PIMRC 2016: Practical Examples of Physical Layer Security

Arsenia Chorti, Univ. Essex

4 September 2016, Valencia

How it looks from outside

"All you need to make a movie is a girl and a gun"

Jean-Luc Godard

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"All you need to do information theory is a log and a lim"

Serio Verdú

The usual suspects



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The usual suspects



Openness of wireless channel: security more challenging

- Eavesdropping
- Denial of service attacks (jamming)

Two flavors of PLS

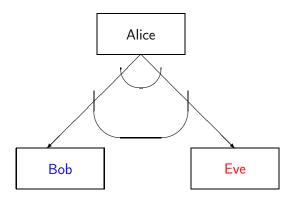
Generate keys from correlated sources through public discussion

Two flavors of PLS

- Generate keys from correlated sources through public discussion
- ② Build encoders for degraded adversarial links: "hide" part of the message in noise

What are the hurdles for PLS to gain trust?

1. The adversaries are not powerful enough (passive)



What real systems look like!

Sport US election Daily Edition

Lifestyle > Tech > News

There are officially more mobile devices than people in the world

The world is home to 7.2 billion gadgets, and they're multiplying five times faster than we are

Zachary Davies Boren | @zdboren | Tuesday 7 October 2014 | 🗇 0 comments









2. The adversary does not have enough observations / diversity

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Antenna vs. cost survey: gains from 7 dBi to 20 dBi for \$25- \$200! [1]

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What about colluding eavesdroppers?

3. Practical issues: Gaussian signalling, long-length encoders

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What about QAM systems?

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What about QAM systems? What about the short codelength regime? What about resource constrained devices?

4. Combining secure systems systems does not guarantee security!

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It is not! [2]

Relevance of PLS in Future Networks (5G, IoT, M2M)

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- 1000-fold increase in throughput (peak)
- Reduced latency < 1 ms (especially for tactile internet)
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- Advanced services (smart*),
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- Massive multi-antenna systems (MIMO), BF and AN
- Small cells: natural PLS setting
- Full duplex: existence of structured interference
- Ad hoc networks, IoT, D2D: secret keys on the fly

Generating secret key from fading coefficients in the presence of active adversaries

Generating keys from correlated sequences

Wireless shared randomness techniques:

Generating keys from correlated sequences

Wireless shared randomness techniques: Generate purely random keys in 3 steps

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Advantage distillation

Generating keys from correlated sequences

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 - Advantage distillation
 - Information reconcilliation

Generating keys from correlated sequences

Wireless shared randomness techniques:

Generate purely random keys in 3 steps

- Advantage distillation
- Information reconcilliation
- Privacy amplification

How good are the PLS keys?

NIST TEST	P-Value
Monobit Frequency	0.739918
Block Frequency	0.739918
Cumulative Sums	0.534146
Runs	0.739918
Longest Run	0.350485
Binary Matrix Rank	0.213309
FFT	0.911413
Non-overlapping Template	0.911413
Overlapping Template	0.534146
Maurer's Universal Test	0.122325
Approximate Entropy	0.739918
Serial	0.739918
Linear Complexity	0.122325

Physical layer authenticated encryption

Replace RSA by PLS key generation in ISO PKE protocol

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- ① PLS key seed SeedGen : $\mathbb{R}^+ \to (0,1)^L \times (0,1)^I$ with (seed,coset) = SeedGen(\mathbf{H}_0)
- ② Sem. sec. hash function $G: (0,1)^L \times (0,1)^I \to (0,1)^{\lfloor H(\text{seed}) \rfloor}$ with output key K=G(seed, coset), K={Ke, Ki}
- $\begin{tabular}{ll} \hline \textbf{3} & Sem. sec. A.E.(encrypt-then-MAC): enc. alg. \\ & cipher=E(Ke, m), dec. alg. m=D(Ke, cipher), sign alg. \\ & t=Sign(Ki, c), ver. alg. v=Ver(Ki, c, t), v \in \{\ c, \bot\} \\ \end{tabular}$

Physical layer authenticated encryption

Replace RSA by PLS key generation in ISO PKE protocol

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- ③ Sem. sec. A.E.(encrypt-then-MAC): enc. alg. cipher=E(Ke, m), dec. alg. m=D(Ke, cipher), sign alg. t=Sign(Ki, c), ver. alg. v=Ver(Ki, c, t), v ∈ { c, \bot }

