Paper reading(18) - CL-686

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Research paper

Title: Model Predictive Idle Speed Control: Design, Analysis, and Experimental Evaluation

Idle Speed: It is the rotational speed of the engine when vehicle is not in motion.

2 controllers: Single input(throttle) and multi-input(throttle and spark timing).

Design Flow

- MPC prediction model is developed, model type and order of model are design parameters selected.
- States are augmented, The cost function weights and the prediction horizon are the design parameters selected. And obtained controller is simulated to evaluate the performance and the relation between closed-loop performance and tuning parameters.
- The MPC controller is synthesized in a suitable way for implementation in automotive ECUs. Computational feasibility of the controller is verified.
- Experimental tests are performed to validate the nominal and non-nominal performance of the closed-loop system

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Engine Model for ISC

$$\dot{N}(t) = \frac{1}{J} \frac{30}{\pi} (M_{\rm e}(t) - M_L(t))$$

$$M_e(t) = \kappa_{\rm spk}(t - t_{ds}) M_{\rm e,\delta}(t - t_d) - M_{\rm fr}(t) - M_{\rm pmp}(t)$$

$$(12b)$$

$$\kappa_{\rm spk}(t) = (\cos(\alpha(t) - \alpha_{MBT}(t)))^{\xi}$$

$$(12c)$$

$$t_d(t) = \frac{60}{N(t)}$$

$$\dot{M}_{\rm e,\delta}(t) = -\gamma_2 \frac{RT_{\rm im}}{V_1} \frac{N(t)}{\gamma_1} M_{\rm e,\delta}(t) + \gamma_2 \frac{RT_{\rm im}}{V_{\rm im}} \gamma_3 \vartheta(t)$$

$$-\frac{\gamma_0 \cdot \gamma_1}{N^2} \dot{N}(t)$$

$$(12e)$$

Figure: Engine model from 1st principle neal idle speed

Engine Model for ISC(Contd...)

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Where, N[RPM]: engine speed, M_e[Nm], M_L[Nm]: engine brake torque, load torque on crankshaft, k_{spk}, \alpha[rad]: torque ratio, spark ignition angle, M_{e,\delta}(t)[Nm]: indicated engine torque, v[deg]: throttle position
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Control oriented model

From the above model,

To remove nonlinearity - control variable is torque ratio k_{spk} Control inputs are throttle position and torque ratio achieved via spark retard.

Linearized model: $Y(s) = G_{thr}(s)e^{st_d}U_1(s) + G_{spk}(s)e^{st_{ds}}U_2(s)$ (Direct result from another reference paper) where.

 $y(t)=N(t)-\bar{N},$ $u_1(t)=v(t)-\bar{v},$ $u_2(t)=k_{spk}(t)-\bar{k_{spk}}$ $u_1,$ u_2 are throttle and spark input.

Control oriented model(Contd...)

Discretized state space model

$$x_{thr}(k+1) = A_{thr}x_{thr}(k) + B_{thr}u_{thr}^{\delta}(k)$$
$$y_{thr}(k) = C_{thr}x_{thr}(k)$$
(1)

$$x_{spk}(k+1) = A_{spk}x_{spk}(k) + B_{spk}u_{spk}^{\delta}(k)$$
$$y_{spk}(k) = C_{spk}x_{spk}(k)$$
 (2)

where, $k \in \mathbb{Z}$, $x_{thr}, x_{spk} \in \mathbb{R}^2$

 $u_{spk}^{\delta}(k),u_{thr}^{\delta}(k)$ are delay free torque ratio and throttle command.

Control oriented model(Contd...)

By cascading a fourth-order airflow delay model with (1), and a first-order spark delay model with (2), the complete linear model of the engine is obtained.

$$x^{p}(k+1) = A^{p}x^{p}(k) + B^{p}u^{p}(k)$$

$$y_{p}(k) = C^{p}x^{p}(k)$$
(3)

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where, for multi input controller

$$\begin{split} \mathbf{x}^p &= \begin{bmatrix} \mathbf{x}_{thr}^f \\ \mathbf{x}_{spk}^f \end{bmatrix}, u^p = \begin{bmatrix} u_{thr} \\ u_{spk} \end{bmatrix}, A^p = \begin{bmatrix} A_{thr}^f & 0 \\ 0 & A_{spk}^f \end{bmatrix}, \\ \mathbf{B}^p &= \begin{bmatrix} B_{thr}^f \\ B_{spk}^f \end{bmatrix}, C^p = \begin{bmatrix} C_{thr}^f & C_{spk}^f \end{bmatrix}, \\ \mathbf{x}_{thr}^f &\in \mathbb{R}^6, \mathbf{x}_{spk}^f \in \mathbb{R}^3 \text{ and } u^p \in \mathbb{R}^2 \end{split}$$

for single input controller

$$x^{p} = x_{thr}^{f}, u^{p} = u_{thr}, A^{p} = A_{thr}^{f}, B^{p} = B_{thr}^{f}, C^{p} = C_{thr}^{f}$$

MPC design

Started with single input (throttle) controller adding the dynamics $q_N(k+1)=q_N(k)+T_s(y_p(k)-r_{y_p}(k))$ where $q_N\in\mathbb{R}$ is the discrete-time integral of the output. $r_{y_p}\in\mathbb{R}$ is the desired speed set-point offset.

