

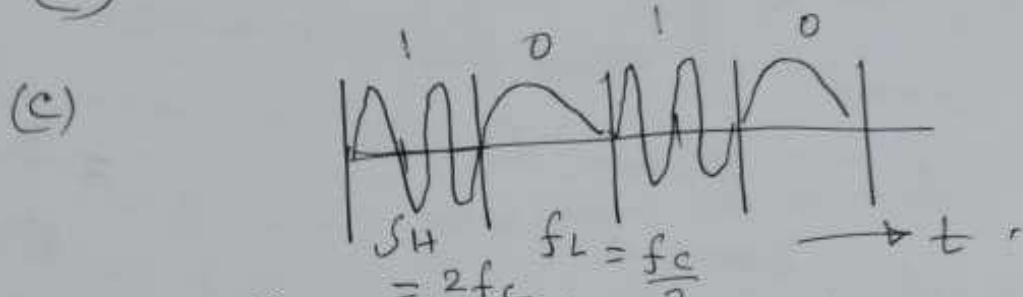
(a) 2 Level FSK: Two frequencies are used.



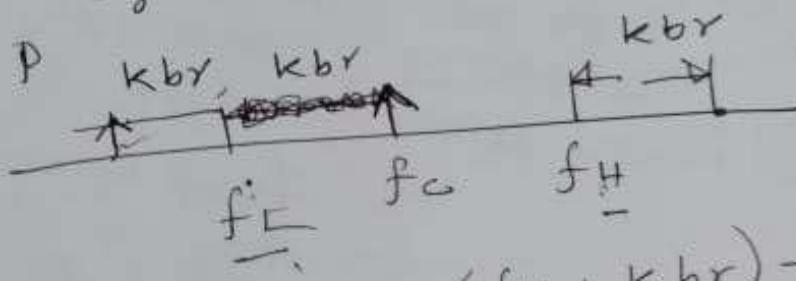
(b) Suppose data = 101010



(b)  $\text{Data} = 101010\ldots$



(c)  $BW \Delta$  channel is above composite analog signal



$$\begin{aligned} BW &= (f_H + k_f b_r) - (f_L - k_f b_r) \\ &= (f_H - f_L) + 2k_f b_r \\ &= b_r + (f_H - f_L) \quad \text{Assume } k_f = \frac{1}{2} \end{aligned}$$

$$\begin{aligned} Br_{fsk} &= BW - (f_H - f_L) \\ &= Br_{ask} \end{aligned}$$

is FSK is not used in digital modulation.

(3) <sup>2 Level</sup> PSK (phase shift keying)

$$f_c(t) = C \sin(2\pi f_c t + \phi_c)$$

Here  $C$  and  $f_c$  constant

$\phi_c$  (phase of the corner)  
is changed by  $L_0 (30^\circ)$   
 $L_1 = 5V^-$

phase ~~corner~~  
~~corner~~ corresponding  
to  $L_0 = 0^\circ$

$$L_1 = 180^\circ (say)$$

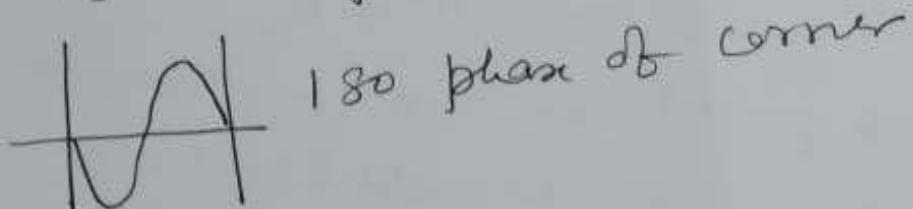
(i) Phase corresponding to  $L_0$

$$\begin{aligned} &= \phi_c + k_p \times 0 \\ &= \phi_c = 0^\circ \quad L_1 \end{aligned} \quad (\text{say } \phi_c = 0)$$

(ii) Phase corresponding to

$$\begin{aligned} (\text{iii}) \text{ signalling element for } L_0 &= \phi_c + k_f \times 5 \\ &= 0 + 180^\circ \quad (\text{by adjusting } k_f) \end{aligned}$$

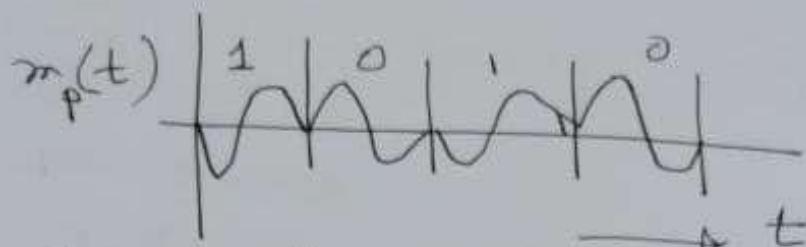
(iv) signalling element for  $L_1$



(v)

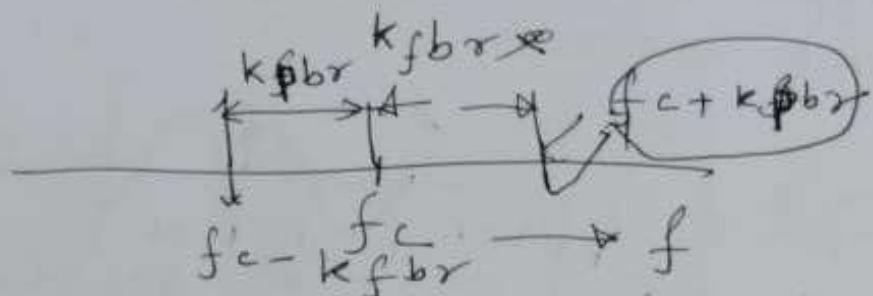
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101010... transmitted



Above is composite signal.