Alice transmits extended ciphertext C = [coset||cipher||t]

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$$Y_{A,i} = \sqrt{p_i}H_i + \sqrt{\gamma_i}G_{A,i} + Z_{A,i}, i = 1, ..., N$$

 $Y_{B,i} = \sqrt{p_i}H_i + \sqrt{\gamma_i}G_{B,i} + Z_{B,i}, i = 1, ..., N, i = 1, ..., N$

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$$\begin{array}{lll} Y_{A,i} & = & \sqrt{p_i} H_i + \sqrt{\gamma_i} G_{A,i} + Z_{A,i}, i = 1, \dots, N \\ Y_{B,i} & = & \sqrt{p_i} H_i + \sqrt{\gamma_i} G_{B,i} + Z_{B,i}, i = 1, \dots, N, \ i = 1, \dots, N \end{array}$$

$$\begin{split} &C(p_i,\gamma_i) = I(Y_{A,i};Y_{B,i}) = \frac{1}{2}\log_2\left(1 + \frac{\sigma_H^2 p_i}{N_{A,i} + N_{B,i} + \frac{N_{A,i}N_{B,i}}{\sigma_H^2 p_i}}\right), \\ &\text{with} \quad N_{A,i} = 1 + \sigma_A^2 \gamma_i, \quad N_{B,i} = 1 + \sigma_B^2 \gamma_i \end{split}$$

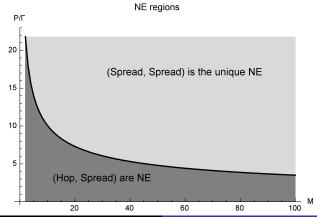
Probabilities $\alpha_i, \beta_i, i \leq N$ for legitimate users and jammer respectively to hop on channel i for Probabilities $\alpha_{N+1}, \beta_{N+1}$ to spread

$$u(\alpha, \beta) = \sum_{i=1}^{N} \left\{ \alpha_i (1 - \beta_i - \beta_{N+1}) C(NP, 0) + \alpha_i \beta_i C(NP, N\Gamma) + \alpha_i \beta_{N+1} C(NP, \Gamma) + \alpha_{N+1} (1 - \beta_i - \beta_{N+1}) C(P, 0) + \alpha_{N+1} \beta_i C(P, N\Gamma) + \alpha_{N+1} \beta_{N+1} C(P, \Gamma) \right\}.$$

Jammer should always spread

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 When unfavorable conditions hop, when favorable conditions spread

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MQAM encoders for the x + y channel

x + y channels

- Wireless network coding systems
- Full duplex

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Creating channel advantage

x + y channels

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Creating channel advantage ... using structured interference

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Creating channel advantage

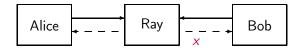
... using structured interference

Gaussian signalling does not offer this opportunity











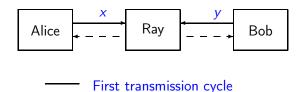
Physical layer network coding ctd'

In PNC we need 2 cycles



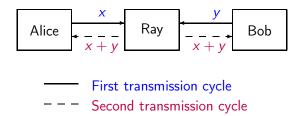
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- Codewords $\in \mathcal{X}_A$, $\mathcal{X}_B \subset \mathcal{C}$ with $M_A = |\mathcal{X}_A|$, $M_B = |\mathcal{X}_B|$, $m_A = \log_2 M_A$, $m_B = \log_2 M_B$



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- Labeling functions $b_A: \mathcal{X}_A \to \mathcal{S}_A \cup \{\varepsilon\}$, $b_B: \mathcal{X}_B \to \mathcal{S}_B \cup \{\varepsilon\}$ where ε represents the empty string



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- $\forall s_A \in \mathcal{S}_A \cup \{\varepsilon\}$ we set $b_A^{-1}(s_A) = \{x_A \mid b_A(x_A) = s_A\}$ and $\forall s_B \in \mathcal{S}_B \cup \{\varepsilon\}$ we set $b_B^{-1}(s_B) = \{x_B \mid b_B(x_B) = s_B\}$