After adding this to eqn (3) of single input controller, state becomes

$$x = \begin{bmatrix} x_p \\ q_N \end{bmatrix}, x \in \mathbb{R}^7$$

$$A = \begin{bmatrix} A_p & 0 \\ T_s C_p & 1 \end{bmatrix}, B = \begin{bmatrix} B_p \\ 0 \end{bmatrix}, C = \begin{bmatrix} C_p & 0 \\ 0 & 1 \end{bmatrix}$$
(4)

Constraints and Cost function

Output constraints: $-200 \le y_p(k) - r_{y_p}(k) \le 200$

Input Constraints: $0 \leq u + u_{FF} \leq 10$, where $u_{FF} = \bar{v}$ and

$$v(t) = u_{FF} + u(t)$$

Cost function:

$$J(y(k), u(k), u(k-1)) = \sum_{i=0}^{h-1} (y(i|k) - r_y)^T Q(y(i|k) - r_y) + \Delta u(i|k) S \Delta u(i|k)$$
(5)

where, (i|k) indicates i steps ahead from k,

 $h \in \mathbb{Z}$ is prediction horizon,

y(k) and u(k) are output and input sequences from 0 to h-1 predicted at step k, $r_v = [r_{v_0}0]^T$ is the output setpoint,

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 $\Delta u(i|k) = u(i|k) - u(i-1|k)$, Q and S are positive definite matrices.

Optimal control problem

$$\min_{\sigma,u(k)} \rho\sigma^{2} + \sum_{i=0}^{h-1} (y(i|k) - r_{y})^{T} Q(y(i|k) - r_{y}) + \Delta u(i|k)^{T} S \Delta u(i|k)$$
s.t.
$$x(i+1|k) = Ax(i|k) + Bu(i|k)$$

$$y(i|k) = Cx(i|k), i = 0, ..., h - 1$$

$$u_{min} \le u(i|k) \le u_{max}, i = 0, ..., h_{u} - 1$$

$$y_{min} - \sigma 1 \le y(i|k) \le y_{max} + \sigma 1, i = 0, ..., h_{c}$$

$$u(i|k) = u((h_{u}) - 1|k), i = h_{u}, ..., h - 1$$

$$\sigma \ge 0$$

$$u(-1|k) = u(k-1), x(0|k) = x(k)$$
(6)

The output constraints are "softened" by the additional optimization variable $\sigma \in \mathbb{R}$ and $\rho > 0$

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Redefining the constraints

The constraints values are computed by the software running in the vehicle ECU.

For single input,

$$y_{min} = \begin{bmatrix} -200 + r_{y_p}(k) \\ -\infty \end{bmatrix}$$

$$u_{min} = -u_{FF}$$

$$y_{max} = \begin{bmatrix} 200 + r_{y_p}(k) \\ +\infty \end{bmatrix}$$

$$u_{max} = 10 - u_{FF}$$

$$(7)$$

For multi-input,

$$y_{min} = \begin{bmatrix} -200 + r_{y_p}(k) \\ -\infty \\ -\infty \end{bmatrix}, u_{min} = \begin{bmatrix} -u_{FF} \\ -0.1 \end{bmatrix}$$

$$y_{max} = \begin{bmatrix} 200 + r_{y_p}(k) \\ +\infty \\ +\infty \end{bmatrix}, u_{max} = \begin{bmatrix} 10 - u_{FF} \\ \Delta k_{spk}^{max}(k) \end{bmatrix}$$
(8)

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Kalman filter

For the idle speed controllers, a stationary Kalman filter has been used for estimating x^p

$$x^{p}(k|k-1) = A^{p}x^{p}(k-1) + B^{p}u^{p}(k-1) x^{p}(k) = x^{p}(k|k-1) + K_{KF}(y_{p}(k) - C^{p}x^{p}(k|k-1))$$
 (9)

where K_{KF} is the stationary Kalman filter gain

 $K_{KF} = PC^{p^T}(C^pPC^{p^T} + R_{KF})^{-1}$ and P is the solution of the algebraic Riccati equation

 $P = A^p P A^{p^T} + QKF - A_p P C^{p^T} (C^p P C^{p^T} + RKF)^{-1} C^p P A^{p^T}$. And Kalman filter is tuned by Q_{KF} and R_{KF} , that represents here a process

Kalman filter is tuned by Q_{KF} and R_{KF} , that represents here a process and measurement noise covariances.

Closed loop simulations

Defined parameters are, $\bar{N} = 600 r/min$, h = 30, $h_c = 3$, $h_u = 3$ $\epsilon_N(t) = \hat{N}(t) - N(t)$, Torque disturbance is of 20Nm

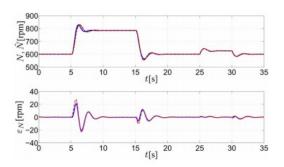


Figure: Upper plot: engine speed vs time, Lower plot: errot in control speeds vs time, nonlinear model (dashed line),continuous-time linearized model (solid line), discrete-time linearized model (dashed-dotted line)

Closed loop simulations(Contd...)

For single-input MPC, spark retard k_{spk} was kept constant and throttle v was the only available degree of freedom

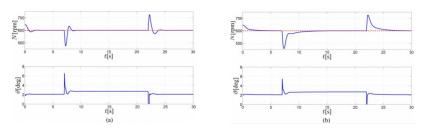


Figure: Single-input MPC with nonlinear model

Part (a) - aggressive tuning - Underdamped behaviour Part (b) - slower overdamped behavior

Closed loop simulations(Contd...)

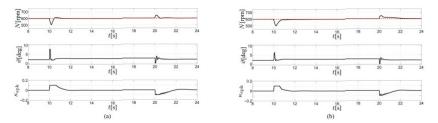


Figure: Multi-input MPC with nonlinear model

In both cases the multi-input controller achieve better disturbance rejection compared to the corresponding single-input MPC in terms of maximum deviation from the setpoint and settling time.