(vi) BW of the channel



BW of channel (PSK)

$$= \text{BW}_{\text{PSK}} = (f_c + k_p br) - (f_c - k_p br)$$

$$= 2 k_p br$$

$$= br \quad (k_p = \frac{1}{2})$$

like ASK.

(vii)

Digital data rate 2 Level PSK

$$= br \times \log_2 \frac{2}{2} \quad (1 \text{ bit/baud})$$

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(22iii) Multi-level PSK - 4 Levels

or 4-PSK.

Four phases say  $\text{bit}^{\text{4PSK}}$

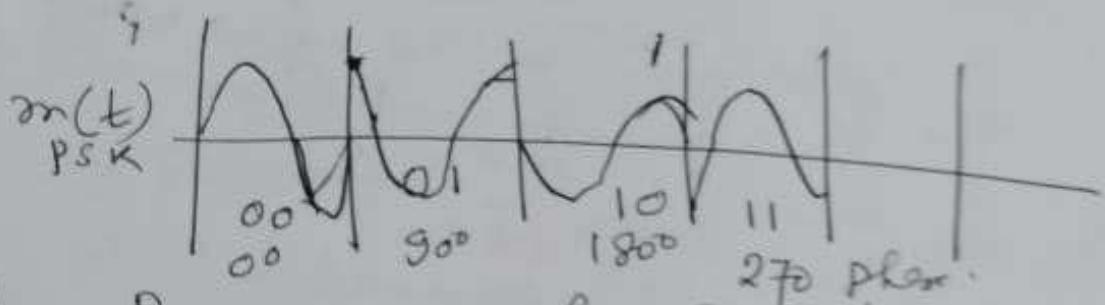
$$0^\circ \Rightarrow 00$$

$$90^\circ \Rightarrow 01$$

$$180^\circ \Rightarrow 10$$

$$270^\circ \Rightarrow 11$$

Suppose we require to send  
bit stream  $00\ 01\ 10\ 11\ 00\dots$ .



(ix)

Band width of channel

will be same as 2-level PSK primarily as it will depend on band rate (effect of phase change for four phase level and for 4-PSK and effect of phase change for 2-PSK is negligible)

(x)

$$\text{Digital Data Rate PSK} = \text{Br} \times \log_2 V$$

$V = \text{No. of phase levels}$

Q9  
\* (x) Limit of  $V$ :

Since noise does not effect Phase  $V_{PSK}$  can be much than  $V_{ASK}$ .

$$2 - PSK \Rightarrow V = 2$$

$$4 - PSK \Rightarrow V = 4$$

$$8 - PSK$$

$$16 - PSK \Rightarrow$$

Common.

④ (i) QAM we combine ASK and PSK to make QAM (Quadrature Amplitude Modulation)

(ii) Suppose Maximum Amplitude level for ASK for a particular noise environment

(iii) Maximum Phase level (depends on the Phase detector detector of Demodulator)  $= V_P = 8$

(iv)  $V = V_A \times V_P = 4 \times 8 = 32$  will create 32-QAM.

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(v) Here assume  $V_A = 2 \left( A/2 \text{ and } A \right)$   
 $V_P = 4 \left( P_1, P_2, P_3, P_4 \right)$   
 Constellation pattern is denoted by  $\begin{pmatrix} 0, 90, 180, \\ 270 \end{pmatrix}$

each point is pair:

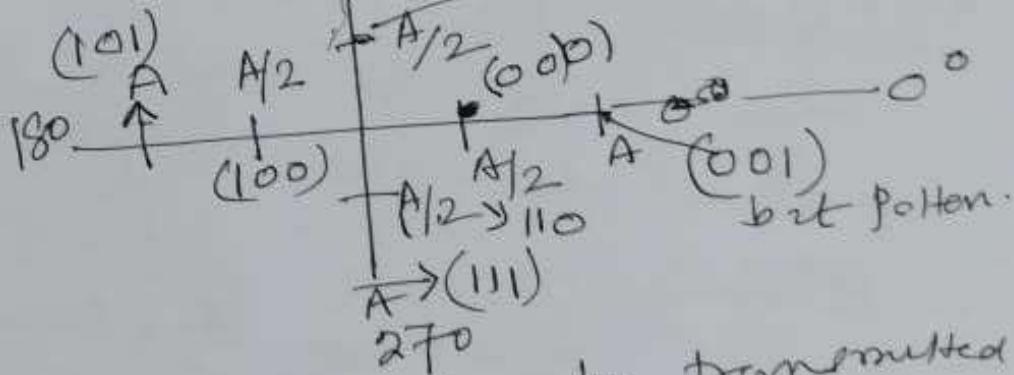
$$(A, P) \text{ ordered pair: } (A, P_1), (A, P_2)$$

$$(A/2, P_1), (A/2, P_3)$$

$$(A, P_2), (A/2, P_3)$$

$$(A/2, P_4), (A, P_4)$$

$$(A/2, P_4), (A, P_4)$$



Now 3 bits can be transmitted  
by one signaling element

(vi) Data Rate  $\rightarrow$   $\text{OSNR} \approx V = 8$  (VA  $\times$  VP)  
 $\text{BW} = B_r \otimes$