Transmission cycles

First Transmission Cycle

- First Transmission Cycle
 - Stochastic encoders $\varphi_A(s_A), \varphi_B(s_B)$ uniformly distributed over $b_A^{-1}(s_A)$ and $b_B^{-1}(s_A)$

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$$Y = X_A + X_B + W$$

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$$Y = X_A + X_B + W$$

- Second Transmission Cycle
 - Ray transmits Z, Alice and Bob observe

$$Y_A = Z + W_A,$$

$$Y_B = Z + W_B$$

W, WA, WB AWGN noise

Upper Bounds on Secrecy Rates

Perfect Secrecy w.r.t. Ray if

- $I(Y; S_A) = 0$, perfect secrecy condition for Alice
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Proposition 1

The perfect secrecy rates are bounded by $R_A^s \leq \widehat{R}_A^s$ and $R_B^s \leq \widehat{R}_B^s$,

$$\widehat{R}_A^s = I(X_A; Y_B | X_B) - I(X_A; Y) + \delta_A$$

$$\widehat{R}_{B}^{s} = I(X_{B}; Y_{A}|X_{A}) - I(X_{B}; Y) + \delta_{B}$$

with
$$\delta_A = H(X_A|Y_B, X_B)$$
 and $\delta_B = H(X_B|Y_A, X_A)$

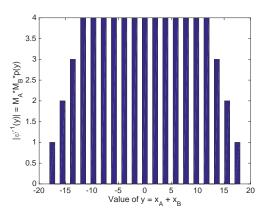


Figure: Pmf of Ray's observation $y = x_A + x_B$ in the noiseless scenario for a 4-PAM modulator and a 16-PAM modulator.

Upper bound in M-PAM systems

Theorem 1

In the noiseless setting with unit channel gains when Alice and Bob employ M_A -PAM and M_B -PAM modulators with $M_B \geq M_A$, we have

$$\widehat{R}^s = m_A \frac{M_B - M_A + 1}{M_B} + \frac{2}{M_A M_B} \sum_{a=1}^{M_A - 1} a \log_2(a).$$

In particular, for fixed M_A we have $\lim_{M_B \to \infty} \widehat{R}^s = m_A$.

Bob's Secret Subset

• Define a subset \mathcal{X}_B^s of \mathcal{X}_B on which Bob will transmit secret bits

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Encoder Construction at Bob cnt'd

Bit Labeling

• Bit labeling for \mathcal{X}_B : perfect binary tree with edges alternately labeled

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- x_B given bit labeling $I(x_B)$ tracing the tree downwards

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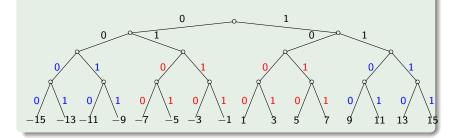
- Bit labeling for \mathcal{X}_B : perfect binary tree with edges alternately labeled
- x_B given bit labeling $I(x_B)$ tracing the tree downwards
- Bob's labeling function $b_B: \mathcal{X}_B \to \mathcal{S}_B \cup \{\varepsilon\}$ defined by

$$b_B(x_B) = \begin{cases} \text{ the last } m_A \text{ bits of } I(x_B) & x_B \in \mathcal{X}_B^s \\ \varepsilon & x_B \notin \mathcal{X}_B^s \end{cases}$$

Bob's Encoder Example

Example:
$$M_A = 4$$
, $M_B = 16$, $\mathcal{X}_B^s = \{-7, \dots, +7\}$, $\mathcal{X}_B = \{-15, \dots, +15\}$

Bob transmits $m_A=2$ bits on \mathcal{X}_B^s with rate $R_B^s=m_A\frac{M_B-2M_A}{M_B}$



$$M_A = 4$$
 and $M_B = 16$

Bob has public and secret bit queues

$$Q_B^p = 10110101, \quad Q_B^s = 1111$$

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- $Q \in \mathcal{X}_B^s \Rightarrow 11$ from $Q_B^s \Rightarrow x_B = +7$ transmitted
- 3 and so on...

time
$$t = 1$$
 $1011 \rightarrow +7 = x_B(1)$

time
$$t = 2$$
 $1101 \rightarrow +11 = x_B(2)$

time
$$t = 3$$
 0111 $\rightarrow -1 = x_B(3)$

Alice's Bit Labeling

With side info, Alice's bit labeling function

$$b_A: \mathcal{X}_A \times \mathcal{X}_B \to \mathcal{S}_A \cup \{\varepsilon\}$$

where \mathcal{X}_A and \mathcal{X}_B M_A -PAM, M_B -PAM and

$$b_A(x_A, x_B) = \begin{cases} I(x_A) & \text{if } |\psi^{-1}(x_A + x_B)| = M_A \\ \varepsilon & \text{otherwise} \end{cases}$$

Theorem 3

Suppose that $M_B \geq M_A$. For the above encoder, $I(S_A; Y) = 0$, $R_A^s = m_A \frac{M_B - M_A + 1}{M_B}$ and for fixed M_A , $\lim_{M_B \to \infty} R_A^s = m_A$.

Example of Alice's Encoder Construction

$$M_A = 4$$
, $M_B = 8$ $Q_A^s = 1011$, $x_B(1) = 7$, $x_B(2) = -5$, $x_B(3) = +5$

		ХB								
		-7	- 5	-3	-1	+1	+3	+5	+7	
	+3	00	00	00	00	00	ε	ε	ε	
$X_{\mathcal{A}}$	+1	ε	01	01	01	01	01	ε	ε	
	-1	ε	ε	11	11	11	11	11	ε	
	-3	ε	ε	ε	10	10	10	10	10	

Alice's encoder

time
$$t = 1$$
 $x_B(1) = +7, \Rightarrow x_A(1) = -3$
time $t = 2$ $x_B(2) = -5, \Rightarrow x_A(2) = \varepsilon$
time $t = 3$ $x_B(3) = +5, \Rightarrow x_A(3) = -1$

Effect of Synchronization Errors

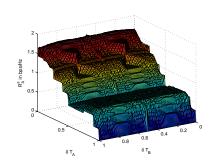
 Synchronization errors ⇒ misalignment of Alice's and Bob's symbols at Ray

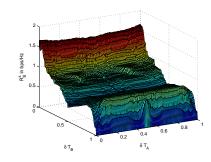
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Advantage in population size

Cooperative multiuser setting

• Network with M parallel Rayleigh subchannels, K legitimate users and E eavesdroppers

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Probability of secrecy outage

Non-cooperative scenario

$$P_{out}^{(nc)}(K, E, \tau)K\Gamma(K) \sum_{n=1}^{E} (-1)^{n+1} {E \choose n} \frac{\Gamma(n2^{-\tau} + 1)}{\Gamma(K + n2^{-\tau} + 1)}$$

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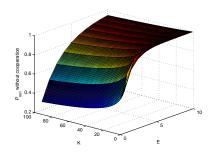
$$P_{out}^{(nc)}(K, E, \tau)K\Gamma(K) \sum_{n=1}^{E} (-1)^{n+1} \binom{E}{n} \frac{\Gamma(n2^{-\tau} + 1)}{\Gamma(K + n2^{-\tau} + 1)}$$

• Fully cooperative scenario (virtual MIMO)

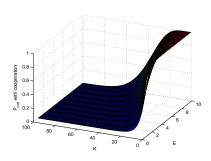
$$P_{out}^{(co)}(K, E, \tau)1 - \frac{\sum_{n=0}^{K-1} {K+E-1 \choose n} 2^{n\tau}}{(1+2^{\tau})^{K+E-1}}$$

$au = 1 ext{ bit/sec/Hz}$

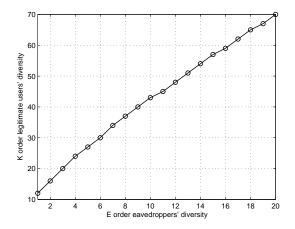
Non-cooperative case $P_{out}^{(nc)}$



Cooperative case $P_{out}^{(co)}$



Required minimum number of K versus E to upper-bound the $P_{out}^{(co)} \leq 1\%$ for $\tau=1$ bit/sec/Hz



Conclusions

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- Mature topic: wireless channels can be used to establish secret keys
- Emerging topic: systems with structured interference

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- Mature topic: wireless channels can be used to establish secret keys
- Emerging topic: systems with structured interference
- Synergy between information theoretic security and crypto necessary
- Nested structures?
- New opportunities: user centric adaptive security

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gracias!

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